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Calderón Strategies for the Convolution Quadrature Time Domain Electric Field Integral Equation

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ABSTRACT In this work, we introduce new integral formulations based on the convolution quadrature method for the time-domain modeling of perfectly electrically conducting scatterers that overcome some of the most critical issues of the standard schemes based on the electric field integral equation (EFIE). The standard time-domain EFIE-based approaches typically yield matrices that become increasingly ill-conditioned as the time-step or the mesh discretization density increase and suffer from the well-known DC instability. This work presents solutions to these issues that are based both on new Calderón strategies and quasi-Helmholtz projectors regularizations. In addition, to ensure an efficient computation of the marching-on-in-time, the proposed schemes leverage properties of the Z-transform—involved in the convolution quadrature discretization scheme—when computing the stabilized operators. The two resulting formulations compare favorably with standard, well-established schemes. The properties and practical relevance of these new formulations will be showcased through relevant numerical examples that include canonical geometries and more complex structures.

INDEX TERMS Boundary element method, Calderón preconditioning, computational electromagnetic, convolution quadrature method, EFIE, time-domain integral equations

I. INTRODUCTION

TIME domain boundary integral equations (TDIEs) are widely used in the simulation of transient electromagnetic fields scattered by perfectly electrically conducting (PEC) objects [1]–[4]. Like their frequency-domain counterparts, the spatial discretization of these equations is often performed via the boundary element method. The time discretization, however, can be tackled in different ways. A popular approach leverages time basis functions either within a Marching-On-in-Time (MOT) scheme [5]–[7] or within a Marching-On-In-Order procedure [8]. The convolution quadrature (CQ) approach [9], [10] is an attractive alternative to these methods in which only space basis functions are explicitly defined. The approach has been applied to several equations in elastodynamics and acoustics [11], [12] and then in electromagnetics [13]. It provides an efficient time-stepping scheme with matrices derived from the space-discretized Laplace domain operators.

Another advantage of the CQ method is the use of implicit schemes (e.g. Runge Kutta methods [14]–[18]), which are generally more stable and typically allow for a better accuracy control of the solution over time [19], [20]. However, the CQ time stepping scheme is solved via a computationally expensive MOT algorithm. Nowadays, fast solvers can reach quasi-linear complexity in time and space [21], [22]. Usually, this fast technology uses iterative solvers, resulting in an overall computational cost that is proportional to the number of iterations which is low for well-conditioned systems. Working with well-conditioned matrices is therefore essential to reduce the computational cost of the solution process, in addition to being necessary to obtain accurate results [23].

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Lamentably, however, the CQ discretized time domain electric field integral equation (EFIE) is plagued by several drawbacks. Indeed, the matrices resulting from the discretization of the EFIE are known to become ill-conditioned for large time steps or at dense mesh discretizations: the condition number of the MOT matrices grows quadratically with the time step and with the inverse of the average mesh edge length. These two phenomena are the CQ counterparts of what for standard MOT schemes are known as the large time step breakdown [24]–[27] and the dense discretization breakdown (or h-refinement breakdown) [28], [29]. Another challenge in handling the CQ EFIE is that it involves operators whose definitions include a time integration. To avoid dealing with this integral, the time-differentiated counterpart of this formulation is often used [13], [30], but this differentiation is subject to a source of instability in the form of spurious linear currents living in the nullspace of the operator that degrades the solution [28], [31]; this phenomenon is known as the direct current instability (DC instability).

In this work, we propose new Calderón-preconditioned and quasi-Helmholtz regularized formulations free from the limitations mentioned above. The Calderón identities they rely on are already a well-established preconditioning approach in both the frequency domain [32] and time domain discretized by the Galerkin method [28], [29], [33], [34] that is extended in this work to convolution quadrature discretizations and complemented with quasi-Helmholtz regularization. The contribution of this paper is twofold: (i) we present a first approach to tackle the regularization of the EFIE operator and to address the DC instability resulting in a new operator that presents no nullspace on simply connected geometries, thus stabilizing the solution, and (ii) we build upon this first regularized form of the EFIE to obtain an equation that, at the price of a higher number of matrix-vector products, is stable in the case of multiply connected geometries.

This article is structured as follows: the time domain formulations of interest are summarized in Section II along with the convolution quadrature method and the boundary element method for spatial discretization; in Section III, the new Calderón and projectors-based preconditioning strategies are presented; finally, Section IV presents the numerical studies that confirm the effectiveness of the different approaches before concluding. Preliminary studies pertaining to this work were presented in the conference contribution [35].

II. Background and Notations

A. Time Domain Integral Formulations

In this work, we consider the problem of time-domain scattering by a perfectly electrically conducting object that resides in free space. The object is illuminated by an electromagnetic field $(\boldsymbol{e}^{\mathrm{inc}}, \boldsymbol{h}^{\mathrm{inc}})(\boldsymbol{r}, t)$ which induces a surface current density $\boldsymbol{j}_\Gamma$ on its boundary Γ that is the solution of the time-domain EFIE

$$\eta_0 \mathcal{T}\left(\boldsymbol{j}_\Gamma\right)(\boldsymbol{r}, t) = -\hat{\boldsymbol{n}}(\boldsymbol{r}) \times \boldsymbol{e}^{\mathrm{inc}}(\boldsymbol{r}, t) \ . \tag{1}$$

Here, $\hat{\boldsymbol{n}}$ is the outpointing normal to Γ and η_0 is the characteristic impedance of the background. The electric field operator $\mathcal{T}$ includes the contributions of the vector and scalar potentials, respectively denoted $\mathcal{T}_{\mathrm{s}}$ and $\mathcal{T}_{\mathrm{h}}$ [30]

$$\mathcal{T}\left(\boldsymbol{f}\right)(\boldsymbol{r}, t) = -\frac{1}{c_0}\frac{\partial}{\partial t}\mathcal{T}_{\mathrm{s}}\left(\boldsymbol{f}\right)(\boldsymbol{r}, t) + c_0 \int_{-\infty}^{t} \mathcal{T}_{\mathrm{h}}\left(\boldsymbol{f}\right)(\boldsymbol{r}, t')\, dt' \,, \tag{2}$$

$$\mathcal{T}_{\mathrm{s}}\left(\boldsymbol{f}\right)(\boldsymbol{r}, t) = \hat{\boldsymbol{n}}(\boldsymbol{r}) \times \iint_{\boldsymbol{r}' \in \Gamma} \left(\mathcal{G}_{\boldsymbol{r}} *_t \boldsymbol{f}\right)(\boldsymbol{r}', t) dS' \,, \tag{3}$$

$$\mathcal{T}_{\mathrm{h}}\left(\boldsymbol{f}\right)(\boldsymbol{r}, t) = \hat{\boldsymbol{n}}(\boldsymbol{r}) \times \nabla \iint_{\boldsymbol{r}' \in \Gamma} \left(\mathcal{G}_{\boldsymbol{r}} *_t \nabla' \cdot \boldsymbol{f}\right)(\boldsymbol{r}', t)\, dS' \,, \tag{4}$$

where c_0 is the speed of light in the background. The temporal convolution product $*_t$ and the temporal Green function $\mathcal{G}$ are defined as

$$\left(f *_t g\right)(t) = \int_{-\infty}^{\infty} f(\tau)\, g(t-\tau)\, d\tau \,, \tag{5}$$

$$\mathcal{G}_{\boldsymbol{r}}(\boldsymbol{r}', t) = \frac{\delta\left(t - \frac{|\boldsymbol{r}-\boldsymbol{r}'|}{c_0}\right)}{4\pi|\boldsymbol{r}-\boldsymbol{r}'|} \,, \tag{6}$$

with δ the time Dirac delta.

B. Marching-On-In-Time with Convolution Quadratures

Let Θ be a placeholder for any of the integral operators previously presented and let $\boldsymbol{k}(\boldsymbol{r}, t)$ be a causal function ($\forall t < 0$, $\boldsymbol{k}(\boldsymbol{r}, t) = \boldsymbol{0}$). With these notations, most time domain integral equation take the form

$$\Theta\left(\boldsymbol{f}_{\mathrm{c}}\right)(\boldsymbol{r}, t) = \boldsymbol{k}(\boldsymbol{r}, t) \ , \tag{7}$$

where $\boldsymbol{f}_{\mathrm{c}}$ is the solution to be solved for. The first step of the Marching-On-in-Time solution scheme with convolution quadratures is to apply the boundary element method [3], [36], [37] as spatial discretization. Assuming separability between the space and time variables, the unknown function $\boldsymbol{f}_{\mathrm{c}}$ is expanded as a linear combination of N_s spatial basis functions such that [38]

$$\boldsymbol{f}_{\mathrm{c}}(\boldsymbol{r}, t) \approx \sum_{n=1}^{N_s} [\boldsymbol{f}_\Gamma]_n(t)\, \boldsymbol{f}_n^{\mathrm{src}}(\boldsymbol{r}) \ , \tag{8}$$

where $\left\{\boldsymbol{f}^{\mathrm{src}}\right\}$ are the source spatial basis functions and their associated time coefficients are stored in the vector $\boldsymbol{f}_\Gamma(t)$. Then, the equation (7) is tested by the spatial basis functions $\left\{\boldsymbol{f}^{\mathrm{tst}}\right\}$ leading to the time-dependent matrix system

$$(\boldsymbol{\theta} * \boldsymbol{f})(t) = \boldsymbol{k}(t) \ , \tag{9}$$

where for n and m in $[\![1, N_s]\!]$, we have,

$$\begin{aligned} (\boldsymbol{\theta} * \boldsymbol{f})_m(t) &= \sum_{n=1}^{N_s} \left(\theta_{m,n} *_t f_n\right)(t) \ , \\ \left[\boldsymbol{\theta}\right]_{m,n}(t) &= \langle \boldsymbol{f}_m^{\mathrm{tst}}, \Theta\left(\delta \boldsymbol{f}_n^{\mathrm{src}}\right)\rangle_\Gamma(t) \ , \\ \left[\boldsymbol{k}_\Gamma\right]_m(t) &= \langle \boldsymbol{f}_m^{\mathrm{tst}}, \boldsymbol{k}\rangle_\Gamma(t) \ , \end{aligned} \tag{10}$$

with

$$\langle \boldsymbol{f}, \boldsymbol{g}\rangle_\Gamma = \iint_{\boldsymbol{r} \in \Gamma} \boldsymbol{f}(\boldsymbol{r}) \cdot \boldsymbol{g}(\boldsymbol{r})\, dS \,. \tag{11}$$

