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# **Electrical characteristics and trap signatures for Schottky barrier diodes on 4H-SiC, GaN-on-GaN, AlGaN/GaN epitaxial substrates**

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#### **Abstract**

The forward and reverse current transport mechanisms, temperature dependence of Schottky barrier height (SBH) and ideality factor, barrier inhomogeneity analysis, and trap parameters for Schottky barrier diodes (SBDs) fabricated on 4H-SiC, GaN-on-GaN and AlGaN/GaN epitaxial substrates are reported. High SBH is identified for Ni/4H-SiC (1.31 eV) and Ti/4H-SiC (1.18 eV) SBDs with a low leakage current density of  $< 10^{-8}$  A/cm<sup>2</sup> at -200 V. Thermally stimulated capacitance (TSCAP) detects the well-known  $Z_{1/2}$  electron trap at  $E_C - 0.65$  eV in both 4H-SiC SBDs, while an additional deep-level trap at  $E_C - 1.13$  eV is found only in Ni/4H-SiC SBDs. The vertical Ni/GaN SBD exhibits a promising SBH of 0.83 eV, and two electron traps at  $E_C$  – 0.18 eV and  $E_C$  – 0.56 eV are identified from deep-level transient Fourier spectroscopy (DLTFS). A peculiar two-diode model behaviour is detected at Metal/GaN/AlGaN/GaN interface of high-electron mobility transistor (HEMT); the first diode (SBH-1 of 1.15 eV) exists at the standard Metal/GaN Schottky junction, whereas the second diode (SBH-2 of 0.72 eV) forms due to the energy difference between the AlGaN conduction band and the heterojunction Fermi level. The compensational Fe-doping-related buffer traps at  $E_C$  – 0.5 eV and  $E_C$  – 0.6 eV are determined in the AlGaN/GaN HEMT, through the drain current transient spectroscopy (DCTS) experiments.

Keywords: Schottky barrier diode, 4H-SiC, GaN, HEMT, current transport, traps.

#### **1. Introduction**

A good quality SBD demands large SBH, high forward current with low voltage drop, low reverse leakage current, high breakdown voltage, and minimal trapping in the epilayer. In practice, the SBH value is determined by the metal work function and the Fermi level pinning effect [1-4]. Low SBH  $( $0.6 \text{ eV}$ ) can only be obtained with Si-based SBDs, resulting$ in high leakage current, low breakdown voltage  $(V_R > -100 V)$ , and off-state power dissipation [5]. For this reason, silicon PiN diodes are preferred for power rectifier applications over SBD structures [5]. On the other hand, high breakdown voltage  $(V_{BR} > 1 \text{ kV})$ , large SBH ( $\Phi_B > 1.3 \text{ eV}$ ), and low leakage current density can be attained using a simple 4H-silicon carbide (4H-SiC) SBD structure, due to the wide bandgap (3.26 eV), high breakdown field, and superior thermal conductivity of 4H-SiC. The 4H-SiC SBD rectifiers have already replaced the silicon PiN diodes up to the voltage range of 5 kV [5]. Ni-based Schottky contacts are widely used on the n-type semiconductors because of its high metal work function (5.15 eV), low price, high thermal stability, good adhesion, and stable nickel silicide formation. Ti-based Schottky contacts on 4H-SiC have also shown promising results ( $\Phi_B$  > 1.1 eV) for rectifier applications [6]. Commercial semiconductor foundries like ON Semiconductor employ Tibased Schottky contacts in their SiC SBD structure [7]. Since Ni- and Ti-based Schottky contacts are utilized in 4H-SiC devices technologies, the Ni/4H-SiC and Ti/4H-SiC SBDs are used for this study.

Gallium nitride (GaN) is another attractive wide bandgap semiconductor (3.4 eV) suitable for mid-range power applications (currently up to 600 V) [8]. The availability of GaN-on-GaN homoepitaxial substrates has given momentum to the device community for developing vertical GaN power devices. Accordingly, vertical Ni/GaN SBD properties are investigated in this work. The AlGaN/GaN HEMT devices are recommended for the RF and microwave electronic systems [8]. Nevertheless, the gate leakage current induces off-state and standby power dissipation, limited gate voltage swing, and undermines the transistor reliability. The gate leakage current must be minimized to improve the off-state HEMT performance [9]. So, the gate current-voltage  $(I_G - V_G)$ characteristics of the AlGaN/GaN HEMT are analyzed at *VDS*  $= 0$  V condition, to understand the gate leakage transport.

Although extensive reports are available on SBD electrical characteristics [9-15], this work systematically investigates the electrical properties of the SBDs fabricated on the different epitaxial substrates (4H-SiC, GaN-on-GaN, and GaN/AlGaN/GaN on SiC) as follows: The current transport processes responsible for the forward and reverse current density are determined by fitting the experimental data. The spatial homogeneity of the barrier is evaluated through temperature-dependent SBH and ideality factor (*n*). The effective SBH is computed from the conventional and modified Richardson plots. *T*<sub>0</sub> anomaly ( $n = 1 + T_0/T$ ) is estimated for non-ideal SBDs. Specifically, a peculiar twodiode model characteristic is detected at metal/GaN/AlGaN/GaN interface of the HEMT. So, this study may provide a complete understanding of the SBD electrical properties of the emerging SiC and GaN semiconductors.

The electrically active traps in the SBD induce charge trapping and detrapping effects during the diode operation. Notably, the traps with deeper energy and higher concentration could reduce the diode forward current, and increase the on-state voltage drop, series resistance (decreasing mobility), and leakage current, thereby degrading

the SBD properties [5, 6]. Thus, the crystal growth and device scientific communities should know the trap signatures (energy level, trap concentration, and capture cross-section) in device to improve the epilayer crystalline quality. In this work, traps in the Ni/4H-SiC SBD, Ti/4H-SiC SBD, Ni/GaN SBD, and AlGaN/GaN HEMT are identified by various characterizations (TSCAP / DLTFS / DCTS). The reported trap parameters can be used to model the trapping-induced changes in the electrical characteristics of the SBDs in commercial device simulators.

#### **2. Experiment**

#### *2.1 SBD and HEMT details*

The fabrication steps for 4H-SiC SBDs, and Ni/GaN SBDs are described elsewhere [16, 17]. The schematic crosssectional diagrams for (a) Ni/4H-SiC SBD, (b) Ti/4H-SiC SBD, (c) vertical Ni/GaN SBD, and (d) GaN/AlGaN/GaN HEMT are depicted in Figure 1. The room-temperature epitaxial layer resistivity of 4H-SiC and GaN-on-GaN wafers is estimated 6.25  $\Omega$ -cm, and 0.45  $\Omega$ -cm, respectively.

