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1 **Resonance frequency measurement with accuracy and stability at the 10⁻¹² level**
2 **in a copper microwave cavity below 26 K by experimental optimization**

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16
17 **Abstract**

18 Single-pressure refractive index gas thermometry (SPRIGT) is a novel primary
19 thermometry technique, developed jointly by TIPC of CAS in China and LNE-Cnam
20 in France. To help obtain a competitive uncertainty of 0.25 mK in thermodynamic
21 temperature measurements, high-stability and low-uncertainty of microwave resonance
22 frequency measurements better than 2 ppb have been demonstrated. This article
23 describes how high-stability and low-uncertainty resonance frequency measurements
24 were achieved using a copper microwave cavity. Microwave measurements were carried
25 out under vacuum and isobarically at helium-4 pressures of (30, 60, 90, and 120) kPa over
26 the temperature range of (5 to 26) K, with good consistency among microwave modes for
27 the thermodynamic temperatures determined. Performance was optimized using an Allan
28 variance analysis. In this way, with an integration time of 3 hours, a stability and accuracy
29 at the 10⁻¹² level were obtained, an almost 20 fold improvement upon our previous result
30 (Zhang *et al. Sci. Bull* 2019; 64: 286-288).

31
32 **Keywords**

33 Stability; Microwave resonance frequency; Resonator; Allan deviation; Primary
34 thermometry; Thermodynamic temperature.

1 Introduction

Single-pressure refractive index gas thermometry (SPRIGT), developed jointly by Technical Institute of Physics and Chemistry of Chinese Academy of Sciences (TIPC-CAS) in China and Laboratoire national de métrologie et d'essais-Conservatoire national des arts et métiers (LNE-Cnam) in France [1], is a type of thermometry methods based on gas polarizability. As a relative primary thermometry technique [2], SPRIGT measurements are conducted on single isobars rather than isotherms, which alleviates the need for accurate absolute pressure measurement and increases the measurement speed ten-fold. With state-of-the-art *ab initio* calculations of helium-4 properties [3-8], a competitive uncertainty of 0.25 mK^{-1} is expected for the measurement of thermodynamic temperature below the neon triple point at 24.5561 K, which is of important strategic significance for the development of both cutting-edge scientific researches and large scientific facilities, such as the Spallation Neutron Source in China and Large Hadron Collider in Europe [9].

Microwave measurements are widely used in the field of gas metrology [10-16]. SPRIGT allows the thermodynamic temperature to be determined from the ratio of gas refractive indices at a single pressure, one measured at the unknown temperature and the other at the reference temperature [1, 17], while the refractive index can be measured from the microwave resonance frequencies in a quasi-spherical high-conductivity resonator. The reference temperature can be the fixed point of neon or a known thermodynamic temperature measured using another absolute primary thermometry method, such as acoustic gas thermometry (AGT). In refractive index gas thermometry (RIGT) [2, 18], microwave measurements are used to determine the gas refractive index $n(p,T)$. From a knowledge of $n(p,T)$ and the pressure p , the thermodynamic temperature T is deduced. RIGT has been successfully implemented at

¹On May 20th, 2019, the Bureau International des Poids et Mesures announced a major revision to the SI, in which the base units, the kelvin, symbol K, was redefined by fixing the value of Boltzmann constant. The practical realizations of the kelvin by primary thermometry are indicated in the “Mise en pratique for the definition of the kelvin in the SI”, in which low-uncertainty primary thermometry is required to promote the realizations of the new kelvin and the spread of high-accuracy, low-temperature metrology. (see <https://www.bipm.org/utis/en/pdf/si-mep/SI-App2-kelvin.pdf>).

1 NRC by Rourke to measure the temperatures 24.5561 K (neon triple point), 54.3584 K
2 (oxygen triple point) and 83.8058 K (argon triple point) which are three ITS-90
3 defining fixed points [19]. Recent progress has also been achieved in the same
4 laboratory on the measurement of the thermodynamic temperature of the triple point of
5 xenon $T = 161.40596$ K by extrapolating the experimentally-determined
6 compressibility at the triple point of water [20]. In a complementary application of
7 RIGT, where the pressure is deduced from the refractive index and temperature
8 measurements, a quantum standard for absolute pressure measurements is currently
9 being developed jointly by LNE in France and INRiM in Italy in the range from 0.2 kPa
10 to 20 kPa by a superconducting microwave cavity [21]. In acoustic gas thermometry
11 [10, 13-16], microwave measurements are used to determine the dimensions or volume
12 of the resonator so that the thermodynamic temperature can be deduced from acoustic
13 measurements. Besides, microwave measurements have also been used in other precise
14 measurements, such as density and critical phenomena of helium [22], frequency [23]
15 and quantum-gas and gravitational physics [24]. These were successfully implemented
16 by using high-quality factor niobium microwave cavities. The aforementioned gas
17 metrology methods (SPRIGT, RIGT, AGT and absolute pressure in a quantum standard)
18 and precise measurements would benefit from any improvement in microwave
19 measurements.

20 The best current primary standards can produce the SI second with a relative
21 standard uncertainty approaching 10^{-16} , while the relative uncertainty of secondary
22 frequency standards based on optical clocks is at the level of 10^{-18} [25]. A convenient
23 way to perform experiments is to use a servo loop to lock the oscillator frequency to
24 the maximum of a resonance. The frequency stability then depends on the shape and
25 sharpness of the resonance, the signal-to-noise ratio, and the type of noise that
26 predominates at the time scale under consideration [26-28]. One of the most commonly
27 used techniques for servo-locking is the sideband method due to Pound and its optical
28 equivalent [28]. By contrast, in SPRIGT, it is essential to perform scans over the
29 resonance lines to check for any change in the width and shape (see Section 3.1). These
30 parameters are used both to measure electrical conductivity and monitor systematic

1 effects. A scan takes typically two minutes. To implement SPRIGT, a system has been
2 built at TIPC-CAS in China, which includes three subsystems, namely a temperature
3 control system, a pressure control system and a microwave system. The temperature
4 and pressure control systems are designed to provide a very stable working environment
5 for the microwave resonator and connecting cables. High-stability temperature control
6 (0.2 mK), pressure control (4 ppm, $1 \text{ ppm} \equiv 10^{-6}$) and microwave resonance
7 frequency (2 ppb, $1 \text{ ppb} \equiv 10^{-9}$) are required for thermodynamic temperature
8 measurements with an uncertainty of 0.25 mK. A simple description of the three
9 subsystems is presented in Section 2.

