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Design and Comparison of Two Advanced Core Control Systems for Flexible Operation of Pressurized Water Reactors

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Abstract

This paper focuses on the design of advanced core control systems for future generations of pressurized water reactors. The objective is to improve the flexibility of nuclear power plants to cope with the rapid growth of renewable energies. In practice, this means that the average coolant temperature, the axial power distribution of the reactor core and the position of the control rods have to be properly regulated during power variations. In previous work, conducted by the same authors, two promising approaches were investigated: 1) fixed-structure gain-scheduled control and 2) nonlinear model predictive control. Here, both methods are tested according to industry standards in an attempt to determine the best one for our problem. To achieve this, two different controllers are designed using a new multipoint kinetic model of the reactor core, which provides an accurate representation of the axial power distribution. The advantages and drawbacks of both design methodologies are discussed and then compared on PWRSimu, an intermediate complexity pressurized water reactor simulator developed by Framatome.

Keywords: Nuclear Power Plants, Gain-Scheduling, Model Predictive Control

Nomenclature

ACT	Average Coolant Temperature
AO	Axial Offset
GSC	Gain-Scheduled Control
MS	Multiple-Shooting
NMPC	Nonlinear Model Predictive Control
NP	Nominal Power
\mathbf{PWR}	Pressurized Water Reactor
\mathbf{SS}	Single-Shooting

1 1. Introduction

In order to address climate change, many countries are 2 seeking to replace fossil fuel power plants by renewable з energy sources [1]. This energy transition poses new chal-4 lenges in terms of management of the electrical network. 5 Since excess electricity cannot currently be stored on a 6 large scale, generation and consumption have to be contin-7 uously balanced to ensure grid stability. For this purpose, 8 a number of power plants must be flexible, i.e., capable of 9 adjusting their power output on demand. However, unlike 10 fossil fuel power plants, which are easily controllable, part 11 of renewable energy sources, such as wind turbines or solar 12

panels, are inherently intermittent. It is therefore essential to enhance the flexibility of conventional power plants to prepare for the upcoming expansion of renewables [2].

A nuclear power plant can be operated either in base-16 load or in load-following mode [3]. In base-load mode, the 17 power output of the plant remains constant regardless of 18 electricity demand. Most nuclear power plants around the 19 world are still operated in base-load mode because it is 20 simpler and more cost-effective. In load-following mode, 21 the power output of the plant is adjusted depending on 22 the forecasted electricity consumption. Typically, the grid 23 operator sends a daily load profile in advance to the plant 24 operator, who will then set the power target of the turbine 25 accordingly. Whether in base-load or in load-following 26 mode, the nuclear power plant can also provide frequency 27 control to the grid. This is achieved by a dedicated con-28 troller that automatically adjusts the power output of the 29 turbine to offset small load imbalances within seconds. 30

Flexible operation of a nuclear power plant is directly 31 related to the design of the core control system [4]. The 32 main task of the core control system is to maintain the 33 average coolant temperature (ACT) and the axial power 34 distribution of the reactor core, or axial offset (AO), within 35 appropriate limits. These limits are defined upstream to 36 ensure acceptable performance levels and safe operation 37 of the plant. During power variations, ACT regulation is 38 achieved by moving several neutron-absorbing control rods 39 inside the reactor core. However, their movements have 40 a detrimental impact on its axial power distribution. Ini-41 tially, the movements of the control rods were mitigated by 42

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adjusting the boron concentration of the primary coolant. 43 This strategy was limiting in terms of flexibility because 44 the boron concentration of the coolant can only change 45 slowly and gradually decreases over the fuel burn-up cycle. 46 For example in mode A, the maximum power variation rate 47 of the plant is typically restricted to 1-2 %NP/min at the 48 beginning of the cycle, and $0.2 \ \%$ NP/min at the end. To 49 overcome these limitations, advanced core control systems 50 usually comprise two separate groups, or banks, of control 51 rods [4]. In this way, one bank of control rods can be used 52 to perform power variations by, e.g., controlling the ACT, 53 while the other can be used to reduce axial power distribu-54 tion disturbances. Thus, flexibility is increased with power 55 variation rates of up to 5 %NP/min. After power varia-56 tions, the xenon distribution of the reactor core will slowly 57 change for several hours until steady-state is reached. Dur-58 ing this time, the evolution of xenon concentration has 59 to be counterbalanced to prevent the AO of the reactor 60 core from drifting away. In highly advanced core control 61 systems, such as mode T [5], this can be done either by 62 moving the control rods or by adjusting the boron concen-63 tration of the primary coolant. The main motivation for 64 using the control rods is to minimize the volume of effluent. 65 Yet, this can decrease the maneuvering capabilities of the 66 plant, as the control rods may not be properly positioned 67 for a quick return to nominal power. Maximum flexibility 68 is achieved when xenon poisoning is compensated by boron 69 concentration adjustments and when the level of insertion 70 of the control rods is controlled. 71

In practice, industrial core control systems manage to 72 meet all the aforementioned requirements using single in-73 put single output Proportional-Integral-Derivative control 74 along with sophisticated logic rules. Even if this approach 75 gives satisfactory results for now, it may become insuf-76 ficient against more stringent grid management criteria. 77 Fortunately, a wide variety of core control systems have 78 already been proposed (see [6] and references therein). 79 Popular methods include LQG/LTR control, H_{∞} control, 80 sliding-mode control, model predictive control and gain-81 scheduled control. Nevertheless, these conceptual core 82 control systems do not tackle all the issues that arise from 83 the flexible operation of nuclear power plants. Hence, 84 this paper proposes to bridge the gap between industry 85 and academia by developing two different advanced core 86 control systems using modern design techniques. In view 87 of the strong operational constraints that are placed on 88 the nuclear industry, two promising approaches have been 80 identified: fixed-structure gain-scheduled control [7, 8] and 90 nonlinear model predictive control [9]. 91

To the best of our knowledge, very few papers have tried to apply these techniques to the control of a nuclear reactor core. For instance, it seems that [7, 8] are the first ones that studied fixed-structure gain-scheduled control in this context. Regarding model predictive control, only a handful of examples have been reported in the scientific literature. In [10, 11], a linear model predictive controller is designed to achieve both ACT and AO regulation of a reactor core that is operated in load-following mode. How-100 ever, the boron concentration of the coolant is adjusted by 101 another module and the nuclear power plant does not pro-102 vide frequency control to the grid. In another paper [12], 103 the power output of a reactor core is controlled by a robust 104 nonlinear model predictive controller. Yet, this theoreti-105 cal work is not industrially relevant as it does not even 106 address AO regulation. By contrast, the main contribu-107 tion of this paper is to conduct a realistic and compre-108 hensive study on the design of nuclear reactor core con-109 trol systems. The objective is to give practical insights 110 and methodological guidelines that could help industrial 111 practitioners working in nuclear engineering. Two mod-112 ern control techniques, which were seldom tested in this 113 field, are compared against the technical specifications of 114 Framatome. In order to ensure consistency with industry 115 standards, the actuators and control objectives are identi-116 cal to those considered in mode T [5]. 117

