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## Optimization of a smart beam for monitoring a connected inaccessible mechanical system : application to bone-implant coupling

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#### 6 Abstract

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This paper deals with the optimization of piezoelectric patches positioning on a beam attached to an inaccessible system. Based on the electro-mechanical coupling coefficients, which are calculated from the mode shapes curvatures, the problem is to provide an optimal positioning of the piezoelectric patches in order to target the modes sensitive to beam boundary conditions. Following the theoretical description of a beam instrumented with collocated piezoelectric patches, a placement optimization strategy is proposed. This strategy lies on the definition of utility functions, based on modal weights and modal sensitivities, calculated from the coupling coefficient. An experimental validation of the method is performed on a concrete case study corresponding to real-time implant stability monitoring. A beam is temporarily rigidly fixed to the implant during its insertion into the bone cavity to allow impaction by the surgeon. The positions of the sensors on the beam are optimized to focus on the beam's modes carrying information on the bone-implant interface, the biomechanical issue being the real-time maximization of implant stability. The originality of the approach lies on the sensor placement optimization on the beam connected to the implant to maximize or minimize simultaneously the amplitude of several modes on the frequency response function at resonances, depending on their sensitivity to the bone-implant interface. The results show very good performance of the piezoelectric placement strategy proposed in this paper, paving the way for new applications of piezoelectric patches design and placement on structures.

Keywords: electro-mechanical coupling, piezoelectric patch optimization, mode observability, smart device, boundary conditions monitoring, bone-implant coupling

#### 7 1. Introduction

- Piezoelectric ceramic transducers bonded to flexible structures are widely employed in the context of
   dynamic measurements. They offer the advantage to be used as sensors or actuators due to to their property
- <sup>10</sup> to convert mechanical strains into electrical voltage and *vice versa*. The modal electro-mechanical coupling

factors (MEMCF) [1, 2] associated to one piezoelectric patch and one mode of vibration of the structure is
 determinant for vibration reduction control [3].

In the literature, there are several examples of multi-modal control, either active or passive with shunted 13 piezoelectric patches on beams [4, 5, 6] and plates [7, 8, 9]. For these applications, the piezoelectric patches 14 are generally uniformly distributed on the structure, considering that every mode has the same weight 15 and the optimization concerns the associated electric circuit. However, in specific applications, the design 16 constraints limit the number of sensor/actuator patches over the structure. Therefore, another strategy for 17 vibration reduction lies in the optimization of the placement and the design of the piezoelectric patch on 18 the structure, as presented in the exhaustive review of Gupta et al. [10]. Among the different criteria used 19 in the literature, maximizing the degree of mode controllability and observability are of particular interest. 20 For instance, Ducarne et al. [11] proposed a parametric study on the placement and geometry of the patches 21 in order to maximize the electro-mechanical coupling coefficient of one specific mechanical vibration mode, 22 to which the shunt is tuned. Others authors studied the same problematic of placement optimization, by 23 considering active control [12, 13]. Moreover, Halim and Moheimani [14] addressed the notions of both 24 spatial and modal controllability, to find the optimal locations of piezoelectric transducers on the plate. The 25 procedure is based on the concept of the spatial  $\mathcal{H}$  norm, used to describe the spatial structural response 26 of the system in an average sense. The optimization relies on the simultaneous vibration reduction over the 27 entire structure, while maintaining modal controllability and observability of the selected vibration modes. 28 In this paper, the specific issue of optimizing the *observability* of a set of modes sensitive to the beam 29 boundary conditions is addressed. Lots of vibro-acoustic applications are concerned by this issue, and among 30 them, this paper focus on a biomedical application. In the orthopedic field, several methods are under 31 development to assist and guide the surgeon during implant insertion into the bone, in order to achieve an 32 optimal implant stability, corresponding to maximizing implant surrounding stresses while avoiding bone 33 fracture risk. Among the different techniques, vibration measurement have proven to be efficient to monitor 34 implant insertion into the bone [15, 16, 17]. Nevertheless, the fixation of sensors or probes on the implant 35 and/or the bone is incompatible with a clinical application [18]. Therefore, several authors studied directly 36 the vibrations of a square shape structure, named ancillary, and used by the surgeon to insert the implant into 37 the bone [19, 20], which can easily be equipped with sensors and allow a good accessibility for measurement. 38 This structure has one free extremity and another one bonded to the implant during its insertion. Due to 39 this fixation, a previous study by our group showed that specific modes of vibration of this square beam 40 structure carry information on the bone-implant system, and in particular on the system rigidity [20]. The 41

<sup>42</sup> issue is then to propose an optimization approach for piezoelectric patch positioning on the square beam
<sup>43</sup> structure, in order to maximize or minimize the *observability* of the modes, depending on their sensitivity
<sup>44</sup> to the implant stability.

Several authors studied the piezoelectric placement optimization in order to target specific modes of 45 interest. For instance, Rosi et al. [21] developed a similar transducer optimization approach for reducing 46 the radiated sound power of thin plates produced by a set of modes. In the field of structural health 47 monitoring, the position of the collocated piezoelectric transducers can be optimized in order to detect 48 damage in thin plate-like structures based on the minimization of the Bayes risk [22]. Interestingly, not only 49 the position but also the local beam stiffness could be adjusted by using the feedback voltage output from the 50 piezoelectric sensors to tune the natural vibration of mode shape sensitive to the existence of a crack [23, 24]. 51 In these examples, the criteria of mode selection differs from the sensitivity to beam boundary conditions. 52 Moreover, in our study, the piezoelectric patches are not shunted since they are used as sensors and not for 53 control purpose. As only the position of the patch on the beam is studied, no optimization algorithm are 54 needed, contrary to what is often done for passive control applications [25, 26, 27]. Eventually, methods for 55 boundary force measurement in beam-like structures have been investigated in the literature by displacement 56 measurement [28] or distributed sensor measurement [29]. However, the issue of our application is to monitor 57 implant stability throughout specific resonance frequency rather than boundary forces measurement. 58

The originality of the present paper comes from the use of an optimization method, based on the modal 59 electro-mechanical coupling factor (MEMCF), in order to maximize the *observability* of a subset of modes 60 sensitive to the beam boundary conditions, related to the implant stability for the considered case study. 61 The choice of this application comes from the need to maximize the initial fixation of orthopedic implants 62 during their insertion into the bone, which is one of the main biomechanical issues related to this surgery. 63 Since this fixation is tricky to maximize for the surgeon without creating a bone fracture, some quantitative 64 methods are underdevelopment to measure in real time implant stability. Some of them are based on the 65 vibro-acoustic characterization of the bone-implant system [17, 30, 20]. In particular, the results of [20], 66 which showed that the implant stability is correlated to specific resonance frequencies of the surgical tool 67 used to insert the implant, named "ancillary" and assimilated to a beam with one free extremity and the 68 other one bounded to the implant being inserted. To improve the vibration measurement method for implant 69 stability characterization by focusing on specific modes of interest, optimal patches positions are determined 70 hereafter on the beam which has one free extremity and the other bounded to the implant being inserted. 71

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The paper is organized in two main sections. Section 2 is dedicated to the description of the smart

<sup>73</sup> beam electromechanical behavior and to the transducers position optimization strategy based on mode <sup>74</sup> observability. In particular, the modal electromechanical coupling between the patch and the beam and the <sup>75</sup> utility functions chosen for the optimization of a subset of modes are defined. In Section 3, an experimental <sup>76</sup> validation of the optimization strategy for implant insertion monitoring is carried out.