10 In this paper, the structure of the quasi-spherical resonator is described and the
11 experimental optimization of microwave measurement is presented. The rest of the
12 paper is structured as follows. In Section 2, we describe the experimental apparatus
13 involving the main instrumentations for microwave, temperature and pressure
14 measurements. Detailed descriptions of the temperature [29, 30] and pressure
15 measurement [31] can be found elsewhere and lie beyond the scope of the present paper.
16 In Section 3, we present how to realize high-stability and low-uncertainty of microwave
17 resonance frequency measurement by optimizing the microwave background, the
18 microwave emission power, mechanical stability, temperature control method, time
19 reference, and microwave signal.

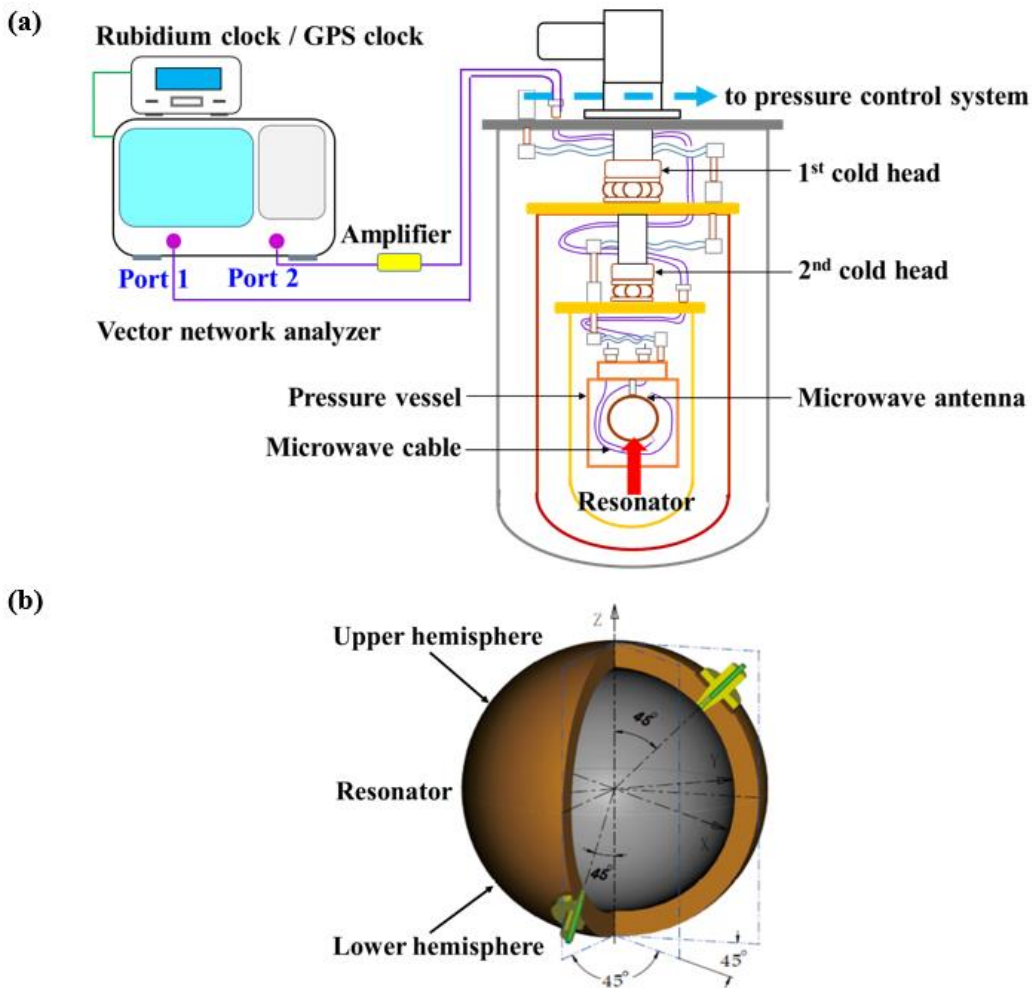
20 **2 Experimental set-up**

21 The apparatus of SPRIGT is shown in Fig. 1a, including microwave, temperature
22 control and pressure control subsystems. For the microwave system, the core element
23 is a copper (Cu-EPT) microwave cavity, *i.e.*, a quasi-spherical resonator (QSR), shown
24 in Fig. 1b. The high-quality factor² QSR was built from two hemispheres whose inner
25 surfaces were machined by precision diamond turning. The dimensions are the same as
26 those of the resonator used at LNE-Cnam for a determination of the Boltzmann constant
27 [11]. The inner shape is designed to be a triaxial ellipsoid defined by

²Quality factors ($Q = f/2g$) for the 4 microwave modes at 5 K ~26 K are: $Q(\text{TM}_{11}) \approx 140\,000$, $Q(\text{TE}_{11}) \approx 210\,000$,
 $Q(\text{TM}_{12}) \approx 170\,000$, $Q(\text{TE}_{13}) \approx 290\,000$.

1
$$\frac{x^2}{a^2} + \frac{y^2}{a^2(1+\varepsilon_2)^2} + \frac{z^2}{a^2(1+\varepsilon_1)^2} = 1 \tag{1}$$

2 with $a = 49.50$ mm, $\varepsilon_1 = 0.001$, and $\varepsilon_2 = 0.0005$, corresponding to the nominal semi-
 3 major axes of the tri-axial ellipsoid of 49.50 mm, 49.75 mm and 50.00 mm on X , Y and
 4 Z axes, respectively. The nominal shell thickness of the resonator is 10.0 mm. Two loop
 5 antennas are connected to a two-port vector network analyzer (Keysight Technologies
 6 N5241A PNA-X), one in each hemisphere used for emission and reception. The
 7 frequency reference of the vector network analyzer is a 10 MHz signal provided by a
 8 rubidium frequency standard (Stanford Research Systems FS725) or, for better stability,
 9 a GPS time and frequency system (Stanford Research Systems FS740, locked to GPS
 10 with OCXO or rubidium frequency standard). For FS740, two kinds of antenna were
 11 used, one indoor, the other outdoor.



12
 13 Fig.1. The structure schematic diagram of SPRIGT. (a) Simplified system diagram. (b) 3D drawing
 14 of the quasi-spherical resonator.