Following on from the work carried out in [8], the first 118 core control system is designed based on a fixed-structure 119 H_2/H_∞ gain-scheduled control approach. The main im-120 provement is that the controller is tuned using a multipoint 121 kinetic model of the reactor core rather than a point ki-122 netic one. In addition, the mathematical formulation of 123 the H_2/H_{∞} synthesis problem has been refined and made 124 more readable. Moreover, the gains are now simultane-125 ously tuned at each operating point and smoothly inter-126 polated as a function of the scheduling variable [13]. This 127 greatly simplifies the design of the controller, as it avoids 128 a posteriori interpolation, or post-processing, of the gains. 129 The second core control system is designed based on a 130 nonlinear model predictive control approach [14, 15]. This 131 time, the multipoint kinetic model of the reactor core is 132 embedded into the controller to predict the future response 133 of the plant. Thus, the control inputs can be calculated 134 online by repeatedly solving a constrained finite horizon 135 optimal control problem. To this end, the model of the 136 reactor core is reduced using singular perturbation theory 137 so that it can be efficiently simulated by the controller. 138

The remainder of this paper is organized as follows: 139 Section II describes the system under study and formal-140 izes the control objectives; Section III presents the design 141 methodology of the fixed-structure H_2/H_{∞} gain-scheduled 142 controller and Section IV presents the design methodol-143 ogy of the nonlinear model predictive controller; Then, 144 the advantages and drawbacks of both techniques are dis-145 cussed in Section V based on a typical load-following 146 MATLAB[®]/Simulink[®] simulation scenario; Finally, con-147 clusions and perspectives are drawn in Section VI. 148

2. Problem Formulation

2.1. System Modeling

To design the core control systems, a new multipoint core kinetic model of a pressurized water reactor (PWR), drawn in Fig. 1, has been developed (a brief overview 153

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of the model is given in [16] Appendix F). Its main ad-154 vantage, compared to the former point kinetic model pre-155 sented in [7, 16], is that the behavior of the AO is more 156 accurately represented since the reactor core is now di-157 vided into six meshes. However, increasing the number of 158 meshes also increases the size of the model. As illustrated 159 in Fig. 1, every new mesh requires five additional states: 160 two average neutron n and delayed neutron c densities, 161 two iodine I and xenon X_e concentrations, and one in-162 let temperature T. For each mesh, neutron dynamics are 163 described by point kinetics equations with one group of de-164 layed neutrons. An adequately chosen constant exchange 165 coefficient has been added to take account of the trans-166 fer of neutrons between meshes. In addition, the iodine 167 and xenon concentrations are calculated by a multipoint 168 iodine-xenon estimator. The rise in coolant temperature, 169 from the bottom to the top of each mesh, is given by a 170 first order differential equation and is proportional to the 171 amount of power generated by nuclear fission. In each 172 mesh, the growth rate of the fission chain reaction is char-173 acterized by a quantity known as reactivity [17, 18]. Crit-174 icality is achieved when the fission chain reaction is stable 175 and self-sustaining, i.e., when the reactivity of the reac-176 tor core equals zero. Reactivity depends on many factors, 177 namely the fuel and coolant temperatures, the xenon con-178 centration of the reactor core, the position of the control 179 rods and the boron concentration of the primary coolant. 180 The resulting feedbacks and underlying interactions be-181 tween all involved variables lead to strong plant nonlin-182 earities and coupling. 183



Fig. 1: Simplified diagram of the multipoint PWR core kinetic model

In order to move from reactor core to plant scale, a very simple steam generator is modeled. Once again, it is assumed that the fall in temperature between the inlet and outlet of the steam generator is given by a first order differential equation and is proportional to the power of 188 the turbine. In fact, whenever the power of the turbine is 189 decreased (resp. increased), the average coolant tempera-190 ture will increase (resp. decrease). As a result, the nuclear 191 power of the reactor core will also decrease (resp. increase) 192 and ultimately reach steady-state because the PWR was 193 designed so that the fission chain reaction can be stabilized 194 by temperature reactivity feedbacks alone (i.e., Doppler ef-195 fect and moderation) [17]. However, it is very likely that, 196 without control, the average coolant temperature will de-197 viate from the limiting conditions of operation. Further-198 more, xenon-induced spatial power oscillations may appear 190 if the axial power distribution of the reactor core is left un-200 controlled. Therefore, the role of the core control system 201 is to ensure that both the ACT and the AO stay within 202 their respective limits before, during and after power vari-203 ations [4]. This can be achieved by slightly altering the 204 reactivity of the reactor core. 205

In mode T [5], the nuclear reaction is controlled by mov-206 ing two separate banks of control rods inside the reac-207 tor core and by adjusting the boron concentration of the 208 primary coolant. Reactivity can be increased (resp. de-209 creased) either by withdrawing (resp. inserting) the con-210 trol rods from (resp. in) the reactor core or by decreasing 211 (resp. increasing) the boron concentration of the coolant. 212 The control rods are mainly used to handle abrupt changes 213 in reactivity but adversely impact the power distribution 214 of the reactor core. In fact, reactivity will mostly decrease 215 in the region of the reactor core where the control rods 216 are being inserted. To alleviate this problem, one bank of 217 control rods, denoted by P_{bank} , is dedicated to ACT reg-218 ulation whereas the other, denoted by $\mathrm{H}_{\mathrm{bank}},$ is dedicated 219 to AO regulation. Conversely, the boron concentration of 220 the primary coolant acts uniformly on the reactor core but 221 takes time to become effective. This is due to the fact that 222 the solutions of boric acid and demineralized water, which 223 are injected to modify the boron concentration, have to 224 flow throughout the volumetric and chemical control cir-225 cuit before reaching the coolant. Hence, boron concentra-226 tion adjustments are mostly used to counterbalance axial 227 xenon oscillations after power variations. To summarize: 228

- ACT regulation is achieved by moving P_{bank}
- AO regulation is achieved by moving H_{bank} during power variations and by adjusting the boron concentration of the primary coolant during xenon oscillations
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In addition, the mode T allows the reactor core to quickly return to its nominal power. This is achieved by controlling the level of insertion of P_{bank} in the reactor core. Last but not least, all three controlled variables can be successfully regulated even when the turbine provides frequency control to the grid. 236 237 238 239 239 239 239