#### 77 2. Equations of the smart device

78 2.1. Geometry and kinematics



Figure 1: a) Geometry of the beam and connection with the inaccessible system b) Piezoelectric patches positions and numbering on the beam and c) A section containing *i*-th patches couple and the associated electric circuit with  $V_i$  the potential difference measured by each patches couple *i* 

<sup>79</sup> Consider an elastic beam of length  $\ell$  with a rectangular section  $b \times h$  (Fig. 1a), on which piezoelectric <sup>80</sup> layers are bonded at  $n_p$  segments along its longitudinal axis x. The beam is fixed at  $x_1 = \ell$  to an external <sup>81</sup> mechanical system. At a segment  $x \in [a_i, b_i]$  ( $i = 1..n_p$ ), the beam is sandwiched between two identical <sup>82</sup> piezoelectric patches of thickness  $h_p$  which are put symmetrically at the upper and the lower surfaces of the <sup>83</sup> beam (i.e at z = h/2 and z = -h/2) (Fig. 1b). Each couple of piezoelectric layers is polarized along  $-e_z$ <sup>84</sup> (Fig. 1c) and thus form a pure bending sensor/actuator.

We assume the piezoelectric layers have the same width b with the beam's section. Each piezoelectric layer is covered on both its upper and lower surfaces by electrodes which are equipotential surfaces. The electric field E assumed to be constant, uniform and parallel to the transverse direction  $e_z$ , is :  $E_z = -V/h_p$ , where V the voltage applied between the electrodes. The purely elastic part of the beam is grounded.

In this study, all forces acting along  $e_y$  are neglected, the system may be described by a 1D Euler-Bernouilli model in the plane (x, z). Moreover, by assuming that the longitudinal and rotation of the beam are negligible, only flexural movement is considered. All components of the displacement field at a point of coordinates (x, y, z) may expressed in terms of the transverse displacement w(x) of the beam's neutral axis:

$$u_x(x, y, z, t) = -zw'(x, t) \tag{1}$$

$$u_y(x, y, z, t) = 0 \tag{2}$$

$$u_z(x, y, z, t) = w(x, t) \tag{3}$$

#### 93 2.2. Consitutive law and governing equations

<sup>94</sup> By using the Euler-Bernoulli beam model, the transverse stress  $\sigma_{zz}$  and the shear strain  $\epsilon_{xz}$  are negligible. <sup>95</sup> Consequently, the only non-zero strain component is:  $\epsilon_{xx} = u'_x = -zw''$  and the full piezoelectric constitutive <sup>96</sup> equation (see Appendix A) reduces to:

$$\sigma_{xx} = Y\epsilon_{xx} - e_{31}E_z,\tag{4}$$

$$D_z = e_{31}\epsilon_{xx} + \varepsilon_{33}E_z. \tag{5}$$

At a point located in the elastic beam (i.e. for  $z \in [-h/2, h/2]$ ):  $Y = Y_b$ ,  $\rho = \rho_b$ ,  $e_{31} = 0$  and  $\varepsilon_{33} = 0$ . Otherwise, in the piezoelectric patches, the mechanical parameters are denoted by  $\rho = \rho_p$  and  $Y = Y_p$ . Since piezoelectric ceramics are transversely isotropic, the elastic parameter  $Y_p$  is an in-plane Young modulus obtained from the corresponding complete elasticity tensor c using the following relation (see Appendix A:

$$Y_p = \frac{(c_{11} - c_{12})((c_{11} + c_{12})c_{33} - 2c_{13}^2)}{c_{11}c_{33} - c_{13}^2}$$
(6)

In the following, the extended Hamilton principle will be used for obtaining the equations of motion and continuity conditions [31]. See Appendix B for more details.

<sup>103</sup> 2.2.1. Equations in a segment containing *i*-th patch  $(x \in [a_i, b_i])$ 

With the presence of the upper and lower piezoelectric patches, the beam has three layers and the kinetic variation  $\delta T^*$  may be expressed by:

$$\delta T^* = \int_{a_i}^{b_i} \left( \int_{-b/2}^{b/2} \int_{-H/2}^{H/2} \rho(\delta \dot{w}) \ \dot{w} \ dy dz \right) dx = \int_{a_i}^{b_i} \bar{m} \ \delta \dot{w} \ \dot{w} dx, \tag{7}$$

where  $H = h/2 + h_p$  and  $\bar{m} = b(\rho_b h + 2\rho_p h_p)$  which is effective mass par unit length of the beam. Similarly, by recalling that the electric potential is fixed ( $\delta E = 0$ ) and by substituting Eqs. (1-5) into (B.10), the potential variation  $\delta W_e^*$  reads:

$$\delta W_e^* = \int_{a_i}^{b_i} \left( \int_{-b/2}^{b/2} \int_{-H/2}^{H/2} (-\delta \epsilon_{xx} \ \sigma_{xx}) \, dy dz \right) dx \tag{8}$$

$$= \int_{a_i}^{b_i} \delta w'' \left(-\mathcal{D} \ w'' + \mathcal{G}V\right) \ dx,\tag{9}$$

109 where

$$\mathcal{D} = \int_{-b/2}^{b/2} \int_{-H/2}^{H/2} Y z^2 \, dy dz = Y_b \, \frac{bh^3}{12} + Y_p \, \frac{bh_p (3h^2 + 6hh_p + 4h_p^2)}{6},\tag{10}$$

$$\mathcal{G} = \int_{-h/2}^{b/2} \int_{-H/2}^{H/2} \frac{e_{31}z}{h_p} \, dy dz = e_{31} \, b(h+h_p) = b(h+h_p) \, d_{31}Y_p, \tag{11}$$

<sup>110</sup> are a bending stiffness and a coupling coefficient, respectively.

111 Then, integrating  $\delta W_e^*$  by parts twice with respect to x gives:

$$\delta W_{e}^{*} = \int_{a_{i}}^{b_{i}} \delta w \left(-\mathcal{D} \ w'' + \mathcal{G}V\right)'' \ dx - \left[\delta w' \left(-\mathcal{D} \ w'' + \mathcal{G}V_{i}\right)\right]_{a_{i}}^{b_{i}} + \left[\delta w \left(-\mathcal{D} \ w'' + \mathcal{G}V_{i}\right)'\right]_{a_{i}}^{b_{i}} \tag{12}$$

$$= \int_{a_i}^{b_i} \delta w M'' dx - [\delta w' M]_{a_i}^{b_i} + [\delta w M']_{a_i}^{b_i}$$
(13)

112 where  $M := -\mathcal{D} w'' + \mathcal{G}V$  is the bending moment in the beam.

