

Systèmes entrée-sortie non linéaires et applications en audio-acoustique

Thomas Hélie

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Séries de Volterra

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Ecole Thématique "Théorie du Contrôle en Mécanique" 2019

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Vito Volterra [1860 (Ancona) - 1940 (Roma)]



Mathématicien et physicien italien connu pour ses travaux sur les équations intégro-différentielles, la dynamique des populations, ... Prédécesseur de Fréchet et Banach, il est considéré comme l'un des fondateurs de l'analyse fonctionnelle. Il s'oppose au régime fasciste (1922) et refuse les honneurs académiques par conviction politique. Il vit alors en grande partie à l'étranger et revient à Rome peu avant sa mort.

Royal Society (1910) - Royal Society of Edinburgh (1913) Un cratère de lune porte son nom.

Avant de commencer, quelques rélénces bibliographiques

- V. Volterra. Theory of Functionnals and of Integral and Integro-Differential Equations. (Dover Pub., 1959).
- R. W. Brockett. Volterra series and geometric control theory. Automatica, 12:167–176, 1976).
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- M. Fliess et al. An algebraic approach to nonlinear functional expansions. (IEEE-TCAS, 30(8):554-570, 1983).



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- F. Lamnabhi-Lagarrigue. Analyse des Systèmes Non Linéaires. (Editions Hermès, 1994).
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Doyle et al. Identification and Control Using Volterra Models. (Springer, 2002).

Hélie & collaborators. Quelques articles personnels joints (2003-2019).





- 2 Séries de Volterra : généralités
- 3 Calcul des noyaux de Volterra d'un système différentiel
- Exercices et applications en audio-acoustique
- 5 Convergence
- Extension en dimension infinie et application



Plan



Préambule

Systèmes linéaires invariants dans le temps (/exemple)

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- Perturbations non linéaires régulières



Outline



Préambule

Systèmes linéaires invariants dans le temps (/exemple)

- Perturbations non linéaires régulières
- Séries de Volterra : généralités
- 3 Calcul des noyaux de Volterra d'un système différentiel
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State-space representation:
$$\mathbf{x}(t) = [\mathbf{z}(t), \mathbf{z}'(t)]^T, \mathbf{x}(0) = [0, 0]^T$$

$$\underbrace{\begin{bmatrix} \mathbf{z}'(t) \\ \mathbf{z}''(t) \end{bmatrix}}_{\mathbf{x}'(t)} = \underbrace{\begin{bmatrix} 0 & 1 \\ -k/m & -a/m \end{bmatrix}}_{\mathbf{A}} \underbrace{\begin{bmatrix} \mathbf{z}(t) \\ \mathbf{z}'(t) \end{bmatrix}}_{\mathbf{x}(t)} + \underbrace{\begin{bmatrix} 0 \\ 1/m \end{bmatrix}}_{\mathbf{B}} \mathbf{u}(t)$$



State-space representation: $\mathbf{x}(t) = [z(t), z'(t)]^{T}, \mathbf{x}(0) = [0, 0]^{T}$

Eq.:
$$\mathbf{x}'(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)$$

Sol.: $\mathbf{x}(t) = \int_0^t e^{\mathbf{A}\tau} \mathbf{B}\mathbf{u}(t-\tau) d\tau$ $\mathbf{y}(t) = [1,0] \mathbf{x}(t)$
 $= \mathbf{C} \mathbf{x}(t)$



$$mz''(t) + az'(t) + kz(t) = f(t)$$

Find the trajectory z(t) with respect to the force f(t)

System with input (u) / output (y) $u := f \longrightarrow \begin{bmatrix} \text{External} & & \\ \text{representation} & & \\ (\text{closed-form ?)} & & \\ y(t) = \int_{0}^{t} h(\tau) u(t-\tau) d\tau \\ \text{Convolution(/filtering) with the} \\ \text{impulse reponse} \\ h(\tau) = C e^{A\tau} B 1_{\tau > 0} \\ \end{bmatrix}$

