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90 nm SRAM Static and Dynamic Mode Real-Time Testing at Concordia Station in Antarctica

G. Tsiligiannis, L. Dilillo, A. Bosio, P. Girard, S. Pravossoudovitch, A. Virazel, P. Cocquerez, J. L. Autran, A. Litterio, F. Wrobel, and F. Saigné

Abstract—In this work, we introduce the experimental setup and the first results of the Real-Time testing platform of High Altitude Memory Test (HAMLET) project installed at the Concordia station in Antarctica (acceleration factor of 9.8 compared to New York City-NYC). The platform is dedicated for the study of the response of 90 nm Commercial Off The Shelf (COTS) Static Random Access Memories (SRAMs) to atmospheric neutrons, operating in both the static and dynamic modes. Based on the up-to-date obtained results, the Soft Error Rate (SER) of the devices is extracted and compared with accelerated SER data. The results reveal the appearance of phenomena such as Single Event Latchups (SELs) during SRAM operation under real conditions, and not only as a result of accelerated overstressing tests. Additionally, results are in agreement with the Joint Electron Device Engineering Council (JEDEC89) standard acceleration factor and the anisotropy of the neutron flux is also studied.

Index Terms—90 nm, atmospheric neutrons, Concordia station, isotropy, real-time, SRAM.

I. INTRODUCTION

TERRESTRIAL radiation effects on electronic components have been studied extensively during the last two decades. The primary reason for this emerging concern lies on

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the reduction of the transistor size that makes electronic devices more sensitive to radiation. Since commercial applications have an increasing requirement for reliable electronic devices, research organizations and industry constantly explore mitigating mechanisms and try to understand the basic phenomena generating failures as a result of natural radiation. To address these targets, radiation testing of electronic devices is one of the most fundamental procedures for reliability assessment.

Radiation testing of electronic devices can be classified in three different categories that work in a complementary way between them: simulation based testing, accelerated testing under beam, and real-time testing (often referred as life testing). Simulation based testing is among the fastest and cheapest methods, since it requires low resources besides the proper models of the studied devices [1], [2]. The drawback of this method is the limitation of the study of the radiation impact to a small number of transistors and not the entire device, because the computing time grows quickly with the complexity of the simulations. Testing in particle accelerator facilities is the most direct and fast method for the investigation of the response of an electronic device in actual radiating environment [3], [4]. The main drawback of this method is the cost, but also the fact that, in some cases, beam irradiation may induce artificial overstress of devices. The elevated flux of particles to which electronic devices are subjected may trigger phenomena that in actual real-time conditions will not occur. Accelerated results have been reported to differ to some degree from real-time results in [5], while in other cases more consistent results are obtained as in [6]. In [7], a comparison between different families of devices is performed showing that accelerated results may be in accordance or not with real-time testing results, depending also on the technology. Finally, real-time testing is considered to be the most reliable method of testing since it requires the components to be tested in their natural environment, while being exposed for a sufficient time to collect statistically significant data. Several studies, in the past and ongoing, are based on the exposure to natural radiation of large arrays of devices in several locations and for long periods (several months). In these devices the soft errors are collected and analyzed. The drawback of real-time testing is that it requires the integration of large scale platforms that can embed numerous components for accelerating the test duration, while at the same time they need to be exposed for a long period in order to collect sufficient data. In [8] a few hundreds of 40 nm SRAMS were tested at the high altitude desert Plateau de Bure (2552 m), in order to record their Soft Error Rate (SER) when exposed to atmospheric neutrons. For the same purpose,

two platforms using 90 nm and 130 nm SRAMs were placed at the Midi-Pyrénées Observatory (OMP, 2885 m) and the city of Puno in Peru (3889 m), as detailed in [9].