In Ni/4H-SiC SBDs, 200 nm thick Ni metal layer was chosen for top Schottky contact, and Ti/Au (50 nm/150 nm) bilayer was used for bottom Ohmic contact. After Ni deposition, the samples underwent for thermal annealing at 450 °C for 30 minutes in Ar ambient. In Ti/4H-SiC SBDs, the same Ti/Au (50nm/150 nm) bilayer was employed for the Schottky contact on the lightly-doped epilayer, as well as the Ohmic contact on the highly-doped back surface. Both the contacts were not annealed in Ti/4H-SiC SBDs. The active area of the Ni/4H-SiC and Ti/4H-SiC SBDs was  $3.8 \times 10^{-2}$  cm<sup>2</sup>.

In vertical Ni/GaN SBDs, Ni/Au (40 nm/200 nm) bimetal layer was considered for the top Schottky contact (200 µm diameter) and the Ti/Al/Ni/Au (30 nm/180 nm/40 nm/200 nm) metal stack was employed for the back Ohmic contact. 200 nm thick Au contact pad was placed on top contact. After metal depositions, no annealing was done in the Ni/GaN SBDs. The active area of the Ni/GaN SBD was  $3.14 \times 10^{-4}$  cm<sup>2</sup>.



**Figure 1.** Schematic cross-section for (a) Ni/4H-SiC SBD, (b) Ti/4H-SiC SBD, (c) Ni/GaN SBD, and (d) GaN/AlGaN/GaN HEMT.

The HEMT device was developed on GaN/AlGaN/GaN heteroepitaxial layer grown on a semi-insulting SiC substrate. The Fe-doping was incorporated into the buffer layer. The Ni/Au bilayer was utilized for the gate Schottky contact, and the regular Ti/Al/Ni/Au-based Ohmic contact was selected for the source and drain electrodes. The silicon nitride (SiN) passivation layer was incorporated in the ungated region of the AlGaN/GaN HEMT structure. The gate length (*LG*) and width  $(W_G)$  of the HEMT were 0.15  $\mu$ m and 200  $\mu$ m, so the active area (*A*) of the Schottky gate diode is  $3 \times 10^{-7}$  cm<sup>2</sup>.

#### *2.2. Electrical characterization*

At first, the forward and reverse *I*-*V* characteristics of the SBDs were measured at room temperature. Then, the *I*-*V* experiments were conducted at different temperatures (from 300 K to 450 K) by varying the sample chuck temperature of the probe station, i.e. current-voltage-temperature (*I*-*V*-*T*) characterization. The gate current-voltage (*IG*-*VG*) characteristics of the HEMT were measured at zero drain voltage ( $V_{DS} = 0$  V), to investigate the gate diode properties.

#### *2.3. Trap characterization*

The traps in the Ni/4H-SiC and Ti/4H-SiC SBDs were identified by the TSCAP spectroscopy. At a low temperature of 120 K  $(T_f)$ , the majority carrier (electron) traps were filled by forward biasing the SBD. After that, sample temperature was increased, and variation in the diode depletion capacitance was measured at a fixed reverse voltage (-40 V) for temperatures ranging from 120 K to 650 K.

DLTFS procedure was used to characterize the traps in the vertical Ni/GaN SBDs. During the DLTFS thermal scan (40 K to 460 K), the traps in the diode are periodically populated by reducing the reverse bias from -5 V to -0.1 V with 100 μs pulse width. At the end of the filling pulse, capacitance emission transients were recorded at -5 V in regular temperature intervals. Since DLTFS software computes the discrete Fourier coefficients for each transient, the emission time constant of the trap was directly extracted from a single DLTFS thermal scan.

The traps in the AlGaN/GaN HEMT were detected by the isothermal DCTS technique. At a fixed temperature, trap filling was achieved by pulsing  $V_{DS}$  from 10 V to 20 V for 100ms duration. Successively upon the filling pulse, drain current emission transients were measured at 10 V over the time interval of 1 μs to 1s. The isothermal DCTS experiments were performed at five different temperatures (300 K, 325 K, 350 K, 375 K, and 400 K) to calculate the trap signatures using the Arrhenius relation.

#### **3. Results and discussion**

#### *3.1. Ni/4H-SiC SBDs*

The forward current density-voltage  $(J_F - V_F)$  characteristics of the Ni/4H-SiC SBD are shown in Figure 2(a). At low

forward voltages ( $V_F$  < 0.38 V), a substantial potential barrier existing at the metal/semiconductor junction retards the electron emission over the barrier, so tunneling current transport  $(J_{TU})$  is estimated for  $J_F$  up to 0.38 V, based on the following expression [2]

$$
J_{TU} = J_{T0} \left( \frac{q(V_F - I_F R_s)}{\eta E_0} \right) \tag{1}
$$

where  $J_{T0}$  is the tunneling saturation current density,  $\eta$  is the fitting parameter, and  $E_0$  defines the barrier transparency during the tunneling [1]. Beyond  $V_F > 0.38$  V, the electrons may surmount the barrier, so the standard thermionic emission (TE) model  $(J_{TE})$  governs  $J_F$ , as predicted by [1, 11, 15]

$$
J_{TE} = J_s \left[ \exp\left(\frac{q(V_F - I_F R_s)}{nkT}\right) - 1\right]
$$
 (2)

$$
J_s = A^* T^2 \exp\left(-\frac{q\Phi_B}{kT}\right) \tag{3}
$$

where  $V_F$  is the forward voltage,  $I_F$  is the forward current,  $q$  is the electronic charge,  $R_s$  is the series resistance,  $J_s$  is the saturation current density,  $A^*$  is the Richardson's constant  $(146 \text{ A } cm^{-2}K^{-2} \text{ for } 4H-SiC)$  [6], *k* is the Boltzmann constant, and  $T$  is the temperature. The SBH  $(1.31 \text{ eV})$  and n  $(1.15 \text{ at}$ 300 K) values for Ni/4H-SiC SBDs are calculated from linear region of  $ln(J_F)$ - $V_F$  plot, using the TE equations (2)-(3).



**Figure 2.** (a) TE and tunneling components in  $J_F$ - $V_F$  characteristics of Ni/4H-SiC SBD at 300 K. (b) *JF*-*V<sup>F</sup>* at different temperatures (300 K to 450 K) is fitted with the TE model.

