

Transient Stabilization of a Synchronous Generator in a Multimachine Power Network

Gilney Damm, Françoise Lamnabhi-Lagarrigue, Riccardo Marino, Cristiano Maria Verrelli

To cite this version:

Gilney Damm, Françoise Lamnabhi-Lagarrigue, Riccardo Marino, Cristiano Maria Verrelli. Transient Stabilization of a Synchronous Generator in a Multimachine Power Network. Françoise Lamnabhi-Lagarrigue; Salah Laghrouche; Antonio Loria; Elena Panteley. Taming Heterogeneity and Complexity of Embedded Control, Wiley, pp.211-225, 2013, 9781905209651. 10.1002/9780470612217.ch13. hal-03471117

HAL Id: hal-03471117 <https://hal.science/hal-03471117>

Submitted on 3 Jan 2022

HAL is a multi-disciplinary open access archive for the deposit and dissemination of scientific research documents, whether they are published or not. The documents may come from teaching and research institutions in France or abroad, or from public or private research centers.

L'archive ouverte pluridisciplinaire **HAL**, est destinée au dépôt et à la diffusion de documents scientifiques de niveau recherche, publiés ou non, émanant des établissements d'enseignement et de recherche français ou étrangers, des laboratoires publics ou privés.

Transient Stabilization of a Synchronous Generator in a Multimachine Power Network

GILNEY DAMM Laboratoire IBISC-CNRS/Université d'Evry Val d'Essonne Laboratoire des Signaux et Systèmes, L2S-CNRS 40 rue du Pelvoux 91020 Evry Cedex FRANCE gilney.damm@ibisc.fr FRANÇOISE LAMNABHI-LAGARRIGUE Plateau de Moulon, Gif sur Yvette 91192 FRANCE lamnabhi@lss.supelec.fr RICCARDO MARINO, CRISTIANO MARIA VERRELLI Universita di Roma Tor Vergata,Dipartimento di Ingegneria Elettronica ´ Via del Politecnico 1, 00133 Rome ITALY

(marino,verrelli)@ing.uniroma2.it

Abstract: The transient stabilization problem of a synchronous generator in a multimachine power system is addressed. A robust adaptive nonlinear feedback control algorithm is designed on the basis of a third order model of the synchronous machine in a multimachine power network which takes into account the effect of transfer conductances and of the remote network dynamics: it does not require the machine internal voltage measurement and relies on the knowledge of only three model parameters and of upper and lower bounds on the uncertain ones. Sufficient conditions to be satisfied by the remaining part of the network for guaranteeing \mathcal{L}_2 and \mathcal{L}_{∞} disturbance attenuation and relative speed regulation are derived which are more general than those required by the single machine-infinite bus approximation.

Key–Words: Power system control, robust nonlinear control, adaptive control, \mathcal{L}_2 and \mathcal{L}_{∞} disturbance attenuation.

1 Introduction

The transient stabilization problem of power networks, which are large-scale interconnected nonlinear systems, consists in the design of a suitable excitation feedback control keeping each generator close to the synchronous speed when mechanical and electrical perturbations, such as load shedding, generation tripping or short circuits, occur. The standard linear controllers which are designed on the basis of linear approximations around operating conditions may not be able to handle the severe disturbances and contingencies typically occurring in power systems so that, in the recent years, several nonlinear control algorithms have been proposed, leading to significant advances in power system control ([6], [8], [11], [12], [4], [9]).

In this paper we extend the analysis presented in [9] by taking into account the effects of transfer conductances and of remote network dynamics: the resulting control for the transient stabilization of a synchronous generator in a multimachine power network, does not assume the availability of the machine internal voltage measurement as well as the knowledge of the model parameters excepting for machine damping and inertia constants and synchronous speed. \mathcal{L}_2 and \mathcal{L}_{∞}

disturbance attenuation and relative speed regulation are guaranteed under certain assumptions on the network dynamics which are more general than those required by the single machine-infinite bus approximation. Simulation results with reference to a 3-machine, 9-bus system illustrate the closed loop performance.

2 Problem Statement and Nonlinear Control Design

A large-scale power system consisting of n generators interconnected through a transmission network is described by the nonlinear third order model in [10] $[1 < i < n]$

$$
\dot{\delta}_{i} = \omega_{i}
$$
\n
$$
\dot{\omega}_{i} = -\frac{D_{i}}{2H_{i}}\omega_{i} + \frac{\omega_{0}}{2H_{i}}P_{mi} - \frac{\omega_{0}}{2H_{i}}P_{ei}
$$
\n
$$
\dot{E}'_{qi} = \frac{k_{ci}}{T'_{d0i}}u_{fi} - \frac{E'_{qi}}{T'_{d0i}} - \frac{(x_{di} - x'_{di})}{T'_{d0i}}I_{di}
$$
\n
$$
P_{ei} = E'^{2}_{qi}G_{ii} + E'_{qi} \sum_{j=1, j \neq i}^{n} \left[E'_{qj}G_{ij}\cos(\delta_{ij}) + E'_{qj}B_{ij}\sin(\delta_{ij})\right]
$$
\n(1)

$$
Q_{ei} = -E'_{qi}^{2} B_{ii} + E'_{qi} \sum_{j=1, j\neq i}^{n} \left[E'_{qj} G_{ij} \sin (\delta_{ij}) \right]
$$