Continuing the effort of real-time testing of electronic devices, in this paper, we present the results achieved by the SRAM-based platform designed and developed at the Laboratoire d'Informatique, de Robotique et de Microélectronique de Montpellier (LIRMM) with the support of Institut d'Électronique et Systèmes (IES) and the Centre National d'Études Spatiales (CNES) under the project High Altitude Memory Test (HAMLET). Under the same project other instruments have been installed at the same location such as silicon diodes and CCD sensors in order to cross validate the results occurring from the monitoring of the neutron flux, however, at the time of this publication the current experiments were the only ones able to provide with statistical data. The platform was placed at the Concordia scientific station in Antarctica [10], in 2013. In the present study, we introduce the design description of the platform, the retrieved results that allowed the estimation of the error rate of the sensing devices (SRAMs), but also the estimation of the atmospheric neutron flux and its characteristics (*e.g.* anisotropy) in this specific location. The results also prove that large-scale events (big clusters of bit-flips) may appear during real-time testing, and not only during accelerated testing. The interest behind this study is that Concordia station provides a unique environment in terms of altitude, longitude and latitude, and more specifically in terms of acceleration factor at ground level.

II. PLATFORM AND ENVIRONMENT SETUP

A. SRAM Based Neutron Detector Module

The test platform is composed of a large number of fundamental modules, in which the sensing elements are 90 nm SRAMs. These modules are composed of four 32 Mbit COTS 90 nm Asynchronous SRAMs, which serve as Single Event collectors. The SRAMs are controlled by a Finite State Machine (FSM), which applies tests in both static and dynamic modes. The FSM is implemented into a FLASH-based FPGA, which is integrated into the same board, as depicted in Fig. 1. The choice of a FLASH based FPGA was made for its inherent robustness against radiation, so that to avoid SEUs occurring to other devices than the SRAM sensing components.

Within each testing module, the FSM applies tests in static mode in three of the four memories by writing a known sequence (solid '1's) and after a certain time threshold (set to twenty minutes), the entire content of the memories is read back. The choice of the solid '1' data background was made because, from a previous study made on the same SRAM device [11], there is a higher sensitivity of the $1 \rightarrow 0$ transition with respect to the $0 \rightarrow 1$ transition (+15%). During the read back, the contents of each word are compared with the expected sequence by the FSM and in the case of a detected upset, the erroneous word is written in a buffer, along with its address and the code of the memory. A master controller regularly downloads the experimental data from the buffer of each detecting module through a serial bus, which utilizes an I²C communication protocol. Between the read-back windows, the FSM applies a dynamic test

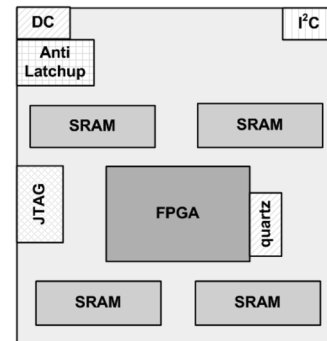


Fig. 1. SRAM based detection module composed of 4 32 Mbit COTS 90 nm Asynchronous bulk SRAM devices. The memories are controlled and tested by an FSM, which is implemented through a FLASH-based FPGA. Additional anti-latchup circuitry is added for the protection of the board. A JTAG interface allows the reprogramming of the FPGA.

$$\begin{aligned} & \uparrow (r1, w0, r0, r0, r0, r0, r0); \\ & \uparrow (r0, w1, r1, r1, r1, r1, r1); \\ & \uparrow (r1, w0, r0, r0, r0, r0, r0); \\ & \downarrow (r0, w1, r1, r1, r1, r1, r1); \\ & \downarrow (r1, w0, r0, r0, r0, r0, r0); \\ & \uparrow (r0, w1, r1, r1, r1, r1, r1) \end{aligned}$$

Fig. 2. Dynamic Stress March test. The test is composed of 6 elements (enclosed in parenthesis), and each element is composed of 7 operations. Each element (thus all the operations composing it) is applied to each word of the memory, and it advances in the memory address span in either an ascending or descending order indicated by the arrow in front of each element.

in the fourth SRAM of the detecting module. The applied test is the “Dynamic Stress” March test [11], in which a sequence of read/write operations is repetitively executed throughout the entire address span of the memory. March tests are algorithms that are utilized during the manufacturing process to reveal functional faults of memories such as coupling and stuck-at faults among others. The scheme describing the Dynamic Stress test is given in Fig. 2.