Equations (1)-(3) designate that  $J_{TE}$  should augment with the increasing temperature, while the tunneling is a temperature-independent process [1]. Figure 2(b) shows that the TE model essentially controls the  $J_F$  characteristics at higher temperatures  $(T > 350 \text{ K})$  in the entire voltage range. The voltage drop across the SBD  $(V_{ON})$  at 10 mA/cm<sup>2</sup> (0.88 V at 300 K and 0.64 V at 450 K) is found to reduce with the temperature due to the decrease in the built-in barrier potential  $(V_{bi})$ . The increased temperature shifts the Fermi level  $(E_F)$ toward the mid-gap energy due to the augmented intrinsic carrier concentration (*ni*), which increases the semiconductor work function ( $\phi_s$ , energy difference between  $E_F$  and vacuum

level increases) [1]. Consequently, *Vbi* decreases at higher temperatures, as per the Schottky-Mott theory equation  $qV_{bi}$  =  $q$  ( $\phi_m$ - $\phi_s$ ) [1-3], where  $\phi_m$  is the metal work function.



**Figure 3.** (a) Temperature induced variations in SBH and *n* of the Ni/4H-SiC SBDs. (b) The conventional and modified Richardson plots for computing effective SBH (*Φeff*) of Ni/4H-SiC SBDs.



**Figure 4.** (a) SBH ( $\Phi_B$ ) vs. *n* is plotted for Ni/4H-SiC and Ti/4H-SiC SBDs (b) *T*<sup>0</sup> anomaly ( $n = 1 + T_0/T$ ) is computed from  $nkT/q$  vs.  $kT/q$ plot for Ni/4H-SiC and Ti/4H-SiC SBDs.

Figure 3(a) displays the variation in the SBH and n of the Ni/4H-SiC SBDs in the temperature range of 300 K to 450 K. It is noticed that SBH increases with the temperature (1.31 eV at 300 K, 1.37 eV at 375 K, and 1.43 eV at 450 K), as conflicting with the expected SBD properties. In general, the electron affinity  $(\chi)$  of the semiconductor increases with the temperature due to the bandgap narrowing effect [1]. As a result, the SBH should decrease with the temperature, according to the expression  $q\Phi_B = q(\phi_m - \chi)$ . Moreover, the electron kinetic energy upsurges with the temperature, promoting the thermionic field emission (TFE). Hence, the TFE current density  $(J_{TFE})$  may be comparable with the  $J_{TE}$ component at higher temperatures [1, 3]; if this is the case, the ideality factor should increase with the temperature. On the contrary, the ideality factor of the Ni/4H-SiC SBD decreases with increasing temperature  $(1.15 \text{ at } 300 \text{ K}, 1.118 \text{ at } 375 \text{ K})$ and 1.1 at 450 K). Thus, the TE current is predominant (relative to TFE) even at high temperatures in the Ni/4H-SiC

SBDs. However, this temperature dependency of the SBD parameters ( $\Phi_B$  increases and n decreases) is often ascribed to the non-ideal SBD properties [10]. Moreover, spatial variation in the SBH (barrier inhomogeneity) is anticipated along the Ni/4H-SiC Schottky junction [4, 10].

The conventional Richardson's plot  $ln(J\sqrt{T^2})$  vs.  $q/kT$  for Ni/4H-SiC SBD is constructed in Figure 3(b), as per the expression derived from  $J_s$  equation [1]

$$
\ln\left(\frac{J_s}{T^2}\right) = -\frac{q}{kT}\Phi_{eff} + \ln(A^*)
$$
\n(4)

The slope of the conventional Richardson's plot yields the effective SBH of 1.06 eV, which is much lower than the SBH value (1.31 eV) at 300 K, as similar to the observations of Roccaforte *et al.* [10]. This reiterates the non-ideal inhomogeneous Schottky barrier behaviour at the Ni/4H-SiC interface. It should be noted that the conventional Richardson analysis is only applicable for ideal cases [1-3] (temperatureindependent SBH and n) and near-ideal SBDs (SBH decreases and n increases with temperature). For the non-ideal Ni/4H-SiC SBD, modified Richardson's plot is created by including the ideality factor temperature dependence in the equation (4), based on the work of Roccaforte *et al.* [10]

$$
\ln\left(\frac{J_s}{T^2}\right) = -\frac{q}{nkT}\Phi_{eff} + \ln(A^*)
$$
\n(5)

The effective SBH ( $\Phi_{\text{eff}}$  = 1.35 eV) is determined from the slope of  $ln(J_s/T^2)$  vs.  $q/nkT$  plot in Figure 3(b). This  $\Phi_{\text{eff}}$  lies between the room temperature and the high-temperature SBH values (1.31-1.43 eV), supporting the literature report [10].

In general, the low SBH (1.31-1.43 eV) patches are only identified from the *I-V* characteristics [10]. Tung's model [18] suggests that a typical non-ideal Schottky interface contains a spatially-homogeneous large barrier  $(\Phi_B^0)$ , which is calculated from the relation between SBH and n [10]. Figure 4(a) plots the measured SBH vs. n for the Ni/4H-SiC SBDs; The linear interpolation of the plot at  $n = 1$  yields  $\Phi_{B-Ni}^0 = 1.66$ eV. The temperature dependence of n for the inhomogeneous SBD is described by  $n = 1 + T_0/T$  [3, 10].  $T_0$  predicts the temperature dependence of n over a wide temperature range and is referred to as " $T_0$  anomaly" [10]. Hence, a high  $T_0$ corresponds to a highly inhomogeneous barrier distributed along the metal/semiconductor interface. In ideal SBDs, n must be unity even after increasing the temperature, as shown in  $kT/q$  vs.  $kT/q$  reference curve in Figure 4(b). Due to the temperature dependency, *nkT*/*q* vs. *kT*/*q* plot occurs in parallel with the ideal curve [10].  $T_0$  anomaly of 44.6 K is extracted for the Ni/4H-SiC SBDs through the linear regression of the *nkT*/*q* vs. *kT*/*q* plot.

