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Asymptotically minimum variance second-order

estimation for non-circular signals with application to

DOA estimation

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Abstract

This paper addresses asymptotically minimum variance (AMV) algorithms within the class of

algorithms based on second-order statistics for estimating direction of arrival (DOA) parameters of

possibly spatially correlated (even coherent) narrowband non-circular sources impinging on arbitrary

array structures. To reduce the computational complexity due to the nonlinear minimization required

by the matching approach, the covariance matching estimation techniques (COMET) is included in

the algorithm. Numerical examples illustrate the performance of the AMV algorithm.

Index terms: second-order statistics-based algorithms, DOA estimation, complex non-circular, asymp-

totically minimum variance.

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1 Introduction

There is considerable literature about second-order statistics-based algorithms for estimating DOA of narrowband sources impinging on an array of sensors. The interest in these algorithms stems from a large number of applications including mobile communications systems [?]. In this application, after frequency down-shifting the sensor signals to baseband, the in-phase and quadrature components are paired to obtain complex signals. And complex non-circular signals [?], for example, binary phase shift keying (BPSK) modulated signals are often used. However, only a few contributions, such as [?],[?],[?],[?] have been devoted to non-circular signals.

The DOA second-order algorithms devoted to complex circular signals rely on the positive definite Hermitian covariance matrix $E(\mathbf{y}_t \mathbf{y}_t^H)$, and naturally they can be used in the context of non-circular signals. Because, the second-order statistical characteristics are also contained in the complex symmetric covariance matrix $E(\mathbf{y}_t \mathbf{y}_t^T)$ for non-circular signals, a potentially performance improvement ought to be obtained if these two covariance matrices are used. In the context of spatially uncorrelated amplitude modulated or BPSK modulated sources impinging on a linear uniform array, a significant performance improvement has been already observed by simulations in [?] and [?] thanks to a MUSIC-like algorithm and a root-MUSIC like algorithm respectively.

To improve the performance of these algorithms and to extend DOA estimation to spatially correlated or even coherent arbitrary non-circular sources and to arbitrary array structures, we propose to consider asymptotically (in the number of measurements) minimum variance algorithms in the class of algorithms based on the two covariance matrices. We extend to complex non-circular processes the result of Porat and Friedlander [?] devoted to the estimating of MA and ARMA parameters of real non-Gaussian processes from sample high-order statistics. After a general lower bound is derived for the covariance of the estimated DOAs, it is shown that a generalized covariance matching algorithm attains this bound. Furthermore, the ideas of COMET [?] are exploited to reduce the dimension of the optimization problem.

The paper is organized as follows. Section 2 presents the asymptotically minimum variance second-order estimator for stationary complex non-circular processes with special attention to the statistics involved. As an application, the estimation of the DOA parameters is considered in Section 3. The asymptotic performance is analyzed in Section 4. Finally, illustrative examples with comparisons with the AMV estimators based on the first covariance matrix only are given in Section 5.

The following notations are used throughout the paper. Matrices and vectors are represented by bold upper

case and bold lower case characters, respectively. Vectors are by default in column orientation, while T, H, * stand for transpose, conjugate transpose, conjugate, respectively. $\mathbf{e}_{K,k}$ is the kth unit vector in \mathcal{R}^K . $\text{vec}(\cdot)$ is the "vectorization" operator that turns a matrix into a vector by stacking the columns of the matrix one below another and $\mathbf{v}(\cdot)$ denotes the operator obtained from $\text{vec}(\cdot)$ by eliminating all supradiagonal elements of the matrix. They are used in conjunction with the Kronecker product $\mathbf{A} \otimes \mathbf{B}$ as the block matrix whose (i,j) block element is $a_{i,j}\mathbf{B}$ and with the vec-permutation matrix \mathbf{K} which transforms $\text{vec}(\cdot)$ to $\text{vec}(\cdot)$ for any square matrix. The notation f(x) = o(x) means that $\lim_{x\to 0} \frac{f(x)}{x} = 0$.

2 Asymptotic minimum variance second-order estimator

We consider a zero-mean strict-sense stationary M-variate complex, possibly non-circular process \mathbf{y}_t whose structured covariance matrices $\mathbf{R}(\Theta) \stackrel{\text{def}}{=} \mathrm{E}(\mathbf{y}_t \mathbf{y}_t^H)$ and $\mathbf{R}'(\Theta) \stackrel{\text{def}}{=} \mathrm{E}(\mathbf{y}_t \mathbf{y}_t^T)$ are parameterized by the real parameter $\Theta \in \mathcal{R}^L$. These covariance matrices are classically estimated by $\mathbf{R}_T = \frac{1}{T} \sum_{t=1}^T \mathbf{y}_t \mathbf{y}_t^H$ and $\mathbf{R}'_T = \frac{1}{T} \sum_{t=1}^T \mathbf{y}_t \mathbf{y}_t^T$ respectively. This parameter is supposed identifiable from $(\mathbf{R}(\Theta), \mathbf{R}'(\Theta))$, in the following sense:

$$\mathbf{R}(\Theta) = \mathbf{R}(\Theta') \ \ \mathrm{and} \ \ \mathbf{R}'(\Theta) = \mathbf{R}'(\Theta') \ \ \Rightarrow \ \ \Theta = \Theta'.$$