The second step is the discretization in time with the convolution quadrature method [9], [10], [12]–[14], [38]. First, the system (9) must be transformed in the Laplace domain [39] and we denote $\boldsymbol{\theta}_{\mathcal{L}}$, $\boldsymbol{f}_{\mathcal{L}}$ and $\boldsymbol{k}_{\mathcal{L}}$ the Laplace transform of $\boldsymbol{\theta}$, $\boldsymbol{f}_{\Gamma}$ and $\boldsymbol{k}_{\Gamma}$. The system (9) is then equivalent to

$$\mathcal{L}\left\{t \mapsto (\boldsymbol{\theta} * \boldsymbol{f})\ (t)\right\}(s) = \boldsymbol{\theta}_{\mathcal{L}}(s)\, \boldsymbol{f}_{\mathcal{L}}(s) = \boldsymbol{k}_{\mathcal{L}}(s)\ , \quad (12)$$

where s, a complex number, is the Laplace parameter. Then, a representation on the Z-domain discretizes the system (12). The Laplace parameter s, in the operator $\boldsymbol{\theta}_{\mathcal{L}}$, is replaced by the matrix-valued parameter

$$\boldsymbol{s}_{\mathrm{cq}}(z) = \frac{1}{\Delta t}\left(\boldsymbol{A} + \frac{\mathbf{1}_p \boldsymbol{b}^T}{z-1}\right)^{-1}\ , \quad (13)$$

diagonalizable for the considered z values, with the following eigenvalue decomposition $\boldsymbol{s}_{\mathrm{cq}}(z) = \boldsymbol{Q}(z)\boldsymbol{\Lambda}(z)\boldsymbol{Q}^{-1}(z)$. The elements of the diagonal matrix $\boldsymbol{\Lambda}(z)$ are the eigenvalues of $\boldsymbol{s}_{\mathrm{cq}}(z)$ and the columns of $\boldsymbol{Q}(z)$ are their associated eigenvectors. The time step size is denoted Δt and the matrix $\boldsymbol{A}$ and the vectors $\mathbf{1}_p$, $\boldsymbol{c}$, $\boldsymbol{b}$ of size p are determined by the implicit scheme used [40]. The discretized Z-domain operator $\boldsymbol{\theta}_{\mathcal{Z}}$ is defined such that for k and l in $[\![1, p]\!]$ and for m and n in $[\![1, N_s]\!]$

$$\left[\boldsymbol{\theta}_{\mathcal{Z}}\right]_{\Phi_{m,k},\Phi_{n,l}}(z) = \left[\boldsymbol{Q}(z)\boldsymbol{\theta}^{\boldsymbol{\Lambda}}_{m,n}(z)\boldsymbol{Q}^{-1}(z)\right]_{k,l}\ , \quad (14)$$

where $\Phi_{\alpha,\beta} = p(\alpha-1)+\beta$ is an appropriate indexing function and $\boldsymbol{\theta}^{\boldsymbol{\Lambda}}_{m,n}(z)$ is a diagonal matrix defined as

$$\left[\boldsymbol{\theta}^{\boldsymbol{\Lambda}}_{m,n}\right]_{k,k}(z) = \left[\boldsymbol{\theta}_{\mathcal{L}}(\boldsymbol{\Lambda}_{k,k})\right]_{m,n}(z)\,. \quad (15)$$

The vectors $\boldsymbol{f}_{\mathcal{Z}}$ and $\boldsymbol{k}_{\mathcal{Z}}$ are the Z-domain representation [40] of the respective time-discretized vectors

$$\begin{aligned}\left[\boldsymbol{k}_i\right]_{\Phi_{m,k}} &= \left[\boldsymbol{k}_{\Gamma}\right]_m\left(\Delta t\,(i + [\boldsymbol{c}]_k)\right)\ ,\\ \left[\boldsymbol{f}_i\right]_{\Phi_{n,k}} &= \left[\boldsymbol{f}_{\Gamma}\right]_n\left(\Delta t\,(i + [\boldsymbol{c}]_k)\right)\ ,\end{aligned} \quad (16)$$

yielding to the following discretization of the system (12)

$$\boldsymbol{\theta}_{\mathcal{Z}}(z)\, \boldsymbol{f}_{\mathcal{Z}}(z) = \boldsymbol{k}_{\mathcal{Z}}(z)\ . \quad (17)$$

Finally, the equivalent time discretized system of (9) is obtained by applying the inverse Z-transform on (17)

$$\mathcal{Z}^{-1}\left\{\boldsymbol{\theta}_{\mathcal{Z}}(z)\boldsymbol{f}_{\mathcal{Z}}(z)\right\}_i = [\boldsymbol{Z}_{\boldsymbol{\theta}} *_s \boldsymbol{f}]_i = \sum_{j=0}^{i} \boldsymbol{Z}_{\boldsymbol{\theta},j}\boldsymbol{f}_{i-j} = \boldsymbol{k}_i\,, \quad (18)$$

where $\boldsymbol{Z}_{\boldsymbol{\theta},i} = \mathcal{Z}^{-1}\left\{z \to \boldsymbol{\theta}_{\mathcal{Z}}(z)\right\}_i$ are the time domain interaction matrices and $*_s$ is the sequence convolution product. The system sequence (18) is rewritten in the following Marching-On-In-Time that can be solved for $\boldsymbol{f}_i$

$$\boldsymbol{Z}_{\boldsymbol{\theta},0}\boldsymbol{f}_i = \boldsymbol{k}_i - \sum_{j=1}^{i} \boldsymbol{Z}_{\boldsymbol{\theta},j}\boldsymbol{f}_{i-j}\,. \quad (19)$$

C. Classic Integral Marching-On-In-Times

In this subsection, the discretization scheme described above will be applied to the specific case of the EFIE. The Rao-Wilton-Glisson (RWG) basis functions $\left\{\boldsymbol{f}^{\mathrm{rwg}}_n\right\}_{N_s}$ [37], [41] are used to expand the current density as

$$\boldsymbol{j}_{\Gamma}(\boldsymbol{r}, t) \approx \sum_{n=1}^{N_s} [\boldsymbol{j}_{\Gamma}]_n(t)\, \boldsymbol{f}^{\mathrm{rwg}}_n(\boldsymbol{r})\ , \quad (20)$$

where the current coefficients are gathered in an unknown vector function of time $\boldsymbol{j}_{\Gamma}(t)$. The EFIE is then tested with rotated RWG basis functions $\left\{\hat{\boldsymbol{n}} \times \boldsymbol{f}^{\mathrm{rwg}}_n\right\}_{N_s}$, leading to the following Marching-On-In-Time

$$\boldsymbol{Z}_{\boldsymbol{T},0}\boldsymbol{j}_i = -\eta_0^{-1}\boldsymbol{e}^{\mathrm{inc}}_i - \sum_{j=1}^{i} \boldsymbol{Z}_{\boldsymbol{T},j}\boldsymbol{j}_{i-j}\,, \quad (21)$$

where the vector sequences $\boldsymbol{j}_i$ and $\boldsymbol{e}^{\mathrm{inc}}_i$, and the time domain interaction matrices $\boldsymbol{Z}_{\boldsymbol{T},i}$ are respectively generated by the convolution quadrature method described in Subsection B of $\boldsymbol{j}_{\Gamma}(t)$ and the following space-discretized vector and matrix

$$\left[\boldsymbol{e}^{\mathrm{inc}}_{\Gamma}\right]_m(t) = \langle \hat{\boldsymbol{n}} \times \boldsymbol{f}^{\mathrm{rwg}}_m, \hat{\boldsymbol{n}} \times \boldsymbol{e}^{\mathrm{inc}}\rangle_{\Gamma}(t)\ , \quad (22)$$

$$\left[\boldsymbol{T}\right]_{m,n}(t) = \langle \hat{\boldsymbol{n}} \times \boldsymbol{f}^{\mathrm{rwg}}_m, \mathcal{T}\left(\delta \boldsymbol{f}^{\mathrm{rwg}}_n\right)\rangle_{\Gamma}(t)\ . \quad (23)$$