The carrier transport processes responsible for the reverse current density-voltage-temperature  $(J_R - V_R - T)$  characteristics of Ni/4H-SiC SBDs are plotted in Figure 5(a). It is found that the trap-assisted tunneling  $(J_{TAT})$  model nearly tracks  $J_R - V_R$  at 300 K, as estimated by the equation [13, 14]

$$
J_{TAT} = A_{TAT} \exp\left(-\frac{4\sqrt{2m_n} (q\phi_t)^{3/2}}{3q\hbar E_m}\right) \tag{6}
$$

where  $A_{TAT}$  is the TAT model fitting parameter,  $q\phi_t$  is the energy location of the trap at  $E_C - E_T$ ,  $m_n$  is the electron effective mass, *ħ* is the reduced Plank's constant, and *E<sup>m</sup>* is the maximum electric field at the Schottky junction. It is accounted that the TAT tunneling takes place via an electron trap at  $E_C$  – 0.65 eV. In fact, this trap  $Z_{1/2}$  is detected in the Ni/4H-SiC SBDs by TSCAP (discussed later). Hence, the selected trap energy at  $E_C - 0.65$  eV in the TAT model has a physical significance. At higher temperatures, *JR-V<sup>R</sup>* properties of the Ni/H-SiC SBDs are fitted by the Poole-Frenkel emission (PFE) model (*JPFE*) [1, 9]

$$
J_{PFE} = A_{PFE} E_m \exp\left(\frac{-q(\phi_t - \sqrt{qE_m/\pi\varepsilon_s})}{kT}\right) \tag{7}
$$

where  $A_{PFE}$  is the PFE model fitting parameter, and  $\varepsilon_s$  is the dielectric constant of the semiconductor. The PFE model (with the same trap at  $E_C - 0.65$  eV) has agreed well with the experimental  $J_R$ - $V_R$  data at 350 K and 400 K, as seen in Figure 5(a). So, it is stated that the trap-related carrier transport models such as TAT and PFE primarily decide the  $J_R - V_R - T$ properties of Ni/4H-SiC SBDs.

The measured TSCAP spectroscopy for Ni/4H-SiC SBD is plotted in Fig. 6(a). It is noticed that the depletion capacitance of the SBD increases with the temperature due to the reduction in *Vbi* [1]. Relative to the without trap filling case, two increasing capacitance steps (T1 and T2) are noticed in the TSCAP spectrum obtained upon the trap filling at *Tf*. The electron emission from a trap at  $E_C$  -  $E_T$  (majority carrier trap in n-type semiconductor) is responsible for the rising capacitance steps in Fig. 6(a) [19, 20]. The mid-step temperature  $(T_{1/2})$  is extracted for each TSCAP step, and the trap activation energy  $(E_a)$  is computed by [19, 20]

$$
E_a = kT_{1/2} \ln \left[ \frac{vkT_{1/2}^2}{q(E_T + 2kT_{1/2})} \right]
$$
 (8)

where *v* is the escape frequency factor ( $v = \sigma_n N_C v_{th}$ ),  $\sigma_n$  is the trap capture cross-section,  $N_c$  is the effective density of states in the conduction band, and  $v_{th}$  is the thermal velocity. The energy level of traps T1 and T2 is found to be  $E_C - 0.65$  eV, and E<sub>C</sub> – 1.13 eV, with  $\sigma_n = 5 \times 10^{-16}$  cm<sup>2</sup> (T1) and  $3 \times 10^{-18}$  cm<sup>2</sup> (T2). The trap concentration  $(N_T)$  for T1 ( $\sim 10^{13}$  cm<sup>-3</sup>) and T2  $(1.3\times10^{13} \text{ cm}^3)$  is estimated by the following equation [19, 20]

$$
N_T \approx 2N_D \frac{\Delta C}{C_R} \tag{9}
$$

where  $N_D$  is the epilayer doping concentration,  $\Delta C$  is the magnitude of the TSCAP signal step, i.e.,  $C_2 - C_1$ , and  $C_R$  is the reverse bias depletion capacitance. The electron trap T1 at  $E_C$  – 0.65 eV might be created due to the omnipresent  $Z_{1/2}$ defect in the n-type 4H-SiC [6]. The other trap T2 at  $E_C - 1.13$ eV may be related to the EH5 defect state [6, 21].



**Figure 5.** Reverse current transport mechanisms (TAT and PFE) in (a) Ni/4H-SiC SBD, and (b) Ti/4H-SiC SBD are identified by fitting the experimental  $J_R-V_R$  data.



**Figure 6.** Measured TSCAP of (a) Ni/4H-SiC SBD, (b) Ti/4H-SiC SBD shows two traps at  $E_C$  – 0.65 eV (T1) and  $E_C$  – 1.13 eV (T2).

#### *3.2. Ti/4H-SiC SBDs*



**Figure 7.** (a) TE and tunneling current components in  $J_F - V_F$ properties of Ti/4H-SiC SBD at 300 K. (b) *JF*-*V<sup>F</sup>* at different temperatures (300 K to 450 K) is fitted with the TE model.

Figure 7(a) depicts the  $J_F$ - $V_F$  characteristics of Ti/4H-SiC SBD at 300 K. The fitting analysis shows that the tunneling is responsible for  $J_F$  at low  $V_F$  < 0.1; after that, the standard TE fully governs  $J_F$ . The SBH and n of the Ti/4H-SiC are computed as 1.18 eV, and 1.17, respectively. Compared to the Ni/4H-SiC SBDs, forward  $J_{TU}$ , and  $V_{ON}$  at 10 mA/cm<sup>2</sup> (0.76V) at 300 K) are lesser in the Ti/4H-SiC SBDs, which may be due to the relatively smaller SBH. The  $J_F$ - $V_F$ -T characteristics of the Ti/4H-SiC SBDs also follow the TE theory, as observed in Figure 7(b). Similar to Ni/4H-SiC SBDs, *VON* at 10 mA/cm<sup>2</sup> decreases with the temperature (0.66 V at 350 K, and 0.48 V at 450 K) because of the reduction in *Vbi*.

Alike Ni/4H-SiC SBDs, SBH of the Ti/4H-SiC SBD increases with the temperature (1.18 eV at 300 K, 1.225 eV at 375 K, and 1.275 eV at 450 K), as noted from Figure 8(a). Likewise, n decreases with the temperature (1.17 at 300 K, 1.137 at 375 K, and 1.11 at 450 K), specifying the barrier inhomogeneity at the Ti/4H-SiC interface [4, 10]. The effective SBH (1 eV) identified from the conventional Richardson plot is lower than the value attained at 300 K, as perceived from Figure 8(b). Therefore, the effective SBH of  $\Phi_{\text{eff}} = 1.3$  eV is calculated from the modified Richardson plot shown in Figure 8(b). The effective SBH (1.3 eV) is higher than the theoretical SBH (1.23 eV) estimated based on the Schottky-Mott theory equation [1-3], by taking the electron affinity of 4H-SiC as 3.1 eV and the Ti work function as 4.33 eV. This observation reveals that the SBH of the Ti/4H-SiC SBD is decided by the Fermi level pinning effect induced by the surface and interface states [1-3]. Indeed, the Fermi level pinning has positively impacted the Ti/4H-SiC interface by increasing the SBH more than the theoretical prediction. The spatially-uniform large barrier ( $\Phi_{B-Ti}^0 = 1.44 \, eV$ ) is determined from SBH vs. n plot in Figure 4(a).  $T_0$  anomaly (51.2 K) for the Ti/4H-SiC SBDs is computed from *nkT*/*q* vs. *kT*/*q* plot in Figure 4(b).  $T_0$  value is slightly higher than the Ni/4H-SiC SBDs, signifying that Ti/4H-SiC interface has a high degree of barrier inhomogeneity [10].