Each March test is composed of several elements, and each element is composed of several operations (in Fig. 2 the operations of each element are enclosed in parenthesis). The March algorithms (Dynamic Stress in our case) are applied during irradiation when the SRAM is exposed to a particle beam or during exposure in the natural environment, repetitively. The Dynamic Stress test is known to induce a Read Equivalent Stress (RES) to the memory cells, by applying multiple read operations in the same word. During a read access, the voltage levels of the node holding logic '1' of the accessed cell are lowered as a result of the bit line discharge by opening the access transistor. The lowering of the voltage of the node storing logic '1' makes the cell more vulnerable to particle induced parasitic currents. At the same time, all the cells belonging to the same word line with the accessed ones will undergo the same stress (lowering of the logic '1' voltage level) since their access transistors will be activated as well [12]. Besides the electric stress induced to the cells during the read operation, according to [13] where memories of similar technology were studied, a low power scheme is applied to the SRAMs which keeps the entire memory in low voltage levels. During dynamic operations (read/write), instead

of powering the entire memory to nominal voltage, well-tap defined electric blocks are powered, inside which the word to be accessed belongs, for their execution. This scheme is applied in order to avoid the occurrence of latchups (thanks to the low voltage preventing the triggering of Single Event Latchup (SEL)), and additionally limit them to the area defined by the electric blocks. By applying several read operations at the same word, the electric block the word belongs to will be powered for a prolonged period, enabling the potential occurrence of micro-latchups in that region.

During the Dynamic Stress test, the read operations are responsible for the sensing of SEUs, and the upsets are immediately transmitted to the local buffer when detected. Within a module, the four memories share address and data buffers, and some of the control signal wires. Thus, the controller can access only one memory at the time. When the three memories that are tested in static mode are accessed for the read-back procedure, the fourth memory that runs the dynamic test is set to standby mode. With the time window between two read-back operations significantly larger (20 minutes) than the duration of the read back time itself (for the 3 memories in static mode, it is less than 1 second), we can state that all four memories share the same experimental time.

Previous experiments made in particle accelerators using dynamic mode testing on SRAMs have revealed higher cross section with respect to static mode testing and the appearance of several phenomena such as Single Event Functional Interrupts (SEFIs) or SELs [13] that were not observed with static mode. Thus, in addition to static test, the proposed test platform runs also dynamic tests with the joint purpose to maximize the observation of events and try to confirm SEL and SEFI occurrence in the natural environment.

Besides the SRAMs and the FPGA, the detecting modules integrate additional circuitry such as a quartz oscillator for the generation of the clock used by the FPGA (running at 50 MHz), a voltage regulator and an anti-latchup component for the protection of the overall module electronics. Finally, a JTAG interface allows the reprogramming of the FPGA device, in order to be able to apply modifications on the setup such as modifying the applied March test, modifying the exposure time window of the memories in static mode, or changing the data background of the memories in static mode.

B. Platform Setup

Each platform integrates 18 SRAM-based modules, making a total of 72 memories and 2304 Mbit capacity (approximately 2Gbit). A dedicated motherboard embeds the bus on which the 18 sensing modules are interconnected, as well as a microcontroller that plays the role of the master of the system. The interconnection bus is serial and uses the I²C protocol. The I²C bus was chosen for its free availability and for low power consumption purpose, since the communication protocol was not speed driven thanks to the low amount of data expected to circulate. The motherboard is also responsible for the power distribution to the detectors and the master microcontroller. The microcontroller is responsible for collecting and time-stamping the SEU information stored in the error buffers of each sensing module and transmitting them by email to our laboratory, in