2.2. Control Objectives

As mentioned before, the goal of an advanced core control system is to keep the ACT, the AO and the position 242 of P_{bank} within appropriate limits during power variations
and while frequency control is active. These limits were determined beforehand by nuclear reactor engineers to comply with genuine performance and safety requirements. In
mode T, their values are given by:

$$|T_a(t) - T_{a,ref} (P_{ref}(t))| \le 1.5 \text{ °C}$$

$$|AO(t) - AO_{ref}| \le 5 \% \qquad (1)$$

$$|P_{bank}(t) - P_{bank,ref} (P_{ref}(t))| \le 30 \text{ steps},$$

where T_a is the average temperature of the coolant, AO 249 is the axial offset of the reactor core, P_{bank} is the posi-250 tion of the first bank of control rods, and $T_{a,ref}$, AO_{ref} 251 and $P_{bank,ref}$ are their respective reference signals. Note 252 that the reference signals $T_{a,ref}$ and $P_{bank,ref}$ are both 253 piecewise linear functions of the load profile P_{ref} whereas 254 AO_{ref} is a constant. Shorthand notation for the deviations 255 of ACT, AO and P_{bank} are: 256

$$\Delta T_{a}(t) = T_{a}(t) - T_{a,ref}$$

$$\Delta AO(t) = AO(t) - AO_{ref}$$

$$\Delta P_{bank}(t) = P_{bank}(t) - P_{bank,ref}(P_{ref}(t))$$

Actuator saturation should also be considered to make the core control system practically workable. In our case, the reactor core is actuated by changing the speeds of both P_{bank} and H_{bank} , and by adjusting the boron concentration of the coolant. The speeds of P_{bank} and H_{bank} are simply limited by:

$$\left|\frac{\mathrm{d}P_{bank}(t)}{\mathrm{d}t}\right| \le 75 \text{ steps/min}$$

$$\left|\frac{\mathrm{d}H_{bank}(t)}{\mathrm{d}t}\right| \le 75 \text{ steps/min},$$
(2)

where H_{bank} is the position of the second bank of control rods. Besides, P_{bank} and H_{bank} are saturated as well since the control rods cannot move beyond their maximum insertion and extraction thresholds:

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$$36 \le P_{bank} \le 1053 \text{ steps}$$

$$9 \le H_{bank} \le 411 \text{ steps.}$$
(3)

The upper and lower bounds of the injected boron concen-270 tration variation rate depend on the current boron con-271 centration of the coolant. More precisely, the higher the 272 boron concentration of the coolant, the lower the efficiency 273 of a boric acid injection. Similarly, the lower the boron 274 concentration of the coolant, the lower the efficiency of 275 a demineralized water injection. This translates into the 276 following inequalities: 277

$$278 \qquad -\frac{10}{M_t}C_b(t) \le \frac{\mathrm{d}C_{b,in}(t)}{\mathrm{d}t} \le \frac{3}{M_t} \left(C_{b,max} - C_b(t)\right), \quad (4)$$

where $C_{b,max}$ is the maximum boron concentration of the coolant, M_t is the total mass of water in the primary cooling circuit, $C_{b,in}$ is the injected boron concentration and C_b is the boron concentration of the coolant.

In our model, it is assumed that the turbine control 283 system has already been designed and is functioning per-284 fectly. Thus, the power of the turbine can be expressed 285 as $P_{turb} = P_{ref} + \delta P_{ref}$ where P_{ref} is the known load 286 profile and δP_{ref} is an unknown reference signal that is 287 added when frequency control is active. Specifically, P_{ref} 288 appears as a ramp-like signal during power variations and 289 as a step-like signal otherwise. The additional term δP_{ref} 290 is norm-bounded by 2.5 %NP and changes about every 291 second. From the perspective of the reactor core, P_{turb} is 292 seen as a disturbance signal that has to be rejected by the 293 core control system. 294

3. Fixed-Structure Gain-Scheduled Controller

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The idea of gain scheduling is to break down the non-296 linear control problem into a finite number of linear sub-297 problems [19, 20]. The main advantage of this divide-and-298 conquer approach is to rely on well-known linear design 299 tools rather than convoluted nonlinear methods. This 300 is particularly interesting given the recent developments 301 in non-smooth optimization for structured robust con-302 trol [21]. 303

3.1. Linearization of the Nonlinear Model

The first step in designing the fixed-structure H_2/H_∞ 305 gain-scheduled controller is to linearize the nonlinear 306 model of the reactor core around several operating 307 points. All were determined using Framatome's certi-308 fied three-dimensional core kinetics computer code, named 309 SMART [22], starting from stationary conditions with 310 equilibrium xenon at 100 %NP. The computation of an op-311 erating point begins by setting the nuclear power of the re-312 actor core at a constant level. Then, the positions of P_{bank} 313 and H_{bank} are updated so that criticality is achieved with 314 minimal ACT and AO deviations. The boron concentra-315 tion of the coolant, on the other hand, is fixed according to 316 the fuel burn-up. In total, 21 operating points were com-317 puted over an evenly spaced grid of power values between 318 100 %NP and 50 %NP (i.e., one point every 2.5 %NP). 319 This number of points is sufficient to obtain a fine-grained 320 description of the original system and to ensure a smooth 321 transition of the gains. 322

Then, the operating points are uniquely defined by a 323 set of scheduling variables that covers the whole operating 324 range of the real plant. Initially, two different schedul-325 ing variables were considered: the position P_{bank} and the 326 known load profile P_{ref} . However, using both variables 327 is irrelevant here, as the closed-loop dynamic of P_{bank} 328 should be directly linked to P_{ref} . The question that re-329 mains is whether to schedule the controller with respect to 330 the internal variable P_{bank} or the exogenous signal P_{ref} . 331 Theoretically speaking, it would be preferable to choose 332 P_{ref} over P_{bank} since the impact of the hidden coupling 333 terms, which appear in the dynamic of the linearized gain-334 scheduled controller [20], is less pronounced when an ex-335 ternal signal is used for scheduling. Yet, in reality, the 336 behavior of the reactor core depends heavily on the position of the control rods. Therefore, just like in [8], P_{bank} is selected as the only scheduling variable.

Subsequently, a collection of 21 linear open-loop models 340 is numerically computed with the linearization algorithm 341 of MATLAB[®]/Simulink[®]. For all $i \in [1, 21]$, the i-th 342 linear open-loop model is obtained by linearizing the non-343 linear model of the plant around the operating point de-344 fined by $\sigma_i = P_{bank,eq}^{(i)}$. The iodine-xenon estimator has been discarded from the linearization process because its 345 346 dynamics are much slower than that of the controlled out-347 puts. Finally, the state-space realization of the i-th lin-348 earized open-loop model is given by: 349

$$\dot{x}_r(t) = A_i x_r(t) + B u(t) + E d(t)$$
$$y = C x_r(t),$$

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where $x_r \in \mathbb{R}^{25}$ is the truncated state vector (i.e., without iodine and xenon dynamics), $u = \begin{bmatrix} \dot{P}_{bank}, \dot{H}_{bank}, \dot{C}_{b,in} \end{bmatrix}^T \in \mathbb{R}^3$ are the control inputs, $d = P_{turb} \in \mathbb{R}$ is the exogenous disturbance, and $y = \begin{bmatrix} T_{in}, T_{out}, AO, P_{bank}, P_{core} \end{bmatrix}^T \in \mathbb{R}^5$ are the measured outputs. Note that the controlled outputs $z = \begin{bmatrix} T_a, AO, P_{bank} \end{bmatrix}^T \in \mathbb{R}^3$ are a subset of the measured output y in the sense that:

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$$z = Ty$$
 with $T = \begin{bmatrix} 1/2 & 1/2 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \end{bmatrix}$.