\n
$$
-E'_{qj} B_{ij} \cos (\delta_{ij}) \right] - x'_{di} \left(I_{di}^{2} + I_{qi}^{2} \right)
$$

\n
$$
I_{di} = -E'_{qi} B_{ii} + \sum_{j=1, j\neq i}^{n} \left[E'_{qj} G_{ij} \sin (\delta_{ij}) \right]
$$

\n
$$
-E'_{qj} B_{ij} \cos (\delta_{ij}) \right]
$$

\n
$$
I_{qi} = E'_{qi} G_{ii} + \sum_{j=1, j\neq i}^{n} \left[E'_{qj} G_{ij} \cos (\delta_{ij}) \right]
$$

\n
$$
+ E'_{qj} B_{ij} \sin (\delta_{ij}) \right]
$$

\n
$$
\delta_{ij} = \delta_{i} - \delta_{j}
$$

\n
$$
V_{ti} = \sqrt{v_{di}^{2} + v_{qi}^{2}} = \sqrt{(x'_{di} I_{qi})^{2} + (E'_{qi} - x'_{di} I_{di})^{2}}
$$

in which (for the *i*-th generator): δ_i (rad) is the power angle, ω_i (rad/s) is the relative angular speed, $E'_{qi}(\text{p.u.})$ is the transient EMF in the quadrature axis, $P_{ei}(\text{p.u.})$ is the active electrical power, $P_{mi}(p.u.)$ is the mechanical input power, Q_{ei} (p.u.) is the reactive electrical power, $I_{di}(\mathbf{p.u.})$ is the direct axis current, $I_{qi}(\mathbf{p.u.})$ is the quadrature axis current, $V_{ti}(\text{p.u.})$ is the terminal voltage, u_{fi} (p.u.) is the input to the SCR amplifier, ω_0 (rad/s) is the synchronous speed, D_i (p.u.) is the damping constant, $H_i(s)$ is the inertia constant, $T'_{d0i}(s)$ is the direct axis transient open circuit time constant, $x_{di}(\text{p.u.})$ is the direct axis reactance, k_{ci} is the gain of the excitation amplifier, $G_{ij}(\text{p.u.})$ and B_{ij} (p.u.) are the *i*-th row and the *j*-th column element of nodal conductance and susceptance matrixes, respectively, at the internal nodes after eliminating all physical buses, which depend on the direct axis transient reactance $x'_{di}(\text{p.u.})$, on the transformer reactance x_{Ti} (p.u.), on the loads and on the transmission line reactance x_{ij} (p.u.) between the *i*-th generator and the *j*-th generator. Direct computation of $P_{ei}(t)$ time derivative $[G_{ij}$ and B_{ij} , $1 \leq j \leq n$, are assumed to be constant] allows us to rewrite the dynamic model (1) as $[I_{qi} \ge c_{Ii} > 0$ ([7])]

$$
\dot{\delta}_i = \omega_i
$$
\n
$$
\dot{\omega}_i = -\frac{D_i}{2H_i}\omega_i + \frac{\omega_0}{2H_i}P_{mi} - \frac{\omega_0}{2H_i}P_{ei} \tag{2}
$$
\n
$$
\dot{\Sigma}_i = k_{ci} \left[I_{qi}^2 + P_{ei}G_{ii}\right] \qquad P_{ei} \qquad G_{ii} P_{ei}^2
$$

$$
\dot{P}_{ei} = \frac{k_{ci}}{T'_{d0i}} \left[\frac{I_{qi}^2 + P_{ei} G_{ii}}{I_{qi}} \right] u_{fi} - \frac{P_{ei}}{T'_{d0i}} - \frac{G_{ii}}{T'_{d0i}} \frac{P_{ei}^2}{I_{qi}^2} \n- \frac{(x_{di} - x'_{di})}{T'_{d0i}} I_{di} I_{qi} - \frac{G_{ii}(x_{di} - x'_{di})}{T'_{d0i}} \frac{I_{di} P_{ei}}{I_{qi}} \n- Q_{ei} \omega_i - B_{ii} \frac{P_{ei}^2}{I_{qi}^2} \omega_i - x'_{di} \left(I_{di}^2 + I_{qi}^2 \right) \omega_i \n+ \frac{P_{ei}}{I_{qi}} R_i
$$

in which δ_i , ω_i , P_{ei} are the state variables, u_{fi} is the control input and the term

$$
R_i = \sum_{j=1, j \neq i}^{n} \left\{ \dot{E}_{qj}^{\prime} \left[G_{ij} \cos \left(\delta_{ij} \right) + B_{ij} \sin \left(\delta_{ij} \right) \right] + \omega_j \left[E_{qj}^{\prime} G_{ij} \sin \left(\delta_{ij} \right) - E_{qj}^{\prime} B_{ij} \cos \left(\delta_{ij} \right) \right] \right\}
$$

represents the effect of the network remote dynamics on the *i*-th generator.