State-space representation: $\mathbf{x}(t) = [z(t), z'(t)]^T, \mathbf{x}(0) = [0, 0]^T$

Eq.:
$$\mathbf{x}'(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)$$

Sol.: $\mathbf{x}(t) = \int_0^t e^{\mathbf{A}\tau} \mathbf{B}\mathbf{u}(t-\tau) d\tau$ $\mathbf{y}(t) = [1,0] \mathbf{x}(t)$
 $= \mathbf{C} \mathbf{x}(t)$

2. AN EXAMPLE: in the LAPLACE domain



$$mz''(t) + az'(t) + kz(t) = f(t)$$

Find the trajectory z(t) with respect to the force f(t)

System with input (\boldsymbol{u}) / output (\boldsymbol{y}) $\boldsymbol{u} := f \longrightarrow \begin{bmatrix} \text{External} \\ \text{representation} \\ (closed-form ?) \end{bmatrix} \longrightarrow \boldsymbol{y} := \boldsymbol{z}$

 $\mathbf{Y}(\boldsymbol{s}) = H(\boldsymbol{s}) \; \boldsymbol{U}(\boldsymbol{s})$

Transfer function (/filter) $H(s) = C(s I_2 - A)^{-1}B$

State-space representation: $\mathbf{x}(t) = [z(t), z'(t)]^T, \mathbf{x}(0) = [0, 0]^T$

Eq.: sX(s) = AX(s) + BU(s)Sol.: $X(s) = (sI_2 - A)^{-1}BU(s)$ Y(s) = [1,0] X(s)= C X(s)

Outline



Préambule

Systèmes linéaires invariants dans le temps (/exemple)

- Perturbations non linéaires régulières



3. What about nonlinear systems ?

Input/Output nonlinear differential system (state x)

Linear case

F(x, u) = Ax + BuG(x, u) = Cx + Du

I/O relation: linear filter

Kernel: $h(t) = \mathbf{C}e^{\mathbf{A}t}\mathbf{B} + \mathbf{D}$ Trsf. fct: $H(s) = \mathbf{C}(s\mathbf{I} + \mathbf{A})^{-1}\mathbf{B} + \mathbf{D}$

<u>Interests:</u> control, spectral analysis, identification, simulation, etc.

$$\mathbf{J} \longrightarrow \boxed{\begin{array}{c} \text{Closed-form} \\ \text{solution?} \end{array}} \longrightarrow \mathbf{y}$$

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3. What about nonlinear systems ?

Input/Output nonlinear differential system (state x)

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$$\rightarrow \boxed{\begin{array}{c} \text{Closed-form} \\ \text{solution?} \end{array}} \rightarrow$$

In general, NO ! But...

Linear case

F(x, u) = Ax + BuG(x, u) = Cx + Du

Weakly nonlinear case

F, G: power series expansions around equilibrium point 0 (nonzero coeff. at order 1)

I/O relation: linear filter

Kernel: $h(t) = \mathbf{C}e^{\mathbf{A}t}\mathbf{B} + \mathbf{D}$ Trsf. fct: $H(s) = \mathbf{C}(s\mathbf{I} + \mathbf{A})^{-1}\mathbf{B} + \mathbf{D}$

<u>Interests:</u> control, spectral analysis, identification, simulation, etc.

Example: a nonlinear spring

$$mz''(t) + az'(t) + \kappa(z(t)) = f(t)$$

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3. What about nonlinear systems ?

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<u>Interests:</u> control, spectral analysis, identification, simulation, etc.

Example: a nonlinear spring

$$mz''(t) + az'(t) + \kappa(z(t)) = f(t)$$

I/O relation: Volterra series

Interests: idem!