If we only consider a distributed transverse load p(x,t) applied to the beam, the virtual work of external force is:

$$\delta W_{ext}^* = \int_{a_i}^{b_i} \delta w \ p \ dx. \tag{14}$$

In order to add the non conservative effects due to the internal damping of the beam, an extended Hamilton
principle approach has to be used:

$$\int_{t_1}^{t_2} \left(\delta T^* + \delta W_e^* + \delta W_{ext} + \delta W_{nc}\right) = 0 \tag{15}$$

where  $\delta W_{nc}$  is the work expended by the the non conservative actions. In the case of internal damping of the beam, we suppose the existence of a linear velocity dependent action and this term reads:

$$\delta W_{nc} = -\int_0^\ell \eta w \delta \dot{w} dx \tag{16}$$

<sup>119</sup> Finally, by applying the extended Hamilton's principle, the equations of motion reads:

$$\bar{m}\ddot{w} + \eta\dot{w} - M'' = p \tag{17}$$

along with the following continuity conditions, where  $[\![f(x)]\!] = f(x^+) - f(x^-)$  is a jump function:

$$\llbracket w \rrbracket = 0, \tag{18}$$

$$\llbracket w' \rrbracket = 0, \qquad (\text{continuity of rotation}), \qquad (19)$$

$$\llbracket \mathcal{D} \ w'' + \mathcal{G} V_i \rrbracket = 0, \qquad (\text{continuity of moment}), \qquad (20)$$

$$\llbracket (\mathcal{D} \ w'' + \mathcal{G}V_i)' \rrbracket = 0, \qquad (\text{continuity of shear force}). \tag{21}$$

As the piezoelectric patch is very thin with respect to the beam, the flexural rigidity  $\mathcal{D}$  and the mass per unit length  $\bar{m}$  of the beam, can be considered as uniform in  $[0, \ell]$ . In addition, the strain may be regarded as uniform over its thickness  $\epsilon_{xx} = -z_m w''$  where  $z_m = (h + h_p)/2$  is the coordinate of center line of the piezoelectric patch. The charge  $Q_i$  of the *i*-th patch is defined as the free electric contained electrode area [12] [11]:

$$Q_i = b \int_{a_i}^{b_i} -D_z \, dx = b \int_{a_i}^{b_i} (e_{31} z_m w'' - \varepsilon_{33} E_z) \, dx \tag{22}$$

$$= \mathcal{G}(w'(b_i) - w'(a_i)) + C_i V_i \tag{23}$$

where  $C_i = \varepsilon_{33}b(b_i - a_i)/h_p$ . If the piezoelectric segment is left open-circuited, then  $Q_i = 0$  and:

$$C_i V_i = -\mathcal{G}(w'(b_i) - w'(a_i)) \tag{24}$$

127 2.2.2. Equation for the whole domain  $(x \in [0, \ell])$ 

For describing the behavior of the beam with length  $\ell$  with  $n_p$  patches, one may use the same expression of the equation (17), with a more general formulation of bending moment:

$$M = -\mathcal{D}w'' + \sum_{i=1}^{n_p} \mathcal{G}V_i(H(x - a_i) + H(x - b_i)), \quad x \in [0, \ell].$$
(25)

where H(.) denotes the Heaviside function.

Moreover, considering the beam with one free extremity and the other mechanically connected to another mechanical system, the boundary conditions at  $x_1 = 0$  and  $x_1 = \ell$  are:

> w''(0) = 0, (no bending moment on the free end), (26) w'''(0) = 0, (no shear force on the free end), (27)  $\mathcal{D} w''(\ell) = M_{ext},$  (bending moment imposed by the external mechanical system), (28)  $\mathcal{D} w'''(\ell) = T_{ext},$  (shear force imposed by the external mechanical system). (29)

133 2.3. Modal expansion

The harmonic motion of the beam may be expressed by N short-circuit eigenmodes  $\omega_k, \phi_k(x)$  (i = 1..N):

$$w(x,t) = \sum_{k=1}^{N} \phi_k(x) q_k(t)$$
(30)

where  $q_k(t)$  is a time-dependent function.  $\omega_k, \phi_k(x)$  may solved from by Eq. (17) with p = 0 and V = 0:

$$-\omega_k^2 \bar{m} \phi_k + \mathcal{D} \phi_k^{(4)} = 0 \tag{31}$$

where <sup>(4)</sup> represents the fourth derivative with respect to the coordinate x. The modes  $\phi_k$  are orthogonal and may be normalized by:

$$\int_0^\ell \phi_j(x)\phi_k(x) = \delta_{jk} \tag{32}$$

By substituting (30) into (25), (17), and multiplying the results by  $\phi_j$ , one obtains N equations (j = 1..N):

$$\bar{m}\ddot{q}_{j}(t) + \omega_{j}^{2}\bar{m}q_{j}(t) + \eta_{j}\dot{q}_{j} - \sum_{i=1}^{n_{p}} g_{ij}V_{i} = F_{j}$$
(33)

139 where  $F_j = \int_0^\ell \phi_j p(x,t) \ dx$ .

The value of the coupling coefficient  $g_{ij}$  of the patch *i* with respect to the mode *j* is defined by:

$$g_{ij}(a_i, b_i) = \mathcal{G} \int_{a_i}^{b_i} \phi_j^{\prime\prime}(x) dx = \mathcal{G} \left( \phi_j^{\prime}(b_i) - \phi_j^{\prime}(a_i) \right)$$
(34)

<sup>141</sup> Similarly, substituting (30) into (24) leads to  $n_p$  equations  $(i = 1..n_p)$ :

$$C_i V_i + \sum_{j=1}^{N} g_{ij} q_j(t) = 0$$
(35)

<sup>142</sup> By seeking the harmonic solutions under the forms:

$$q_j(t) = \tilde{q}_j(\omega)e^{i\omega t}, \qquad V_i(t) = \tilde{V}_i(\omega)e^{i\omega t}, \qquad F_j(t) = \tilde{F}_j(\omega)e^{i\omega t}, \tag{36}$$

the final system of modal equations (Eqs. 37, 38) reads:

$$-\omega^2 \bar{m}\tilde{q}_j + \omega_j^2 \bar{m}\tilde{q}_j + \mathrm{i}\omega\eta_j\tilde{q}_j - \sum_{i=1}^{n_p} g_{ij}\tilde{V}_i = \tilde{F}_j, \quad (j = 1..N)$$
(37)

$$C_i \tilde{V}_i + \sum_{j=1}^N g_{ij} \tilde{q}_j = 0, \quad (i = 1..n_p)$$
 (38)

#### <sup>144</sup> 2.4. Optimization of the transducers positions

In this section, an optimization procedure is described for positioning the piezoelectric patch on a structure so that the measurement of specific modes is maximized. The aim is to find the utility function that represents the piezoelectric patch sensitivity as a function of the specific modes of interest. More specifically, we intend to find the optimal placement  $a_i$  for the patch such as the coupling  $g_{ij}$  between the *i*-th patch and the *j*-th mode is maximized.