In this paper, we restrict our analysis to a particular power system PS , which consists of a group G_M of $n-1$ generators [tied together by a strong network of transmission lines] which is linked to a single generator g_m (referred to as r-th generator) by a comparatively weak set of tie lines: the transient stabilization problem of the synchronous generator g_m in the multimachine power network PS is addressed. Let us state the theoretical result of the paper (the proof is omitted).

Theorem*: Consider the* r*-th generator dynamics. Denote by* δ_{rs} *the pre-fault constant value for the power angle* δ_r *and assume that for all* $t \geq 0$

- *i)* for each $j \neq r$, $\delta_j(t)$ and $E'_{qj}(t)$ are continu*ous functions of time t and boundedness of* $\delta_r(t)$ *,* $\omega_r(t)$, $P_{er}(t)$ implies boundedness of $E'_{qj}(t)$;
- \dot{u})_{a)} there exist μ_r , ν_r , ρ_r non-negative reals, $\psi_{\mu r}(\cdot)$, $\psi_{\nu r}(\cdot)$, $\psi_{\rho r}(\cdot)$ K_{∞} *functions and* $g_r(t)$ *bounded non-negative real-valued function of time* t *such that*

$$
|R_r(t)| \le \sup_{0 \le \tau \le t} \{g_r(\tau)\} + \mu_r \psi_{\mu r} \Big(\max_{0 \le \tau \le t} \{ |\delta_r(\tau) - \delta_{rs}| \} \Big) + \nu_r \psi_{\nu r} \Big(\max_{0 \le \tau \le t} \{ |\omega_r(\tau)| \} \Big) + \rho_r \psi_{\rho r} \Big(\max_{0 \le \tau \le t} \{ |P_{er}(\tau) - P_{mr}(\tau)| \} \Big);
$$

 $\delta\delta_{lb}$ there exist α_{lr} known non-negative reals and $\varphi_{lr}(\cdot)$ *known* K_{∞} *functions such that for* $l =$ μ, ν, ρ

$$
\psi_{lr}(\cdot) \leq \alpha_{lr} + \varphi_{lr}(\cdot) \doteq \bar{\varphi}_{lr}(\cdot)
$$

- *iii*) $P_{er}(t) \ge c_{Pr} > 0$;
- *iv) the unknown (possibly time-varying) parame*ters $T'_{d0r}(t)$, $x_{dr}(t)$, $x'_{dr}(t)$, $k_{cr}(t)$ and the un*known constant parameter* Grr *are within the corresponding known positive physical bounds* (T'_{d0rm}, T'_{d0rM}) , (x_{drm}, x_{drM}) , (x'_{drm}, x'_{drM}) , (k_{crm}, k_{crM}) , (G_{rrm}, G_{rrM}) ;
- *v*) the unknown constant parameter B_{rr} is within the *corresponding known physical bounds* (Brrm*,* B_{rrM});

vi) $|\dot{P}_{mr}(t)| \leq \dot{P}_{Mr}$ *with known* \dot{P}_{Mr} .

Then the nonlinear adaptive feedback control algorithm $[k_{\delta r}, k_{\omega r}, k_{pr}, k_{per}, k_{\omega pr}, k_r, k_{Rr}, \beta_{jr}, \beta_{xr},$ $\beta_{Br}, \varepsilon_{jr}, \varepsilon_{xr}, \varepsilon_{Br}$ ($1 \leq j \leq 6$) are positive control *parameters]*

$$
u_{fr} = \frac{I_{qr}}{\left(\hat{\theta}_{5r}I_{qr}^2 + \hat{\theta}_{6r}P_{er}\right)} v_r
$$

$$
-\frac{k_r I_{qr}\left(2I_{qr}^4 + 2P_{er}^2\right)}{4\left(\hat{\theta}_{5r}I_{qr}^2 + \hat{\theta}_{6r}P_{er}\right)^2\left(\theta_{5rm}I_{qr}^2 + \theta_{6rm}P_{er}\right)} \cdot v_r^2 (P_{er} - P_{er}^*)}.
$$

$$
v_r = -\frac{5}{4}k_{pr}(P_{er} - P_{er}^*) + \frac{\omega_0}{2H_r}(\omega_r - \omega_r^*)
$$

\n
$$
-\frac{k_{Rr}}{4}\frac{P_{er}^2}{I_{qr}^2}(P_{er} - P_{er}^*) + \frac{5D_rk_{\delta r}}{4\omega_0}\omega_r + \frac{2H_r}{\omega_0}\omega_r
$$

\n
$$
-\frac{k_r}{4}(P_{er} - P_{er}^*)\left[P_{er}^2 + I_{dr}^2I_{qr}^2 + \frac{I_{dr}^2P_{er}^2}{I_{qr}^2}\right]
$$