France. For the communication between the microcontroller and the detecting modules, a Round-Robin protocol is implemented, during which the Master microcontroller sends in a cyclic manner a request for communication to each detecting module. Following, each detecting module responds to the master by transmitting the contents of its buffer in case it is not empty, or by just acknowledging its functioning. In the case after a request a timeout occurs (the detecting module does not respond to the communication query of the master) the master resets the nonresponding module. If the communication is restored, the Round-Robin continues its operation. In the case that communication is not restored, the device is deactivated by power switching-off and labeled as nonoperational. Between the executions of two Round-Robins, a certain time window elapses for the collection of the upsets. This time window is set to twenty minutes (same as the time window of the static test), in order to allow a significant amount of time to collect SEUs, but at the same time, to not overload the buffers in case of a major event such as micro-SEL or SEFI. Approximately, 2-3 emails are sent daily to our laboratory with the recorded upsets and the state of each detecting module (operational/not_operational). The model of the Round-Robin and the diagnostics procedures are part of the Finite State Machine implemented in the master and they have been modeled initially in UML language and tested under the SPIN verification tool [14], proving that no *livelocks* or *deadlocks* are generated during execution. Further details on the diagnostics procedures and the models that have been used can be found in the extended study presented in [15]. The final implementation of the master controller has been performed in C language as part of the microcontroller configuration. Due to connectivity problems, as a result of the Concordia station accessing internet through a satellite and being limited on power consumption, communications are not always guaranteed. However, the data, which are meant to be transferred, are not lost in case of email sending failure, and the microcontroller assures their proper transmission. Fig. 3 illustrates a functional scheme of the test platform.

In the Concordia station, we installed two identical test platforms, for a total amount of 4608 Mbit. In order to explore anisotropy of atmospheric neutrons in Concordia, one platform has been mounted horizontally (Machine X), while the other one vertically (Machine Y).

C. Concordia Station

The Concordia station is located at an altitude of 3233 m in Antarctica (75°06'S–123°21'E) in the department called Dome C of the Antarctic Plateau [10]. It is one of the coldest places on Earth, with temperatures ranging from -25°C down to -80°C . Due to its location, a very high flux of atmospheric neutrons is expected (according to JEDEC standard, an acceleration factor of 9.81 is expected with respect to the NYC measurements [16]). The room in which the instruments operated was under constant temperature of 10°C .

III. RESULTS AND ANALYSIS

During the analysis of the retrieved results, certain guidelines have been followed in order to make the results more reliable and avoid misinterpretations. One of the most important

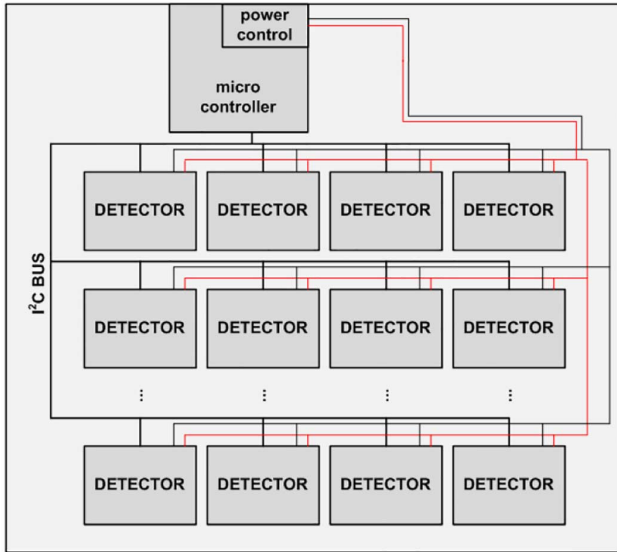


Fig. 3. SRAM based detector platform. In this platform, 18 detecting modules are mounted, and each one of them integrates four 32 Mbit 90 nm SRAMs. A microcontroller is responsible for the data collection of each detector via an I²C bus. A motherboard interconnects all the detectors with the microcontroller and distributes the power.

aspects of the study is the proper calculation of the occurring upsets, since they represent the major metric. For this reason, a neighboring scheme was adopted to classify Multiple Cell Upsets (MCUs), SEFIs and other types of large-scale events such as *Single Event Latchup*. In Table I, we provide both the event counts and the total number of corrupted bits as they were recorded. This scheme follows a ± 3 -bit neighboring policy, in which corrupted bit-cells that have a physical distance smaller or equal than three bits are considered the result of the same event. The timestamp has been taken into account as well for the clustering procedure, but it is rather unlikely to observe two events occurring in such a small topological distance (equal to three bits), since the event rate is approximately 1-2 events per day for 2304 Mbit. The software that was used for this purpose has been extensively successfully utilized and validated with data coming from accelerated tests. The data in Table I have been processed to obtain the graphs in Fig. 4 and Fig. 5.