4. From a qualitative point of view...

A few comparisons:

| | closed-form sol. | distortions | self-oscillations |
|----------|------------------|-------------|--------------------|
| Case | w.r.t. input | | bifucations, chaos |
| General | no | yes | yes |
| Volterra | yes | yes | no |
| Linear | yes | no | no |

4. From a qualitative point of view...

A few comparisons:

| | closed-form sol. | distortions | self-oscillations |
|----------|------------------|-------------|--------------------|
| Case | w.r.t. input | | bifucations, chaos |
| General | no | yes | yes |
| Volterra | yes | yes | no |
| Linear | yes | no | no |

Interest of Volterra series:

- Natural distortions for high amplitudes
- Possible extensions to partial differential equations
- Audio-acoustics: large dynamics (/fortissimo)

5. What is the idea ?

For a Weakly Nonlinear System ...

| $\mathbf{x}'(t) = \mathbf{F}(\mathbf{x}(t), \mathbf{u}(t))$ | $\boldsymbol{F}(X,U) = \sum_{m,n} \frac{D_{m,n} \boldsymbol{F}(0,0)}{m! p!} (X,\ldots,X,U,\ldots,U)$ |
|--|--|
| $\boldsymbol{y}(t) = \boldsymbol{G}(\boldsymbol{x}(t), \boldsymbol{u}(t))$ | $\boldsymbol{G}(X,U) = \sum_{m,n} \frac{D_{m,n} \boldsymbol{G}(0,0)}{m! n!} (X,\ldots,X,U,\ldots,U)$ |

Consider the input as a **perturbation** of the system. Mark it with $\eta \in (0,1)$: $u(t) = \eta v(t)$

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For a Weakly Nonlinear System ...

 $\begin{aligned} \mathbf{x}'(t) &= \mathbf{F}(\mathbf{x}(t), \mathbf{u}(t)) \qquad \mathbf{F}(X, U) = \sum_{m,n} \frac{D_{m,n} \mathbf{F}(0,0)}{m!n!} (X, \dots, X, U, \dots, U) \\ \mathbf{y}(t) &= \mathbf{G}(\mathbf{x}(t), \mathbf{u}(t)) \qquad \mathbf{G}(X, U) = \sum_{m,n} \frac{D_{m,n} \mathbf{G}(0,0)}{m!n!} (X, \dots, X, U, \dots, U) \end{aligned}$

Consider the input as a perturbation of the system. Mark it with $\eta \in (0,1)$: $u(t) = \eta v(t)$

(i) Introduce $\mathbf{x}(t) = \sum_{n} \eta^{n} \mathbf{x}_{n}(t)$ and $\mathbf{y}(t) = \sum_{n} \eta^{n} \mathbf{y}_{n}(t)$ (ii) Inject these series expansions in the system equations (iii) Sort equations w.r.t. η^{n}

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We obtain a sequence of linear ODEs, indexed w.r.t. n

For a Weakly Nonlinear System ...

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Consider the input as a perturbation of the system. Mark it with $\eta \in (0,1)$: $u(t) = \eta v(t)$

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We obtain a sequence of linear ODEs, indexed w.r.t. n

(iv) <u>Solution</u>: Each x_n is a **multiple convolution** of *n* repeated versions of the input and a computable multivariate kernel \longrightarrow Volterra kernel

Plan



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- 2 Séries de Volterra : généralités
 - Domaine temporel
 - Domaine fréquentiel et de Laplace

3 Calcul des noyaux de Volterra d'un système différentiel

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Time domain

- Definition and examples
- A convergence criterion
- Non-uniqueness of kernels
- Remark on time-varying systems

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(with several possible functional settings)

Definition

A system $\underbrace{U}_{\{h_n\}}$ is defined by the Volterra series $\{h_n\}_{n\geq 1}$ if

$$\begin{aligned} \mathbf{y}(t) &= \sum_{n=1}^{+\infty} \int_{\mathbb{R}^n} h_n(\tau_1, \dots, \tau_n) \, u(t-\tau_1) \dots \, u(t-\tau_n) \, \mathrm{d}\tau_1 \dots \, \mathrm{d}\tau_n \\ &= \int_{\mathbb{R}} h_1(\tau_1) \, u(t-\tau_1) \, \mathrm{d}\tau_1 \end{aligned}$$

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Example

• Linear filters: $h_n = 0$, if $n \ge 2$.