1 The QSR has been closed successfully at room temperature using a microwave
2 method by monitoring the change of relative excess half-width [32]. The resonator was
3 designed to be used under vacuum and then with helium gas. For this reason, to avoid
4 contamination, Argon (purity of 99.999 %, supplied by Beijing AP BAIF Gases
5 Industry Co. Ltd.) was flowed through it at a continuous rate of 80 SCCM³ until the
6 resonator was coupled in the pressure vessel of the cryostat. After putting the resonator
7 into the pressure vessel of the cryostat, a preliminary test on the performance of the
8 quasi-spherical resonator under vacuum was implemented without using any
9 optimization methods involved in the present work. Frequency stability and uncertainty
10 of 0.02 ppb and 0.03 ppb were realized by using 0 dBm microwave emission power
11 with an integration time of 3 h [17], equivalently 0.063 ppb and 0.095 ppb for -10 dBm
12 microwave emission power [1].

13 For the temperature control system, a cryogen-free cryostat was developed for
14 SPRIGT in our previous work [29], using a two-stage GM type pulse tube cryocooler
15 as the cooling resource. Based on multi-layer radiation shields combined with the
16 thermal-resistance method, gas type heat switch and PID control method, the pressure
17 vessel temperature stabilities of 0.021 mK ~ 0.050 mK were realized with an integration
18 time of 0.8 s in the temperature range from 5 K to 25 K. Based on further investigation
19 on thermal characterization of the cryostat, the pressure vessel temperature stabilities
20 were controlled at 0.019 mK in the same range [30]. In this work, the temperature of
21 the resonator was measured by a rhodium iron thermometer (Tinsley, SN226242)
22 coupled to an AC resistance bridge (ASL F900). The 10 Ω standard reference resistance
23 (Tinsley 5685A, SN1580409) was placed in an oil bath (Aikom Instruments MR 5100-
24 L) with temperature stability better than 1 mK.

25 For the pressure control system, the detailed content is not emphasized in this work
26 since it has been presented in our previous work [31]. The uncertainty contribution of
27 pressure to measurements of thermodynamic temperature has also been reported in our

³SCCM = standard cubic centimeters per minute, corresponding to $1.6667 \times 10^{-8} \text{ m}^3 \cdot \text{s}^{-1}$ in the International System of Units (SI). We define the volumetric flow 1 sccm as the flow of 1 cubic centimeter per minute of argon at the pressure 101.325 kPa and temperature 20°C.

1 previous work [1] and Fig. 2 in the review paper of RIGT [2]. To fill helium-4 gas
 2 (purity of 99.9999 %, supplied by Air Liquide) into the resonator and keep a single
 3 pressure, a gas compensation loop was built at room temperature to compensate for the
 4 leak of the piston gauge. The gas pressure was measured by an absolute-pressure piston
 5 gauge (Fluke PG 7601) and conducted by maintaining the piston at a constant height
 6 using a laser interferometer (KEYENCE LK-G80). Pressure stabilities of 0.0032 Pa at
 7 30 kPa and 0.002 Pa at 90 kPa were realized with an integration time of 1.4 s at room
 8 temperature, corresponding to relative stabilities of 0.1 ppm and 0.02 ppm, respectively
 9 [31].

10 **3 Experimental optimization and discussions**

11 To realize high-stability and low-uncertainty microwave resonance frequency
 12 measurements at low temperatures, several experimental optimizations were
 13 implemented by using Allan analysis of variance in the present work. Detailed results
 14 are listed as below.

15 **3.1 Microwave background polynomial**

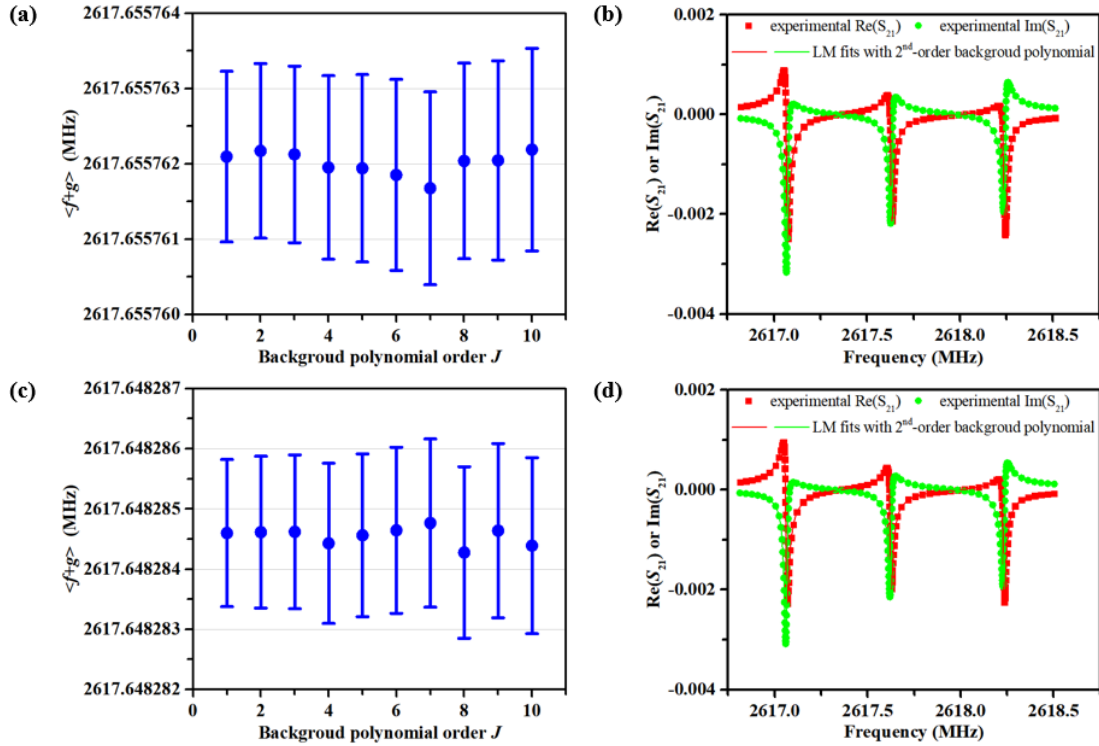
16 In this work, the microwave resonance frequencies f_n and half-widths g_n were
 17 determined from non-linear least-squares fitting (LM method [33]) of the measured
 18 complex scattering parameters S_{21} as a function of frequency [34]. Here we rewrite
 19 equation 15 in the literature to the following form

$$20 \quad S_{21} = \sum_n \frac{A_n f}{f^2 - (f_n + i g_n)^2} + \sum_{j=0}^J B_j (f - f_*)^j \quad (2)$$

21 where the fitting parameters were the complex constants A_n , B_j and the three complex
 22 resonance frequencies $(f_n + i g_n)$, one for each component of the triplet. In equation 2, J
 23 is the background polynomial order, f is the source frequency and f_* is an arbitrary
 24 constant, here we made the same choice as the literature [34], *i.e.*, $f_* = f_x$, to avoid
 25 numerical problems in the fitting program.