Through the count of the failing bits and the study of their position in the physical bitmap of the memory, we were able to identify Single Bit Upsets (SBUs), MCUs, and other major events such as micro-SEs. Table I provides the information concerning the SER as calculated, considering the events count and the total count of corrupted bits (SER_{event} and SER_{raw} respectively), while in parenthesis information regarding the standard error of the Poisson distribution is provided.

In order to better understand the obtained results, it is essential to normalize the SER with a known reference position, such as the NYC. Considering the acceleration factor of 9.81 given by JEDEC standard, the SER of 14858FIT/Mbit ($\pm 7.5\%$) at Concordia for Machine Y will correspond to an SER of 1514FIT/Mbit at NYC and the SER of 13754FIT/Mbit ($\pm 7.9\%$) of Machine X corresponds to an SER of 1402FIT/Mbit at NYC. By comparing these results to recent results obtained with accelerated tests, we can have a more global figure of the SER response

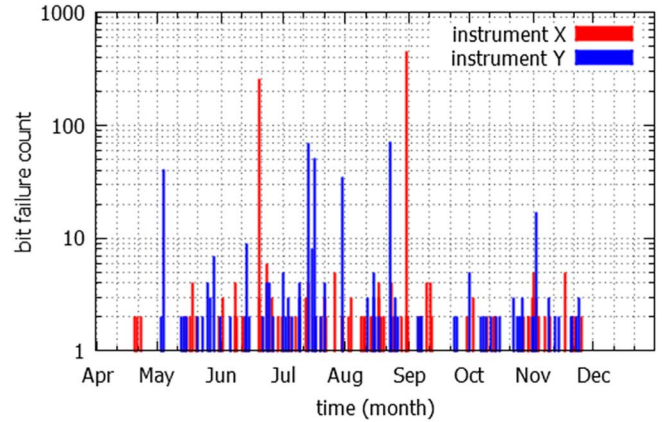


Fig. 4. Bit failure count plot for the two test platforms. Considering the average number of received failures w.r.t. the number of days the machines were operational, a failure rate of 1.05 bits/day is calculated for the 2304 Mbits.

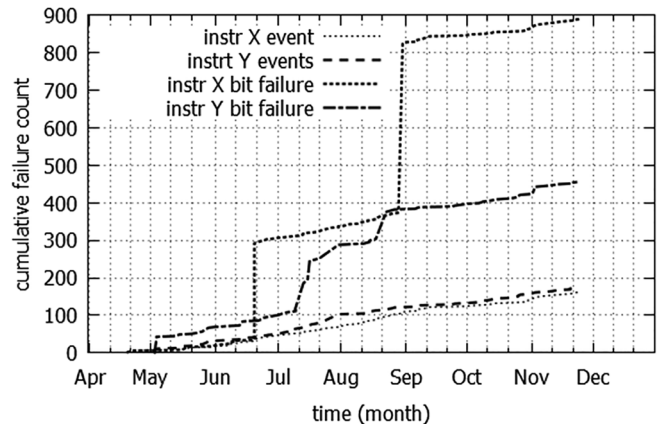


Fig. 5. Cumulative plots of the bit failures (total number of corrupted bits) and of the events (SBUs are counted as one event and MCUs as well) for both machines. Considering the results of Fig. 3, the steep parts of the plots are explained from the occurrence of major events.