Definition

A system \underbrace{u}_{h_n} is defined by the Volterra series $\{h_n\}_{n\geq 1}$ if

$$\mathbf{y}(t) = \sum_{n=1}^{+\infty} \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t - \tau_1) ... u(t - \tau_n) d\tau_1 ... d\tau$$

$$= \sum_{n=1}^{+\infty} \alpha_n (u(t))^n$$
 Power series

Example

- Linear filters: $h_n = 0$, if $n \ge 2$.
- Memoryless fct: $h_n(\tau_1, ..., \tau_n) = \alpha_n \, \delta(\tau_1, ..., \tau_n), \qquad (\delta: \text{Dirac}).$

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Example

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• General case: n=1 (linear contrib.), n=2 (quadratic), etc.

Definition

A system $\underbrace{U}_{\{h_n\}}$ is defined by the Volterra series $\{h_n\}_{n\geq 1}$ if

$$\mathbf{y}(t) = \sum_{n=1}^{\infty} \underbrace{\int_{\mathbb{R}^n} h_n(\tau_1, \dots, \tau_n) u(t-\tau_1) \dots u(t-\tau_n) d\tau_1 \dots d\tau_n}_{\text{of multiple convolutions}}$$

Example

- Linear filters: $h_n = 0$, if $n \ge 2$.
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- General case: n=1 (linear contrib.), n=2 (quadratic), etc.

A system is causal if

$$\begin{array}{rcl} \tau < 0 & \Rightarrow & h(\tau) = 0 & (\text{linear}) \\ \exists k \text{ s.t. } \tau_k < 0 & \Rightarrow & h_n(\tau_1, ..., \tau_k, ..., \tau_n) = 0 & (\text{Volterra}) \end{array}$$

RECALL: definition of a Volterra series

$$\mathbf{x}(t) = \sum_{n=1}^{+\infty} \mathbf{x}_n(t) \text{ with } \mathbf{x}_n(t) = \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t-\tau_1) ... u(t-\tau_n) d\tau_1 ... d\tau_n$$

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Bounded Input Bounded Output (BIBO) result $(\|\boldsymbol{u}\|_{\infty} = \sup_{t \in \mathbb{R}} |\boldsymbol{u}(t)|)$

$$|\mathbf{x}_n(t)| = \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t-\tau_1) ... u(t-\tau_n) d\tau_1 ... d\tau_n$$

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$$\begin{aligned} \mathbf{x}_{n}(t) &= \left| \int_{\mathbb{R}^{n}} h_{n}(\tau_{1},...,\tau_{n}) \, u(t-\tau_{1})... \, u(t-\tau_{n}) \, \mathrm{d}\tau_{1}... \, \mathrm{d}\tau_{n} \right| \\ &\leq \left| \int_{\mathbb{R}^{n}} h_{n}(\tau_{1},...,\tau_{n}) \right| \underbrace{\left| u(t-\tau_{1}) \right| \ldots \left| u(t-\tau_{n}) \right|}_{\mathbf{U}(t-\tau_{n})} \, \mathrm{d}\tau_{1}... \, \mathrm{d}\tau_{n} \end{aligned}$$

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Bounded Input Bounded Output (BIBO) result

$$(||u||_{\infty} = \sup_{t \in \mathbb{R}} |u(t)|)$$

$$|\mathbf{x}_{n}(t)| = \int_{\mathbb{R}^{n}} h_{n}(\tau_{1},...,\tau_{n}) u(t-\tau_{1})...u(t-\tau_{n}) d\tau_{1}...d\tau_{n}$$

$$\leq \int_{\mathbb{R}^{n}} |h_{n}(\tau_{1},...,\tau_{n})| \underbrace{|u(t-\tau_{1})|...|u(t-\tau_{n})|}_{\leq ||u||_{\infty}...||u||_{\infty}} d\tau_{1}...d\tau_{n}$$

$$||\mathbf{x}_{n}||_{\infty} \leq \underbrace{\int_{\mathbb{R}^{n}} |h_{n}(\tau_{1},...,\tau_{n})| d\tau_{1}...d\tau_{n}}_{=||h_{n}||_{1}} (L^{1}\operatorname{-norm})$$