Since the control inputs u are the time derivatives of the state variables P_{bank} , H_{bank} and $C_{b,in}$, the open-loop state matrix A_i and the control matrix B can be decomposed as:

$$A_i = \begin{bmatrix} A_{1i} & A_{2i} \\ \mathbf{0} & \mathbf{0} \end{bmatrix}, \qquad B = \begin{bmatrix} \mathbf{0} \\ I_3 \end{bmatrix}$$

Thus, the poles of the i-th linearized open-loop model are 364 the union of the eigenvalues of A_{1i} plus three controllable 365 poles located at s = 0. For the chosen set of model param-366 eters, it was observed that every matrix A_{1i} is Hurtwitz 367 stable. This numerical assessment is consistent with real-368 ity insofar as the reactor core was designed to be stable 369 under normal conditions. Hence, it can be concluded that 370 the linear open-loop models are all stabilizable. 371

372 3.2. Architecture of the Controller

To achieve asymptotic tracking of a class of reference 373 inputs in the presence of another class of disturbances, a 374 model of both exogenous signals should be embedded in 375 the controller. Then, this model should be duplicated as 376 many times as the number of outputs to be controlled [23]. 377 Here, the reference inputs and the disturbance belong to 378 the same class of signals that are constant and piecewise 379 linear. Thus, the controller should comprise at least two 380 integrators per controlled output channel to achieve the 381 desired closed-loop performance. However, an integrator 382 is already included in every input-output transfer func-383 tion of the linearized open-loop models. Hence, only one 384



Fig. 2: Block diagram of the linear time-invariant closed-loop system computed at the i-th operating point

extra integrator needs to be added per controlled output channel. This can be realized with the following linear time-invariant output feedback controller:

$$\begin{split} \dot{x}_{I}(t) &= K_{I}\left(z_{ref}(t) - z(t)\right) \\ u(t) &= K_{P}\left(y_{ref}(t) - y(t)\right) + x_{I}(t), \end{split}$$

where $y_{ref} \in \mathbb{R}^5$ are the reference signals of the measured 389 outputs and $z_{ref} \in \mathbb{R}^3$ are the reference signals of the con-390 trolled outputs. This controller can be seen as a multiple-391 input multiple-output version of a Proportional-Integral 392 controller with gain matrices $K_P \in \mathbb{R}^{3 \times 5}$ and $K_I \in \mathbb{R}^{3 \times 3}$. 303 The proportional part of the controller is then filtered to 394 provide robustness against high-frequency noise and un-395 modeled dynamics: 396

$$\begin{split} \dot{x}_{I}(t) &= K_{I}(z_{ref}(t) - z(t)) \\ \dot{x}_{P}(t) &= \text{diag}\left(\frac{1}{a}, \frac{1}{a}, \frac{1}{b}\right) \left(-x_{P}(t) + K_{P}\left(y_{ref}(t) - y(t)\right)\right) \quad \text{397} \\ u(t) &= x_{P}(t) + x_{I}(t), \end{split}$$

where a > 0 and b > 0 are the time constants of the firstorder low-pass filters that are applied to x_P . The time constants of the first and second filters are chosen equal since the dynamic of P_{bank} is expected to be similar to that of H_{bank} . A block diagram of the linear time-invariant closed-loop system that is computed at the i-th operating point, $i \in [1, 21]$, is shown in Fig. 2, where:

$$\begin{aligned} G_{d \to y}^{(i)}(s) &= C \left(sI_{25} - A_i \right)^{-1} E \\ G_{u \to y}^{(i)}(s) &= C \left(sI_{25} - A_i \right)^{-1} B \\ C_y(s) &= \text{diag} \left(\frac{1}{as+1}, \frac{1}{as+1}, \frac{1}{bs+1} \right) K_P \\ C_z(s) &= \frac{1}{s} K_I, \end{aligned}$$

and $e = z - z_{ref}$ is the tracking error:

$$e = \left[\Delta T_a, \Delta AO, \Delta P_{bank}\right]^T. \tag{5}$$

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The signals $d_u \in \mathbb{R}^3$ and $d_y \in \mathbb{R}^5$ are unknown additive disturbances that may appear on the actuators and on the sensors of the plant. The input and the output sensitivity functions:

$$S_{u}^{(i)}(s) = \left[I_{5} + (C_{y}(s) + TC_{z}(s))G_{u \to y}^{(i)}(s)\right]^{-1}$$

$$S_{y}^{(i)}(s) = \left[I_{3} + G_{u \to y}^{(i)}(s)(C_{y}(s) + TC_{z}(s))\right]^{-1},$$
⁴¹²

are given by the closed-loop transfer functions from d_u to 413 u and from d_y to y respectively. 414

3.3. Tuning and Gain-Scheduling 415

The conventional approach [19, 20] for scheduling the 416 gains of the controller would be to: 1) synthesize a lin-417 ear time-invariant controller at each operating point using, 418 e.g., a fixed-structure H_2/H_{∞} synthesis algorithm and 2) 419 smoothly interpolate the gains of the controller as a func-420 tion of the scheduling variable P_{bank} . However, indepen-421 dently tuning the controller at each operating point may 422 lead to undesirable jumps in the values of the gains once 423 they have been interpolated. To prevent this, the gains can 424 be tuned against multiple neighboring models rather than 425 just one [8]. Here, the tuning and scheduling stages are 426 blended together using the Gain Surface Tuning method 427 presented in [13]. First, the gain matrices K_P , K_I and the 428 time constants a, b are written as quadratic polynomials 429 of the scheduling variable $\sigma = P_{bank}$: 430

$$K_P(\sigma) = K_{P0} + K_{P1}\sigma + K_{P2}\sigma^2$$
$$K_I(\sigma) = K_{I0} + K_{I1}\sigma + K_{I2}\sigma^2$$
$$a(\sigma) = a_0 + a_1\sigma + a_2\sigma^2$$
$$b(\sigma) = b_0 + b_1\sigma + b_2\sigma^2,$$

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with coefficients $K_{Pj} \in \mathbb{R}^{3 \times 5}$, $K_{Ij} \in \mathbb{R}^{3 \times 3}$, $a_j \in \mathbb{R}$ and $b_j \in \mathbb{R}$ for all $j \in \{1, 2, 3\}$. Then, these coefficients are 432 433 tuned to ensure proper closed-loop performance at each 434 operating point. Specifically, let $\mathbf{K} \in \mathbb{R}^{78}$ be the vector of 435 tunable parameters. For each linearized open-loop model, 436 the performance objectives are represented by the follow-437 ing frequency-domain criteria: 438

$$\begin{split} \min_{\mathbf{K}} & \left\| \operatorname{diag}\left(\beta_{1},\beta_{2},\beta_{3}\right) H_{z_{ref} \to e}^{(i)}(\mathbf{K},s) \frac{1}{s^{2}} \right\|_{2} \\ & \left\| W_{11} H_{d \to e_{1}}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| W_{12} H_{d \to e_{2}}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| W_{13} H_{d \to e_{3}}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| W_{21} H_{d \to u_{1}}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| W_{22} H_{d \to u_{2}}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| W_{23} H_{d \to u_{3}}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| \gamma H_{d_{u} \to u}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \left\| \gamma H_{d_{u} \to u}^{(i)}(\mathbf{K},s) \right\|_{\infty} \leq 1 \\ & \operatorname{Re}(p^{(i)}(\mathbf{K})) \leq -\alpha, \end{split}$$

where, for all $i \in [1, 21]$, $H_{u \to y}^{(i)}$ denotes the closed-loop 440 transfer function from input signal u to output signal y for 441 the i-th plant model, $\|\cdot\|_2$ and $\|\cdot\|_{\infty}$ are the H_2 and H_{∞} 442 norms, and $p^{(i)}$ are the closed-loop poles of the i-th plant 443 model. Having two integrators in the loop transfer func-444 tion means that $H_{z_{ref} \to e}^{(i)}$ has two zeros at s = 0. Hence, its 445