However, the time integral contribution of this operator $\mathcal{T}$ involves an unbounded number of non-vanishing matrices $\boldsymbol{Z}_{\boldsymbol{T},i}$ (21), leading to a prohibitive quadratic complexity with the number of time steps [42]. Historically, the time differentiated formulation is preferred because it is not afflicted by this drawback [13], [30], and leads to the following MOT [30]

$$\boldsymbol{Z}_{\dot{\boldsymbol{T}},0}\boldsymbol{j}_i = -\eta_0^{-1}\dot{\boldsymbol{e}}^{\mathrm{inc}}_i - \sum_{j=1}^{N_{\mathrm{conv}}} \boldsymbol{Z}_{\dot{\boldsymbol{T}},j}\boldsymbol{j}_{i-j}\,, \quad (24)$$

where $\dot{\boldsymbol{e}}^{\mathrm{inc}}_n$ and $\dot{\boldsymbol{Z}}_n$ are respectively the time domain vectors and interaction matrices generated by the convolution quadrature method described in Subsection B of

$$\left[\dot{\boldsymbol{e}}^{\mathrm{inc}}_{\Gamma}\right]_m(t) = \langle \hat{\boldsymbol{n}} \times \boldsymbol{f}^{\mathrm{rwg}}_m, \hat{\boldsymbol{n}} \times \frac{\partial}{\partial t}\boldsymbol{e}^{\mathrm{inc}}\rangle_{\Gamma}\,, \quad (25)$$

$$\left[\dot{\boldsymbol{T}}\right]_{m,n}(t) = \langle \hat{\boldsymbol{n}} \times \boldsymbol{f}^{\mathrm{rwg}}_m, \frac{\partial}{\partial t}\mathcal{T}\left(\delta \boldsymbol{f}^{\mathrm{rwg}}_n\right)\rangle_{\Gamma}\,. \quad (26)$$

D. EFIE DC instability

The electric field integral operator suffers from the DC instability: since for all constant-in-time solenoidal current $\boldsymbol{j}_{\mathrm{cs}}$ we have $\nabla \cdot \boldsymbol{j}_{\mathrm{cs}} = 0$ and $\frac{\partial}{\partial t}\boldsymbol{j}_{\mathrm{cs}} = \boldsymbol{0}$, we can conclude that

$$\mathcal{T}\left(\boldsymbol{j}_{\mathrm{cs}}\right) = \boldsymbol{0}\,. \quad (27)$$

Therefore, the EFIE solution is only determined up to a constant solenoidal current [30]. Its time differentiated counterpart inherits these drawbacks and amplifies the DC instability by further adding linear in time solenoidal currents to the nullspace. This latter deteriorates the late time simulation in which spurious currents grow exponentially in the operator nullspace [43]. This behaviour is predicted by the polynomial eigenvalues analysis of the MOT: a stable

MOT has all its eigenvalues inside the unit circle in the complex plane while a MOT that suffers from the DC instability has some eigenvalues that cluster around 1 [44]. The eigenvalue distribution of the time differentiated EFIE MOT is represented in Figure 1 in which such a cluster is clearly visible around 1.

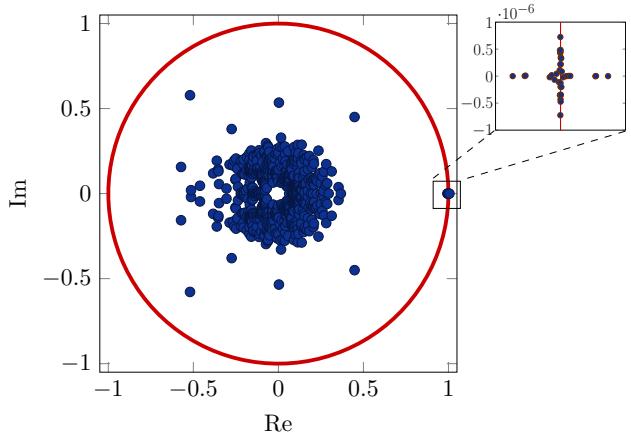

FIGURE 1. Polynomial eigenvalues of the TD-EFIE MOT scheme on the unit sphere, $N_s = 270$ and $\Delta t = 3$ ns

E. Quasi-Helmholtz projectors

Previous works show that the electric field integral equation discretized in space using RWG basis functions can be stabilized by the quasi-Helmholtz projectors [30], [34]. These projectors are formed from the star-to-rwg transformation matrix, denoted $\boldsymbol{\Sigma}$ and defined in [45], which maps the discretized current into the non-solenoidal contributions [46], [47]. The quasi-Helmholtz projectors on the non-solenoidal space and its complementary (the one on solenoidal/quasi-harmonic space) are respectively

$$\boldsymbol{P}^{\Sigma} = \left(\boldsymbol{\Sigma}\left(\boldsymbol{\Sigma}^T\boldsymbol{\Sigma}\right)^{+}\boldsymbol{\Sigma}^T\right) \quad \text{and} \quad \boldsymbol{P}^{\Lambda H} = \boldsymbol{I} - \boldsymbol{P}^{\Sigma}\,, \tag{28}$$

where $^{+}$ denotes the Moore–Penrose pseudoinverse [45].

III. Calderón preconditioning of the EFIE

EFIE formulations based on the quasi-Helmholtz projectors cure the DC instability and the conditioning at large time steps. However, these formulations still suffer from a dense discretization breakdown. One appealing strategy could be to apply standard preconditioning schemes to the Marching-On-In-Time matrices directly to cure the matrix conditioning issues. However, the solution currents would remain unaltered and subject to DC instabilities as the original scheme. This is why the preconditioning has to be performed on the continuous equations to build a new operator without nullspace and then discretize the formulation to obtain a well-conditioned scheme. In this part, Calderón preconditioning strategies are proposed to cure the DC instability and the conditioning breakdowns.

A. A Convolution Quadrature Calderón time-domain EFIE

Calderón preconditioners are based on the Calderón identity [29], [48]

$$\mathcal{T}^2 = \mathcal{T} \circ \mathcal{T} = -\mathcal{I}/4 + \mathcal{K}^2\,, \tag{29}$$

where $\circ$ is the composition operator, $\mathcal{I}$ is the identity operator and $\mathcal{K}$ is defined as

$$\mathcal{K}\left(\boldsymbol{f}\right)(\boldsymbol{r},t) = \hat{\boldsymbol{n}}\left(\boldsymbol{r}\right) \times \nabla \times \iint_{\boldsymbol{r}'\in\Gamma} \left(\mathcal{G}_{\boldsymbol{r}} *_t \boldsymbol{f}\right)(\boldsymbol{r}',t)dS'\,. \tag{30}$$

The operator $-\mathcal{I}/4 + \mathcal{K}^2$ is a well-behaved operator for increasing discretization densities. As a consequence, with a proper discretization, $\mathcal{T}^2$ is well-conditioned for large time steps and dense meshes for simply connected structures [45]. In practice, a discretization of $\mathcal{T}^2$ is used in which the right EFIE operator is discretized with RWG basis functions and the left preconditioner is discretized with Buffa-Christiansen (BC) basis functions $\left(\boldsymbol{f}_n^{\text{bc}}\right)_{N_s}$ [49]–[51]

$$\left[\mathbb{T}\right]_{m,n}(t) = \langle \hat{\boldsymbol{n}} \times \boldsymbol{f}_m^{\text{bc}}, \mathcal{T}\left(\delta \boldsymbol{f}_n^{\text{bc}}\right)\rangle_{\Gamma}(t)\,. \tag{31}$$

The preconditioning leads to the following space-discretized formulation

$$\left(\mathbb{T}\boldsymbol{G}_m^{-1} * \left(\boldsymbol{T} * \boldsymbol{j}_{\Gamma}\right)\right)(t) = -\eta_0^{-1}\left(\mathbb{T}\boldsymbol{G}_m^{-1} * \boldsymbol{e}_{\Gamma}^{\text{inc}}\right)(t)\,, \tag{32}$$

where the matrix $\boldsymbol{G}_m$ is the mixed gram matrix linking the the two discretizations

$$\left[\boldsymbol{G}_m\right]_{m,n} = \langle \hat{\boldsymbol{n}} \times \boldsymbol{f}_m^{\text{rwg}}, \boldsymbol{f}_n^{\text{bc}}\rangle_{\Gamma}\,. \tag{33}$$

Then, the convolution quadrature leads to the MOT scheme

$$\begin{aligned}\left[\boldsymbol{Z}_{\mathbb{T}}\widetilde{\boldsymbol{G}}_m^{-1} *_s \boldsymbol{Z}_{\boldsymbol{T}}\right]_0 \boldsymbol{j}_i = -\eta_0^{-1}\left[\boldsymbol{Z}_{\mathbb{T}}\widetilde{\boldsymbol{G}}_m^{-1} *_s \boldsymbol{e}^{\text{inc}}\right]_i \\ - \sum_{j=1}^{i}\left[\boldsymbol{Z}_{\mathbb{T}}\widetilde{\boldsymbol{G}}_m^{-1} *_s \boldsymbol{Z}_{\boldsymbol{T}}\right]_j \boldsymbol{j}_{i-j}\,,\end{aligned} \tag{34}$$

where $\boldsymbol{Z}_{\mathbb{T},i}$ are the time domain interaction matrices of the space-discretized operator $\mathbb{T}(t)$ (31) generated by the convolution quadrature method described in Subsection B, the sequence convolution quadrature product $*_s$ is the discretization of the space-discretized temporal convolution product $*$ and

$$\widetilde{\boldsymbol{G}}_m = \boldsymbol{G}_m \otimes \boldsymbol{I}_p\,. \tag{35}$$