#### 150 2.4.1. Modal sensitivity

We compute  $\tilde{q}_j$  from Eq. (37), then inject the result into Eq. (38). By neglecting the quadratic terms with respect to the coupling term  $g_{ij}$ , we obtain the so called sensor equation for each patch *i*:

$$\tilde{V}_i = -\sum_{j=1}^N \frac{g_{ij}\tilde{F}_j}{C_i\bar{m}(\omega_j^2 - \omega^2 + i\omega\eta_j/\bar{m})}$$
(39)

In our case the external force is a point force applied at a specific point of the beam  $x_f$ , with amplitude  $\tilde{F}_0(\omega)$ , i.e.  $\tilde{p}(x,\omega) = F_0(\omega)\delta(x-x_f)$ . This means that:

$$\tilde{F}_j = \int_0^\ell \tilde{p}(x,\omega)\phi_j(x)dx = \int_0^\ell \tilde{F}_0(\omega)\delta(x-x_f)\phi_j(x)dx = \tilde{F}_0\phi_j(x_f)$$
(40)

<sup>155</sup> then the voltage measured on each piezo is:

$$\tilde{V}_i = -\sum_{j=1}^N g_{ij} \frac{\tilde{F}_0(\omega)\phi_j(x_f)}{C_i \bar{m}(\omega_j^2 - \omega^2 + i\omega\eta_j/\bar{m})}$$
(41)

<sup>156</sup> Finally, we can define the transfer function at the i-th patch as:

$$H_i(\omega) = \frac{\tilde{V}_i(\omega)}{\tilde{F}_0(\omega)} = -\sum_{j=1}^N \frac{g_{ij}\phi_j(x_f)}{C_i\bar{m}(\omega_j^2 - \omega^2 + i\omega\eta_j/\bar{m})}$$
(42)

<sup>157</sup> This transfer function is at the core of the optimization process.

The contribution of the j-th mode to the frequency response function Eq. (42) is:

$$G_{ij}(\omega) = \left| \frac{g_{ij}\phi_j(x_f)}{C_i \bar{m}(\omega_j^2 - \omega^2 + i\omega\eta_j/\bar{m})} \right|$$

In order to estimate the sensitivity of each piezo patch *i* to mode *j*, with respect to the position  $a_i$ , we can introduce the following norm  $\|.\|_{\infty}$ , that in our case corresponds to evaluating the contribution of the *j*-th mode at resonance frequency  $\omega_j$ :

$$\kappa_{ij} = \left\| G_{ij}(\omega_j) \right\|_{\infty} = \left| \frac{g_{ij}\phi_j(x_f)}{C_i i \omega_j \eta_j} \right|$$
(43)

<sup>162</sup> For each given patch i, a normalised form of modal sensitivity to mode j is given by:

$$\mathcal{M}_{ij} = \frac{\kappa_{ij}}{\max_{j=1..N} \kappa_{ij}} \tag{44}$$

where  $\mathcal{M}_{ij}$  represents the control authority of the *i*-th patch fixed over the *j*-th mode. Since the aim is to maximize the sensitivity of the piezoelectric sensor in regard to the detection of one or several mode(s) rather than using the patch as an actuator to control the mode(s), in this paper, it is named *modal sensitivity* in comparison with notion of *modal controllability* used in Halim and Moheimani [14].

#### 167 2.4.2. Definition of a new utility function

Different utility functions can be defined starting from the concept of modal sensitivity. The idea is to obtain modal filtering, which is achieved by adding a weight in the optimization for specific modes. In this way it is possible to "see" or "hide" a mode [21]. To this end we define a vector  $\boldsymbol{\xi}$  of modal weights, whose the component  $\xi_j$  corresponds to weight of the *j*-th mode:

$$\xi_{j} = \begin{cases} 1 \text{ if the } j\text{-th mode has to be seen} \\ -1 \text{ if the } j\text{-th mode has to be hidden} \\ 0 \text{ if the } j\text{-th mode is not relevant} \end{cases}$$
(45)

Then, for each value of  $a_i, b_i$  of the transducer and assuming that  $x_f$  is fixed, the utility function is defined as:

$$\mathcal{F}(a_i, b_i, x_f) = \sum_{j=1}^N \xi_j \mathcal{M}_{ij}(a_i, b_i, x_f)$$
(46)

and the optimal positions  $\left\{\hat{a}_i, \hat{b}_i\right\}$  is defined by:

$$\left\{\hat{a}_{i}, \hat{b}_{i}\right\} = \arg\max(\mathcal{F}(a_{i}, b_{i}, x_{f}))$$
(47)

Several transducers can be optimized at the same time, by choosing local maxima of the transfer function.
In this case, constraints on transducers sizes have to be considered to avoid the overlap of two patches. The
practical application of the method will be detailed in the case study presented in Section 3.

#### 178 3. Case study : optimal transducers positioning for implant stability monitoring

The optimization method described in Section 2.4 is applied to a square beam named ancillary hereafter, and connected to an inaccessible mechanical system corresponding to an orthopedic implant into a bone mimicking phantom (Fig. 2). The ancillary is a part fixed to the implant during the surgery in order to enable the surgeon to apply hammer impact at one of its extremity, leading to implant insertion. The aim of this case study is to present and validate a procedure of piezoelectric patches placement optimization for monitoring the stability of an implant which is not directly accessible for vibration measurement. The optimization principle lies on focusing on specific modes of vibration of the ancillary, which are sensitive to implant stability into the bone, and to hide modes which are not of biomechanical interest.

#### 187 3.1. Geometry

A stainless steel beam of square section was designed with the geometrical and material properties given in Table C.4 (see Appendix C.1). The bottom threaded extremity was rigidly fixed in the implant and the top extremity was kept free. The implant was a cementless femoral stem made of titanium alloy (TiAl6Al4V) and coated with hydroxyapatite (CERAFIT RMIS, Ceraver, Roissy, France). The implant was fully inserted into a bone mimicking phantom (ORTHObones, 3B Scientific, Hamburg, Germany) embedded in polymer resin in order to enable clamping to the vibration table, similarly as what was done in [20, 32].



Figure 2: Experimental case study configuration. The same notations as the ones shown in Figure 1 are indicated : accessible instrumented length  $\ell$ , axis directions and coordinates origin O. The ancillary (a) corresponds to the accessible part, the implant (femoral stem) (b) and the bone (femur) (c) represent the inaccessible part.

#### 194 3.2. Optimization algorithm

<sup>195</sup> The experimental optimization approach consists in the following successive steps described in Fig. 3:

1. Modal characterization of the ancillary without piezoelectric sensors. The modes of vibration

<sup>197</sup> of the ancillary are measured in the frequency range [0.4 12.8] kHz by Experimental Modal Analysis

(EMA). The experimental mode shapes  $\phi_j^e$ , the resonance frequencies  $f_j = 2\pi\omega_j$  and the damping coefficients  $\beta_j = \eta_j/\bar{m}$  are retrieved.

2. Piezoelectric patches placement optimization. Different utility functions are defined from Eq. (46) based on the modal sensitivity  $\mathcal{M}_{ij}$  calculated for each position  $a_i$  of piezo i and mode j. Considering the information carried by each mode j on the implant stability, a vector  $\boldsymbol{\xi}$  is defined. For each utility function  $\mathcal{F}$ , the optimal piezoelectric patch position  $\{\hat{a}_i, \hat{b}_i\}$  is determined by Eq. (47). One patch is associated with one utility function and  $n_p$  patches are used.