To consider the asymptotic performance of a second-order algorithm, we adopt a functional analysis which consists in recognizing that the whole process of constructing an estimate Θ_T of Θ is equivalent to defining a functional relation linking this estimate Θ_T to the statistics $(\mathbf{R}_T, \mathbf{R}_T')$ from which it is inferred. This functional dependence is denoted

$$(\mathbf{R}_T, \mathbf{R}_T') \longmapsto \Theta_T = \operatorname{alg}(\mathbf{R}_T, \mathbf{R}_T').$$

By assumption, $\Theta = \operatorname{alg}(\mathbf{R}(\Theta), \mathbf{R}'(\Theta))$, so the different algorithms alg(.) constitute distinct extensions of the mapping $(\mathbf{R}(\Theta), \mathbf{R}'(\Theta)) \longmapsto \Theta$ generated by any unstructured Hermitian matrix \mathbf{R}_T and complex symmetric matrix \mathbf{R}_T' .

To extend the ideas of Porat and Friedlander [?] concerning asymptotically minimum variance second-order estimators, to complex non-circular processes, two conditions must be satisfied. First, the covariance $\mathbf{C}_{r'}(\Theta)$ of the asymptotic distribution of $(\mathbf{R}_T, \mathbf{R}_T')$ must be regular. Second, the involved second-order algorithm considered as a mapping must be complex differentiable w.r.t. $(\mathbf{R}_T, \mathbf{R}_T')$ at the point $(\mathbf{R}(\Theta), \mathbf{R}'(\Theta))$. While these two conditions are satisfied for a second-order algorithms based on \mathbf{R}_T only, none of these two conditions are satisfied in our

situation for the following reasons. First, because \mathbf{R}_T' is symmetric, the rank of $\mathbf{C}_{r'}(\Theta)$ which is the rank of the set of the entries of $(\mathbf{R}_T, \mathbf{R}_T')$ is not full. Consequently $\mathbf{C}_{r'}(\Theta)$ is singular. Second, because \mathbf{R}_T' is complex non Hermitian, an algorithm considered as a mapping, is not complex differentiable w.r.t. \mathbf{R}'_T at point $\mathbf{R}'(\Theta)$.

To satisfy these two conditions, we must eliminate the common terms in \mathbf{R}_T' and add complex conjugate associated terms. Below, we consider the equivalent to $(\mathbf{R}_T, \mathbf{R}_T')$ statistics \mathbf{s}_T constituted by $\mathbf{r}_T \stackrel{\text{def}}{=} \text{vec}(\mathbf{R}_T)$, $\tilde{\mathbf{r}}_T' \stackrel{\text{def}}{=} \mathbf{v}(\mathbf{R}_T') \text{ and } \tilde{\mathbf{r}}_T'^* \stackrel{\text{def}}{=} \mathbf{v}(\mathbf{R}_T'^*),$

$$\mathbf{s}_T \stackrel{\mathrm{def}}{=} \left(egin{array}{c} \mathbf{r}_T \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ & & \ &$$

and the associated mapping:

$$\mathbf{s}_T \longmapsto \Theta_T = \operatorname{alg}(\mathbf{s}_T).$$

 $\mathbf{r}(\Theta)$, $\tilde{\mathbf{r}}'(\Theta)$ and $\mathbf{s}(\Theta)$ are defined in the same way from $\mathbf{R}(\Theta)$ and $\mathbf{R}'(\Theta)$. Because $\text{vec}(\mathbf{R}_T^T) = \text{vec}(\mathbf{R}_T^T) = \text{Kvec}(\mathbf{R}_T)$,