26 Fig.2 shows the fitting results of TM11 microwave mode for different background
 27 polynomial orders at 5 K and 24.5 K under vacuum. With the polynomial order

1 increasing, the average fitted triplet frequency $\langle f + g \rangle$ and the uncertainty $u(\langle f + g \rangle)$
 2 have a little change, as shown in Fig. 2a and Fig. 2c; the maximum relative changes on
 3 $\langle f + g \rangle$ are about 2.0×10^{-10} and 1.9×10^{-10} for 5 K and 24.5 K, respectively. These
 4 changes are less than half of the relative standard uncertainty of $\langle f + g \rangle$, indicating that
 5 increasing the microwave background polynomial order cannot obviously reduce their
 6 influence on the average fitted triplet frequency and the uncertainty in our system.
 7 Considering other microwave modes and those measured at room temperatures, a 2nd-
 8 order background polynomial is competent in this work. Table 1 lists the fitted
 9 frequencies and half-widths of the triplet for 2nd-order background polynomial. Fig. 2b
 10 and Fig. 2d compare the results of the measured and calculated components of the
 11 scattering parameters S_{21} , where good agreement was found.
 12



13 Fig.2. Fitting results of TM11 microwave mode for different background polynomial orders under vacuum: (a) $\langle f + g \rangle$ against microwave background polynomial order J at 5 K. (b) Real and
 14 imaginary components of the parameter S_{21} at 5 K, measured and fitted using 2nd-order background
 15 polynomial. (c) $\langle f + g \rangle$ against microwave background polynomial order J at 24.5 K. (d) Real and
 16 imaginary components of the parameter S_{21} at 24.5 K, measured and fitted values using 2nd-order
 17 background polynomial.
 18
 19

1 Table 1. Fitted values of the triplet with 2nd-order background polynomial.

T / K	f_1 / MHz	g_1 / MHz	f_2 / MHz	g_2 / MHz	f_3 / MHz	g_3 / MHz
5	2617.0688336(11)	0.0093815(11)	2617.6300911(15)	0.0093725(15)	2618.2397203(16)	0.0098880(16)
24.5	2617.0614026(12)	0.0094027(12)	2617.6226012(16)	0.0093916(16)	2618.2321434(17)	0.0099124(17)

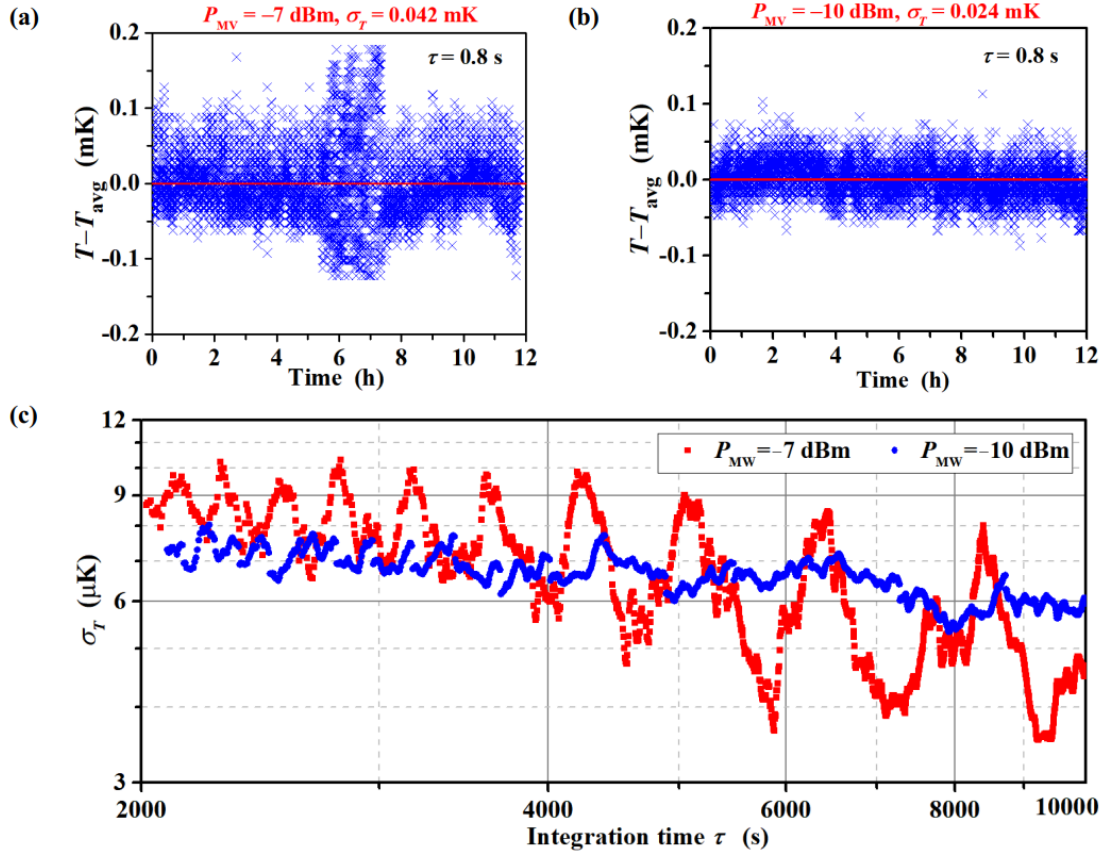
2 Besides, the background noise from cables and feed-throughs also influences the
 3 microwave measurement. To reduce the noise as much as possible, one can use some
 4 low-loss cables and feed-throughs, covering the working frequency range. For further
 5 improvement, the optimal combination of cables and feed-throughs, even the working
 6 microwave modes can be selected by measuring and analyzing the S_{21} parameters.

7 **3.2 Microwave emission power**

8 In this work, the microwave signal was generated by a two-port N5241A vector
 9 network analyzer. The signal was transmitted from port 1 of the vector network analyzer,
 10 passing through the cables, feed-throughs, and resonator, later received by port 2. Since
 11 the relative loss of the cables and feed-throughs are nearly constant at stable operating
 12 conditions, the microwave emission power has a direct influence on the measurement
 13 inside the resonator. If the power is too small, the microwave signal may be drowned
 14 in the background noise and thus leads to a relatively larger uncertainty of the fitted
 15 frequency or even no triplet. On the contrary, the microwave signal will heat the
 16 resonator too much, this will increase the temperature control instability and in turn
 17 reduce the frequency stability. Thus, it is necessary to optimize the microwave emission
 18 power.