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RECALL: definition of a Volterra series

$$\mathbf{x}(t) = \sum_{n=1}^{+\infty} \mathbf{x}_n(t) \text{ with } \mathbf{x}_n(t) = \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t-\tau_1) ... u(t-\tau_n) d\tau_1 ... d\tau_n$$

Bounded Input Bounded Output (BIBO) result

$$(||u||_{\infty} = \sup_{t \in \mathbb{R}} |u(t)|)$$

$$|x_n(t)| = \left| \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t - \tau_1) ... u(t - \tau_n) d\tau_1 ... d\tau_n \right|$$

$$\leq \int_{\mathbb{R}^n} \left| h_n(\tau_1, ..., \tau_n) \right| \underbrace{|u(t - \tau_1)| \dots |u(t - \tau_n)|}_{\leq ||u||_{\infty} \dots ||u||_{\infty}} d\tau_1 ... d\tau_n$$

$$||x_n||_{\infty} \leq \underbrace{\int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n)}_{= ||h_n||_1} \underbrace{(L^1\text{-norm})}_{(L^1\text{-norm})}$$

Hence,
$$\|x\|_{\infty} \leq \sum_{n=1}^{+\infty} \|x_n\|_{\infty} \leq \sum_{n=1}^{+\infty} \|h_n\|_1 (\|u\|_{\infty})^n.$$

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A convergence criterion (2/2)

(see e.g. [Boyd, 1984])

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Gain bound function φ

Define $\varphi(z) = \sum_{n \ge 1} \|h_n\|_1 z^n$ with convergence radius ρ at z = 0.

Theorem (BIBO result)

If $\|u\|_{\infty} < \rho$, then the Volterra series expansion of x is normally convergent and

 $\|\mathbf{X}\|_{\infty} \leq \boldsymbol{\varphi}(\|\mathbf{U}\|_{\infty}) < +\infty.$

A convergence criterion (2/2)

(see e.g. [Boyd, 1984])

Gain bound function φ

Define $\varphi(z) = \sum_{n \ge 1} \|h_n\|_1 z^n$ with convergence radius ρ at z = 0.

Theorem (BIBO result)

If $\|u\|_{\infty} < \rho$, then the Volterra series expansion of x is normally convergent and

$$\|\mathbf{X}\|_{\infty} \leq \boldsymbol{\varphi}(\|\mathbf{U}\|_{\infty}) < +\infty.$$

Moreover, the truncation error is bounded:

$$\left\|\sum_{n=N+1}^{+\infty} x_n\right\|_{\infty} \leq \sum_{n=N+1}^{+\infty} \|h_n\|_1 \left(\|u\|_{\infty}\right)^n$$

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Remark:

Permuting variables
$$\tau_k$$
 in $\mathbf{y}(t) = \sum_{n=1}^{+\infty} \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t - \tau_1) ... u(t - \tau_n) d\tau_1 ... d\tau_n$

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leaves the output y unchanged.

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leaves the output \mathbf{y} unchanged.

Example

 $h_2(\tau_1, \tau_2)$, $h_2(\tau_2, \tau_1)$, but also $\alpha h_1(\tau_1, \tau_2) + (1 - \alpha)h_2(\tau_2, \tau_1)$ define the same Input-Ouput system.

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Definition: A kernel is symmetrical if

for all permutations π , $h_n(\tau_{\pi(1)}, ..., \tau_{\pi(n)}) = h_n(\tau_1, ..., \tau_n)$

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Symmetrical versions of Volterra kernels $SYM[h_n]$ are unique

$$\mathsf{SYM}[h_n](\tau_1,...,\tau_n) = \frac{1}{n!} \sum_{\pi} h_n(\tau_{\pi(1)},...,\tau_{\pi(n)})$$

Other unique versions (triangular kernels, regular kernels) are also available.