\n
$$
+\frac{P_{er}^4}{I_{qr}^4}\left(1 + \omega_r^2\right) + \omega_r^2\left(I_{dr}^2 + I_{qr}^2\right)^2
$$

\n
$$
-\frac{k_r}{4}\left(\frac{5}{4}k_{\omega r} + \frac{5}{4}k_{\delta r} + \frac{5}{4}k_{per}
$$

\n
$$
+\frac{k_r}{4} + \frac{1}{k_r} + \frac{\omega_0^2}{16H_r^2}k_{\omega pr}
$$

\n
$$
+\frac{1}{k_{\omega pr}}\right)^2(P_{er} - P_{er}^*) + \hat{\theta}_{1r}P_{er} + \hat{\theta}_{2r}I_{dr}I_{qr}
$$

\n
$$
+\hat{\theta}_{3r}\frac{I_{dr}P_{er}}{I_{qr}} + \hat{\theta}_{4r}\frac{P_{er}^2}{I_{qr}^2} + \hat{B}_{rr}\frac{P_{er}^2}{I_{qr}^2}\omega_r
$$

\n
$$
+\hat{x}_{dr}'\left(I_{dr}^2 + I_{qr}^2\right)\omega_r + Q_{er}\omega_r - \frac{5k_{\delta r}}{4\omega_0}\left[D_r\omega_r
$$

\n
$$
-\omega_0(\hat{P}_{mr} - P_{er})\right]
$$

\n
$$
-\frac{2H_r}{\omega_0}\left(\frac{5}{4}k_{\omega r} + \frac{1}{k_{\omega pr}}\right)\left(\frac{5}{4}k_{\omega r}
$$

\n
$$
+\frac{D_r}{2H_r}\right)(\omega_r - \omega_r^*) + (\delta_r - \delta_{rs}) + \frac{\omega_0}{2H_r}(P_{er
$$

$$
+\bar{\varphi}_{\rho r}^{2}\left(\max_{0\leq\tau\leq t}\left\{[P_{er}(\tau)-P_{er}^{*}(\tau)]\right\}\right)
$$

$$
+P_{mrM}-P_{mrm}+\frac{\dot{P}_{Mr}}{k_{per}}+\frac{2H_{r}}{\omega_{0}}\left(\frac{5}{4}k_{\omega r}\right)
$$

$$
+\frac{1}{k_{\omega pr}}+\frac{5}{4}k_{\delta r}\right)\max_{0\leq\tau\leq t}\left\{[\omega_{r}(\tau)-\omega_{r}^{*}(\tau)]\right\}
$$

$$
+\left[\frac{2H_{r}}{\omega_{0}}\left(1+\frac{25}{16}k_{\delta r}^{2}\right)+\frac{5D_{r}k_{\delta r}}{4\omega_{0}}\right]\cdot
$$

$$
\cdot\max_{0\leq\tau\leq t}\left\{\left[\delta_{r}(\tau)-\delta_{rs}\right]\right\}+1\right]
$$

$$
\omega_{r}^{*} = -\frac{5}{4}k_{\delta r}(\delta_{r}-\delta_{rs})
$$
(3)
$$
P_{er}^{*} = \frac{2H_{r}}{\omega_{0}}\left[\frac{5}{4}k_{\omega r}(\omega_{r}-\omega_{r}^{*})+\frac{5}{4}k_{\delta r}\omega_{r}
$$

$$
-\frac{D_{r}}{2H_{r}}\omega_{r}^{*}+(\delta_{r}-\delta_{rs})
$$

$$
+\frac{1}{k_{\omega pr}}(\omega_{r}-\omega_{r}^{*})\right]+\hat{P}_{mr}
$$

 ω_r^*

$$
\hat{P}_{mr} = \phi_r + \frac{2H_r}{\omega_0} \left(\frac{5}{4} k_{per} + \frac{k_r}{4} + \frac{1}{k_r} \right)
$$
\n
$$
+ \frac{\omega_0^2}{16H_r^2} k_{\omega pr} \right) \omega_r
$$
\n
$$
\dot{\phi}_r = \left(\frac{5}{4} k_{per} + \frac{k_r}{4} + \frac{1}{k_r} + \frac{\omega_0^2}{16H_r^2} k_{\omega pr} \right) \left[-\phi_r \right.
$$
\n
$$
+ \frac{D_r}{\omega_0} \omega_r + P_{er} - \frac{2H_r}{\omega_0} \left(\frac{5}{4} k_{per} + \frac{k_r}{4} \right)
$$
\n
$$
+ \frac{1}{k_r} + \frac{\omega_0^2}{16H_r^2} k_{\omega pr} \right) \omega_r
$$
\n
$$
\phi_r(0) = \hat{P}_{mr}(0) - \frac{2H_r}{\omega_0} \left(\frac{5}{4} k_{per} + \frac{k_r}{4} + \frac{1}{k_r} \right)
$$
\n
$$
+ \frac{\omega_0^2}{16H_r^2} k_{\omega pr} \right) \omega_r(0),
$$
\n
$$
P_{mrm} \leq \hat{P}_{mr}(0) \leq P_{mrM}
$$
\n
$$
\dot{\hat{\theta}}_{1r} = \text{Proj} \left[-\frac{P_{er}(P_{er} - P_{er}^*)}{\beta_{1r}}, \hat{\theta}_{1r}, \theta_{1rm},
$$