The Kronecker product $\otimes \boldsymbol{I}_p$ is required to match with the convolution quadrature method where $\boldsymbol{I}_p$ is the identity matrix of size p. Unfortunately, the MOT in (34), involves operators with temporal integrations leading to a time consuming MOT. A more favorable scheme can be obtained by noticing the following commutative properties

$$\frac{\partial}{\partial t}\mathcal{T}_{\text{s/h}}\left(\boldsymbol{f}\right)(\boldsymbol{r},t) = \mathcal{T}_{\text{s/h}}\left(\frac{\partial}{\partial t}\boldsymbol{f}\right)(\boldsymbol{r},t)\,, \tag{36}$$

$$\int_{-\infty}^{t}\mathcal{T}_{\text{s/h}}\left(\boldsymbol{f}\right)(\boldsymbol{r},t')\,dt' = \mathcal{T}_{\text{s/h}}\left(\int_{-\infty}^{\cdot}\boldsymbol{f}\right)(\boldsymbol{r},t)\,, \tag{37}$$

and the cancellation property $\mathcal{T}_{\rm h}^2 = \mathbf{0}$ [32], we have

$$\begin{aligned}\mathcal{T}^2 &= \left(-\frac{1}{c_0}\frac{\partial}{\partial t}\mathcal{T}_{\rm s} + c_0\int_{-\infty}^{\cdot}\mathcal{T}_{\rm h}\right)\circ\left(-\frac{1}{c_0}\frac{\partial}{\partial t}\mathcal{T}_{\rm s} + c_0\int_{-\infty}^{\cdot}\mathcal{T}_{\rm h}\right)\\ &= c_0^{-2}\frac{\partial^2}{\partial t^2}\mathcal{T}_{\rm s}^2 - \mathcal{T}_{\rm s}\circ\mathcal{T}_{\rm h} - \mathcal{T}_{\rm h}\circ\mathcal{T}_{\rm s}\,.\end{aligned} \tag{38}$$

This is advantageous since, besides not involving any time integration contribution, the operator $c_0^{-2}\frac{\partial^2}{\partial t^2}\mathcal{T}_{\rm s}^2 - \mathcal{T}_{\rm s}\mathcal{T}_{\rm h} - \mathcal{T}_{\rm h}\mathcal{T}_{\rm s}$ has no nullspace for simply connected geometries leading to a DC-stable discretization ("dottrick TDEFIE") [52]. By extending the previous notations on $\mathcal{T}_{\rm s}$ and $\mathcal{T}_{\rm h}$

$$\begin{aligned}\left[\boldsymbol{T}_{\rm s/h}\right]_{m,n}(t) &= \langle\hat{\boldsymbol{n}}\times\boldsymbol{f}_m^{\rm rwg}, \mathcal{T}_{\rm s/h}\left(\delta\boldsymbol{f}_n^{\rm rwg}\right)\rangle_\Gamma(t)\,,\\ \left[\mathbb{T}_{\rm s/h}\right]_{m,n}(t) &= \langle\hat{\boldsymbol{n}}\times\boldsymbol{f}_m^{\rm bc}, \mathcal{T}_{\rm s/h}\left(\delta\boldsymbol{f}_n^{\rm bc}\right)\rangle_\Gamma(t)\,,\end{aligned} \tag{39}$$

the proposed space-discretized operator is denoted

$$\mathrm{T}_c = c_0^{-2}\frac{\partial^2}{\partial t^2}\mathbb{T}_{\rm s}\boldsymbol{G}_m^{-1} * \boldsymbol{T}_{\rm s} - \mathbb{T}_{\rm s}\boldsymbol{G}_m^{-1} * \boldsymbol{T}_{\rm h} - \mathbb{T}_{\rm h}\boldsymbol{G}_m^{-1} * \boldsymbol{T}_{\rm s}\,, \tag{40}$$

yielding to the following space-discretized formulation

$$\left(\mathrm{T}_c * \boldsymbol{j}_\Gamma\right)(t) = -\eta_0^{-1}\left(\mathbb{T}\boldsymbol{G}_m^{-1} * \boldsymbol{e}_\Gamma^{\rm inc}\right)(t)\,. \tag{41}$$

The right-hand side operator still involves a temporal integration in (41). However, given the commutative properties (37), the temporal integral on the scalar potential $\mathcal{T}_h$ is evaluated with the incident field

$$c_0\left(\int_{-\infty}^{\cdot}\mathbb{T}_{\rm h}\boldsymbol{G}_m^{-1} * \boldsymbol{e}_\Gamma^{\rm inc}\right)(t) = c_0\left(\mathbb{T}_{\rm h}\boldsymbol{G}_m^{-1} * \boldsymbol{e}_\Gamma^{\rm prim}\right)(t)\,, \tag{42}$$

where

$$\begin{aligned}\left[\boldsymbol{e}_\Gamma^{\rm prim}\right]_m(t) &= \langle\hat{\boldsymbol{n}}\times\boldsymbol{f}_m^{\rm rwg}, \int_{-\infty}^{t}\hat{\boldsymbol{n}}\times\boldsymbol{e}^{\rm inc}(t')dt'\rangle_\Gamma\,,\\ \left[\boldsymbol{e}_i^{\rm prim}\right]_{\boldsymbol{\Phi}_{m,k}} &= \left[\boldsymbol{e}_\Gamma^{\rm prim}\right]_m\left(\Delta t\left(i+[\boldsymbol{c}]_k\right)\right)\,.\end{aligned} \tag{43}$$

Therefore, the previous MOT is rewritten as

$$\boldsymbol{Z}_{\mathrm{T_c},0}\boldsymbol{j}_i = -\eta_0^{-1}\sum_{j=0}^{N_{\rm conv}}\left(\boldsymbol{Z}_{\mathrm{T_s},j}\boldsymbol{e}_{i-j}^{\rm inc} + \boldsymbol{Z}_{\mathrm{T_h},j}\boldsymbol{e}_{i-j}^{\rm prim}\right) - \sum_{j=1}^{N_{\rm conv}}\boldsymbol{Z}_{\mathrm{T_c},j}\boldsymbol{j}_{i-j}\,, \tag{44}$$

where the time domain interaction matrices $\boldsymbol{Z}_{\mathrm{T_c},i}$, $\boldsymbol{Z}_{\dot{\mathrm{T}}_{\rm s},i}$ and $\boldsymbol{Z}_{\mathrm{T_h},i}$ are respectively generated by the convolution quadrature method described in Subsection B of the space-discretized operators $\mathrm{T_c}$, $\dot{\mathrm{T}}_{\rm s} = -c_0^{-1}\frac{\partial}{\partial t}\mathbb{T}_{\rm s}\boldsymbol{G}_m^{-1}$ and $\mathrm{T_h} = c_0\mathbb{T}_{\rm h}\boldsymbol{G}_m^{-1}$. As in (34), the interaction matrix sequence $\boldsymbol{Z}_{\mathrm{T_c},i}$ involves computationally expensive sequence convolution products $*_s$, however, the convolution quadrature method allows the substitution of the sequence convolution products $*_s$ by matrix multiplications in the Z-domain, that can be evaluated at a lesser cost. By extending the notations of the convolution quadrature described in Subsection B on the space-discretized operators $\boldsymbol{T}_{\rm s/h}$ and $\mathbb{T}_{\rm s/h}$ and by using the Z-domain properties, the matrix sequence $\boldsymbol{Z}_{\mathrm{T_c},i}$ is equal to

$$\begin{aligned}\boldsymbol{Z}_{\mathrm{T_c},i} = c_0^{-2}\mathcal{Z}^{-1}&\left\{\overline{\boldsymbol{s}}_{\rm cq}^2\,\mathbb{T}_{\rm s,\mathcal{Z}}\widetilde{\boldsymbol{G}}_m\boldsymbol{T}_{\rm s,\mathcal{Z}}\right\}\\ &- \mathcal{Z}^{-1}\left\{\mathbb{T}_{\rm h,\mathcal{Z}}\widetilde{\boldsymbol{G}}_m\boldsymbol{T}_{\rm s,\mathcal{Z}} + \mathbb{T}_{\rm s,\mathcal{Z}}\widetilde{\boldsymbol{G}}_m\boldsymbol{T}_{\rm h,\mathcal{Z}}\right\}\,,\end{aligned} \tag{45}$$

where the matrix $\overline{\boldsymbol{s}}_{\rm cq}(z) = \boldsymbol{I}_{N_s}\otimes\boldsymbol{s}_{\rm cq}(z)$ is the Z-discretization of the time derivative and $\boldsymbol{I}_{N_s}$ is the identity matrix of size N_s. The formulation (44) is a good candidate to obtain a stable current solution, however, the proposed operator $\mathcal{T}^2$ has static nullspaces for multiply connected geometries [33]. As such, (40) is still subject to DC-instabilities for multiply connected geometries. The polynomial eigenvalue analysis on a sphere and on a torus (respectively Figure 2 and Figure 3) illustrate this phenomenon. While all the eigenvalues cluster in 0 in the spherical case, an analysis on a torus highlights four eigenvalues of this MOT clustered around 1, corresponding to the four constant regime solutions [33].

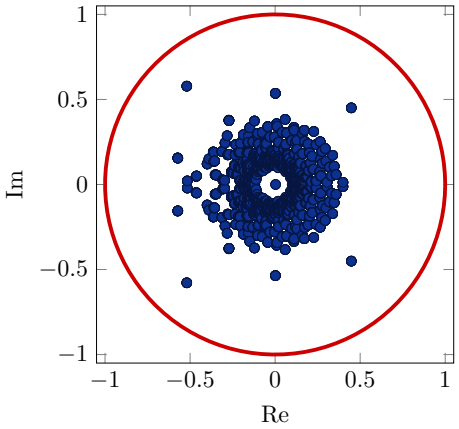

FIGURE 2. Polynomial eigenvalues of the Calderón EFIE MOT scheme on the unit sphere, $N_s = 270$ and $\Delta t = 3$ ns.