 H_2 norm can be shaped with a double integrator weighting 446 function which, in this context, is interpreted as a refer-447 ence signal generator. In fact, minimizing the frequency-448 weighted H_2 norm of $H_{z_{ref} \to e}^{(i)}$ ensures both asymptotic 449 tracking and disturbance rejection of step-like and ramp-450 like exogenous signals. The parameters $(\beta_1, \beta_2, \beta_3) \in \mathbb{R}^3_{>0}$ 451 are chosen such that ACT regulation takes precedence over 452 AO regulation which, in turn, takes precedence over P_{bank} 453 regulation. The scalar weights $(W_{11}, W_{12}, W_{13}) \in \mathbb{R}^3_{>0}$ and 454 $(W_{21}, W_{22}, W_{23}) \in \mathbb{R}^3_{>0}$ are used to limit the maximum set-455 point deviation and the maximum control effort that are 456 induced by a change in the power of the turbine. The rea-457 son why the controller is tuned against several single-input, 458 single-output H_{∞} constraints, rather than one multiple-459 input, multiple-output H_{∞} constraint, is to make the op-460 timization problem more readable. Finally, the parameters 461 $\gamma \in \mathbb{R}_{>0}$ and $\alpha \in \mathbb{R}_{>0}$ are selected to provide a sufficient 462 level of input-robustness and stability to the closed-loop 463 system.

4. Nonlinear Model Predictive Controller

The principle of model predictive control is to use a 466 model of the plant to predict and optimize its future be-467 havior by repeatedly solving an on-line optimal control 468 problem over a finite horizon [14, 15]. This control tech-469 nique, also known as receding horizon control, has proven 470 to be very successful in the process industry because it 471 can handle large scale, multi-input multi-output nonlin-472 ear systems that are subject to actuator and state con-473 straints [24]. 474

4.1. Model Reduction 475

The nonlinear state-space representation of the full-476 order PWR model can be written as: 477

$$\begin{aligned} x(t) &= f_0 \left(x(t), u(t), d(t) \right) \\ x(t_0) &= x_0, \end{aligned}$$

465

where $x \in \mathbb{R}^{37}$ is the full state vector, $x_0 \in \mathbb{R}^{37}$ is the initial 479 state, $u \in \mathbb{R}^3$ are the control inputs, and $d \in \mathbb{R}$ is the ex-480 ogenous disturbance. In the sequel, it is assumed that the 481 state is perfectly measured. Since the model is intended 482 to be used in a model predictive control framework, spe-483 cial attention is given to its computational complexity. In 484 practice, many numerical simulation runs are performed 485 to solve the optimal control problem at each time step. 486 However, the multiple time scale behavior of the model, 487 ranging from fast neutron to slow xenon dynamics, ren-488 ders it numerically stiff [25]. Although the full-order model 489 can be simulated using powerful integration solvers, such 490 as [26] or [27], model reduction is still conducted to make 491 the optimal control problem more tractable. State estima-492 tion, which is outside the scope of the paper, is another 493 motivation for reducing the order of the model. In the 494 future, a state observer should be designed to implement 495 the NMPC controller [28]. This task would obviously be 496

easier to achieve with a reduced-order model. Therefore, 497 singular perturbation theory [29] is employed to remove 498 the dynamics that are not essential regarding the control 499 objectives. A well-known approximation in nuclear reac-500 tor physics is the prompt-jump approximation [17, 18]. 501 Whenever a fission reaction occurs, two types of neutrons 502 are emitted: prompt neutrons, which are directly emit-503 ted from fission, and delayed neutrons, which are emit-504 ted a moment later from the radioactive decay of fission 505 products. Under normal operating conditions, the reac-506 tor core is predominantly governed by the dynamic of de-507 layed neutrons. Consequently, prompt neutron dynamics 508 can be reasonably approximated by steady-state equations 509 without any significant loss of accuracy [30]. The result-510 ing reduced-order model is given by a set of semi-explicit 511 differential-algebraic equations: 512

$$\begin{aligned} \dot{x}_d(t) &= f\left(x_d(t), x_a(t), u(t), d(t)\right) \\ & 0 &= g\left(x_d(t), x_a(t), u(t), d(t)\right) \\ & x_d(0) &= x_{d0}, \end{aligned}$$

where the state is divided into a slow differential part $x_d \in \mathbb{R}^{31}$ and a fast algebraic part $x_a \in \mathbb{R}^6$.

516 4.2. Formulation of the Optimal Control Problem

At each time step, the NMPC algorithm yields a se-517 quence of control inputs by solving a finite horizon optimal 518 control problem that is initialized with the current state of 519 the plant. Usually, only the first component of the control 520 input sequence is applied open-loop to the plant until a 521 new one is calculated at the next time step. The greatest 522 strength of NMPC is that all the previously defined core 523 control specifications can be directly incorporated into the 524 finite horizon optimal control problem: 525

$$\min_{\overline{u}(\cdot)} \int_{t}^{t+NT_{s}} \left(\left\| e\left(\overline{x}_{d}(\tau), \overline{x}_{a}(\tau), \overline{d}(\tau)\right) \right\|_{Q}^{2} + \left\|\overline{u}(\tau)\right\|_{R}^{2} \right) \mathrm{d}\tau$$

subject to:

526

$$\forall \tau \in [t, t + NT_s] \begin{cases} c\left(\overline{x}_d(\tau), \overline{x}_a(\tau), \overline{u}(\tau), d(\tau)\right) \leq 0\\ \dot{\overline{x}}_d(\tau) = f\left(\overline{x}_d(\tau), \overline{x}_a(\tau), \overline{u}(\tau), \overline{d}(\tau)\right)\\ 0 = g\left(\overline{x}_d(\tau), \overline{x}_a(\tau), \overline{u}(\tau), \overline{d}(\tau)\right)\\ \overline{x}_d(t) = x_d(t), \end{cases}$$
(6)

where $N \ge 1$ is the length of the prediction horizon, $T_s > 0$ 527 is the sampling time of the controller, and $x_d(t)$ is the dif-528 ferential state of the plant that is measured at time $t \ge 0$. 529 The core control specifications (1) - (4) are represented by 530 the inequality constraints map $c: \mathbb{R}^{31} \times \mathbb{R}^6 \times \mathbb{R}^3 \times \mathbb{R} \to \mathbb{R}^{16}$ 531 and the tracking error (5) is given by the nonlinear map 532 $e: \mathbb{R}^{31} \times \mathbb{R}^6 \times \mathbb{R} \to \mathbb{R}^3$. The weighted 2-norms $\|e\|_Q$ and 533 $\|u\|_R$ are defined by $\|e\|_Q^2 \coloneqq e^T Q e$ and $\|u\|_R^2 \coloneqq u^T R u$ where $Q \in \mathbb{R}^{3 \times 3}$ and $R \in \mathbb{R}^{3 \times 3}$ are symmetric positive-534 535 definite matrices. The bar subscript denotes internal 536

model variables that are used within the controller to predict the behavior of the real plant. In particular, the disturbance signal is written as $d(t) = \overline{d}(t) + \delta d(t)$ where \overline{d} is the known load profile that is sent in advance by the grid dispatcher to the plant and $|\delta d| \leq 2.5$ %NP is the unknown norm-bounded term that appears when frequency control is active.