 $\Big], \theta_{1rm} \leq \hat{\theta}_{1r}(0) \leq \theta_{1rM}$

 $\Big], \ \theta_{2rm} \leq \hat{\theta}_{2r}(0) \leq \theta_{2rM}$

 $\frac{F_{er}-F_{er} }{\beta_{2r}}, \hat{\theta}_{2r}, \theta_{2rm},$

 $\frac{F_{\text{eff}}-F_{\text{er}}}{I_{qr}\beta_{3r}}, \hat{\theta}_{3r}, \theta_{3rm},$

 $\theta_{1rM}, \varepsilon_{1r}$

 $\theta_{2rM}, \varepsilon_{2r}$

 $\dot{\hat{\theta}}_{2r}$ = Proj $\left[-\frac{I_{dr}I_{qr}(P_{er}-P_{er}^*)}{\beta}\right]$

 $\dot{\hat{\theta}}_{3r}$ = Proj $\left[-\frac{I_{dr}P_{er}(P_{er}-P_{er}^*)}{I_{eq}}\right]$

$$
\theta_{3rM}, \varepsilon_{3r} \Bigg], \ \theta_{3rm} \leq \hat{\theta}_{3r}(0) \leq \theta_{3rM}
$$
\n
$$
\hat{\theta}_{4r} = \Pr \{ \text{Proj} \Bigg[-\frac{P_{er}^2 (P_{er} - P_{er}^*)}{I_{qr}^2 \beta_{4r}}, \hat{\theta}_{4r}, \theta_{4rm},
$$
\n
$$
\theta_{4rM}, \varepsilon_{4r} \Bigg], \ \theta_{4rm} \leq \hat{\theta}_{4r}(0) \leq \theta_{4rM}
$$
\n
$$
\hat{\theta}_{5r} = \Pr \{ \text{obj} \Bigg[\frac{I_{qr}^2 v_r (P_{er} - P_{er}^*)}{\left(\hat{\theta}_{5r} I_{qr}^2 + \hat{\theta}_{6r} P_{er} \right) \beta_{5r}}, \hat{\theta}_{5r}, \theta_{5rm},
$$
\n
$$
\theta_{5rM}, \varepsilon_{5r} \Bigg], \ \theta_{5rm} \leq \hat{\theta}_{5r}(0) \leq \theta_{5rM}
$$
\n
$$
\hat{\theta}_{6r} = \Pr \{ \text{obj} \Bigg[\frac{P_{er} v_r (P_{er} - P_{er}^*)}{\left(\hat{\theta}_{5r} I_{qr}^2 + \hat{\theta}_{6r} P_{er} \right) \beta_{6r}}, \hat{\theta}_{6rm},
$$
\n
$$
\theta_{6rM}, \varepsilon_{6r} \Bigg], \ \theta_{6rm} \leq \hat{\theta}_{6r}(0) \leq \theta_{6rM}
$$
\n
$$
\hat{x}_{dr}' = \Pr \{ \text{obj} \Bigg[-\frac{\left(I_{dr}^2 + I_{qr}^2 \right) \omega_r (P_{er} - P_{er}^*)}{\beta_{2r}}, \hat{x}_{dr}',
$$
\n
$$
x_{drm}', x_{drM}', \varepsilon_{sr} \Bigg], \ x_{drm}' \leq x_{dr}'(0) \leq x_{drM}
$$
\n
$$
\hat{B}_{rr} = \Pr \{ \text{obj} \Bigg[-\frac{P_{er}^2 \omega_r (P_{er} - P_{er}^*)}{\beta_{2r}} \hat{\theta}_{r}, \hat{B}_{rr}, \theta_{rrm},
$$
\n
$$
\theta_{1rm} = \frac{1}{T_{d0rm}M}, \ \theta_{1rm} \le
$$

$$
Proj[\zeta, \hat{z}_r, z_{rm}, z_{rm}, \varepsilon_{zr}] = \begin{cases} \zeta \zeta_{p1} & \text{if } C_a \\ \zeta \zeta_{p2} & \text{if } C_b \\ \zeta & \text{otherwise} \end{cases}
$$

$$
\zeta_{p1} = \left[1 - \frac{z_{rm}^2 - \hat{z}_r^2}{z_{rm}^2 - (z_{rm} - \varepsilon_{zr})^2}\right]
$$
\n
$$
\zeta_{p2} = \left[1 - \frac{\hat{z}_r^2 - z_{rm}^2}{(z_{rm} + \varepsilon_{zr})^2 - z_{rm}^2}\right]
$$
\n
$$
0 \le z_{rm}(z_{rm} - \varepsilon_{zr}), \quad 0 \le z_{rm}(z_{rm} + \varepsilon_{zr})
$$
\n
$$
\mathcal{C}_a: \hat{z}_r < z_{rm} \& \zeta < 0, \quad \mathcal{C}_b: \hat{z}_r > z_{rm} \& \zeta > 0
$$