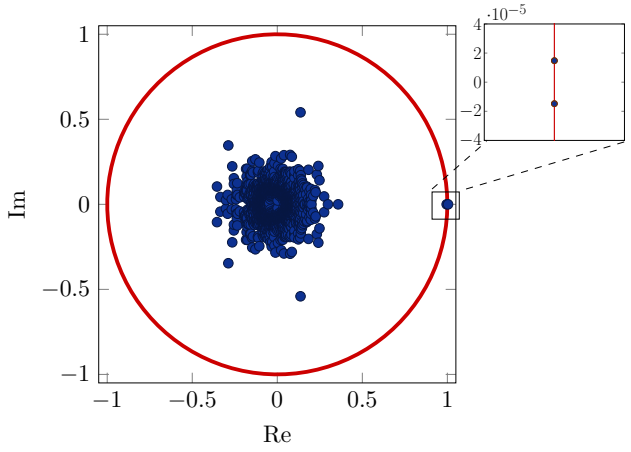

FIGURE 3. Polynomial eigenvalues of the Calderón EFIE MOT scheme on a torus with a inner and outer radii respectively equal to 0.2 m and 0.5 m, $N_s = 387$ and $\Delta t = 3$ ns. Near 1, four eigenvalues are clustered, superposed two by two on this figure.

B. A Convolution Quadrature Calderón time-domain EFIE regularized with quasi-Helmholtz projectors

The previous Calderón formulation is perfectly adapted to simply connected geometries, ensuring that the new operator

has no nullspace. However, on multiply-connected geometries, the harmonic subspace is non-empty, thus enlarging the nullspace of $\mathcal{T}$ which is a new source of DC instability in (44)[33]. The discretized EFIE operators can be regularized using the quasi-Helmholtz projectors to address this issue. Because the regularization is based on projectors, it does not compromise the h-refinement regularizing effect of the original Calderón scheme. The regularized EFIE space-discretized operators are

$$\boldsymbol{T}^{\text{reg}} = \left(\frac{c_0}{a}\int_{-\infty}^{\cdot} \boldsymbol{P}^{\Lambda H} + \boldsymbol{P}^{\Sigma}\right) * \boldsymbol{T} * \left(\boldsymbol{P}^{\Lambda H} + \frac{a}{c_0}\frac{\partial}{\partial t}\boldsymbol{P}^{\Sigma}\right), \quad (46)$$

$$\mathbb{T}^{\text{reg}} = \left(\frac{c_0}{a}\int_{-\infty}^{\cdot} \mathbb{P}^{\Sigma H} + \mathbb{P}^{\Lambda}\right) * \mathbb{T} * \left(\mathbb{P}^{\Sigma H} + \frac{a}{c_0}\frac{\partial}{\partial t}\mathbb{P}^{\Lambda}\right), \quad (47)$$

where the BC quasi-Helmholtz projectors are defined with the loop-to-RWG transformation matrix $\boldsymbol{\Lambda}$ [45] such that

$$\mathbb{P}^{\Lambda} = \left(\boldsymbol{\Lambda}\left(\boldsymbol{\Lambda}^T\boldsymbol{\Lambda}\right)^{+}\boldsymbol{\Lambda}^T\right) \quad \text{and} \quad \mathbb{P}^{\Sigma H} = \boldsymbol{I} - \mathbb{P}^{\Lambda}, \quad (48)$$

and where the scaling $\frac{c_0}{a}$ with a defined as the maximal diameter of the scatterer, ensures consistent dimensionality and helps reduce the conditioning further. This application of the projectors is equivalent to differentiating the non-solenoidal contributions on the left and in time integrating the solenoidal contributions on the right of each EFIE operator [30]. The regularized Calderón operator in the space-discretized time domain is

$$\mathrm{T}_{\mathrm{c}}^{\text{reg}} = \mathbb{T}^{\text{reg}}\boldsymbol{G}_m^{-1} * \boldsymbol{T}^{\text{reg}}. \quad (49)$$

At first sight, (46) and (47) seem to involve unpractical temporal integrals. However, the problematic contributions in the regularized EFIE operator $\mathbf{T}^{\text{reg}}$ will vanish, since $\boldsymbol{P}^{\Lambda H}\boldsymbol{T}_{\mathrm{h}} = \boldsymbol{T}_{\mathrm{h}}\boldsymbol{P}^{\Lambda H} = \mathbf{0}$ and $\boldsymbol{P}^{\Sigma}\boldsymbol{T}_{\mathrm{h}}\boldsymbol{P}^{\Sigma} = \boldsymbol{T}_{\mathrm{h}}$, and we have

$$\begin{aligned}\boldsymbol{T}^{\text{reg}} = &-a^{-1}\boldsymbol{P}^{\Lambda H}\boldsymbol{T}_{\mathrm{s}}\boldsymbol{P}^{\Lambda H} - c_0^{-1}\boldsymbol{P}^{\Lambda H}\frac{\partial}{\partial t}\boldsymbol{T}_{\mathrm{s}}\boldsymbol{P}^{\Sigma}\\ &- c_0^{-1}\boldsymbol{P}^{\Sigma}\frac{\partial}{\partial t}\boldsymbol{T}_{\mathrm{s}}\boldsymbol{P}^{\Lambda H} - \frac{a}{c_0^2}\boldsymbol{P}^{\Sigma}\frac{\partial^2}{\partial t^2}\boldsymbol{T}_{\mathrm{s}}\boldsymbol{P}^{\Sigma} + a\boldsymbol{T}_{\mathrm{h}}.\end{aligned} \quad (50)$$

Similarly, the dual EFIE operator simplifies as

$$\begin{aligned}\mathbb{T}^{\text{reg}} = &-a^{-1}\mathbb{P}^{\Sigma H}\mathbb{T}_{\mathrm{s}}\mathbb{P}^{\Sigma H} - c_0^{-1}\mathbb{P}^{\Sigma H}\frac{\partial}{\partial t}\mathbb{T}_{\mathrm{s}}\mathbb{P}^{\Lambda}\\ &- c_0^{-1}\mathbb{P}^{\Lambda}\frac{\partial}{\partial t}\mathbb{T}_{\mathrm{s}}\mathbb{P}^{\Sigma H} - \frac{a}{c_0^2}\mathbb{P}^{\Lambda}\frac{\partial^2}{\partial t^2}\mathbb{T}_{\mathrm{s}}\mathbb{P}^{\Lambda} + a\mathbb{T}_{\mathrm{h}}.\end{aligned} \quad (51)$$

The space-discretized formulation of the regularized Calderón EFIE is

$$\mathrm{T}_{\mathrm{c}}^{\text{reg}} * \boldsymbol{y}_{\Gamma} = -\eta_0^{-1}\mathbb{T}^{\text{reg}}\boldsymbol{G}_m^{-1} * \boldsymbol{R} * \boldsymbol{e}_{\Gamma}^{\text{inc}}, \quad (52)$$

where

$$\boldsymbol{R} = \frac{c_0}{a}\int_{-\infty}^{\cdot} \boldsymbol{P}^{\Lambda H} + \boldsymbol{P}^{\Sigma} \quad \text{and} \quad \boldsymbol{j}_{\Gamma} = \left(\boldsymbol{P}^{\Lambda H} + \frac{a}{c_0}\frac{\partial}{\partial t}\boldsymbol{P}^{\Sigma}\right)\boldsymbol{y}_{\Gamma}. \quad (53)$$

The right-hand side of (52) has a temporal integral which is directly evaluated on the incident field to avoid quadratic complexity with the number of time steps. This leads to the following MOT

$$\begin{aligned}\boldsymbol{Z}_{\mathrm{T}_{\mathrm{c}}^{\text{reg}},0}\boldsymbol{y}_i = &-\eta_0^{-1}\sum_{j=0}^{N_{\text{conv}}} \boldsymbol{Z}_{\mathrm{T}^{\text{reg}},j}\left(\widetilde{\boldsymbol{P}}^{\Sigma}\boldsymbol{e}_{i-j}^{\text{inc}} + \frac{c_0}{a}\widetilde{\boldsymbol{P}}^{\Lambda H}\boldsymbol{e}_{i-j}^{\text{prim}}\right)\\ &- \sum_{j=1}^{N_{\text{conv}}} \boldsymbol{Z}_{\mathrm{T}_{\mathrm{c}}^{\text{reg}},j}\boldsymbol{y}_{i-j},\end{aligned} \quad (54)$$

where $\widetilde{\boldsymbol{P}}^{\Lambda H} = \boldsymbol{P}^{\Lambda H}\otimes I_p$, $\widetilde{\boldsymbol{P}}^{\Sigma} = \boldsymbol{P}^{\Sigma}\otimes I_p$, the vector sequence $\boldsymbol{y}_n$ is the time discretization of $\boldsymbol{y}_{\Gamma}(t)$, and the time domain interaction matrices $\boldsymbol{Z}_{\mathrm{T}_{\mathrm{c}}^{\text{reg}},i}$ and $\boldsymbol{Z}_{\mathrm{T}^{\text{reg}},i}$ are respectively generated by the convolution quadrature method described in Subsection B of the space-discretized operators $\mathrm{T}_{\mathrm{c}}^{\text{reg}}$ and $\mathrm{T}^{\text{reg}} = \mathbb{T}^{\text{reg}}\boldsymbol{G}_m^{-1}$. Once the computation of $\boldsymbol{y}_n$ is done, the current $\boldsymbol{j}$ still has to be evaluated. The convolution quadrature discretization of the time derivative is

$$\begin{aligned}\boldsymbol{Z}_{\frac{\partial}{\partial t},i} &= \mathcal{Z}^{-1}\left\{z \to \overline{\mathbf{s}}_{\text{cq}}(z)\right\}_i\\ &= \Delta t^{-1}\overline{\boldsymbol{A}^{-1}}\delta_{i,0} - \Delta t^{-1}\overline{\boldsymbol{A}^{-1}\mathbf{1}_p\boldsymbol{b}^T\boldsymbol{A}^{-1}}\delta_{i,1},\end{aligned} \quad (55)$$