19 Fig. 3 shows the optimized results of the power at 5.0 K under vacuum based on
 20 the TM11, TE11, TM12 and TE13 microwave modes, which were employed in our
 21 preliminary measurements [17]. A bigger power is easier to increase the temperature
 22 instability, as shown in Fig. 3a and Fig. 3b, where the temperature stabilities are about
 23 0.042 mK and 0.024 mK for microwave emission power $P_{\text{MW}} = -7$ dBm and $P_{\text{MW}} =$
 24 -10 dBm, respectively. Fig. 3c plots the temperature Allan standard deviation against
 25 the integration time for these two powers. It can be seen that there is an oscillation of
 26 the temperature Allan standard deviation for $P_{\text{MW}} = -7$ dBm, which is coupling with
 27 the measurements of the 4 microwave modes, measured in the TM11-TE11-TM12-

1 TE13 loop. While for $P_{\text{MW}} = -10$ dBm, no oscillation was observed on the temperature
 2 Allan standard deviation, which means the microwave power has negligible disturbance
 3 effect on temperature stability. Thus the optimal microwave emission power was found
 4 to be $P_{\text{opt}} = -10$ dBm. In the present work, we use one single microwave emission
 5 power P_{opt} during all the low-temperature microwave frequency measurements no
 6 matter whether the resonator is under vacuum or filled with high-purity helium-4 gas.



7

8 Fig.3. Microwave emission power optimization at 5.0 K under vacuum. (a) Temperature evolution
 9 with $P_{\text{MW}} = -7$ dBm. (b) Temperature evolution with $P_{\text{MW}} = -10$ dBm. (c) Temperature Allan
 10 standard deviation σ_T as a function of integration time for $P_{\text{MW}} = -7$ dBm and $P_{\text{MW}} = -10$ dBm.

11

12 **3.3 Mechanical stability, temperature control method, time reference,** 13 **and signal intensity**

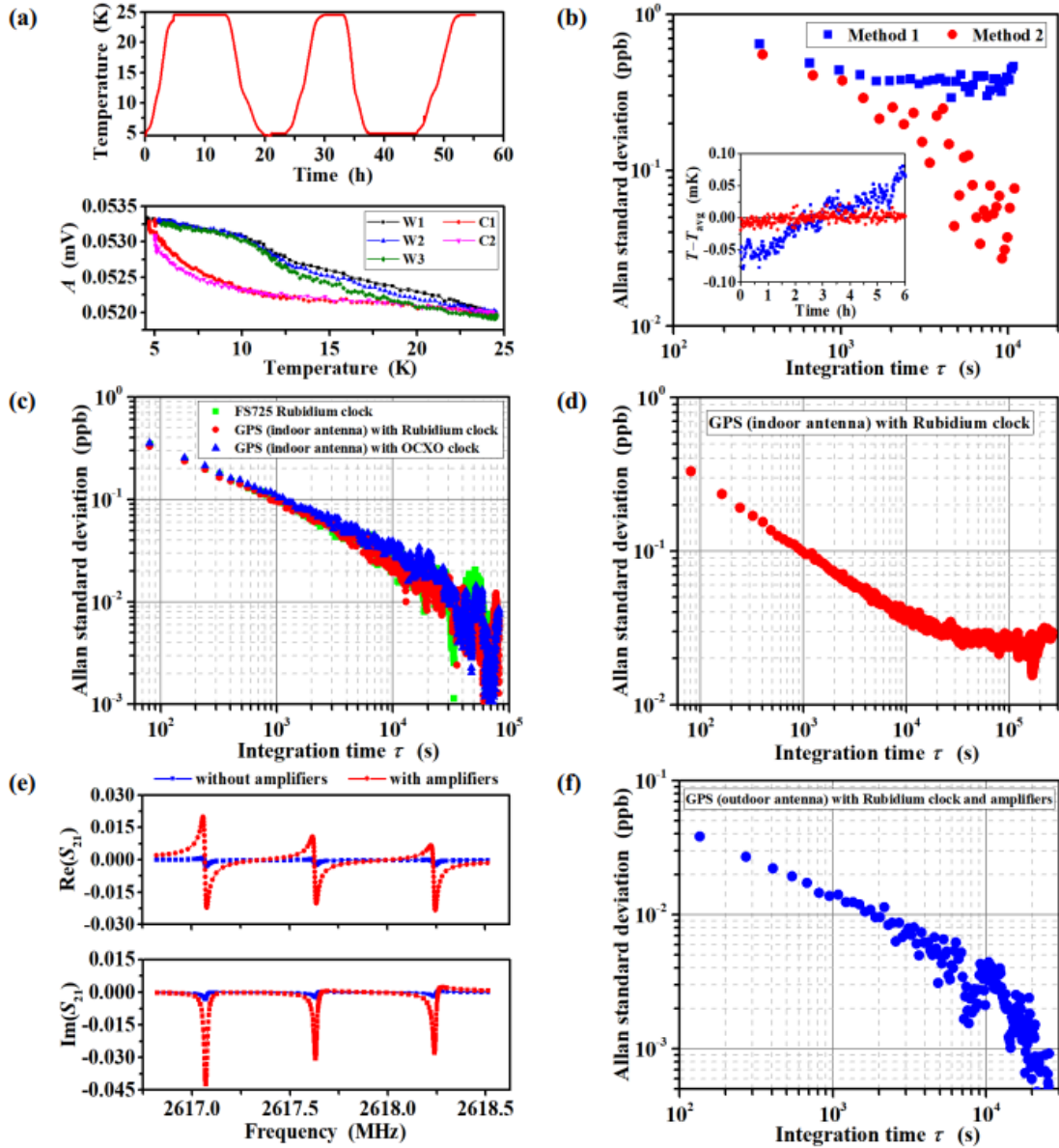
14 In this section, mechanical stability, temperature control method, time reference and
 15 signal intensity were also studied in this work under vacuum unless otherwise specified.