**Figure 8.** (a) Temperature dependent (300-450 K) SBH and n values for Ti/4H-SiC SBDs. (b) The conventional and modified Richardson plots for the Ti/4H-SiC SBDs.

Figure 5(b) illustrates that the PFE model (*J<sub>PFE*</sub>) fairly predicts  $J_R$ - $V_R$  of Ti/4H-SiC SBDs at different temperatures

(300 K to 400 K). The TSCAP in Fig. 6(b) reveals only a single trap T1 at  $E_C$  – 0.65 eV ( $Z_{1/2}$ ) in Ti/4H-SiC SBDs, so the deep-level trap T2 is not detected. Overall, the Ti/4H-SiC SBDs have shown promising results such as reasonably high SBH of 1.18 eV, low  $V_{ON}$ , and low  $J_R$  (on par with Ni/4H-SiC SBDs), for rectifier applications.

#### *3.3. Vertical Ni/GaN SBDs*

The  $J_F$ - $V_F$ -T characteristics of vertical Ni/GaN SBDs are plotted in Figure 9(a). It is found that the standard  $J_{TE}$  controls the forward-biased Ni/GaN SBDs over the temperature range of 293 K to 393 K. Thus, the forward  $J_{\tau U}$  is negligible in the Ni/GaN SBDs. The SBH and n at 293 K are found to be 0.83 eV, and 1.1, respectively. Due to the high  $\gamma = 4$  eV, a low SBH value is obtained for the Ni/GaN SBDs, in comparison with the 4H-SiC SBDs.  $V_{ON}$  (0.36 V at 10 mA/cm<sup>2</sup>) is lower than in 4H-SiC SBDs. As anticipated, *VON* at 10 mA/cm<sup>2</sup> decreases with the temperature due to the decreased  $V_{bi}$ . Conversely,  $V_{ON}$ at 180 A/cm<sup>2</sup> increases with the temperature (the augmented series resistance effect is not visible in the semi-log  $J_F$ - $V_F$  plot) due to the mobility degradation [5].



**Figure 9.** (a)  $J_F$ - $V_F$ -T of the vertical Ni/GaN SBDs are fitted by the TE model. (b) Changes in the SBH and n of the Ni/GaN SBD in the temperature range of 293 K to 393 K.

It is shown in Figure 9(b) that the SBH decreases with the increasing temperature (0.83 eV at 293 K, 0.815 eV at 353 K, and 0.8 eV at 393 K), as opposed to the 4H-SiC SBDs. The temperature-caused reduction in SBH is correlated with the bandgap narrowing effect in the GaN, as per the equation [1]

$$
E_g(T) \approx E_g(0) - \frac{\alpha T^2}{T + \beta} \tag{10}
$$

where  $E_g(T)$  is the temperature dependency of the bandgap,  $E_g$  (*0*) is the bandgap of the semiconductor at 0 K,  $\alpha$  and  $\beta$  are the material-dependent parameters. The temperature-induced changes in the SBH  $(\Delta \Phi_B)$  and the bandgap of GaN  $(\Delta E_g)$  are compared in Figure 10(a). The parameter values of  $E_g$  (0) = 3.47 eV,  $\alpha = 9.09 \times 10^{-4}$  eV/K, and  $\beta = 836$  K are chosen based on the Sentaurus TCAD material parameter files [22]. The temperature-induced variations in  $\Delta \Phi_B$  and  $\Delta E_g$  follow a

similar trend, signifying that these physical processes (parameter variations) are interrelated, although the values slightly differ. Accordingly, the bandgap narrowing effect of GaN is considered the prime reason for the reduction in the SBH upon increasing the temperature. The ideality factor of the Ni/GaN SBD increases with the temperature (1.1 at 293 K, 1.144 at 353K, and 1.18 at 393 K), possibly due to the notable increase in the forward TFE current density [1, 3].



**Figure 10.** (a) Correlation of temperature-induced changes in SBH (Δ*ΦB*) and bandgap of GaN (Δ*Eg*). (b) Conventional Richardson's plot for Ni/GaN SBDs and the inset shows the SBH vs. n plot.

The effective SBH (0.88 eV) for the Ni/GaN SBD is determined from the conventional Richardson plot in Figure 10(b). As the temperature dependence of the SBH and ideality factor obeys the nearly-ideal behaviour, modified Richardson plot and *T<sup>0</sup>* anomaly analysis are not required for the GaN SBDs. In fact, the modified Richardson plot underestimates the effective SBH (0.81 eV), as n reduces with the temperature.  $\Phi_B^0$  for the Ni/GaN SBD is extracted as 0.87 eV, from the SBH vs. n plot [10] in the inset of Figure 10(b). It is noted that  $\Phi_B^0$  value (0.87eV) closely resembles the effective SBH (0.88 eV) of the Ni/GaN SBDs.



**Figure 11.** (a) The SBL and TFE components in the reverse current density of the Ni/GaN SBDs at 293 K. (b) The TFE model fairly predicts  $J_R$ - $V_R$  in the temperature range of 313 K to 373 K.