is bounded and guarantees the closed loop system (2)-(3) to satisfy, in terms of $y_r(t) = [\delta_r(t) \delta_{rs}$, $\omega_r(t)$, $P_{er}(t) - P_{mr}(t)$ ^T, $x_r(t) = [y_r(t)^T$, $P_{mr}(t) - \hat{P}_{mr}(t)$]^T, $w_{dr}(t) = [\hat{P}_{mr}(t), \theta_{1rM} - \theta_{1rm} + \theta_{1rM})$ $\varepsilon_{1r},\theta_{2rM}-\theta_{2rm}+\varepsilon_{2r},\theta_{3rM}-\theta_{3rm}+\varepsilon_{3r},\theta_{4rM} \theta_{4rm} + \varepsilon_{4r}, \, {\rm max} \{ \theta_{5rM} - \theta_{5rm} + \varepsilon_{5r}~, \theta_{6rM} - \theta_{6rm}$ $+\varepsilon_{6r}\}, x'_{drM} - \dot{x'_{drm}} + \varepsilon_{xr}B_{rrM} - B_{rrm} + \varepsilon_{Br},$ $\sup_{0\leq \tau\leq t}\{g_r(\tau)\},\mu_r,\nu_r,\rho_r]^T$, the following proper*ties:*

(S1) exponential convergence and L[∞] *disturbance attenuation, i.e.*

$$
||y_r(t)||^2 \leq h_{1r}e^{-c_r t} + \frac{1}{k_r} \gamma_{1r}(||w_{dr}(\cdot)||_{\infty})
$$

holds for all $t \geq 0$ *, where* $h_{1r}(x_r(0)) \geq 0$ *, c_r* > 0 *and* $\gamma_{1r}(r)$ *is a class* \mathcal{K}_{∞} *function*;

 $(S2)$ \mathcal{L}_2 *disturbance attenuation, i.e.*

$$
\int_0^T \|y_r(\tau)\|^2 d\tau \le h_{2r} + \frac{1}{k_r} \int_0^T \gamma_{2r}(\|w_{dr}(\tau)\|) d\tau
$$

holds for any given $T > 0$ *, where* $h_{2r}(x_r(0)) \geq$ 0 *and* $\gamma_{2r}(r)$ *is a class* \mathcal{K}_{∞} *function.*

If, in addition to previous assumptions, there exist a non-negative real t_0 *and positive reals* M_{Rr} *,* $\gamma_{\delta r}$ *,* $\gamma_{\omega r}$ *,* γ_{pr} *such that*

vii) $P_{mr}(t)$, $T'_{d0r}(t)$, $x_{dr}(t)$, $x'_{dr}(t)$, $k_{cr}(t)$ are con*stant for all* $t \geq t_0$;

$$
\begin{array}{ll}\n\text{viii)} & \lim_{t \to \infty} \int_{t_0}^t R_r^2(\tau) d\tau \leq & M_{Rr} \\
& + \gamma_{\delta r} \lim_{t \to \infty} \int_{t_0}^t [\delta_r(\tau) - \delta_{rs}]^2 d\tau \\
& + \gamma_{\omega r} \lim_{t \to \infty} \int_{t_0}^t [\omega_r(\tau)]^2 d\tau \\
& + \gamma_{pr} \lim_{t \to \infty} \int_{t_0}^t [P_{er}(\tau) - P_{mr}(\tau)]^2 d\tau;\n\end{array}
$$

ix) min
$$
\left\{ \left(\frac{k_{\delta r}}{4} - \frac{\bar{\gamma}_{\delta r}}{k_{Rr}} \right), \left(\frac{k_{\omega r}}{4} - \frac{\bar{\gamma}_{\omega r}}{k_{Rr}} \right), \left(\frac{k_{pr}}{4} - \frac{4\gamma_{pr}}{k_{Rr}} \right), \left(\frac{k_{per}}{4} - \frac{4\gamma_{pr}}{k_{Rr}} \right), \left(\frac{k_{per}}{4} - \frac{4\gamma_{pr}}{k_{Rr}} \right) \right\} > 0 \text{ with } \bar{\gamma}_{\delta r} = \gamma_{\delta r} + \frac{25}{8} \gamma_{\omega r} k_{\delta r}^2 +
$$

\n $4 \left[\frac{2H_r}{\omega_0} \left(\frac{25}{16} k_{\delta r}^2 + 1 \right) + \frac{5}{4} \frac{D_r k_{\delta r}}{\omega_0} \right]^2 \gamma_{pr} \text{ and } \bar{\gamma}_{\omega r} =$
\n $2\gamma_{\omega r} + 4 \left[\frac{2H_r}{\omega_0} \left(\frac{5}{4} k_{\omega r} + \frac{5}{4} k_{\delta r} + \frac{1}{k_{\omega pr}} \right) \right]^2 \gamma_{pr},$

then controller (3) guarantees the additional property:

(S3) asymptotic convergence, i.e.

$$
\lim_{t \to \infty} \left\| [\delta_r(t) - \delta_{rs}, \omega_r(t), P_{er}(t) - P_{mr}] \right\| = 0.
$$

Remark 1: The proposed control algorithm (3) relies on the locally measured signals: δ_r , ω_r , P_{er} , I_{dr} , I_{qr} , Q_{er} (the method for measuring the power angle δ_r can be found in [3], [2] and [5]).