Transcription of the continuous optimal control problem 544 into a finite-dimensional nonlinear programming problem 545 is commonly achieved using either direct single-shooting or 546 direct multiple-shooting [31, 32]. Here, both options will 547 be implemented and tested thereafter. In a direct method, 548 the continuous-time control inputs are approximated with 549 a finite number of parameters. Most of the time, the con-550 trol input sequence $\overline{\mathbf{u}} = \{\overline{u}[0], \dots, \overline{u}[N-1]\}$ is discretized 551 with a piecewise constant parameterization: 552

$$\forall k \in \llbracket 0, N-1
rbracket, \ \forall au \in [t_k, t_{k+1}], \ \overline{u}(au) = \overline{u}[k],$$
 553

where $t_k = t + kT_s$. The key difference between singleshooting and multiple-shooting is the parameterization of the state trajectory.

In single-shooting, the state trajectory is computed in one shot, from t_0 to t_N , outside of the optimization problem. In fact, for a given control input sequence $\overline{\mathbf{u}}$ and a fixed initial state $x_d(t)$, the state trajectory can be entirely determined by simulating the model of the plant:

$$\forall k \in \llbracket 0, N-1 \rrbracket, \ \begin{pmatrix} \overline{x}_d(t_{k+1}) \\ \overline{x}_a(t_{k+1}) \end{pmatrix} = \phi_{ss}\left(t_{k+1}; x_d(t), \overline{\mathbf{u}}\right),$$

where $\phi_{ss} : \mathbb{R} \times \mathbb{R}^{31} \times \mathbb{R}^{3 \times N} \to \mathbb{R}^{37}$ is a numerical integration operator that returns the solution of the initial value problem:

$$\forall \tau \in [t_0, t_N] \begin{cases} \bar{x}_d(\tau) = f\left(\bar{x}_d(\tau), \bar{x}_a(\tau), \bar{u}(\tau), d(\tau)\right) \\ 0 = g\left(\bar{x}_d(\tau), \bar{x}_a(\tau), \bar{u}(\tau), \bar{d}(\tau)\right) \\ \bar{x}_d(t_0) = x_d(t), \end{cases}$$

at time t_{k+1} . Therefore, the optimization problem can be reduced to: 567

$$\min_{\overline{\mathbf{u}}} \sum_{k=0}^{N-1} \left\| e\left(\phi_{ss}(t_{k+1}; x_d(t), \overline{\mathbf{u}}), \overline{d}(t_{k+1})\right) \right\|_Q^2 + \left\|\overline{u}[k]\right\|_R^2$$

569

subject to:

$$\forall k \in \llbracket 0, N-1 \rrbracket, \ c\left(\phi_{ss}(t_{k+1}; x_d(t), \overline{\mathbf{u}}), \overline{u}[k], \overline{d}(t_{k+1})\right) \leq 0$$

Hence, the advantage of using single-shooting is to remove 570 the equality constraints from the optimization problem. 571 Moreover, the state trajectory is always guaranteed to 572 be consistent with the dynamical equations of the model. 573 However, using single-shooting can increase the computa-574 tional complexity of the optimization problem because the 575 state equations are repeatedly propagated into both the 576 cost function and the constraints. This is especially true 577 for nonlinear and/or unstable systems.

In multiple-shooting, the state trajectory is cut into 579 N pieces which are independently computed over each 580 time interval $[t_k, t_{k+1}]$. These pieces are then reassembled within the optimization problem by imposing equality constraints on the differential and algebraic states:

$$\min_{\overline{\mathbf{x}}_{\mathbf{d}},\overline{\mathbf{x}}_{\mathbf{a}},\overline{\mathbf{u}}} \sum_{k=0}^{N-1} \left\| e\left(\overline{x}_{d}[k+1],\overline{x}_{a}[k],\overline{d}(t_{k+1})\right) \right\|_{Q}^{2} + \left\|\overline{u}[k]\right\|_{R}^{2}$$
subject to:

$$\forall k \in \llbracket 0, N-1 \rrbracket,$$

$$\begin{cases} c\left(\overline{x}_{d}[k+1],\overline{x}_{a}[k],\overline{u}[k],\overline{d}(t_{k+1})\right) \leq 0 \\ \left(\overline{x}_{d}[k+1]\right) - \phi_{ms}\left(t_{k+1};\overline{x}_{d}[k],\overline{u}[k]\right) = 0 \\ \overline{x}_{d}[0] - x_{d}(t) = 0, \end{cases}$$

where $\phi_{ms} : \mathbb{R} \times \mathbb{R}^{31} \times \mathbb{R}^6 \to \mathbb{R}^{37}$ is a numerical integration operator that returns the solution of the initial value problem:

$$\forall \tau \in [t_k, t_{k+1}] \begin{cases} \dot{\overline{x}}_d(\tau) = f\left(\overline{x}_d(\tau), \overline{x}_a(\tau), \overline{u}(\tau), \overline{d}(\tau)\right) \\ 0 = g\left(\overline{x}_d(\tau), \overline{x}_a(\tau), \overline{u}(\tau), \overline{d}(\tau)\right) \\ \overline{x}_d(t_k) = x_d(t_k), \end{cases}$$

at time t_{k+1} . The drawback of multiple-shooting is 589 that the dimension of the optimization problem is sub-590 stantially increased since the differential and algebraic 591 state sequences $\overline{\mathbf{x}}_{\mathbf{d}} = \{\overline{x}_d[0], \ldots, \overline{x}_d[N]\}$ and $\overline{\mathbf{x}}_{\mathbf{a}} =$ 592 $\{\overline{x}_a[0],\ldots,\overline{x}_a[N-1]\}$ become decision variables. How-593 ever, this is not a real issue because the structure of the 594 resulting nonlinear program is block-sparse, provided that 595 the decision variables are arranged as follows: 596 597

$$\overline{x}_{d}[0]^{T}, \overline{u}[0]^{T}, \overline{x}_{a}[0]^{T}, \dots,$$

$$\overline{x}_{d}[N-1]^{T}, \overline{u}[N-1]^{T}, \overline{x}_{a}[N-1]^{T}, \overline{x}_{d}[N]^{T})^{T}.$$

In fact, it has often been reported that efficient sparsity 601 exploitation can actually improve the rate of convergence 602 of the Newton-type algorithm which is used to solve the 603 nonlinear program [33]. Furthermore, having $\overline{\mathbf{x}}_{\mathbf{d}}$ and $\overline{\mathbf{x}}_{\mathbf{a}}$ as 604 decision variables is beneficial for initializing the problem 605 since a priori information about the state trajectory can 606 be supplied to the algorithm. Last but not least, multiple-607 shooting is better suited for highly nonlinear and/or unsta-608 ble plants because integration is broken down into short 609 time intervals, meaning that the nonlinearity is equally 610 distributed over the shooting nodes. 611