Figure 11(a) depicts  $J_R$ - $V_R$  characteristics of the Ni/GaN SBDs at 293 K. It is perceived that the Schottky barrier lowering effect (SBL) contributes to  $J_R$  up to -50 V; beyond that voltage, thermionic field emission (TFE) entirely governs the  $J_R$ - $V_R$  properties. The increase in  $J_R$  due to the SBL effect  $(J<sub>SBL</sub>)$  is expressed by [1-3, 14]

$$
J_{SBL} = A^* T^2 \exp\left(-\frac{q(\Phi_B - \Delta\phi)}{kT}\right) \tag{11}
$$

$$
\Delta \phi = \sqrt{\frac{qE_m}{4\pi\varepsilon_s}}\tag{12}
$$

where Δ*ϕ* is the SBL-induced reduction in SBH. The TFE model ( $J_{\text{TFE}}$ ) is symbolized based on the Sze's equation [1]

$$
J_{TFE} = \frac{A^* T}{k} \sqrt{\pi E_{00} q \left[ V_R + \frac{\Phi_B}{\cosh^2(E_{00}/kT)} \right]}
$$
  
 
$$
\times \exp\left(\frac{-q \Phi_B}{E_0}\right) \exp\left(\frac{qV_R}{\varepsilon}\right)
$$
 (13)

$$
\varepsilon' = \frac{E_{00}}{(E_{00}/kT) - \tanh(E_{00}/kT)}
$$
(14)

The  $J_R$ - $V_R$ -T characteristics (from 293 K to 373 K) are fairly predicted by the TFE model, as illustrated in Figure 11(b). These observations reveal that the TFE is the primary current transport in the reverse-biased Ni/GaN SBDs. In literature, Ren *et al.* [14] demonstrated nearly-ideal *I-V* characteristics for Ni/GaN SBD with  $SBH = 0.97$  eV, and  $n = 1.04$ . The SBH and n values for Au/GaN SBDs are reported [23, 24] in the range of 0.83-0.9 eV, and 1.06-1.15, respectively. The current results suggest that the studied Ni/GaN SBDs used exhibit quasi-ideal characteristics, such as  $SBH = 0.83$  eV,  $n = 1.1$ , and typical temperature dependence of SBH and n.



**Figure 12.** (a) Measured DLTFS for Ni/GaN SBD (first order Fourier sine coefficient b<sub>1</sub> used). (b) Arrhenius plots for traps E1 and E2.

The DLTFS represented in first-order Fourier sine coefficient  $b_1$  for the Ni/GaN SBD is shown in Figure 12 (a). Two positive DLTFS peaks E1 and E2 are produced due to the electron emission from the trap at  $E_C - E_T$  (based on our DLTFS measurement setup) [25]. The Arrhenius plots for E1 and E2 are made in Figure 12(b).  $E_a$ , and  $\sigma_n$  of the traps are computed from the slope and intercept of the Arrhenius plot as per the following expression [26]

$$
\ln(\tau_n T^2) = \frac{E_a}{kT} - \ln\left(\frac{\sigma_n N_C v_{th}}{gT^2}\right)
$$
 (15)

The  $b_1$  signal magnitude ( $\Delta C$ ) for E1 and E2 is extracted [23] and  $N_T$  for the traps is estimated using the equation (9). Accordingly, trap signatures such as  $E_a$ ,  $\sigma_n$ , and  $N_T$  are determined for E1 (E<sub>C</sub> – 0.18 eV,  $1.3 \times 10^{-18}$  cm<sup>2</sup>,  $4 \times 10^{13}$  cm<sup>-3</sup>) and E2 (E<sub>C</sub> – 0.56 eV,  $10^{-15}$  cm<sup>2</sup>,  $8 \times 10^{14}$  cm<sup>-3</sup>). The shallow electron trap E1 at  $E_C$  – 0.18 eV is ascribed to the nitrogenvacancy  $(V_N)$  defect in the GaN epilayer [27]. Other electron trap E2 at  $E_C$  – 0.56 eV may be related to the nitrogen-antisite  $(N_{Ga})$  intrinsic defect in the GaN layers [28].

#### *3.4. GaN/AlGaN/GaN HEMT*

Figure 13(a) shows the forward  $J$ *G*- $V$ *G* characteristics (for  $V_{DS} = 0$  V) of the AlGaN/GaN HEMT at 298 K, 348 K, and 398 K. The fitting analysis reveals the three different regions of operation in the forward  $J_G-V_G-T$  properties. The forward current density is low and nearly unchanged with the applied voltage in region-1 ( $V_G$  < 0.7 V), so the saturation current density is the only possible mechanism in this region. Thereafter, two linear regions are observed in Figure 13(a) with dissimilar slopes, revealing the presence of two nonidentical SBHs at metal/GaN/AlGaN/GaN interface [29, 30]. Note that, the series resistance effect produces a linear region in the linearly scaled *JF-V<sup>F</sup>* characteristics at higher current densities [1,8], but it does not induce a linear segment in the semilog  $J_F$ - $V_F$  and  $ln(J_F)$  vs.  $V_F$  plots. This point confirms that metal/GaN/AlGaN/GaN system has the two-diode model properties, as reported by Chen *et al.*[29] and Greco *et al.*[30].



**Figure 13.** (a) *JG-V<sup>G</sup>* characteristics of AlGaN/GaN HEMT show three different regions of operation during forward bias. (b) Temperature dependence of  $\Phi_{BI}$  and  $n_1$  for the gate diode-1.

In the two-diode model [29, 30], the first diode (SBH-1) corresponds to the regular metal/GaN Schottky barrier junction; while, the second diode (SBH-2) forms due to the energy difference between the AlGaN conduction band edge and the Fermi level of the AlGaN/GaN heterointerface, as illustrated in Figure 14(a). Using the TE model, the SBH at 298 K for the first and second diodes of the HEMT gate is calculated as 1.15 eV ( $\Phi_{B1}$ ), and 0.72 eV ( $\Phi_{B2}$ ), respectively. Theoretical SBH of 1.15 eV is computed for the Ni/GaN cap Schottky barrier gate diode-1, by taking  $\chi_{GAN} = 4$  eV and  $\phi_{m-Ni}$  $= 5.15$  eV. Similarly, the ideality factor for the gate diode-1 and 2 is calculated 1.67 (n<sub>1</sub>), and 4.36 (n<sub>2</sub>). Since the gate diode-2 is associated with the unusual AlGaN/GaN interface, a largely deviated ideality factor of  $4.36$  (n<sub>2</sub>) is attained.



Figure 14. (a) Schematic conduction band diagram at equilibrium shows SBH-1 at metal/semiconductor interface and SBH-2 at the AlGaN/GaN heterointerface. (b) The equivalent circuit for the two diode model associated with the metal/GaN/AlGaN/GaN interface.



**Figure 15.** (a) The conventional and modified Richardson's plots for the Schottky gate diode-1. (b) The SBH-2 ( $\Phi_{B2}$ ) and n<sub>2</sub> variations with the temperature for the gate diode-2.