3. Piezoelectric patches placement validation: The amplitude ratios at  $\omega_j$  are calculated between the frequency response functions  $H_i$ , measured successively with each piezoelectric patch *i*. The results are compared with what was expected from the definition of each utility function  $\mathcal{F}$ .



Figure 3: Experimental procedure for piezoelectric patches placement optimization and protocol validation.

#### <sup>208</sup> 3.3. Modal characterization of the ancillary

Modal analysis measurements. The ancillary modal features were determined by performing EMA following 209 the same protocol described in [20] and shown in Fig. 4. Only the mode shapes oscillating along  $e_z$ , were 210 measured, since the properties of the mode shapes along  $e_{u}$  were shown to not evolve with implant insertion 211 into the bone in a previous study [20]. The structure was excited by hammer impacts (modal hammer 212 8204, Brüel & Kjaer, Naerum, Denmark) at 13 locations  $x_f$  equally spaced of 10 mm beginning at 5 mm 213 from the ancillary top surface. At each location, the structure was excited 5 times along  $e_z$  (see Fig. 4). 214 The z-accelerations consecutive to each impact were recorded by a tri-axial accelerometer (365A01PCB 215 Piezotronics, Depew, NY, USA) glued at the ancillary top extremity, as shown in Fig. 4. The excitation 216 impact was lower than 50N to ensure it does not modify the implant position and consequently the boundary 217

<sup>218</sup> conditions of the beam during the modal features measurement. A dedicated data acquisition module (BK

<sup>219</sup> Connect, Brüel & Kjaer, Naerum, Denmark) with a sampling rate of 51.2 kHz was used to record the

acceleration signals during 0.25 s. We assume that the weight of the accelerometer (1 g) and the weight and

rigidity of the piezoelectric patches do not influence the structure dynamic behavior.



Figure 4: a) Picture and b) scheme of the Experimental Modal Analysis set-up with  $\ell = 125$  mm.

Modes selection. An example of two frequency response functions (frf) measured in the range [0.4 12.8] kHz for an impact at  $x_f = 5$  mm along  $e_z$ , is presented in Fig. 5. The two frf correspond to two different implant insertion stages, one at beginning of insertion (in grey) and one for the implant fully seated (in black). The choice to show frf corresponding to an impact location at  $x_f = 5$  mm is explained because it is a good compromise to see all the modes on the frequency range [0.4 12.8] kHz, due to the proximity to the free extremity, where the mode shapes are close.

Five modes were selected from the *frf* of implant fully inserted, for the patch placement optimization analysis, denoted A, B, C, D and E hereafter. Table 1 gathers the modal features of the five selected modes obtained for the implant fully seated : resonance frequency f and damping factor  $\beta$ , calculated for each resonance frequency with the 3 dB method. The resonance frequencies of modes A and B are in good agreement with what was found in [20]. These two resonance frequencies increase during implant insertion into the bone and reach values of 2.6 kHz and 3.3 kHz once the implant fully seated, which corresponds to the behavior of modes 2Y and  $2Y_b$ , identified in [20].

Mode shapes. The experimental modal shapes  $\phi_j^e$  of the five selected modes presented in Table 1 and in Fig. 5 are shown in Fig. 6. The five modes are of bending nature, with either two nodes (modes A, B and C) are



Figure 5: Frequency response functions, *frf*, measured by the accelerometer for an impact at  $x_f = 5$  mm along  $e_z$  at beginning (in grey) and end (in black) of implant insertion. The letters from A to E indicate the modes selected for the patch placement optimization analysis.

$\mathbf{Mode}\ \#$	A	B	C	D	E
f (Hz)	2609	3391	4160	6157	10956
$\beta$ (-)	2.19	1.20	1.74	1.86	0.64

Table 1: Names and modal features (resonance frequency f, damping factor  $\beta$ ) of the modes selected for the piezoelectric patches placement optimization procedure.

three nodes for higher resonance frequencies (modes D and E). This result confirms that mode A and B correspond to modes 2Y and  $2Y_b$  in [20]. Each mode shape  $\phi_j^e$  was normalized by its Euclidean norm  $||.||_2$ . Each discrete experimental mode shape  $\phi_j^e$ , was fitted using the general equation of normal beam shapes adapted from Eq. (31):

$$\phi_j^{(4)} = \frac{\omega_j^2 \bar{m}}{\mathcal{D}} \phi_j \tag{48}$$

where  $\lambda_j^{(4)} = \omega_j^2 \bar{m} / \mathcal{D}$ . Hence, the general solution of Eq. (48) is given by:

$$\phi_j = C_1 \sin(\lambda_j x) + C_2 \cos(\lambda_j x) + C_3 \sinh(\lambda_j x) + C_4 \cosh(\lambda_j x)$$
(49)

where the constant  $C_{1,2,3,4}$  are determined from the boundary conditions expressed by Eq. (26) to (29). In our case study, considering a free end at x = 0, the bending moment and shear force were both equal to zero at this end (Eq. (26) and Eq. (27)) which gives  $C_1 = C_3$  and  $C_2 = C_4$ . Then, for each mode j, the values of  $C_1$  and  $C_2$  were determined using an error minimization algorithm in Matlab. The mode shapes were fitted in order to further calculate the coupling coefficient  $g_{ij}$  from the mode shape curvature  $\phi_j''$  (Eq. 34), which requires a continuous shape.

#### 242 3.4. Piezoelectric patches placement optimization

This section is dedicated to the definition of one or more utility function(s)  $\mathcal{F}$  aiming at finding patches positions  $\hat{a}_i$   $(i = 1..n_p)$  on the ancillary such as a combination of modes  $\phi_j$  (j = 1..N) are visible or hidden



Figure 6: Experimental (o) and fitted (-) mode shapes  $\phi_j$  of modes A to E, vibrating along  $e_z$ .

when measuring the frequency response function  $H_i$  by the piezoelectric patch *i*. The choice to see or hide one or several mode(s) is based on the mode resonance frequency sensitivity to the implant position into the bone, corresponding to the so-called "inaccessible part". The modes  $\phi_j$  considered for the optimization analysis are modes #*A*, *B*, *C*, *D* and *E*.

Piezoelectric patch properties. The piezoelectric transducers that are used for resonance frequency functions measurement are square shaped with a length b = h = 10 mm and a thickness  $h_p = 0.2$  mm. The geometry is fixed and constant over the patches. The detailed properties or given in Table C.5 in the Appendix C.2. The same configuration as presented in Section 2.1 and Fig. 1 with two piezoelectric patches put symmetrically at the upper and lower surfaces of the beam is considered hereafter.

Modal sensitivity. As described in Section 2.4, the normalized modal sensitivity  $\mathcal{M}_{ij}$  defined by Eq. (44) is calculated for each mode j and patch i by varying the position  $a_i$  of the patch. As the length  $\ell$  of the instrumented beam is 125 mm and the patches are 10 mm length,  $a_i \in [0-115]$  mm, in order to avoid the patch to exceed the accessible part of the structure. The spatial resolution for varying the position  $a_i$  of the patch and calculating  $\mathcal{M}_{ij}$  is 1 mm.