612 4.3. Practical Implementation Issues

The nonlinear program that is transcribed using single 613 shooting is composed of 3N decision variables and 16N614 inequality constraints. On the other hand, the nonlin-615 ear program that is transcribed using multiple-shooting is 616 composed of 40N+31 decision variables, 37N+31 equality 617 constraints and 16N inequality constraints. In both cases, 618 it is expected that the resolution time of the optimization 619 problem becomes greater than the duration T_s in between 620

two sampling instants because the length $N \ge 1$ of the 621 prediction horizon has to be large enough to ensure closed-622 loop stability. Yet, we still want to recalculate the control 623 inputs at each time step so the plant does not remain in 624 open-loop for too long. Therefore, the optimization rou-625 tine is deliberately terminated once the resolution time ex-626 ceeds T_s seconds. Even if the solution is sub-optimal [34], 627 we believe that it should have sufficiently converged to 628 provide an acceptable level of closed-loop performance. 629

Consequently, a prediction-based delay compensation 630 scheme [35] is incorporated into the NMPC algorithm to 631 account for the non-negligible resolution time $0 < \tau \leq T_s$ 632 of the optimization problem. To illustrate, let $\mu_N(x_d(t))$ 633 be the feedback value given by the (sub-optimal) control 634 input sequence $\overline{\mathbf{u}}^*(x_d(t))$ that (almost) solves the optimal 635 control problem with initial state $x_d(t)$. Because of the 636 computational delay, $\mu_N(x_d(t))$ is applied to the plant at 637 time $t+\tau$ instead of time t. Thus, the greater the mismatch 638 between $x_d(t)$ and $x_d(t+\tau)$, the greater the difference be-639 tween the expected and the real behavior of the closed-loop 640 system. Knowing the maximum resolution time $\tau_{max} = T_s$ 641 of the optimization problem, the idea of [35] is to solve 642 the optimal control problem in advance, starting from an 643 estimate $\overline{x}_d(t+T_s)$ of the future state in place of the cur-644 rently measured state $x_d(t)$. The resulting feedback value 645 $\mu_N(\overline{x}_d(t+T_s))$, available at time $t+\tau$, is then applied 646 to the plant accordingly at time $t + T_s$. The next state 647 estimate can be computed straightforwardly by simulat-648 ing the model of the plant from the newly measured state 649 $x_d(t+T_s)$ with input $\mu_N(\overline{x}_d(t+T_s))$. Note that, since the 650 state is estimated in an open-loop fashion, it is almost cer-651 tain that the real state $x_d(t+T_s)$ will differ from $\overline{x}_d(t+T_s)$. 652 However, it is still relevant to use the delay-compensated 653 input, as some degree of robustness is expected against 654 estimation errors [36]. 655

5. Simulation Results





656

657

Fig. 3: Normalized power of the turbine with frequency control

To assess and compare the performance of the two core control systems, a typical 8-hour load-following 656

scenario is conducted on PWRSimu, an intermediate 660 complexity PWR simulator developed by Framatome in 661 MATLAB^(R)/Simulink^(R). As shown in Fig. 3, this sce-662 nario involves two power variations (from 100 %NP to 663 50 %NP and vice versa) that are carried out at a rate of 664 5 %NP/min with frequency control. The full-order non-665 linear PWR model is simulated on Simulink with the stiff 666 adaptive-step integration solver ode15s [27]. 667

The fixed-structure gain-scheduled controller has been tuned with the following set of parameters:

$$\begin{array}{cccc} \beta_1 = 10/1.5 & \beta_2 = 5/5 & \beta_3 = 2/30 \\ W_{11} = 10/1.5 & W_{12} = 10/5 & W_{13} = 10/30 \\ W_{21} = 10/1.25 & W_{22} = 10/1.25 & W_{23} = 10/0.04 \\ \gamma = 0.5 & \alpha = 0.0001, \end{array}$$

using Systume (MATLAB Control System Toolbox¹⁸) [21]. In order to model the whole closed-loop system in Simulink, the quadratic polynomials $K_P(\sigma)$, $K_I(\sigma)$, $a(\sigma)$ and $b(\sigma)$ are approximated by feeding the scheduling variable $\sigma = P_{bank}$ into 1-D lookup table blocks with linear interpolation and clip extrapolation options.

Tuning of the NMPC controller has been achieved in two steps. First, the matrices Q and R of the cost function (6) are selected diagonal with normalized weights:

$$q_{1} = \frac{\varepsilon_{1}}{(e_{1,max})^{2}} \qquad q_{2} = \frac{\varepsilon_{2}}{(e_{2,max})^{2}} \qquad q_{3} = \frac{\varepsilon_{3}}{(e_{3,max})^{2}}$$
$$r_{1} = \frac{\nu_{1}}{(u_{1,max})^{2}} \qquad r_{2} = \frac{\nu_{2}}{(u_{2,max})^{2}} \qquad r_{3} = \frac{\nu_{3}}{(u_{3,max})^{2}},$$

680

where, for all $i \in \{1, 2, 3\}$, $e_{i,max}$ and $u_{i,max}$ denote the maximum acceptable values of e_i and u_i (see Bryson's rule [37]). Then, the tunable parameters $(\varepsilon_1, \varepsilon_2, \varepsilon_3) \in \mathbb{R}^3_{>0}$ and $(\nu_1, \nu_2, \nu_3) \in \mathbb{R}^3_{>0}$ are used to penalize, in order of importance, the deviations of ACT, AO and P_{bank} without unduly restricting the control effort. In the end, the weighting matrices are given by:

688
$$Q = \operatorname{diag}\left(\frac{30}{1.5^2}, \frac{20}{5^2}, \frac{15}{30^2}\right), R = \operatorname{diag}\left(\frac{0.01}{1.25^2}, \frac{0.01}{1.25^2}, \frac{0.01}{0.04^2}\right)$$

The sampling time $T_s = 60$ s and the prediction horizon 689 N = 10 are chosen so as to balance on-line complexity 690 and closed-loop stability. The NMPC algorithm is im-691 plemented in MATLAB R2019a with CasADi v3.5.5 [38]. 692 This open-source software provides several building blocks 693 which are especially useful for solving large-scale nonlinear 694 programing problems efficiently. Practically speaking, the 695 key asset of CasADi is its ability to quickly compute Jaco-696 bians and exploit their sparsity pattern using algorithmic 697 differentiation and graph coloring techniques. Within this 698 framework, the reduced-order PWR model is simulated 699 with the stiff adaptive-step integration solver IDAS [26] 700 and the optimization problem is solved with the primal-701 dual interior point solver IPOPT [39] using an Intel[®] Core^{π} 702 i3-6100U processor with 16 GB of RAM. The closed-loop 703 interactions between the controller and the plant are rep-704 resented by the sequential execution of both MATLAB® 705 and Simulink[®] files. 706

 10^{2} Final value of the cost function 10 10 10 10^{-2} 10⁻³ 10 10⁻⁵ 3 5 0 2 4 6 7 8 Time (hours)

Fig. 4: Semi-log plot of the final value of the cost function (NMPC single-shooting vs NMPC multiple-shooting)