Remark 2: Let $\Sigma_r(y_{r1}, y_{r2}, y_{r3}, R_r)$ be the system consisting of the j-th generators ($j \neq r$) with inputs the $y_r(t)$ vector components $y_{r1}(t)$, $y_{r2}(t)$, $y_{r3}(t)$ and output the function $R_r(t)$. Less restrictive conditions than those required in the single machine-infinite bus approximation are to be satisfied by the remaining part of the network: assumption ii) $_a$) is satisfied if system $\Sigma_r(y_{r1}, y_{r2}, y_{r3}, R_r)$ is Input to Output Stable with gain functions $\psi_{\mu r}(\cdot)$, $\psi_{\nu r}(\cdot)$, $\psi_{\rho r}(\cdot)$; assumption ii) $_{b)}$ requires the knowledge of suitable functions $\overline{\varphi}_{\mu r}(\cdot), \overline{\varphi}_{\nu r}(\cdot), \overline{\varphi}_{\rho r}(\cdot)$ which majorize the gain functions $\psi_{\mu r}(\cdot), \psi_{\nu r}(\cdot), \psi_{\rho r}(\cdot)$; assumption viii) is satisfied if system $\Sigma_r(y_{r1}, y_{r2}, y_{r3}, R_r)$ has finite \mathcal{L}_2 gains, bounded from above by the positive numbers $\sqrt{\gamma_{\delta r}/3}$, $\sqrt{\gamma_{\omega r}/3}, \sqrt{\gamma_{pr}/3}.$

3 Simulation Results

In this section we will test by simulation the control algorithm (3) with reference to each generator of the popular Western System Coordinating Council (WSCC) 3-machine, 9-bus system described in [10] and [1] $[D_i = 0, 1 \le i \le 3]$, with the aim of numerically checking the conservativeness of the conditions provided by the theorem, of testing the performance and of verifying the robustness of the proposed control law with respect to unmodelled dynamics: the simulation is carried out by using the two-axis model (see [10]) from which the model (1) has been derived. The initial conditions for the state variables are computed by systematically solving the load-flow equations of the network and by computing the values of the algebraic variables. For $1 \leq i \leq 3$, the functions $\bar{\varphi}_{\mu i}(\cdot),\bar{\varphi}_{\nu i}(\cdot),\bar{\varphi}_{\rho i}(\cdot)$ are set equal to $id(\cdot)_{\left| [0,\infty)\right. }$ $(id(\cdot))$ is the identity function], the control parameters are chosen as $k_{\delta i} = k_{\omega i} = 0.01, k_{pi} = 720,$ $k_{\text{pei}} = k_{\omega p i} = 1, k_i = 0.001, k_{\text{R}i} = 0.1, \hat{\beta}_{1i} = \beta_{2i}$ $\beta_{3i} = \beta_{4i} = \beta_{5i} = \beta_{6i} = \beta_{xi} = \beta_{Bi} = 48000,$ $\varepsilon_{1i} = \varepsilon_{2i} = \varepsilon_{3i} = \varepsilon_{4i} = \varepsilon_{5i} = \varepsilon_{6i} = \varepsilon_{xi}$ $\varepsilon_{Bi} = 0.00001$, while $P_{Mi} = 1$ and the known upper and lower bounds on the model parameters values are $k_{c1m} = 0.6, k_{c1M} = 1.4, x_{d1m} = 0.1, x_{d1M} = 0.2,$ $x'_{d1m} = 0.05$ p.u., $x'_{d1M} = 0.07$ p.u., $T'_{d01m} = 6$ p.u.,