TABLE I
EXPERIMENTAL DATA SUMMARY

	Machine X	Machine Y
Experimental Period	10/4/2013-26/11/2013	
Operational Days	213	
Total number of corrupted bits	889($\pm 3.4\%$)	458($\pm 4.6\%$)
Total number of events	162($\pm 7.9\%$)	175($\pm 7.5\%$)
Events of the three memories in Static mode	139($\pm 8.4\%$)	144($\pm 8.3\%$)
Events of the memory in Dynamic mode	23($\pm 20\%$)	31($\pm 17\%$)
SBUs	138	128
MCUs 2-4 cells	21	29
MCUs with ≥ 5 corrupted cells	3	18
SER_{raw} (FIT/Mbit)	75479($\pm 3.4\%$)	38885($\pm 4.6\%$)
SER_{event} (FIT/Mbit)	13754($\pm 7.9\%$)	14858($\pm 7.5\%$)

of the memories under study. Based on accelerated under beam experiments for the same memories, the cross section was found to be $1.25 \cdot 10^{-13}$ cm²/bit when irradiated at the ISIS facilities with neutrons that approach the atmospheric spectra of energies from 10-800 MeV [17]. This cross section corresponds

to an SER of 1742FIT/Mbit at NYC with a standard error of 16%. These results reveal that the difference between accelerated tests and life testing can be small (less than 20% which is part of the expected statistical error). Nevertheless, past studies have shown different results as in [5] or [6] where accelerated tests have given slightly higher estimation of the device error rate. As mentioned in [6] several reasons can explain a relevant deviation between accelerated testing and real-time testing such as errors induced by the flux calculation in the accelerator facilities, errors induced by the acceleration factor calculation, variability, aging etc.

Observing the results from Table I, we can observe the difference between the occurring upsets when the memories operate in Static mode and in Dynamic mode. The results show that Static mode testing provides more events with respect to Dynamic mode testing, and specifically the Dynamic Stress March algorithm that was chosen. This result goes against previous experimental data from accelerated testing, during which the Dynamic Stress March algorithm proved to provide the highest cross section among different dynamic tests and the static test. The difference in the number of occurring events is 30–50% lower for the Dynamic Stress test with respect to the static. The difference on the occurring events in both platforms between static and dynamic mode, can be attributed to two main reasons. The first reason is the lack of sufficient statistical data that can justify these results. When comparing the data occurring from dynamic and static mode testing, the lack of sufficient data occurring during dynamic mode testing, does not allow the generation of a solid result. Future measurements will reveal a more solid outcome. The second reason is that during dynamic mode testing, two effects work in opposite directions with respect to the SRAM cell robustness. Although the Dynamic Stress algorithm induces the RES to the cells of the array making them weaker, at the same time, a portion of the SRAM cell array defined by the electric blocks is powered at nominal V_{DD} . The powering of the cells at nominal V_{DD} makes them more robust with respect to cells that remain in low power mode. Considering the results collected so far, the low power mode during static mode testing seems to dominate the sensitivity with respect to the RES induced during Dynamic Stress test. Additional data collected from the Concordia station will shed more light on this part of the study. The overall SER of both Machine X and Y was calculated by considering both the tests in static and dynamic mode. This scheme was considered for the reason of data accumulation in the short period of exposure so far, but also of a more realistic behavior of a system where SRAM devices operate in both the static and dynamic mode.

A. Neutron Isotropy

Comparing the results from the two instruments, we can also observe that instrument X (horizontally placed), has a smaller failure rate (−9.2%) with respect to instrument Y (vertically placed). Such a small variance can be related either to a statistical error or to an actual minor effect of flux anisotropy. Further collection of data is necessary to strengthen the validity of the low effect of anisotropy.

B. Deceptive Failures Due to Permanent Defects

As already described above, interesting effects were observed during testing. Instrument Y presented some repetitive failures that affected a certain number of cells and that did not occur systematically or periodically. For example, some bits in a few consecutive words have been failing only few times during the 8-month period of the experiment. These events were not systematic, but they appeared as deceptive faults. This probably means that permanent reduction of the reliability of a group of cells has occurred, but this implies the normal function of these cells for most of the time for both retention and access. On the other hand, some concurrent stress conditions such as the dynamic access, noise on the power line and/or particle hit (that singularly would not be able to trigger events) may stimulate this weakness and cause errors. The lowering of reliability may be the result of device aging, electromagnetic interferences, defective devices (considering the total amount of devices used for these platforms), or radiation-induced defects. Currently, the platforms have not been retrieved yet, and thus we cannot provide any additional information.