 $\mathbf{r}(\Theta), \tilde{\mathbf{r}}'(\Theta) \text{ and } \mathbf{s}(\Theta) \text{ are defined ...}$ $\mathbf{s}^* = \mathbf{P}\mathbf{s}, \text{ where } \mathbf{P} \text{ is the permutation matrix } \begin{pmatrix} \mathbf{K} & \mathbf{O} & \mathbf{O} \\ \mathbf{O} & \mathbf{O} & \mathbf{I} \\ \mathbf{O} & \mathbf{I} & \mathbf{O} \end{pmatrix}. \text{ Consequently, any mapping alg(.) differentiable}$

w.r.t. $(\Re(\mathbf{s}), \Im(\mathbf{s}))$ becomes differentiable w.r.t. \mathbf{s} alone if $\delta \mathbf{s}$ is structured as $\delta \mathbf{s} = \begin{pmatrix} \delta \mathbf{r} \\ \delta \tilde{\mathbf{r}} \\ \delta \tilde{\mathbf{s}}'^* \end{pmatrix}$, in which case

$$\operatorname{alg}[\mathbf{s}(\Theta) + \delta \mathbf{s}] = \operatorname{alg}[\mathbf{s}(\Theta)] + [\mathbf{D}_s, \mathbf{D}_s^*] \begin{bmatrix} \delta \mathbf{s} \\ \delta \mathbf{s}^* \end{bmatrix} + o(\delta \mathbf{s}) = \Theta + \mathbf{D}_s^{\operatorname{alg}} \delta \mathbf{s} + o(\delta \mathbf{s})$$

where \mathbf{D}_s and \mathbf{D}_s^* denote the Jacobian matrices of this differential at point $\mathbf{s}(\Theta)$, with $\mathbf{D}_s^{\mathrm{alg}} \stackrel{\mathrm{def}}{=} \mathbf{D}_s + \mathbf{D}_s^* \mathbf{K}$. And because $alg[\mathbf{s}(\Theta)] = \Theta$ for all Θ , we have with $\mathbf{S} \stackrel{\text{def}}{=} \frac{d\mathbf{s}(\Theta)}{d\Theta}$:

$$\operatorname{alg}[\mathbf{s}(\Theta + \delta\Theta)] = \operatorname{alg}[\mathbf{s}(\Theta) + \mathbf{S}\delta\Theta + o(\delta\Theta)] = \Theta + \mathbf{D}_s^{\operatorname{alg}}\mathbf{S}\delta\Theta + o(\delta\Theta) = \Theta + \delta\Theta.$$

Therefore $\mathbf{D}_s^{\text{alg}}$ is a left inverse of \mathbf{S} :

$$\mathbf{D}_s^{\text{alg}}\mathbf{S} = \mathbf{I}_L,\tag{2.1}$$

and this time, the rank of the set of the entries of \mathbf{s}_T is generally $M^2 + M(M+1)$ and so, the covariance $\mathbf{C}_s(\Theta)$ of the asymptotic distribution of \mathbf{s}_T is a Hermitian positive definite matrix. Therefore, we obtain by application of theorem 2 of [?], extended to the complex case:

Theorem 1 The asymptotic covariance matrix \mathbf{C}_{Θ} of an estimator of Θ given by an arbitrary second-order algorithm is bounded below by the real symmetric matrix $(\mathbf{S}^H \mathbf{C}_s^{-1}(\Theta) \mathbf{S})^{-1}$:

$$\mathbf{C}_{\Theta} = \mathbf{D}_{s}^{\mathrm{alg}} \mathbf{C}_{s}(\Theta) (\mathbf{D}_{s}^{\mathrm{alg}})^{H} \ge \left(\mathbf{S}^{H} \mathbf{C}_{s}^{-1}(\Theta) \mathbf{S}\right)^{-1}.$$
(2.2)

Proof: From (??), we get

$$0 \leq \left[\mathbf{D}_{s}^{\mathrm{alg}} - (\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\mathbf{S})^{-1}\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\right]\mathbf{C}_{s}(\Theta)\left[\mathbf{D}_{s}^{\mathrm{alg}} - (\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\mathbf{S})^{-1}\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\right]^{H}$$

$$= \mathbf{D}_{s}^{\mathrm{alg}}\mathbf{C}_{s}(\Theta)(\mathbf{D}_{s}^{\mathrm{alg}})^{H} - \mathbf{D}_{s}^{\mathrm{alg}}\mathbf{S}(\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\mathbf{S})^{-1} - (\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\mathbf{S})^{-1}\mathbf{S}^{H}(\mathbf{D}_{s}^{\mathrm{alg}})^{H} + (\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\mathbf{S})^{-1}$$

$$= \mathbf{D}_{s}^{\mathrm{alg}}\mathbf{C}_{s}(\Theta)(\mathbf{D}_{s}^{\mathrm{alg}})^{H} - (\mathbf{S}^{H}\mathbf{C}_{s}^{-1}(\Theta)\mathbf{S})^{-1},$$

and furthermore, because $\mathbf{s}^* = \mathbf{P}\mathbf{s}$ implies $\mathbf{S}^* = \mathbf{P}\mathbf{S}$ and $\mathbf{C}_s^T(\Theta) = \mathbf{P}\mathbf{C}_s(\Theta)\mathbf{P} \ (\mathbf{S}^H\mathbf{C}_s^{-1}(\Theta)\mathbf{S})^T = \mathbf{S}^T(\mathbf{C}_s^{-1}(\Theta))^T\mathbf{S}^* = \mathbf{S}^H\mathbf{P}(\mathbf{C}_s^T(\Theta))^{-1}\mathbf{P}\mathbf{S} = \mathbf{S}^H\mathbf{C}_s^{-1}(\Theta)\mathbf{S}$, the Hermitian matrix $(\mathbf{S}^H\mathbf{C}_s^{-1}(\Theta)\mathbf{S})^{-1}$ is real symmetric.