 $T'_{d01M} = 12$ p.u., $G_{11m} = 0.5$ p.u., $G_{11M} = 1.5$ p.u., $P_{m1m}^{301m} = 0.3 \text{ p.u., } P_{m1M} = 1.1 \text{ p.u., } B_{11m} = -6 \text{ p.u.,}$ $B_{11M} = -1$ p.u., $k_{c2m} = 0.6$, $k_{c2M} = 1.4$, $x_{d2m} =$ 0.7, $x_{d2M} = 1.1$, $x'_{d2m} = 0.09$ p.u., $x'_{d2M} = 0.14$ p.u., $T'_{d02m} = 4$ p.u., $T'_{d02M} = 8$ p.u., $G_{22m} = 0.3$ p.u., $G_{22M} = 0.6$ p.u., $P_{m2m} = 1.2$ p.u., $P_{m2M} = 2$ p.u., $B_{22m} = -6$ p.u., $B_{22M} = -1$ p.u., $k_{c3m} = 0.6$, $k_{c3M} = 1.4, x_{d3m} = 1.2, x_{d3M} = 1.5, x'_{d3m} = 0.1$ p.u., $x'_{d3M} = 0.3$ p.u., $T'_{d03m} = 4$ p.u., $T''_{d03M} = 8$ p.u., $G_{33m} = 0.1$ p.u., $G_{33M} = 0.4$ p.u., $P_{m3m} = 0.4$ p.u., $P_{m3M} = 1.2$ p.u., $B_{33m} = -6$ p.u., $B_{33M} = -1$ p.u.. The goal of the simulation is to verify the effect of a three-phase fault occurring near bus 7 at the end of line 5-7 at $t = 0.001$ s, which is cleared in five cycles (0.083 s) by opening line 5-7. The same simulation is performed by testing separately (with reference to each generator of WSCC) the proposed controller and both the IEEE-Type I and the fast exciters with PSS (the exciter data are taken from [10], while $K_0 = \tau_0 = 24, \tau_1 = \tau_3 = 0.5, \tau_2 = \tau_4 = 0.05$ and the saturating limits are ± 0.1 in the PSS diagram in [1]). The results are reported in Figures 1-3: the proposed controller restores the synchronous speeds, while the linear controllers either cannot survive the severe fault or can only mantain the system stable.

Figure 1: Proposed control [Generator 1 (solid), Generator 2 (dot), Generator 3 (dashed)]; a) P_{mi} b) $\delta_i - \delta_{is}$ c) ω_i d) P_{ei} e) V_{ti} f) u_{fi}

4 Conclusions

A robust adaptive nonlinear feedback control (3) has been designed on the basis of a third order model

Figure 2: IEEE-Type I exciter with PSS [Generator 1 (solid), Generator 2 (dot), Generator 3 (dashed)]; a) P_{mi} b) $\delta_i - \delta_{is}$ c) ω_i d) P_{ei} e) V_{ti} f) u_{fi}

(1) of a synchronous generator in a multimachine power network with nontrivial conductances. Only three model parameters and upper and lower bounds on the uncertain ones are required to be known. \mathcal{L}_2 and \mathcal{L}_{∞} disturbance attenuation and asymptotic regulation properties *(S1)*-*(S3)* are guaranteed under suitable conditions to be satisfied by the network dynamics which generalize those required by the single machine-infinite bus approximation.

Acknowledgements: The research was supported by MIUR and through a European Community Marie Curie Fellowship (in the framework of the CTS), contract number: HPMT-CT-2001-00278.

References:

- [1] P.–M. Anderson, and A.–A. Fouad, *Power system control and stability,* Wiley-IEEE Press, 2002.
- [2] E. Barrera-Cardiel, and N. Pastor-Gomez, Microcontroller-based power-angle instrument for a power systems laboratory, *IEEE Pow. Eng. Soc. Sum. Meet.* 2, 1999, pp. 1008–1012.
- [3] Y. Chen, C. Zhang, Z. Hu, and X. Wang, A new approach to real time measurement of power angles of generators at different locations for stability control, *Pow. Eng. Soc. Win. Meet.* 2, 2000, pp. 1237–1242.
- [4] G. Damm, R. Marino, and F. Lamnabhi-Lagarrigue, Adaptive nonlinear output feedback

Figure 3: Fast exciter with PSS [Generator 1 (solid), Generator 2 (dot), Generator 3 (dashed)]; a) P_{mi} b) $\delta_i - \delta_{is}$ c) ω_i d) P_{ei} e) V_{ti} f) u_{fi}

for transient stabilization and voltage regulation of power generators with unknown parameters, *Int. J. Rob. Nonlin. Contr.* 14, 2004, pp. 833– 855.

- [5] F.–P. de Mello, Measurement of synchronous machine rotor angle from analysis of zero sequence harmonic components of machine terminal voltage, *IEEE Trans. Pow. Del.* 9, 2004, pp. 1770–1777.
- [6] L. Gao, L. Chen, Y. Fan, and H. Ma, A nonlinear control design for power systems, *Aut.* 28, 1992, pp. 975–979.
- [7] Y. Guo, D.–J. Hill, and Y. Wang, Nonlinear decentralized control of large-scale power systems, *Aut.* 36, 2000, pp. 1275–1289.
- [8] R. Marino, An example of nonlinear regulator, *IEEE Trans. Aut. Contr.* 29, 1984, pp. 276–279.
- [9] R. Marino, T. Shen, and C.–M. Verrelli, Robust adaptive transient stabilization of a synchronous generator with parameter uncertainty, *Eur. J. Contr.*, to appear.
- [10] P.–W. Sauer, and M.–A. Pai, *Power system dynamics and stability,* Prentice Hall, 1997.
- [11] Y. Wang, and D.–J. Hill, Robust nonlinear coordinated control of power systems, *Aut.* 32, 1996, pp. 611–618.
- [12] Y. Wang, D.–J. Hill, R.–H. Middleton, and L. Gao, Transient stabilization of power systems with an adaptive control law, *Aut.* 30, 1994, pp. 1409–1413.