where $\delta_{i,0}$ is the Kronecker delta, $\overline{\boldsymbol{A}^{-1}} = I_{N_s}\otimes\boldsymbol{A}^{-1}$ and $\overline{\boldsymbol{A}^{-1}\mathbf{1}_p\boldsymbol{b}^T\boldsymbol{A}^{-1}} = I_{N_s}\otimes\boldsymbol{A}^{-1}\mathbf{1}_p\boldsymbol{b}^T\boldsymbol{A}^{-1}$ [30]. Therefore, the current solution is obtained as

$$\begin{aligned}\boldsymbol{j}_i &= \left[\left(\widetilde{\boldsymbol{P}}^{\Lambda H} + \frac{a}{c_0}\boldsymbol{Z}_{\frac{\partial}{\partial t}}\widetilde{\boldsymbol{P}}^{\Sigma}\right) *_s \boldsymbol{y}\right]_i\\ &= \widetilde{\boldsymbol{P}}^{\Lambda H}\boldsymbol{y}_i + \frac{a}{\Delta t c_0}\widetilde{\boldsymbol{P}}^{\Sigma}\left[\overline{\boldsymbol{A}^{-1}}\boldsymbol{y}_i - \overline{\boldsymbol{A}^{-1}\mathbf{1}_p\boldsymbol{b}^T\boldsymbol{A}^{-1}}\boldsymbol{y}_{i-1}\right].\end{aligned} \quad (56)$$

IV. Results

To test the effectiveness of the proposed schemes, simulations have been realized with different geometries, excited by a Gaussian pulse plane wave

$$\boldsymbol{e}^{\text{inc}}(\boldsymbol{r},t) = A_0\exp\left(-\frac{\left(t-\frac{\hat{\boldsymbol{k}}\cdot\boldsymbol{r}}{c}\right)^2}{2\sigma^2}\right)\hat{\boldsymbol{p}}, \quad (57)$$

$$\boldsymbol{h}^{\text{inc}}(\boldsymbol{r},t) = \frac{1}{\eta_0}\hat{\boldsymbol{k}}\times\boldsymbol{e}^{\text{inc}}(\boldsymbol{r},t), \quad (58)$$

where $\sigma = 6/(2\pi f_{\text{bw}})$, $\hat{\boldsymbol{p}} = \hat{\boldsymbol{x}}$, $\hat{\boldsymbol{k}} = -\hat{\boldsymbol{z}}$, $A_0 = 1\,\text{V}\,\text{m}^{-1}$ and f_{bw} is the frequency bandwidth. Notice that this frequency bandwidth is proportional to the maximal frequencies excited by the pulse Gaussian. In this work, the Runge-Kutta Radau IIA method of stage 2 is used for all simulations [17], [18]. The time step size Δt has been chosen equal to $(\psi f_{\max})^{-1}$ where $f_{\max}$ is the upper frequency of the excitation and $\psi = 3$ is an oversampling parameter.

A. Canonical geometries

To illustrate the key properties of the newly proposed schemes, namely the Calderón preconditioned MOT (44) and the Calderón preconditioned MOT regularized by the quasi-Helmholtz-projectors (54), they are compared in the case of modelling of canonical scatterers to other formulations present on the literature: the EFIE MOT schemes (MOT EFIE) (21), the time-differentiated one (MOT TD-EFIE) (24), the MOT regularized by the quasi-Helmholtz-projectors

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(MOT qH-EFIE) [30]. In this subsection, the excitation parameters have been chosen not to excite the first resonant mode of the geometries.

The first set of numerical tests were performed on the unit sphere, discretized with 270 RWG functions. The intensity of the resulting currents at one point of the geometry are shown in Figure 4. As expected, the time differentiated and the non-differentiated EFIEs are the only formulations suffering from DC-instabilities on this simply connected scenario. In addition, the condition number of the matrices to invert for each MOT are presented in Figure 5 and Figure 6 with respect to the time step size Δt and the mesh density h^{-1}. The standard EFIE formulation and its time-differentiated counterpart suffer from ill-conditioning at large time steps while the stabilized ones remain well-conditioned. Instead, only the Calderón preconditioned formulations presented in this work remain well-conditioned for dense discretizations (Figure 6).

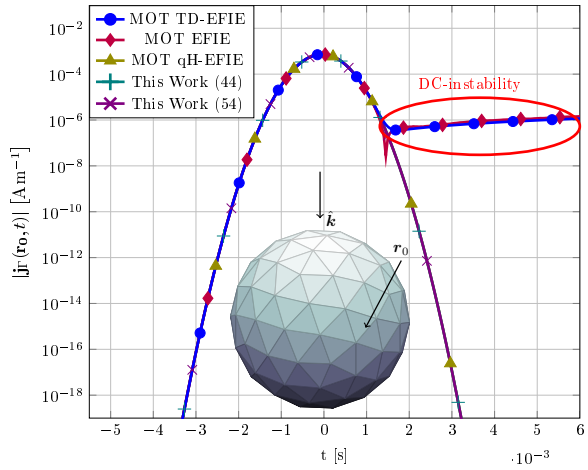

FIGURE 4. Evolution in time of the current intensity at $\boldsymbol{r}_0 = (-0.36, 0.89, 0.11)$m of the sphere where $f_{\text{bw}} = 25$ kHz with parameters $N_s = 270$ and $\Delta t = 4.5 \times 10^4$ ns.

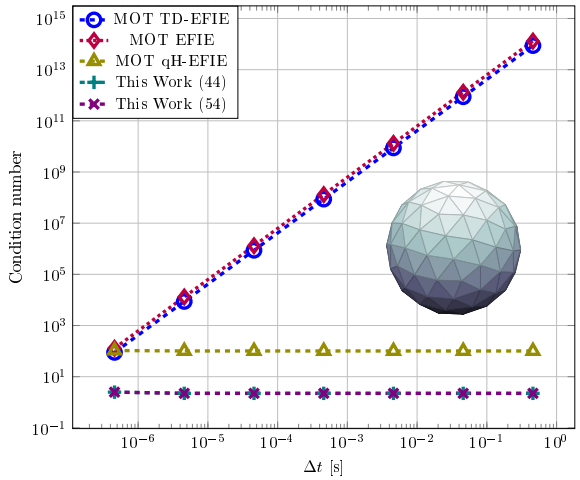

FIGURE 5. Condition number of the EFIE MOT matrices with respect to the time step size Δt ($N_s = 270$) on a sphere.

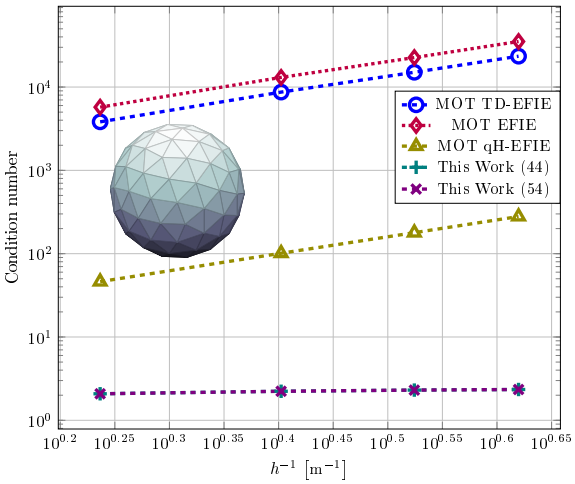

FIGURE 6. Condition number of the EFIE MOT matrices with respect to the mesh size h ($\Delta t = 45$ ns) on a sphere.

The second set of tests focused on the stability of the different formulations when modelling multiply connected scatters, here a torus with inner radius of 0.2 m and outer radius of 0.5 m. The current densities at the probe point are shown in Figure 7 and the conditioning studies are represented in Figure 8 and Figure 9. In line with the polynomial eigenvalue analysis of the non-regularized Calderón EFIE formulation (Figure 3), the formulation (44) suffers from DC instability. Moreover, the static nullspace of the continuous operator deteriorates the condition number of the matrix to invert for large time steps (Figure 9). However, the newly proposed regularized Calderón formulation (54) is stable and remains well-conditioned at large time steps and dense meshes for this geometry.

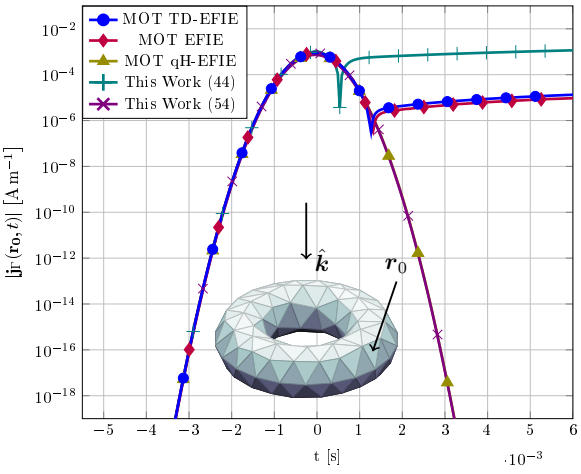

FIGURE 7. Evolution in time of the current intensity at $\boldsymbol{r}_0 = (0.62, -0.13, 0.11)$m of the torus where $f_{\text{bw}} = 25$ kHz with parameters $N_s = 387$ and $\Delta t = 4.5 \times 10^4$ ns.