For each mode j, the results show at least one minimum and maximum on the modal sensitivity function  $\mathcal{M}_{ij}$  (see Fig. 7) depending on the patch position  $a_i$ . The higher frequency modes D and E (6157 Hz and 10956 Hz, respectively) have two local maxima which offer two possibilities for patch placement whereas lower frequency modes A, B and C (2609 Hz, 3391 Hz and 4160 Hz, respectively) only hold one maximum.



Figure 7: Modal sensitivities  $\mathcal{M}_{ij}$  depending on the patch position  $a_i$  on the ancillary calculated for each mode A to E.

Choice of transducers positions. The strategy employed for the definition of a utility function  $\mathcal{F}$  is to use 263 one transducer associated to multiple modes, as explained in the section 2.4.2. Different utility functions 264 can be defined, each function being associated with an optimal patch position  $\{\hat{a}_i, \hat{b}_i\}$ , defined by Eq. (47). 265 The comparison of the *frf* measured at the beginning of implant insertion and once the implant is fully 266 seated (see Fig. 5) shows that 4 modes (namely, modes #A, B, C and E) over the 5 selected for the study, 267 are sensitive to implant stability with a frequency shift of more than 150 Hz to higher frequencies between 268 the beginning and end of implant insertion. The sensitivity of the modes A and B is in good agreement 269 with what was obtained in [20]. 270

From these results and in order to demonstrate the performance of the optimization method, three examples of utility functions  $\mathcal{F}$  are defined with Eq. (46), for which the modal weights  $\xi_j$  associated to each mode j are shown in Table 2. Considering these examples, each utility function  $\mathcal{F}$  is associated with one position  $\hat{a}_i$  of the patch i such as:

•  $\mathcal{F}_1$ : The resulting position  $\hat{a}_1$  of the patch assumes to be the best compromise to see all the modes A to E on the frequency response function ( $\xi_j=1$ ). This configuration acts as a *reference case*, which will allow the comparison with other configurations where modes should be filtered ( $\mathcal{F}_2$ ,  $\mathcal{F}_3$ ), in order to evaluate the efficiency of the optimization method.

•  $\mathcal{F}_2$ : The resulting position  $\hat{a}_2$  of the patch assumes to be the best compromise to hide mode  $D(\xi_4=-1)$ while letting the other modes visible  $(\xi_{j\neq 4}=1)$ . This configuration enables to focus only on the modes 281 sensitive to implant insertion (namely, modes A, B, C and E), since the resonance frequency of mode 282 D has shown to not significantly change during implant insertion (see Fig. 5).

•  $\mathcal{F}_3$ : The resulting position  $\hat{a}_3$  of the patch assumes to be the best compromise to filter mode  $C(\xi_3=-1)$ while ensuring visibility of mode  $B(\xi_2=1)$ . The other modes are not concerned by the optimization configuration  $(\xi_{1,4,5}=0)$ . This configuration enables to focus on the tracking of the resonance frequency of mode B for implant insertion monitoring, by avoiding confusing with mode C, due to their close resonance frequencies.

Modes	A	B	C	D	E	${\hat{a}_i, \hat{b}_i}$ (mm)
$\mathcal{F}_1$	1	1	1	1	1	$\{40,  50\}$
$\mathcal{F}_2$	1	1	1	-1	1	$\{78, 88\}$
$\mathcal{F}_3$	1	1	-1	1	1	$\{91, 101\}$

Table 2: Modal weights  $\xi_j$  defined for the utility functions  $\mathcal{F}_{1,2,3}$  and resulting optimal piezoelectric positions  $\hat{a}_1$ ,  $\hat{a}_2$ ,  $\hat{a}_3$ . The reference case corresponding to  $\mathcal{F}_1$  is highlighted in bold.

Figure 8b shows the three utility functions  $\mathcal{F}_1$ ,  $\mathcal{F}_2$ ,  $\mathcal{F}_3$  calculated for the weight coefficients  $\xi_j$  defined in Table 2. The modal sensitivities  $\mathcal{M}_{ij}$  are represented in Fig. 8a for j = 1..5. For each utility function, only one maximum is identified, corresponding to the position  $\hat{a}_i$  of the piezoelectric patch defined by Eq. (47). Note that to avoid overlap, for each patch i, we verified that  $|\hat{a}_i - \hat{a}_{i+1}| > b$ , in order to allow the bounding of the 3 transducers on the same beam.

#### <sup>293</sup> 3.5. Validation of optimization placement protocol

The structure has been assembled following the optimization procedure described in the previous section. The piezoelectric patches have been bonded on the surface of the beam using a one-component silver-filled epoxy resin (Elecolit 3043, *Panacol, Germany*) cured at 160°C during 30 min. The resulting smart ancillary is shown in Fig. 9.

To assess the validity of the patches placement optimization procedure, a frequency analysis was per-298 formed in order to compare the frequency response functions  $H_i$  measured by each patch i. To do so, the 299 ancillary was excited by a modal hammer, impacting the structure at  $x_f = 5 \text{ mm}$  along  $e_z$ . The choice 300 of the position of the source  $x_f$  is explained by the similarity of the mode shapes near the free extremity 301 (see Fig. 6), which provides optimization results independent of  $x_f$ . However, as shown in Eq. (42), it 302 would have been possible to study and optimize the position of the excitation source  $x_f$  in the calculation 303 of the utility functions  $\mathcal{F}$ . The system dynamic response was measured successively by one of the three 304 piezoelectric patches, the two others being short-cut during the acquisition time. The sampling frequency 305



Figure 8: a) Modal sensitivities  $\mathcal{M}_{ij}$  as a function of the patch position  $a_i$  calculated for j = 1..5 (namely, modes A to E) and b) corresponding utility functions  $\mathcal{F}_{1,2,3}$  calculated for  $\boldsymbol{\xi} = [1 \ 1 \ 1 \ 1 \ 1]$ ,  $\boldsymbol{\xi} = [1 \ 1 \ 1 \ -1 \ 1]$  and  $\boldsymbol{\xi} = [0 \ 1 \ -1 \ 0 \ 0]$ , respectively, with the optimal positions  $\hat{a}_i$  corresponding to the maximum of each utility function  $\mathcal{F}$ .



Figure 9: Image of the instrumented ancillary equipped with piezoelectric patches #1, #2, #3 bonded at locations  $\hat{a}_1 = 40$  mm,  $\hat{a}_2 = 78$  mm and  $\hat{a}_3 = 91$  mm, calculated from the optimization procedure described in Section 3.4

and the acquisition time were the same as for the modal characterization, leading to a frequency resolution

$$_{307}$$
  $df = 4$  Hz.

The frequency response function  $H_1$  obtained by the piezoelectric patch #1, placed such as the utility function allow the visibility of all the set of modes A to E, is compared with  $H_2$  measured by the piezoelectric patch #2 in Fig. 10a and with  $H_2$  measured by the piezoelectric patch #3 in Fig. 10b. The amplitudes at resonances measured by the three piezoelectric patches are given in Table 3 together with the amplitude's variation at resonance obtained for patch #2 and patch #3 with respect to the reference patch #1.