In order to choose between single-shooting and multiple-708 shooting, both transcription methods were implemented 709 and tested. As can be seen in Fig. A.7 and A.9, the 710 ACT deviation and the position deviation of P_{bank} are 711 almost identical for both methods. However, the AO 712 deviation, shown in Fig. A.8, is significantly smaller in 713 multiple-shooting because the behavior of H_{bank} , shown 714 in Fig. A.11, is less erratic than in single-shooting. The 715 same phenomenon can be observed, but to a lesser extent, 716 for P_{bank} in Fig. A.10 and for the boron concentration in 717 Fig. A.12. The reason behind this, as Fig. 4 demonstrates, 718 is that the optimization problem converges more quickly 719 to the solution when transcribed in multiple-shooting. In 720 fact, it turned out that further increasing the resolution 721 time in multiple-shooting had no effect on the final value 722 of the cost function. This suggests that the optimal solu-723 tion can always be found within the allocated time, which 724 is why multiple-shooting is selected over single-shooting. 725

Overall, the performance of the NMPC controller is bet-726 ter than that of the gain-scheduled controller. As can be 727 seen in Fig. A.13, A.14 and A.15, both controllers man-728 age to keep the ACT, the AO and the position of P_{bank} 729 close to their reference signals. Yet, as can be observed in 730 Fig. A.16, A.17 and A.18, the gain-scheduled controller 731 tends to overuse the actuators in comparison with the 732 NMPC controller. This shows that the bandwidth of the 733 controller does not need to be particularly large even when

Table 1: Comparison between the gain-scheduled controller and the nonlinear model predictive controller

GSC	NMPC
Several linear models	One nonlinear model
Frequency domain design	Time domain design
Output-feedback control	State-feedback control
Robust design	Nominal design
Pre-computed gains	Online calculations
Continuous-time control	Discrete-time control
No anticipation	Preventive actions



Fig. 5: Fixed-structure gain-scheduled control approach

frequency control is active. However, the NMPC controller 735 may struggle to reject other sudden and unexpected dis-736 turbances, as the control inputs are only updated every 737 minute. To give a concrete example, note that the prin-738 cipal effect of frequency control is to cause ACT fluctu-739 ations of about ± 0.8 °C. A closer look at Fig. A.13 re-740 veals that these fluctuations cannot be properly mitigated 741 by the NMPC controller because they are unpredictable 742 and of relatively high frequency. The gain-scheduled con-743 troller, on the other hand, gives better results since it pro-744 vides continuous feedback to the plant and was designed 745 using robust control methods. As a matter of fact, ro-746 bustness is still an open issue in the field of model predic-747 tive control [15]. Therefore, it should be mentioned that 748 the reduced-order PWR model, which is embedded in the 749 NMPC controller, derives from the one that is used for 750 validation. Hence, it is not surprising that the NMPC 751 controller performs better than the gain-scheduled con-752 troller, which was designed off-line based on simpler linear 753 models. This is both a strength and a weakness of non-754 linear model predictive control. While it is indeed better 755 to use a very detailed and accurate model for prediction, 756 it can also become an issue if the computation time of 757 the NMPC control law gets too long. Worse, using an ex-758 cessively complex model can even make the optimization 759 problem intractable. Besides, it is also good to recall that 760 the NMPC controller is a state-feedback controller whereas 761 the gain-scheduled controller is an output-feedback one. 762



Fig. 6: Nonlinear model predictive control approach

Unfortunately, some state variables, such as the xenon concentration of the reactor core, cannot be physically measured. Thus, the NMPC controller cannot be practically implemented unless a nonlinear state observer is designed. This not only increases the overall complexity of the core control system but also raises additional concerns about its robustness.

6. Conclusion

In this paper, two advanced core control systems have 771 been designed for flexible operation of future PWRs. The 772 first one is a fixed-structure H_2/H_{∞} gain-scheduled con-773 troller and the second one is an NMPC controller. The 774 design methodologies of both controllers have been de-775 tailed and their performances have been compared on the 776 intermediate complexity pressurized water reactor simula-777 tor of Framatome, named PWRSimu. The gain-scheduled 778 controller is well adapted to cope with unexpected distur-779 bances and model uncertainties. Moreover, its architec-780 ture is readable and easy to implement. However, some 781 improvements should be made to avoid overusing the ac-782 tuators (a dead-zone nonlinearity could be taken into ac-783 count in the design of the controller using, e.g., the circle 784 criterion). By contrast, the NMPC controller can handle a 785 wide range of constraints and can easily deal with distur-786 bances that are known in advance. Its predictive capabil-787 ities make it possible to achieve good performance while 788 limiting the control effort. However, it is very difficult to 789 analyze its robustness, especially when used together with 790 a state-observer. In future work, the controllers will be 791 combined in a hierarchical manner to overcome the lim-792 itations of both design strategies. The NMPC controller 793 could be located at an upper level to provide a feedforward 794 action to the plant as well as reference output trajectories 795 to the gain-scheduled controller. Then, the gain-scheduled 796 controller could be used at a lower level to reject unknown 797 disturbances and to ensure that the closed-loop system 798 stays close to the reference trajectory despite model un-799 certainties. This hierarchical scheme is also motivated by 800

the multiple time scales of the plant. In fact, the upper 801 layer could anticipate future load variations and forecast 802 the evolution of xenon concentration while the lower layer 803 could simply focus on disturbance rejection and setpoint 804 tracking. It would also make sense to leave the calcula-805 tion of boron concentration adjustments to the upper layer 806 since the evolution of the boron concentration is substan-807 tially slower than the movements of the control rods. 808

809 Conflict of interest

810 None declared.

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814 Appendix A. Simulation Results



Fig. A.7: Average coolant temperature deviation (NMPC singleshooting vs NMPC multiple-shooting)



Fig. A.8: Axial offset deviation (NMPC single-shooting vs NMPC multiple-shooting)



Fig. A.9: Position deviation of ${\rm P}_{\rm bank}$ (NMPC single-shooting vs NMPC multiple-shooting)



Fig. A.10: Position of $\mathbf{P}_{\mathrm{bank}}$ (NMPC single-shooting vs NMPC multiple-shooting)



Fig. A.11: Position of ${\rm H}_{\rm bank}$ (NMPC single-shooting vs NMPC multiple-shooting)



Fig. A.12: Boron concentration of the primary coolant (NMPC single-shooting vs NMPC multiple-shooting)



Fig. A.15: Position deviation of P_{bank} (Gain-Scheduled Controller vs NMPC multiple-shooting)



Fig. A.13: Average coolant temperature deviation (Gain-Scheduled Controller vs NMPC multiple-shooting)



Fig. A.16: Position of $\mathbf{P}_{\mathrm{bank}}$ (Gain-Scheduled Controller vs NMPC multiple-shooting)



Fig. A.14: Axial offset deviation (Gain-Scheduled Controller vs NMPC multiple-shooting)



Fig. A.17: Position of ${\rm H}_{\rm bank}$ (Gain-Scheduled Controller vs NMPC multiple-shooting)



Fig. A.18: Boron concentration of the primary coolant (Gain-Scheduled Controller vs NMPC multiple-shooting)

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