Remark on time varying systems

A definition is also available for time-varying systems:

$$y(t) = \sum_{n=1}^{+\infty} \int_{\mathbb{R}^n} g_n(t,\theta_1,\dots,\theta_n) u(\theta_1)\dots u(\theta_n) d\theta_1\dots d\theta_n$$

Time-Invariant (TI) case and link with kernels h_n

TI case:
$$y(t) = \sum_{n=1}^{+\infty} \int_{\mathbb{R}^n} h_n(\tau_1, ..., \tau_n) u(t-\tau_1) ... u(t-\tau_n) d\tau_1 ... d\tau_n$$

Kernels g_n of a TI system are such that $(\theta_k = t - \tau_k)$

$$g_n(t,t-\tau_1,...,t-\tau_n)=h_n(\tau_1,...,\tau_n)$$

does not depend on t

Causal kernels g_n

$$\exists k \text{ s.t. } \theta_k > t \Rightarrow g_n(t, \theta_1, ..., \theta_k, ..., \theta_n) = 0$$

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Outline



2 Séries de Volterra : généralités

- Domaine temporel
- Domaine fréquentiel et de Laplace

3 Calcul des noyaux de Volterra d'un système différentiel

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- Exercices et applications en audio-acoustique
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- Extension en dimension infinie et application



Laplace(/Fourier) domain and analogies with linear systems

Laplace domain (or Fourier domain with $s = 2i\pi f$)

Transfer function:
$$H(s) = \int_{\mathbb{R}} h(\tau) e^{-s\tau} d\tau$$
 (lin.)

Transfer kernel:
$$H_n(s_{1:n}) = \int_{\mathbb{R}^n} h_n(\tau_{1:n}) e^{-(s_1\tau_1 + \ldots + s_n\tau_n)} d\tau_1 \ldots d\tau_n$$
 (Volt.)
denoting $(s_{1:n}) = (s_1, \ldots, s_n)$ and $(\tau_{1:n}) = (\tau_1, \ldots, \tau_n)$.

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| Causal stable system: NO poles (and NO singularities) | | | | |
|---|-------------------------------|------------|--|--|
| of <i>H</i> (<i>s</i>) | for $\Re \mathbf{e}(s) > 0$ | (linear) | | |
| of <i>H_n(s_{1:n}</i>) | for $\Re \mathbf{e}(s_k) > 0$ | (Volterra) | | |

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| of <i>H_n(s_{1:n}</i>) | for $\Re e(s_k) > 0$ | (Volterra) | | |

Input/Output relation

 $u \rightarrow \text{system} \rightarrow y$

Transfer function: Y(s) = H(s) U(s) (lin.)

Transfer kernel: more complex (Volt.)

(next part: use interconnection laws)

A result on periodic signals

Analytic input signal
$$u(t) = ae^{i\omega t}$$

 $u(t) = ae^{i\omega t} \longrightarrow [\{h_n\}] \longrightarrow y(t) = \sum_{n=1}^{+\infty} a^n H_n(i\omega, ..., i\omega) e^{in\omega t}$

Periodic input signals / Fourier series

$$u(t) = \sum_{k=-\infty}^{+\infty} u_k e^{ik\omega t} \longrightarrow [\{h_n\}] \longrightarrow y(t) = \sum_{k=-\infty}^{+\infty} y_k e^{ik\omega t}$$

with $y_k = \sum_{n=1}^{+\infty} \sum_{\substack{k_1, \dots, k_n = -\infty \\ k_1 + \dots + k_n = k}}^{+\infty} u_{k_1} \dots u_{k_n} H_n(ik_1\omega, \dots, ik_n\omega)$

Distortion coefficient for $u(t) = a \cos(\omega t)$

 $D(a,\omega) = \sum_{n=2}^{+\infty} |y_n|^2 / |y_1|^2$: closed-form solution w.r.t. a, ω, H_n .

In summary:

A Volterra series ...

- catches distortions (memory combined with nonlinearities)
- sorts the nonlinear responses w.r.t. the degree n of homogenous contributions of u
- generalizes the convolution principle
- can be described by transfer kernels in the frequency domain (as filters).
- is usually non unique but uniquely defined versions are available (useful for identification purposes)

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