1 3.3.1 Mechanical stability

2 When cooling the resonator from room temperature down to low temperatures (5
3 K to 25 K) for each independent run in this work, it is important to keep the resonator
4 in good mechanical stability before doing the microwave measurements. A simple way
5 to release the mechanical stress is to do temperature cycling in the objective temperature
6 range, for example, cooling and warming between 25 K and 5 K in this work. Because
7 releasing mechanical stress needs some time. Besides, it is also good for temperature
8 and frequency measurements. At the same set point for temperature control, after the
9 temperature cycling, it is easier to realize highly stable temperature measurements,
10 which has been observed in our system. The temperature stability, in turn, will improve
11 the frequency measurement as shown in Section 3.3.2 below. When the temperature of
12 the resonator reached the objective temperatures (5 K and 24.5 K), temperature
13 regulations together with microwave measurements were implemented for several
14 hours. If the frequencies of each microwave mode had no obvious change anymore
15 (changing within microwave measurement uncertainty) at the objective temperatures,
16 then it can be considered that the mechanical stress of the resonator is completely
17 released.

18 A cooling-warming temperature cycling for 55 hours was carried out in this work.
19 The result is shown in Fig. 4a, where the cooling process and warming process are
20 denoted as C and W, respectively, in the figure legend, and the behind number is the
21 sequence of the temperature cycle. The average amplitude of $|A_p|$ in equation 2 of the
22 cooling process and warming process, denoted as A , was used to indicate the frequency
23 change with the temperature change. They do not completely overlap mainly because
24 of a big difference in temperature change speed between them. This will be improved
25 later in our work by controlling the cooling and warming speed automatically. Generally,
26 after several cooling and warming cycles, microwave measurements can be
27 implemented in the resonator when the temperature and the filled gas pressure are stable
28 simultaneously. This is the case for the present work, where obvious resonance
29 frequency changes were not observed for each microwave mode at the two objective

1 temperatures, respectively.



2
 3 Fig.4. Optimized results of mechanical stability, temperature control, time reference and signal on
 4 TM11 microwave mode under vacuum unless otherwise specified. (a) Evolution of resonator
 5 temperature and average amplitude of $|A_p|$ in equation 2 during temperature cycle. (b) Frequency
 6 Allan standard deviation with two temperature control methods at $T = 15.0$ K and $p = 30$ kPa. (c)
 7 Two-days frequency stability at 24.5 K with different time references. (d) Seven-days frequency
 8 stability at 5.0 K for FS740 time reference locked to GPS with Rubidium atomic clock configured
 9 indoor antenna. (e) Real and imaginary components of $|S_{21}|$ of the triplet at 5.0 K with and without
 10 the two microwave amplifiers. (f) Frequency Allan standard deviation at 5.0 K using FS740 locked
 11 to GPS with Rubidium atomic clock configured outdoor antenna and amplifiers.

12 3.3.2 Temperature control method

13 Due to thermal expansion effect of the resonator, its dimensions will change if the

1 temperature is not stable. The microwave resonance frequency is proportional to the
2 reciprocal of the radius of the resonator, thus the temperature stability has a direct
3 influence on microwave measurements. Fig. 4b shows the influence on microwave
4 resonance frequency stability of TM11 mode for two temperature control methods at T
5 = 15.0 K and $p = 30$ kPa. Method 1 uses an 8½ digit multimeter (Keithley 2002) with
6 Cernox sensor as in our previous work [30], while method 2 uses an AC resistance
7 bridge (ASL F18) with a calibrated rhodium-iron sensor as in CCT-K1 [35]. The
8 temperature stabilities of the two methods are 38 μ K and 8 μ K with an integration time
9 of 33.6 s, respectively. The frequency stability of method 2 is better than that of method
10 1 by about an order of magnitude with an integration time $\tau = 3$ h. Good temperature
11 stability can improve the microwave resonance frequency stability. Thus, in our later
12 measurements, method 2 was adopted all the time to maintain a good frequency stability.

13 3.3.3 Time reference

14 Usually, the vector network analyzer is connected to an external time reference,
15 such as rubidium atomic clock or GPS clock, to realize high-accuracy and high-stability
16 frequency measurements. In this work, rubidium atomic clock (FS725) and GPS clock
17 (FS740 locked to GPS) were used. While the long-term stability of GPS is excellent,
18 its short-term stability is rather poor in comparison to modern oscillators [36]. So in
19 this work, two different oscillators, an OCXO and a rubidium atomic frequency
20 standard, were configured for FS740.

21 Fig. 4c shows the typically measured stability of microwave resonance frequency
22 of TM11 mode at 24.5 K in two days with the three kinds of time references, namely,
23 FS725 (rubidium atomic clock), FS740 locked to GPS with rubidium atomic clock and
24 with OCXO clock, with GPS configured indoor antenna. The frequency stability of
25 FS740 locked to GPS with rubidium atomic clock configured indoor antenna and that
26 of the FS725 are comparatively equivalent, both stabilities are better than that of OCXO
27 clock. A resonance frequency stability of 0.02 ppb was realized using FS740 locked to
28 GPS with rubidium atomic clock configured indoor antenna and FS725 with an
29 integration time $\tau = 3$ h in the two-days microwave measurements. The SPRIGT

1 measurement at different temperatures and pressures may take one or two months, the
2 time reference stability of the reference should be tested at this level to check if any
3 other effect may perturb the measurement.

4 Furthermore, long-term microwave measurements at 5 K in seven days were
5 carried out by FS740 locked to GPS with rubidium atomic clock configured indoor
6 antenna. The stability result is plotted in Fig. 4d. A good resonance frequency stability
7 of 0.04 ppb was achieved with $\tau = 3$ h.

8 **3.3.4 Signal intensity**

9 Amplifier is usually used to enhance the microwave signal intensity as shown in
10 literature [37], where the amplitude of $|S_{21}|$ was amplified by nearly ten times with the
11 help of a set of two 10 dBm amplifiers (Model ZX60-14012L manufactured by Mini-
12 Circuits, bandwidth from 0.3 MHz to 14 GHz) connected in series at room temperature.
13 The uncertainty was reduced by a factor of 5 - 12 even with a low-cost vector network
14 analyzer.