The two-diode model of Chen *et al.* [29] and Greco *et al.* [30] describes that the diodes are serially connected in a backto-back arrangement at the metal/GaN/AlGaN/GaN interface. The equivalent circuit for the two-diode model of the HEMT gate structure is visualized in Figure 14(b). The applied  $V_G$  to the metal/GaN/AlGaN/GaN system may equal to the sum of the voltage drop across the diode-1 and diode-2 (i.e.,  $V_G = V_I$  $+ V_2$ ) [27, 28]. For 0.7 V <  $V_G$  < 1.25 V, the gate diode-1 is forward-biased due to the applied voltage drop  $V_I$ . The diode-1 current density (*JGD1*) in region-2 can be represented by the TE model equation  $(V_G \gg kT/q)$ , neglecting the series resistance effect [29, 30]

$$
J_{GD1} = A^* T^2 \exp\left(\frac{-q\Phi_{B1}}{kT}\right) \exp\left(\frac{qV_1}{n_1kT}\right) = J_{s1} \exp\left(\frac{qV_1}{n_1kT}\right) \quad (16)
$$

where  $J_{sI}$  is the saturation current density for the diode-1, and  $V_I$  is the voltage drop across the diode-1. In region-3, the

applied  $V_G$  ( $> 1.25$  V) may drop across the diode-2, as a result, the diode-2 is reverse biased. It is considered that the reverse current density of the gate diode-2  $(J_{GD2})$  is roughly equal to the reverse saturation current density [29, 30]

$$
J_{GD2} \approx A^* T^2 \exp\left(\frac{-q\Phi_{B2(V_2)}}{kT}\right) \tag{17}
$$

If the diode-2 is reverse-biased, the resulting current  $(I_{D2})$  is directed from the cathode to the anode terminal of the diode-2. Consequently, the diode-2 current direction  $(I_{D2})$  is similar to that of *ID1*. Hence, a negative sign is not required for the diode-2 current. Since *VG* changes the Fermi level position and modulates the 2DEG density at the AlGaN/GaN heterointerface, the SBH-2  $(\Phi_{B2})$  becomes a bias-dependent. It is worth noting that the applied gate voltage does not modify the SBH-1 at the metal/GaN interface [1]. The SBH-2 voltage dependence,  $\Phi_{B2(V_2)}$ , is symbolized as per the first-order Taylor series expansion [29, 30]

$$
\Phi_{B2(V_2)} = \Phi_{B2(0)} - \left(\frac{\partial \Phi_{B2}}{\partial V_2}\right) V_2 = \Phi_{B2(0)} - \left(\frac{1}{n_2}\right) V_2 \tag{18}
$$

where  $\Phi_{B2(0)}$  is the zero-bias SBH for the diode-2,  $\partial \Phi_{B2}$  /  $\partial V_2$  defines the bias voltage dependency of the SBH-2, and  $V_2$  is the voltage drop across the diode-2. The gate voltageinduced change in the SBH-2  $(\partial \Phi_{B2} / \partial V_2)$  is related to the reciprocal of the diode-2 ideality factor (1/*n2*). Accordingly, it is presumed that the SBH-2 variation is small compared to the  $V_2$  change, thereby resulting in a high n value ( $n_2 = 4.36$ ) for the diode-2. After substituting  $\Phi_{B2(V_2)}$  into equation (17),  $J_{GD2}$ 

expression is rewritten as [29, 30]

$$
J_{GD2} = A^* T^2 \exp\left(\frac{-q\Phi_{B2(0)}}{kT}\right) \exp\left(\frac{qV_2}{n_2 kT}\right) = J_{s2} \exp\left(\frac{qV_2}{n_2 kT}\right) (19)
$$

The above equation is valid only during the reverse bias operation of the diode-2; however this expression is identical to the forward-biased diode-1 equation (TE model) [29, 30]. The diode-2 current also flows from the higher potential to the lower potential of the bias supply, as represented in Figure 14(b). Thus, the forward  $J_G$ - $V_G$  properties of the HEMT gate can be modelled using the TE model with two different SBHs.

Since the gate diode-1 is related to the standard metal/GaN Schottky interface, the temperature dependence of the SBH-1  $(\Phi_{BI})$  and n<sub>1</sub> are discussed first. It is perceived in Figure 13(b) that SBH-1 increases and  $n_1$  decreases with the temperature, indicating the Schottky barrier inhomogeneity at the Ni/GaN cap interface. As anticipated for the non-ideal Schottky junction [10], the conventional Richardson plot undervalues  $(0.8 \text{ eV})$  the effective SBH. Furthermore, Figure 15(a) specifies that the modified Richardson analysis yields an unrealistic effective SBH (3 eV) for the metal/GaN Schottky junction. It is found that the temperature-induced changes in  $n_1$  of the diode-1 do not follow the  $T_0$  anomaly expression. From the SBH-1 vs.  $n_1$  plot (not shown), the spatially-uniform

high barrier for the metal/GaN cap interface is calculated as 1.45 eV, which is much higher than the theoretical SBH (1.15 eV) projected by the Schottky-Mott equation. Figure 15(b) plots the SBH-2 ( $\Phi_{B2}$ ) and n<sub>2</sub> variations with the temperature for the gate diode-2. SBH-2 slightly decreases and  $n_2$  increases with the temperature. The heterojunction Fermi level may move downwards as the temperature increases [1]. In this condition, the SBH-2 should increase with the temperature; a contradictory nature is observed in Figure 15(b). In addition, the diode-2 is associated with the peculiar AlGaN/GaN interface system (not a typical metal/semiconductor junction); hence diode-2 parameters are not further investigated.

The reverse  $J_G$ - $V_G$  characteristics of the HEMT gate diode at different temperatures are plotted in Figures 16(a) and (b). For low reverse voltages ( $V_G$  > -1 V), the leakage current sign is obtained in another direction (semilog plot ignores negative values), possibly due to the instrument noise (current is very low). In the intermediate range (-2 V <  $V_G$  < -1 V),  $J_G$ - $V_G$ properties are fitted with the TAT model, as seen in Figure 16(a). However, beyond that voltage ( $V_G$  < -2 V), the  $J_{TAT}$ overestimates the gate leakage current density. So, the reverse  $J_G$ - $V_G$  fitting is accomplished by Fowler-Nordheim tunneling (FNT) current density equation (*JFNT*) [1, 9]

$$
J_{FNT} = A_{FNT} E_m^2 \exp\left(-\frac{4\sqrt{2m_n} (q \Phi_B)^{3/2}}{3q \hbar E_m}\right) \tag{20}
$$

where *AFNT* is the FNT model fitting parameter. The electric field at the AlGaN/GaN heterojunction computed from the TCAD simulations [22] is used in TAT and FNT models.

Figures 16(a) and (b) the FNT model reasonably predicts the reverse  $J_G$  for  $V_G$  < -2 V over the wide temperature range of 293 K to 373 K. Therefore, it is considered the FNL transport plays a significant role in deciding the reverse gate current properties of the GaN/AlGaN/GaN HEMT.