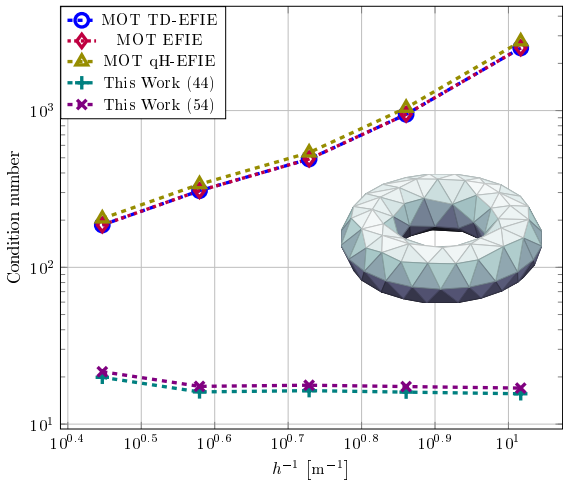

FIGURE 8. **Condition number with respect to the mesh size h ($\Delta t = 4.5$ ns) on a torus of the EFIE MOT schemes.**

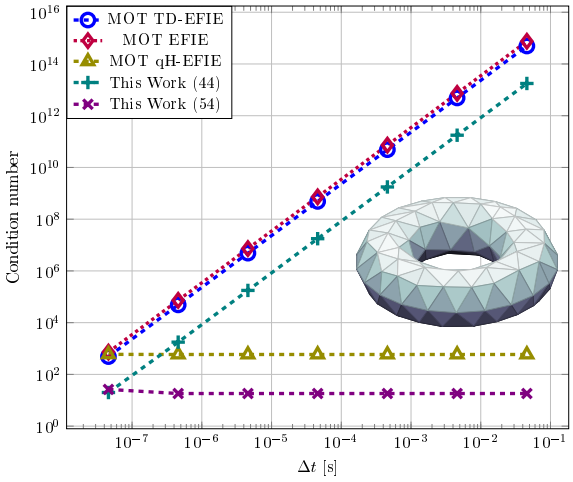

FIGURE 9. **Condition number with respect to the time step size Δt ($N_s = 387$) on a torus of the EFIE MOT schemes.**

B. Non-canonical geometries

The final set of numerical tests is dedicated to more complex test structures (Figure 10). In addition, instead of direct solver we rely on the iterative solver GMRES with different relative target tolerances ϵ [53] . For practical reason, the maximum number of iteration has been limited to 200 without restart.

All the structures have been illuminated by a pulse Gaussian plane wave with $f_{\text{bw}} = 1.6$MHz. Table 1 shows the condition number and the number of iterations needed. The Calderón preconditioned formulation (CP-EFIE) (44) and the Calderón preconditioned formulation regularized by the quasi-Helmholtz-projectors (qH-CP-EFIE) (54) are the only formulations requiring less than 200 iterations to converge at each time steps. As expected, the condition number of the CP-EFIE is high on non-simply connected geometries because of the presence of the operator DC nullspace on these structures. Although, in this case, the number of iterations remains low because the nullspace results in isolated elements in the spectrum, the solution is corrupted by DC instability arising from this nullspace. This phenomenon is absent for the qH-CP-EFIE formulation which is free from high conditioning or DC instability and yields stable solutions up to the target precision of the iterative solver.

Finally, the setup time of the different formulations for the cube with multiple holes scatterer and the corresponding times required to compute a single time step are listed

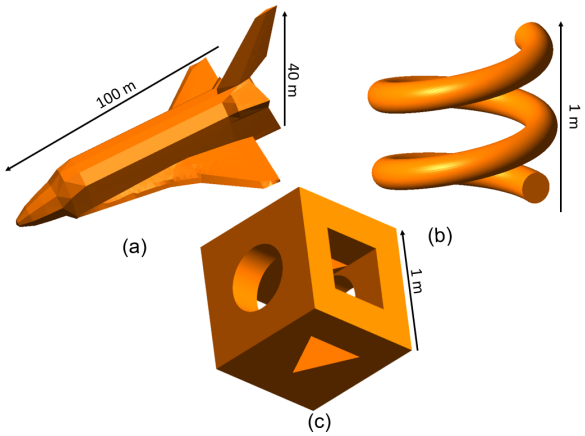

FIGURE 10. **Test structures. (a) space shuttle; (b) compression spring; (c) cube with multiple holes.**

Space shuttle: $N_s = 4311$				
	Cond(Z_0)	N_{iter}		
Tolerance ϵ	$\emptyset$	10^{-3}	10^{-6}	10^{-10}
MOT TD-EFIE	$4.3 \cdot 10^6$	> 200	> 200	> 200
MOT EFIE	$6.5 \cdot 10^6$	> 200	> 200	> 200
MOT qH-EFIE	$5.3 \cdot 10^4$	> 200	> 200	> 200
This work (44)	38	15	28	41
This work (54)	41	4	18	33
Compression spring: $N_s = 3906$				
	Cond(Z_0)	N_{iter}		
Tolerance ϵ	$\emptyset$	10^{-3}	10^{-6}	10^{-10}
MOT TD-EFIE	$2.9 \cdot 10^6$	> 200	> 200	> 200
MOT EFIE	$4.3 \cdot 10^6$	> 200	> 200	> 200
MOT qH-EFIE	$1.6 \cdot 10^3$	57	154	> 200
This work (44)	158	8	31	50
This work (54)	133	2	19	39
Cube with multiple holes: $N_s = 3267$				
	Cond(Z_0)	N_{iter}		
Tolerance ϵ	$\emptyset$	10^{-3}	10^{-6}	10^{-10}
MOT TD-EFIE	$1.2 \cdot 10^9$	> 200	> 200	> 200
MOT EFIE	$1.8 \cdot 10^9$	> 200	> 200	> 200
MOT qH-EFIE	$6.6 \cdot 10^3$	50	196	> 200
This work (44)	$9.5 \cdot 10^6$	5	20	82
This work (54)	50	1	11	25

TABLE 1. **Condition number and maximum number of iterations of the different formulations with different relative tolerances.**

Cube with holes: N_s = 3267			
	Setup time	Computation time for a single time step	DC stable
MOT TD-EFIE	10 min	69 s	No
MOT EFIE	10 min	78 s	No
MOT qH-EFIE	19 min	3.4 s	Yes
This work (44)	134 min	0.84 s	No
This work (54)	138 min	0.40 s	Yes

TABLE 2. Computation times of the different formulations with $\epsilon = 10^{-6}$.

in Table 2. As expected, the preconditioned formulations require a longer setup time, which is rapidly compensated by a faster computation of the time steps. The setup time of these formulations can be significantly reduced by leveraging fast solvers [54]; this, however, goes beyond the scope of this work.

Conclusion

In this paper, novel Calderón preconditioned techniques have been presented for the time domain Electric Field Integral Equations solved with Marching-On-In-Time with convolution quadratures. These formulations eliminate the DC-instability for simply and multiply connected geometries. In addition, they cure the *h*-refinement and large time step breakdowns and generate well-conditioned Marching-On-In Time. Finally, numerical results on complex geometries showcased the effectiveness of the proposed schemes.

References

[1] A. F. Peterson, S. L. Ray, R. Mittra, I. of Electrical, and E. Engineers, *Computational methods for electromagnetics*. IEEE press New York, 1998, vol. 351.

[2] J.-M. Jin, *Theory and computation of electromagnetic fields*. John Wiley & Sons, 2011.

[3] W. C. Gibson, *The method of moments in electromagnetics*. Chapman and Hall/CRC, 2021.

[4] D. Colton and R. Kress, *Integral equation methods in scattering theory*. SIAM, 2013.

[5] C. L. Bennett Jr, *A technique for computing approximate electromagnetic impulse response of conducting bodies*. Purdue University, 1968.

[6] S. M. Rao and D. R. Wilton, "Transient scattering by conducting surfaces of arbitrary shape," *IEEE Transactions on Antennas and Propagation*, vol. 39, no. 1, pp. 56–61, 1991.

[7] F.-J. Sayas, *Retarded potentials and time domain boundary integral equations: A road map*. Springer, 2016, vol. 50.

[8] Y.-S. Chung, T. K. Sarkar, B. H. Jung, *et al.*, "Solution of time domain electric field integral equation using the laguerre polynomials," *IEEE Transactions on Antennas and Propagation*, vol. 52, no. 9, pp. 2319–2328, 2004.

[9] C. Lubich, "Convolution quadrature and discretized operational calculus. i," *Numerische Mathematik*, vol. 52, no. 2, pp. 129–145, 1988.

[10] C. Lubich, "Convolution quadrature and discretized operational calculus. ii," *Numerische Mathematik*, vol. 52, no. 4, pp. 413–425, 1988.

[11] M. Schanz and H. Antes, "A new visco-and elastodynamic time domain boundary element formulation," *Computational Mechanics*, vol. 20, no. 5, pp. 452–459, 1997.

[12] L. Banjai, M. Messner, and M. Schanz, "Runge–kutta convolution quadrature for the boundary element method," *Computer methods in applied mechanics and engineering*, vol. 245, pp. 90–101, 2012.

[13] X. Wang and D. S. Weile, "Implicit runge-kutta methods for the discretization of time domain integral equations," *IEEE transactions on antennas and propagation*, vol. 59, no. 12, pp. 4651–4663, 2011.

[14] C. Lubich and A. Ostermann, "Runge-kutta methods for parabolic equations and convolution quadrature," *mathematics of computation*, vol. 60, no. 201, pp. 105–131, 1993.