15 In this work, with the same method and two amplifiers like the literature [37],
16 measurements were carried out at 5 K under vacuum. Fig. 4e shows the comparison
17 results of the real and imaginary components of $|S_{21}|$ with and without the two amplifiers.
18 With amplifiers, the amplitudes were amplified by about 10 times and in turn, the fitted
19 relative standard uncertainty of resonance frequency was reduced by a factor of 10
20 (from 0.3 ppb to 0.031 ppb). This is because the amplifiers increase the resolution of
21 the peaks of the triple. A similar situation appears in TE11, TM12, and TE13 microwave
22 modes, with the fitted relative standard uncertainty of resonance frequency reduced by
23 a factor of 5, 5, and 10, respectively. It is a simple and effective way to solve signal
24 problems, such as high-loss of microwave cable and feed-through, even low-cost vector
25 network analyzer. Besides, low-temperature microwave amplifiers are expected to have
26 a better performance than room-temperature amplifiers, and this will be one solution of
27 ultra-high accuracy microwave measurements in the future work.

28 **3.3.5 Overall performance**

29 In Section 3.3.3, the microwave measurements were carried out by using the indoor

1 antenna, which was placed outside a window of the laboratory building, facing north
2 against another building. Actually, the FS740 was supplied with two types of antenna,
3 indoor antenna and outdoor antenna. According to the user's manual of the FS740 time
4 reference, best results can be realized if the antenna has a clear unobstructed view of
5 the sky. It is highly recommended to put the outdoor antenna on the roof of the building,
6 within which the FS740 is located. Because it improves the quality and reliability of
7 the GPS signals as more satellites will typically be visible and with a higher signal-to-
8 noise ratio. This can improve the long-term stability of the FS740 time reference by a
9 factor of three. This configuration was used for the following test described in this
10 Section.

11 Combining the above optimized results, namely, second-order background
12 polynomial, optimized power $P_{\text{opt}} = -10$ dBm, good mechanical stability, temperature
13 control using an AC resistance bridge and rhodium-iron resistance sensor, the time
14 reference FS740 locked to GPS with rubidium atomic clock but configured outdoor
15 antenna and two room-temperature amplifiers connected in series, preliminary stability
16 measurements on TM11 mode were implemented at 5 K under vacuum. Fig. 4f plots
17 the Allan standard deviation of the resonance frequency. With the same time reference,
18 FS740 locked to GPS with rubidium atomic clock, the stability for outdoor antenna
19 with amplifiers is about an order of magnitude better than that for only indoor antenna
20 as shown in Fig. 4d and Fig. 4f. High-stability of 3.7 parts per trillion (ppt, 1 ppt \equiv
21 $10^{-12} \equiv 10^{-3}$ ppb) and low-uncertainty of 5.2 ppt were realized with an integration
22 time of 3 h. The frequency stability and uncertainty in the present work are nearly 20
23 times better than those in our previous work [17], 0.064 ppb and 0.095 ppb with the
24 same integration time and $P_{\text{opt}} = -10$ dBm [1].

25 In SPRIGT, the thermodynamic temperature uncertainty has many influence
26 factors, frequency is one of the major factors as shown in our previous work [1] and the
27 RIGT review paper [2]. The improvement in frequency measurement accuracy can
28 reduce its influence on thermodynamic temperature measurement by a factor of 20,
29 which is good for realizing the high-accuracy measurement of thermodynamic
30 temperature by SPRIGT. The above optimization methods have the potential to be used

1 in gas metrology and other research fields, where high-stability and low-uncertainty
2 microwave measurements are necessary.

3 **3.4 Results with pressures**

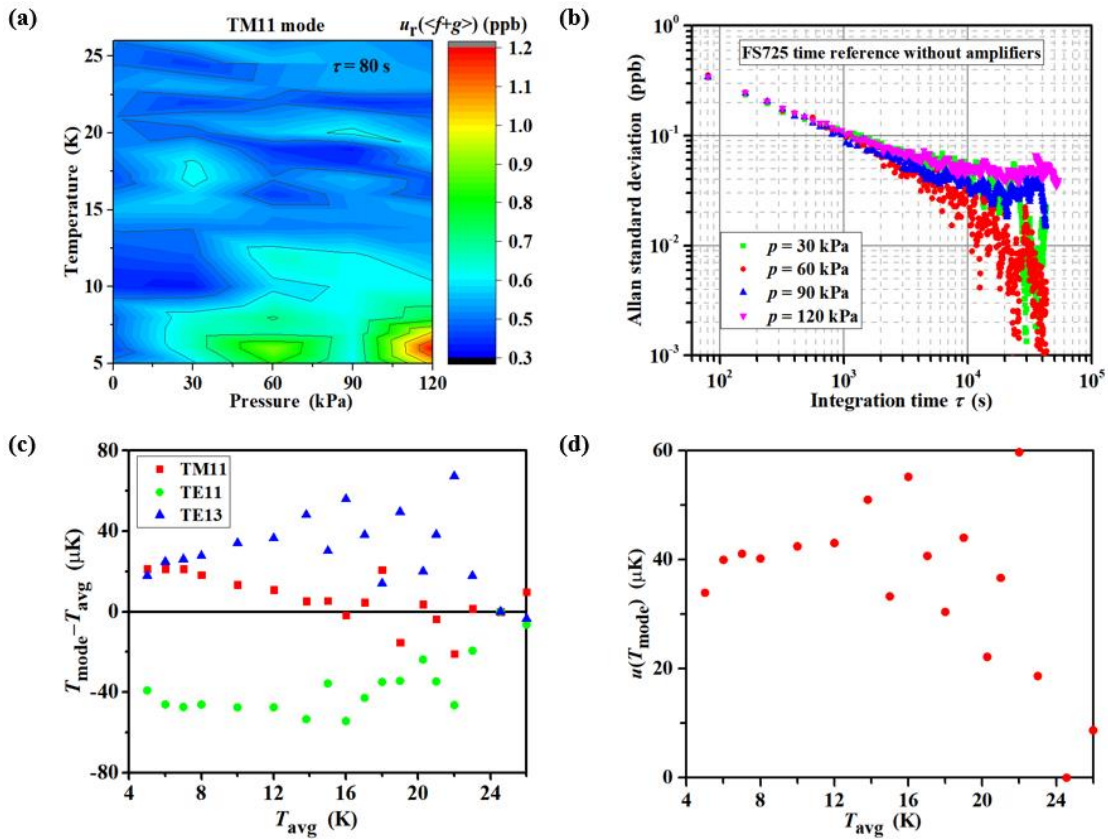
4 After many experimental optimizations under vacuum, preliminary microwave
5 measurements were performed from 5 K to 26 K with the resonator filled by high-purity
6 helium-4 at the following conditions: 1) fitting S_{21} with 2nd order background
7 polynomial; 2) setting microwave emission power $P = P_{\text{opt}} = -10$ dBm; 3) doing
8 temperature cycle to make the resonator with a good mechanical stability; 4) controlling
9 temperature by AC resistance bridge and rhodium iron sensor; 5) using time reference
10 FS725; and 6) without the two amplifiers. The measurements were implemented on
11 isobars and the results are described in the [Fig. 5](#).