C. Large-Scale Errors

Another interesting observation that was made as part of the real-time testing that we performed at Concordia was the occurrence of a few micro-SEs. As can be seen from Fig. 4, there are some points that indicate a large increment of the failing bits, in a single day. This increment is not the result of multiple MCUs, but rather the occurrence of an SEL that was limited in a small region of the memory. Fig. 6 gives an example of such an event, where each pixel in black corresponds to a corrupted bit, and each pixel in white corresponds to a noncorrupted bit. In [13], similar micro-SEs have been observed in the same SRAMs while irradiating them with a neutron beam with a spectrum that approaches the atmospheric one. These latchups do not propagate to the entire memory, but they affect a significant number of cells (few hundreds) belonging to a rectangular region defined by the presence of well-taps [13]. As previously explained, thanks to the adopted low-power scheme, the largest part of the memory array is kept in retention mode with low voltage, and only the selected electric block (where the read/write access is made) is powered with nominal voltage. The electric block is defined with respect to the n- and p-well tap positioning. According to [18] only the nominal voltage allows the triggering of latchup that cannot spread to the rest of the memory array because of the well-tap barrier and the low voltage in the unselected blocks. The observation of such events during real-time testing is of critical importance, since it demonstrates that they may occur not only under accelerated conditions, but also in real-time. Moreover, such events do not require the power reset of the device, since at the end of the operation cycle the affected block is fed with low voltage and the latchup is removed. On the memory bitmap, these SELs appear as large scale MCUs, as shown in Fig. 5.

D. Environmental Phenomena

For the platforms operating in Antarctica, there has been the opportunity to observe additional phenomena that in normal conditions, they would not face. The most important one has



Fig. 6. Physical bitmap representation of the memory under test. In this figure, a SEL is propagated to a well-tap defined region of the memory. Pixels in white denote noncorrupted bit-cells, while pixels in black correspond to corrupted cells.

been the *aurora australis* (southern light), which occurs several times every year. Aurora Australis is the light emitted when charged particles (electrons and protons) enter the Earth's atmosphere and collide with other atoms. During the experiments, several aurora australis occurred at Concordia, and, in particular, one episode lasted for many hours between the days 14th and 15th of July 2013. During these events the SEU sensing platforms were operational (no communication or power supply problems), but the recorded data do not show any difference in the error rate *w.r.t.* normal conditions. Such an observation shows that these particles did not reach the devices, since their charge has been absorbed by the atmosphere in higher altitudes and/or the structure of the building and the device packaging.

A phenomenon that could potentially induce a significant difference on the performed measurements, are the Ground Level Enhancements (GLEs)[19]. These events are the result of Solar Energetic Particle (SEP) events, during which particles of energies up to the GeV level reach the ground level as a result of the SEPs collision to the earth's atmosphere. Such particles can potentially induce SEUs to the memory devices, thus increase the observed rate of errors. However, during the period of exposure the data have been collected (April-November 2013), no Ground Level Enhancement (GLE) events have been reported. The two latest GLE events have occurred in May 2012 and January 2014 according to [20].

IV. CONCLUSION

In this paper, we introduced the experiment made through an SEU detecting platform for real time testing that operated at the Concordia Station in Antarctica. The retrieved results validated the acceleration factor of the referenced atmospheric neutron flux in the Concordia station, as well as the error rate of the SRAMs under study for such environments. Considering the flux acceleration factor that was found with respect to the one of NYC, similar experiments placed in Pic Du Midi for example, would require either more exposure time (at least $\times 2$) so that enough statistical data are gathered, or larger memory arrays. Interesting phenomena have been recorded such as the occurrence of micro-SEs, proving that these events are not occurring only in accelerated testing environments, but they appear also in real-life applications. The different orientation of the sensing devices (SRAMs) during experiments has demonstrated a good isotropic distribution of the neutron flux. Finally, it was shown that Southern Lights do not have an actual impact on the number of recorded events.

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