[15] A. V. Oppenheim, J. R. Buck, and R. W. Schafer, *Discrete-time signal processing. Vol. 2*. Upper Saddle River, NJ: Prentice Hall, 2001.

[16] F. Chipman, "A-stable runge-kutta processes," *BIT Numerical Mathematics*, vol. 11, no. 4, pp. 384–388, 1971.

[17] C. Runge, "Über die numerische auflösung von differentialgleichungen," *Mathematische Annalen*, vol. 46, no. 2, pp. 167–178, 1895.

[18] W. Kutta, "Beitrag zur naherungsweisen integration totaler differentialgleichungen," *Z. Math. Phys.*, vol. 46, pp. 435–453, 1901.

[19] J. C. Butcher, *Numerical methods for ordinary differential equations*. John Wiley & Sons, 2016.

[20] L. Skvortsov, "Accuracy of runge-kutta methods applied to stiff problems," *Computational mathematics and mathematical physics*, vol. 43, no. 9, pp. 1320–1330, 2003.

[21] L. Banjai and M. Kachanovska, "Fast convolution quadrature for the wave equation in three dimensions," *Journal of Computational Physics*, vol. 279, pp. 103–126, 2014.

[22] T. Maruyama, T. Saitoh, T. Bui, and S. Hirose, "Transient elastic wave analysis of 3-d large-scale cavities by fast multipole bem using implicit runge–kutta convolution quadrature," *Computer Methods in Applied Mechanics and Engineering*, vol. 303, pp. 231–259, 2016.

[23] G. H. Golub and C. F. Van Loan, *Matrix computations*. JHU press, 2013.

[24] F. P. Andriulli, H. Bagci, F. Vipiana, G. Vecchi, and E. Michielssen, "Analysis and regularization of the td-efie low-frequency breakdown," *IEEE transactions on antennas and propagation*, vol. 57, no. 7, pp. 2034–2046, 2009.

[25] N.-W. Chen, K. Aygün, and E. Michielssen, "Integral-equation-based analysis of transient scattering and radiation from conducting bodies at very low frequencies," *IEE Proceedings-Microwaves, Antennas and Propagation*, vol. 148, no. 6, pp. 381–387, 2001.

[26] F. P. Andriulli, K. Cools, I. Bogaert, and E. Michielssen, "On a well-conditioned electric field integral operator for multiply connected geometries," *IEEE transactions on antennas and propagation*, vol. 61, no. 4, pp. 2077–2087, 2012.

[27] I. Bogaert, K. Cools, F. P. Andriulli, and H. Bağcı, "Low-frequency scaling of the standard and mixed magnetic field and müller integral equations," *IEEE Transactions on Antennas and Propagation*, vol. 62, no. 2, pp. 822–831, 2013.

[28] F. P. Andriulli, K. Cools, F. Olyslager, and E. Michielssen, "Time domain calderón identities and their application to the integral equation analysis of scattering by pec objects part ii: Stability," *IEEE transactions on antennas and propagation*, vol. 57, no. 8, pp. 2365–2375, 2009.

[29] K. Cools, F. P. Andriulli, F. Olyslager, and E. Michielssen, "Time domain calderón identities and their application to the integral equation analysis of scattering by pec objects part i: Preconditioning," *IEEE Transactions on Antennas and Propagation*, vol. 57, no. 8, pp. 2352–2364, 2009.

[30] A. Dély, F. P. Andriulli, and K. Cools, "Large time step and dc stable td-efie discretized with implicit runge–kutta methods," *IEEE Transactions on Antennas and Propagation*, vol. 68, no. 2, pp. 976–985, 2019.

[31] Y. Shi, H. Bağcı, and M. Lu, "On the static loop modes in the marching-on-in-time solution of the time-domain electric field integral equation," *IEEE Antennas and Wireless Propagation Letters*, vol. 13, pp. 317–320, 2014.

[32] F. P. Andriulli, K. Cools, H. Bagci, *et al.*, "A multiplicative calderon preconditioner for the electric field integral equation," *IEEE Transactions on Antennas and Propagation*, vol. 56, no. 8, pp. 2398–2412, 2008.

[33] K. Cools, F. P. Andriulli, F. Olyslager, and E. Michielssen, "Nullspaces of mfie and calderón preconditioned efie operators applied to toroidal surfaces," *IEEE Transactions on Antennas and Propagation*, vol. 57, no. 10, pp. 3205–3215, 2009.

[34] Y. Beghein, K. Cools, and F. P. Andriulli, "A dc-stable, well-balanced, calderón preconditioned time domain electric field integral equation," *IEEE Transactions on Antennas and Propagation*, vol. 63, no. 12, pp. 5650–5660, 2015.

[35] P. Cordel, A. Dély, A. Merlini, and F. P. Andriulli, "Calderón preconditioners for the td-efie discretized with convolution quadratures," in *2022 IEEE International Symposium on Antennas and Propagation and USNC-URSI Radio Science Meeting (AP-S/URSI)*, IEEE, 2022, pp. 1672–1673.

[36] R. F. Harrington and J. L. Harrington, *Field computation by moment methods*. Oxford University Press, Inc., 1996.

[37] S. Rao, D. Wilton, and A. Glisson, "Electromagnetic scattering by surfaces of arbitrary shape," *IEEE Transactions on antennas and propagation*, vol. 30, no. 3, pp. 409–418, 1982.

[38] L. Banjai and F. Sayas, "Integral equation methods for evolutionary pde," *Springer Ser. Comput. Math., Springer, Cham*, 2022.

[39] L. Schwartz, *Mathematics for the physical sciences*. Courier Dover Publications, 2008.

[40] A. Dély, A. Merlini, K. Cools, and F. P. Andriulli, "Convolution quadrature time domain integral equation methods for electromagnetic scattering," *Advances in Time-Domain Computational Electromagnetic Methods*, pp. 321–359, 2022.

[41] P.-A. Raviart and J.-M. Thomas, "A mixed finite element method for 2-nd order elliptic problems," in *Mathematical aspects of finite element methods*, Springer, 1977, pp. 292–315.

[42] Y. Beghein, K. Cools, and D. De Zutter, "A temporal galerkin discretization of the charge-current continuity equation," in *2013 International Conference on Electromagnetics in Advanced Applications (ICEAA)*, IEEE, 2013, pp. 628–631.

[43] P. W. van Diepen, R. J. Dilz, A. P. Zwamborn, and M. C. van Beurden, "The role of jordan blocks in the mot-scheme time domain efie linear-in-time solution instability," *Progress In Electromagnetics Research B*, vol. 95, no. 95, p. 123, 2022.

[44] S. Walker, M. Bluck, and I. Chatzis, "The stability of integral equation time-domain scattering computations for three-dimensional scattering; similarities and differences between electrodynamic and elastodynamic computations," *International Journal of Numerical Modelling: Electronic Networks, Devices and Fields*, vol. 15, no. 5-6, pp. 459–474, 2002.

[45] S. B. Adrian, A. Dély, D. Consoli, A. Merlini, and F. P. Andriulli, "Electromagnetic integral equations: Insights in conditioning and preconditioning," *IEEE Open Journal of Antennas and Propagation*, 2021.

[46] S. B. Adrian, A. Dély, D. Consoli, A. Merlini, and F. P. Andriulli, "Electromagnetic integral equations: Insights in conditioning and preconditioning," *IEEE Open Journal of Antennas and Propagation*, vol. 2, pp. 1143–1174, 2021. DOI: 10.1109/OJAP.2021.3121097.

[47] J.-C. Nédélec, *Acoustic and electromagnetic equations: integral representations for harmonic problems*. Springer, 2001, vol. 144.

[48] G. C. Hsiao and R. E. Kleinman, "Mathematical foundations for error estimation in numerical solutions of integral equations in electromagnetics," *IEEE transactions on Antennas and Propagation*, vol. 45, no. 3, pp. 316–328, 1997.

[49] F. P. Andriulli, K. Cools, I. Bogaert, H. Bagci, P. Yla-Ojiala, and E. Michielssen, "Analysis and discretization of the yukawa-calderon preconditioned cfie," in *28th Annual Review of Progress in Applied Computational Electromagnetics (ACES-2012)*, Curran Associates, Inc., 2012, pp. 454–463.

[50] A. Buffa and S. Christiansen, "A dual finite element complex on the barycentric refinement," *Mathematics of Computation*, vol. 76, no. 260, pp. 1743–1769, 2007.

[51] K. Cools, F. Andriulli, D. De Zutter, and E. Michielssen, "Accurate and conforming mixed discretization of the mfie," *IEEE antennas and wireless propagation letters*, vol. 10, pp. 528–531, 2011.

[52] F. P. Andriulli, K. Cools, F. Olyslager, and E. Michielssen, "The "dottrick tdefie": A dc stable integral equation for analyzing transient scattering from pec bodies," in *2008 IEEE Antennas and Propagation Society International Symposium*, IEEE, 2008, pp. 1–4.

[53] Y. Saad and M. H. Schultz, "Gmres: A generalized minimal residual algorithm for solving nonsymmetric linear systems," *SIAM Journal on scientific and statistical computing*, vol. 7, no. 3, pp. 856–869, 1986.

[54] H. Bagci, F. P. Andriulli, K. Cools, F. Olyslager, and E. Michielssen, "A calderón multiplicative preconditioner for the combined field integral equation," *IEEE Transactions on Antennas and Propagation*, vol. 57, no. 10, pp. 3387–3392, 2009. DOI: 10.1109/TAP.2009.2029389.

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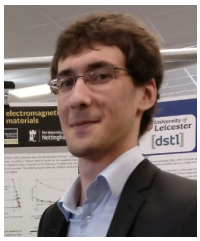

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