12 [Fig. 5a](#) shows the total relative standard uncertainty of $\langle f+g \rangle$ of TM11 mode at
13 temperatures from 5 K to 26 K and pressures up to 120 kPa with $\tau = 80$ s⁴. Most
14 uncertainties are located within 0.9 ppb, except the p - T region with pressures near 120
15 kPa and temperatures near 5 K. The main reason is that at lower temperatures, the
16 pressure stability inside the resonator becomes a little worse than that at higher
17 temperatures, which disturbed the frequency stability by the change of helium-4 density
18 inside the resonator. [Fig. 5b](#) plots a one-day microwave measurement of TM11 mode
19 at 24.5 K with different pressures. The frequency stabilities are about (0.03, 0.02, 0.04,
20 0.05) ppb for (30, 60, 90, 120) kPa with $\tau = 3$ h, respectively, and the temperature
21 stabilities are about (6.9, 8.3, 7.9, 6.6) μK with $\tau = 33.6$ s. Due to the pressure stability,
22 60 kPa has the best frequency stability in the one-day stability measurement. These are
23 still better than our previous work (0.063 ppb under vacuum provided with $P_{\text{opt}} = -10$
24 dBm) even the pressure instability increases the microwave frequency measurement

⁴It is usually used a constant integration time for all the microwave modes. Figure 5(a) only shows the uncertainty of TM11 microwave mode for different pressures and temperatures. However, for TE13 mode, the uncertainties are very near 2 ppb. Besides, the integration time was also chosen to be in the same “time” as the other instruments, such as the resistance bridge. The integration time of 20 000 s is used to check and characterize the stability of the measurements.

1 instability. The stability and uncertainty still have room for improvement, provided later
 2 using the time reference, FS740 locked to GPS with rubidium atomic clock configured
 3 outdoor antenna, and microwave amplifiers.

4 Thermodynamic temperatures calculated from different modes were determined
 5 for the isobar $p = 60$ kPa, based on the SPRIGT principle [1]. Fig. 5c and Fig. 5d
 6 presents the differences between mode thermodynamic temperatures T_{mode} and the
 7 average values T_{avg} , ($T_{\text{mode}} - T_{\text{avg}}$), at temperatures from 5 K to 26 K, where microwave
 8 measurements with $\tau = 80$ s were used. Good mode consistency was observed on (T_{mode}
 9 $- T_{\text{avg}}$) within ± 80 μK , and a small relative standard uncertainty component for the
 10 thermodynamic temperatures, from mode consistency, was realized with $u(T_{\text{mode}})$
 11 values less than 60 μK ⁵.



12
 13 Fig.5. Microwave measurements from 5 K to 26 K at pressures up to 120 kPa. (a) Total relative
 14 standard uncertainty of $\langle f + g \rangle$ for TM11 mode with $\tau = 80$ s. (b) One-day frequency stability
 15 measurements at 24.5 K for (30, 60, 90, 120) kPa. (c) Thermodynamic temperature differences

⁵Uncertainty budget for SPRIGT has been reported in our previous work [1] and the review paper of RIGT [2]. Since this work is focused on microwave measurements, it will be conducted in our future work.

1 ($T_{\text{mode}} - T_{\text{avg}}$) for TM11, TE11 and TE13 microwave modes at 60 kPa. (d) Standard uncertainty
2 component of SPRIGT thermodynamic temperature from mode consistency.

3 **4 Conclusion**

4 The object of this study was the optimization of microwave frequency
5 measurements at low temperatures. The measurements in question concern the
6 resonance frequencies of a high quality factor, tri-axial ellipsoidal copper resonator
7 either under vacuum or else filled with high-purity helium-4 gas (at pressures from
8 30 kPa to 120 kPa). The ratio of frequencies with and without gas yields the refractive
9 index from which temperature can be determined with high accuracy once the effects
10 of pressure are taken into account. This is called refractive index gas thermometry
11 (RIGT). The application here concerns a novel variant of the technique exploiting single
12 isobars such that pressure-dependent effects almost cancel – single-pressure RIGT or
13 SPRIGT. Microwave resonances are detected using pairs of antennas linked to a vector
14 network analyzer.

15 The study dealt with both experimental parameters and data analysis. Frequency
16 was characterized using the two-sample (Allan) variance widely used in time and
17 frequency metrology. To minimize self-heating yet with an acceptable signal-to-noise
18 ratio, a microwave power of -10 dBm was found to be optimal. A slight **improvement**
19 of the quality of the fitted lineshape compared with our previous work [17] was
20 obtained using a quadratic function for the background rather than a linear one. The use
21 of temperature cycling between 5 K and 25 K turned out to improve the reproducibility
22 of frequency measurements. Fine control of the resonator temperature is achieved using
23 a calibrated rhodium-iron resistance thermometer read using an AC resistance bridge
24 instead of an 8½ digit multimeter. The local frequency reference was a GPS referenced
25 commercial rubidium clock. It was essential to use a roof-based outdoor antenna for the
26 GPS system. Frequency resolution was also improved by the introduction of two
27 microwave pre-amplifiers in series, albeit close to the network analyzer located at room
28 temperature. Increased performance is expected with the installation of cryogenically
29 cooled pre-amplifiers located closer to the microwave resonator.

1 In this way, for an integration time of three hours, microwave resonance frequency
2 measurements with a stability at the 10^{-12} level were demonstrated, which represents an
3 almost 20 fold enhancement compared with those of our previous work [17]. The
4 present observations could help in other high-stability, low-uncertainty frequency
5 measurements in copper microwave cavities, not only for improving thermodynamic
6 temperature measurement in SPRIGT, but also for other precise measurements, where
7 high-stability and low-uncertainty microwave measurements are necessary.

8 The final uncertainty achieved using SPRIGT in our laboratory is equal to or better
9 than 0.17 mK in the range 5 K to 25 K. An extensive description of the uncertainty
10 budget will be presented in a future publication. As far as frequency measurements are
11 concerned (*i.e.*, the subject of the present manuscript), had they not been possible at the
12 10^{-12} level, the subtle influence of microwave cables and feed-thoughts would perhaps
13 not have been detected as early as it was. The effect of thermal cycling is not one of
14 uncertainty but rather of potential mechanical instability.

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