Figure 10: a) Comparison of the frequency response functions  $H_1$  and  $H_2$  measured by the piezoelectric patch #1 and the piezoelectric patch #2 and b) comparison of  $H_1$  and  $H_3$  measured by the piezoelectric patch #1 and the piezoelectric patch #3. A zoom on the target modes by the utility functions  $\mathcal{F}_2$  and  $\mathcal{F}_3$  is included in the graphs.

Modes	A	B	C	D	E
patch $\#1 (V/N)$	0.898	0.467	0.231	0.427	0.330
patch $\#2 (V/N)$	0.851 (-5.24%)	0.337 (-27.8%)	0.136~(-41.3%)	0.060 (-85.9%)	0.386 (+17.0%)
patch #3 $(V/N)$	0.600~(-33.2%)	0.152 (-67.4%)	0.054~(-76.8%)	0.234~(-45.2%)	0.156~(-52.7%)

Table 3: Amplitudes at resonances of modes A to E measured by patches #1, #2, #3 together with the percentage of amplitude's decrease at resonance indicated in brackets, with regard to measurements from patch #1. The patch #1 corresponds to the reference case and is highlighted in bold.

The amplitude decrease of 85.9% of mode D measured by the piezoelectric patch #2 with regard to 313 patch #1, validates the position optimization procedure since the utility function  $\mathcal{F}_2$  aimed to hide mode 314 D. Similarly, the frequency measurement with the patch #3 leads to a decrease of 76.9% of the amplitude 315 of the mode C with regard to patch #1, which was expected by the utility function  $\mathcal{F}_3$ . Nevertheless, 316 the unintended associated decrease of the amplitude of mode B show the limitation of the utility function, 317 defined by weight coefficient  $\xi_j$  and patch position  $a_i$ . This unintended decrease is explained by the strong 318 similarity between the shapes of mode B and C (see Fig. 6). Therefore, another condition based on a 319 minimum visibility of the modes associated with  $\xi_i=1$  should be implemented in the utility function to 320 ensure mode visibility. It could also be relevant to study the effect of the geometrical parameters of the 321 piezoelectric patch as the length b or the thickness  $h_p$ , both implied in the calculation of the coupling 322 coefficient  $g_{ij}$  and therefore  $\mathcal{M}_{ij}$ . 323

#### 324 4. Conclusion and discussions

This paper proposes a method for optimizing the positioning of piezoelectric patches on a beam attached 325 to an unreachable structure. The issue of maximizing the observability on the frequency response function 326 of a set of modes, which is sensitive to the beam boundary conditions, was addressed. For that purpose, 327 an optimization strategy was developed, based on the definition of utility functions, expressed in terms of 328 modal weight and in terms of the so-called modal observability, for which the expression has been exhibited. 329 It has been shown that the modal observability depends on the electro-mechanical coupling coefficient, 330 calculated from the patches geometrical and material properties, and the beam mode shapes curvatures. 331 The originality of the paper comes from the use of the modal electro-mechanical coupling factor (MEMCF) 332 to optimize sensors placement on the beam in order to improve beam boundary condition monitoring. An 333 original application for implant stability measurement was proposed. 334

In the particular case developed in this paper, one piezoelectric patch is associated to several modes, 335 whose the observability has to be improved. Therefore, the optimal piezoelectric patches positions are the 336 result of a compromise obtained from the utility functions, aiming at improving the observability of a set 337 of modes sensitive to beam boundary conditions, while minimizing the observability of the modes which are 338 not sensitive to them. In the experimental case study, the particular issue of monitoring implant stability 339 by means of a smart beam was investigated. The smart beam, which is basically a tool used by the surgeon 340 to insert the implant, was equipped with patches for which the placement was determined to optimize the 341 mode observability on the FRF, depending on the sensitivity to the stability, that was demonstrated in 342 a previous study [20]. The mode observability measured by the position-optimized patches was in good 343 agreement with what was expected from the utility functions definition. The amplitude of the modes that 344 should be hidden on the FRF because they are not sensitive to implant stability decreased up to 85% with 345 regard to the reference position. The results of this study open the path towards the conception of a smart 346 tool optimized for the real-time monitoring of implant stability during the insertion of an implant by the 347 surgeon. 348

However, since the optimization strategy is based on the modal observability and in particular the coupling coefficient, the positions strongly depend on the mode shapes curvatures. For low frequency modes, the placement possibilities are limited because of the few number of nodes, and therefore close mode shapes will tend for close piezoelectric patches positions. Moreover, another drawback of the study is that optimization of the dimensions of the piezoelectric patches was not studied, whereas the electro-mechanical coefficient also depends on the thickness or the length of the patch. This aspect was studied in [11] and could be easily implemented in the strategy proposed in this paper. Another way of optimization, would be
to use specific forms of piezoelectric patches, enabling accurate boundary force measurement, as proposed in
[33]. Nevertheless, this method is more complex to implement regarding both the piezoelectric patch design
and the post-processing analysis.

The method proposed in this paper is particularly adapted for applications where mode shapes change 359 with boundary conditions, as it relies on mode shape curvatures. It could be applied for other configurations 360 of beam-like structures with bending modes that should be tuned, based on the modal observability concept. 361 Also, a relevant perspective will be to investigate the placement optimization of the excitation source, and 362 in particular, by using one of the piezoelectric patch as an actuator. This outlook is of particular interest 363 concerning the biomedical application presented in the case study, since it would lead to an automation 364 of the measurement by preventing from the use of a modal hammer as a source of excitation that is not 365 convenient in a clinical environment. Then, as investigated by several authors [23, 24, 6], another strategy 366 approach could be to use of a network of sensors along the beam with different control signals to tune specific 367 modes. The main advantage of this approach would be to conceive one unique beam for several applications. 368

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#### <sup>374</sup> Appendix A. Constitutive equations of piezoelectric materials

*Constitutive equations.* Assuming a linear piezoelectric material, the 3D constitutive equations may be written in matrix form as follows [34]:

$$\begin{pmatrix} \boldsymbol{\sigma} \\ \boldsymbol{D} \end{pmatrix} = \begin{bmatrix} \mathbf{c} & -\mathbf{e}^T \\ \mathbf{e} & \boldsymbol{\varepsilon} \end{bmatrix} \begin{pmatrix} \boldsymbol{\epsilon} \\ \boldsymbol{E} \end{pmatrix}$$
(A.1)

where  $\boldsymbol{\sigma} = (\sigma_{xx}, \sigma_{yy}, \sigma_{zz}, \sigma_{yz}, \sigma_{xz}, \sigma_{xy})^T$  is the stress vector;  $\boldsymbol{D} = (D_x, D_y, D_z)^T$  is the electric displacement vector;  $\boldsymbol{\epsilon} = (\epsilon_{xx}, \epsilon_{yy}, \epsilon_{zz}, 2\epsilon_{yz}, 2\epsilon_{xz}, 2\epsilon_{xy})^T$  is the strain vector; and  $\boldsymbol{E} = (E_x, E_y, E_z)^T$  is the electric field vector. The matrices  $\mathbf{c}, \mathbf{e}, \boldsymbol{\epsilon}$  are the matrices with elastic, electro-mechanic and dielectric constant entries.