**Figure 16.** (a) The reverse *JG-V<sup>G</sup>* characteristics of Schottky gate diode at two different temperatures of 298 K and 348 K. (b) The reverse *JG-V<sup>G</sup>* is fitted with the FNT model at 323 K and 373 K.



**Figure 17.** (a) Derivative of DCTS of AlGaN/GaN HEMT shows two traps F1 and F2. (b) Arrhenius plots for F1 and F2.

Two distinct positive peaks F1 and F2 are identified in the derivative DCTS for AlGaN/GaN HEMT in Figure 17(a). Our

earlier simulation studies [24] suggest that the electron trapping in the GaN buffer layer can induce a positive DCTS peak. The emission time constant  $(\tau_n)$  decreases with temperature, and its temperature dependency follows the Arrhenius equation (15). Hence,  $E_a$  and  $\sigma_n$  of the traps F1 (E<sub>C</sub>  $-0.5$  eV,  $5\times10^{-16}$  cm<sup>2</sup>) and F2 (E<sub>C</sub>  $-0.6$  eV,  $10^{-16}$  cm<sup>2</sup>) are identified from the Arrhenius plots in Figure 17(b). Based on our earlier findings and literature reports, the traps  $F1$  at  $E_C$  – 0.5 eV and F2 at  $E_C$  – 0.6 eV are attributed to the Fe-dopingrelated acceptor states in the GaN buffer region [24].

The electrical parameters (SBH, n,  $V_{ON}$  at 10 mA/cm<sup>2</sup>,  $\Phi_{eff}$ ,  $\Phi_B^0$ ,  $T_0$  anomaly for 4H-SiC SBDs), dominant forward  $J_F$  and reverse  $J_R$  transport models, trap signatures  $(E_T, \sigma_n, N_T)$  for Ni/4H-SiC SBD, Ti/4H-SiC SBD, Ni/GaN SBD, metal/GaN/AlGaN/GaN HEMT gate diode-1 and diode-2 are summarized in Table 1.

**Table 1.** Electrical parameters, dominant forward  $J_F$  and reverse  $J_R$  models, and trap signatures for various SBD devices

Parameter	Ni/4H-SiC SBD	Ti/4H-SiC SBD	Ni/GaN SBD	<b>HEMT</b>	<b>HEMT</b>
				gate diode-1	gate diode-2
$SBH$ (eV)	1.31 (at $300 \text{ K}$ )	1.18 (at $300 \text{ K}$ )	$0.83$ (at 293 K)	$1.15$ (at 298 K)	$0.72$ (at 298 K)
$\mathbf n$	$1.15$ (at 300 K)	$1.17$ (at 300 K)	1.1 (at $293 K$ )	1.67 (at 298 K)	4.36 (at 298 K)
$V_{ON}$ at 10 mA/cm <sup>2</sup> (V)	$0.88$ (at 300 K)	$0.76$ (at 300 K)	$0.36$ (at 293 K)	1.1 (at 298 K) from forward $J_G-V_G$	
$\Phi_{\text{eff}}$ (eV)	1.35	1.3	0.88	*	$\ast$
$\Phi_R^0$ (eV)	1.66	1.44	0.87	*	$\ast$
$T_0$ anomaly (K)	44.6	51.2	$\ast$	$\ast$	$\ast$
dominant $J_F$ model	TE (all $T$ )	TE (all $T$ )	TE (all $T$ )	TE (all $T$ )	TE (all $T$ )
dominant $J_R$ model	<b>TAT at 300 K</b>	PFE for	TFE for	FNL for 298 K $\leq T \leq 398$ K	
	PFE for $T \geq 350$ K	300 K $\leq T \leq 400$ K	$293K \leq T \leq 373K$		
trap-1 signatures	T1 at $E_C$ – 0.65 eV,	T1 at $E_C$ – 0.65 eV,	E1 at $E_C$ – 0.18 eV,	F1 at $E_C$ – 0.5 eV,	
$(E_T, \sigma_n, N_T)$	~5×10 <sup>-16</sup> cm <sup>2</sup> ,	$\sim 5 \times 10^{-16}$ cm <sup>2</sup> ,	$1.3\times10^{-18}$ cm <sup>2</sup> ,	$\sigma_n = 5 \times 10^{-16}$ cm <sup>2</sup>	
	$\sim 10^{13}$ cm <sup>-3</sup>	$\sim 10^{13}$ cm <sup>-3</sup>	$4\times10^{13}$ cm <sup>-3</sup>		
trap-2 signatures	T2 at $Ec - 1.13$ eV,	T <sub>2</sub> is not detected	E2 at $E_C$ – 0.56 eV,	F2 at $E_C$ – 0.6 eV,	
$(E_T, \sigma_n, N_T)$	~3×10 <sup>-18</sup> cm <sup>2</sup> .		$10^{-15}$ cm <sup>2</sup> .	$\sigma_n = 10^{-16}$ cm <sup>2</sup>	
	$\sim 1.3 \times 10^{13}$ cm <sup>-3</sup>		$8\times10^{14}$ cm <sup>-3</sup>		

\*not applicable for these devices

#### **4. Conclusion**

The forward and reverse current transport models, temperature dependence of SBH and n, *T<sup>0</sup>* anomaly for nonideal diodes, effective SBH, and trap parameters for Ni/4H-SiC, Ti/4H-SiC, Ni/GaN SBDs, and Schottky gate diode of AlGaN/GaN HEMT are presented. Both 4H-SiC SBDs demonstrated excellent electrical properties, high SBH (> 1eV), low *J<sup>R</sup>* (< 10-8 A/cm<sup>2</sup> at -200 V), low *VON* (< 0.9 V at 10 mA/cm<sup>2</sup>), and  $V_{BR}$  < -200 V for rectifier applications. Two deep-level traps at  $E_C$  – 0.65 eV and  $E_C$  – 1.13 eV are detected in the 4H-SiC SBDs. The Ni/GaN SBDs exhibited quasi-ideal

characteristics with  $n = 1.1$ , SBH of 0.83 eV, and two electron traps at  $E_C$  – 0.18 eV and  $E_C$  – 0.56 eV are identified. The Schottky gate of the HEMT has shown a two-diode model behaviour, such that, diode-1 corresponds to the regular metal/GaN Schottky barrier junction; while diode-2 forms at the unusual AlGaN/GaN heterointerface due to the energy difference between the Fermi level and AlGaN conduction band. The Fe-doping-related traps at  $E_C - 0.5$  eV and  $E_C - 0.6$ eV are determined in the HEMTs by the DCTS technique.

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