$$\mathbf{c} = \begin{bmatrix} c_{11} & c_{12} & c_{13} & 0 & 0 & 0\\ c_{21} & c_{22} & c_{23} & 0 & 0 & 0\\ c_{31} & c_{32} & c_{33} & 0 & 0 & 0\\ 0 & 0 & 0 & c_{44} & 0 & 0\\ 0 & 0 & 0 & 0 & c_{55} & 0\\ 0 & 0 & 0 & 0 & 0 & c_{66} \end{bmatrix}, \mathbf{e} = \begin{bmatrix} 0 & 0 & 0 & 0 & e_{15} & 0\\ 0 & 0 & 0 & e_{24} & 0 & 0\\ e_{31} & e_{32} & e_{33} & 0 & 0 & 0 \end{bmatrix}, \boldsymbol{\varepsilon} = \begin{bmatrix} \varepsilon_{11} & 0 & 0\\ 0 & \varepsilon_{22} & 0\\ 0 & 0 & \varepsilon_{33} \end{bmatrix}, \quad (A.2)$$

Alternatively, the equation (A.1) may be written by:

$$\begin{pmatrix} \epsilon \\ D \end{pmatrix} = \begin{bmatrix} \mathbf{s} & \mathbf{d} \\ \mathbf{d}^T & \boldsymbol{\varepsilon} \end{bmatrix} \begin{pmatrix} \boldsymbol{\sigma} \\ E \end{pmatrix}$$
(A.3)

where  $\mathbf{s} := \mathbf{c}^{-1}$  is the compliance matrix and  $\mathbf{d} := \mathbf{s}\mathbf{e}^T$  contains the piezoelectric constants. Note that  $Y_b = 1/s_{11}$ .

#### 377 Appendix B. Extended Hamilton principle

*Energy density function.* The total stored energy density in a unit volume of material is the sum of the mechanical work and of the electrical work [31]:

$$dW_e(\boldsymbol{\epsilon}, \boldsymbol{D}) = (d\boldsymbol{\epsilon})^T \boldsymbol{\sigma} + (d\boldsymbol{D})^T \boldsymbol{E}$$
(B.1)

Note that for a conservative system,  $W_e = \int_{(0,0)}^{(\epsilon,D)} dW_e$  which leads to the constitutive equation:

$$\boldsymbol{\sigma} = \frac{\partial W_e}{\partial \boldsymbol{\epsilon}}, \quad \text{and} \quad \boldsymbol{E} = \frac{\partial W_e}{\partial \boldsymbol{D}}$$
 (B.2)

The coenergy density function is defined by:

$$W_e^*(\boldsymbol{\epsilon}, \boldsymbol{E}) = \boldsymbol{E}^T \boldsymbol{D} - W_e(\boldsymbol{\epsilon}, \boldsymbol{D})$$
(B.3)

By calculating the total differential of  $dW_e^*$  and by using the (A.1), we can express  $W_e^*$  as:

$$W_e^*(\boldsymbol{\epsilon}, \boldsymbol{E}) = \frac{1}{2} \boldsymbol{E}^T \boldsymbol{\varepsilon} \boldsymbol{E} + \boldsymbol{\epsilon}^T \mathbf{e} \boldsymbol{E} - \frac{1}{2} \boldsymbol{\epsilon}^T \mathbf{c} \boldsymbol{\epsilon}$$
(B.4)

It follows that:

$$\boldsymbol{\epsilon} = \frac{\partial W_e^*}{\partial \sigma}, \quad \text{and} \quad \boldsymbol{S} = \frac{\partial W_e^*}{\partial \boldsymbol{E}}$$
 (B.5)

Hamilton's principle. We assume the admissibility requirements which impose that the virtual displacements  $\delta u$  must be compatible with the the kinematic of the system, and the admissible flux linkage variation  $\delta \lambda$  must be compatible with Kirchhoff's voltage rule. The Hamilton's principle reads [2]:

$$\int_{t_1}^{t_2} \left(\delta T^* + \delta W_e^* + \delta W_{ext}\right) = 0$$
(B.6)

where

$$T^* = \frac{1}{2} \int_{\Omega} \rho \dot{\boldsymbol{u}}^T \dot{\boldsymbol{u}} \, dV \tag{B.7}$$

$$W_e^* = \frac{1}{2} \int_{\Omega} \left( \boldsymbol{E}^T \boldsymbol{\varepsilon} \boldsymbol{E} + 2\boldsymbol{\epsilon}^T \mathbf{e} \boldsymbol{E} - \boldsymbol{\epsilon}^T \mathbf{c} \boldsymbol{\epsilon} \right) dV$$
(B.8)

and which follows that:

$$\delta T^* = \int_{\Omega} \rho \delta \dot{\boldsymbol{u}}^T \dot{\boldsymbol{u}} \, dV \tag{B.9}$$

$$\delta W_e^* = \int_{\Omega} \left( \delta \boldsymbol{E}^T \boldsymbol{D} - \delta \boldsymbol{\epsilon}^T \boldsymbol{\sigma} \right) dV \tag{B.10}$$

The term  $\delta W_{ext}$  represents the virtual work of the external forces and of the external applied currents:

$$\delta W_{ext} = \int_{\Omega} \delta \boldsymbol{u}^T \boldsymbol{f}_v dV + \int_{S_f} \delta \boldsymbol{u}^T \boldsymbol{f}_s dS - \int_{S_q} \delta \phi^T \bar{q} dS, \tag{B.11}$$

where  $f_v$  is the volume force,  $f_s$  is the surface forces applied on  $S_f$ ,  $\bar{q}$  is surface charge density applied on  $S_q$ .

#### <sup>380</sup> Appendix C. Properties of the ancillary and the piezoeletric patches

381 Appendix C.1. Ancillary

Properties of the ancillary				
Coefficient	Value	Description		
A	$10 \times 10 \text{ mm}^2$	Section of the ancillary		
h	$10\mathrm{mm}$	Thickness		
l	$125\mathrm{mm}$	Length of the ancillary (squared		
		part only)		
$Y_b$	$193\mathrm{GPa}$	Young's modulus		
ho	$8000  {\rm kg  m^{-3}}$	Density		
$\mu$	0.3	Poisson's ratio		

Table C.4: Dimensions and mechanical properties of the ancillary

### 382 Appendix C.2. Piezoelectric patches

Properties of the piezoelectric ceramic			
Coefficient	Value	Description	
$h_p$	$0.2\mathrm{mm}$	Thickness	
b	$10\mathrm{mm}$	Length	
$d_{31}$	$-180 \times 10^{-12} \mathrm{C/N}$	Coupling coefficient	
$Y_p$	$62.3\mathrm{GPa}$	In-plane Young's Modulus	
$ ho_p$	$7800{ m kg}{ m m}^{-3}$	Density	
$\mu_p$	0.35	Poisson's ratio	

Table C.5: Dimensions and mechanical properties of the piezoelectric ceramic element

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