Furthermore, we prove that this lowest bound is asymptotically tight, i.e., there exists an algorithm alg(.) whose covariance of the asymptotic distribution of Θ_T satisfies (??) with equality. Therefore, theorem 3 of [?] extends to the complex non-circular case.

Theorem 2 The following nonlinear least square algorithm is an AMV second-order algorithm.

$$\Theta_T = \arg\min_{\alpha \in \mathcal{T}^L} [\mathbf{s}_T - \mathbf{s}(\alpha)]^H \mathbf{C}_s^{-1}(\alpha) [\mathbf{s}_T - \mathbf{s}(\alpha)]. \tag{2.3}$$

Proof: By a perturbation analysis, $\Theta_T = \Theta + \delta\Theta$ is associated with $\mathbf{s}_T = \mathbf{s}(\Theta) + \delta\mathbf{s}$ (with $\delta\mathbf{s}$ strutured). If $V(\alpha) \stackrel{\text{def}}{=} [\mathbf{s}(\Theta) - \mathbf{s}(\alpha)]^H \mathbf{C}_s^{-1}(\alpha) [\mathbf{s}(\Theta) - \mathbf{s}(\alpha)]$ and $V_T(\alpha) \stackrel{\text{def}}{=} [\mathbf{s}_T - \mathbf{s}(\alpha)]^H \mathbf{C}_s^{-1}(\alpha) [\mathbf{s}_T - \mathbf{s}(\alpha)]$, we have: $\frac{dV(\alpha)}{d\alpha}|_{\alpha=\Theta} = \mathbf{0}$ and $\frac{dV_T(\alpha)}{d\alpha}|_{\alpha=\Theta+\delta\Theta} = \mathbf{0}$. Expanding these two derivatives, we straightforwardly obtain: $(\mathbf{S}^H \mathbf{C}_s^{-1}(\Theta)\mathbf{S} + \mathbf{S}^T \mathbf{C}_s^{-1}(\Theta)^* \mathbf{S}^*) \delta\Theta + o(\delta\Theta) = \mathbf{S}^H \mathbf{C}_s^{-1}(\Theta) \delta\mathbf{s} + \mathbf{S}^T \mathbf{C}_s^{-1}(\Theta)^* \delta\mathbf{s}^* + o(\delta\mathbf{s})$. Consequently the algorithm (??) satisfies:

$$alg[\mathbf{s}(\Theta) + \delta \mathbf{s}] = \Theta + \left(\mathbf{S}^{H} \mathbf{C}_{s}^{-1}(\Theta) \mathbf{S} + \mathbf{S}^{T} \mathbf{C}_{s}^{-1}(\Theta)^{*} \mathbf{S}^{*}\right)^{-1} \left(\mathbf{S}^{H} \mathbf{C}_{s}^{-1}(\Theta), \mathbf{S}^{T} \mathbf{C}_{s}^{-1}(\Theta)^{*}\right) \begin{pmatrix} \delta \mathbf{s} \\ \delta \mathbf{s}^{*} \end{pmatrix} + o(\delta \mathbf{s})$$

$$= \Theta + \left(\mathbf{S}^{H} \mathbf{C}_{s}^{-1}(\Theta) \mathbf{S}\right)^{-1} \mathbf{S}^{H} \mathbf{C}_{s}^{-1}(\Theta) \delta \mathbf{s} + o(\delta \mathbf{s}),$$

by using $\mathbf{S}^* = \mathbf{P}\mathbf{S}$ and $\mathbf{C}_s^T(\Theta) = \mathbf{C}_s^*(\Theta) = \mathbf{P}\mathbf{C}_s(\Theta)\mathbf{P}^T$ in the second equality. Consequently the derivative of the mapping alg(.) involved by (??) is $\mathbf{D}_s^{\mathrm{alg}} = \left(\mathbf{S}^H \mathbf{C}_s^{-1}(\Theta)\mathbf{S}\right)^{-1}\mathbf{S}^H \mathbf{C}_s^{-1}(\Theta)$ and $\mathbf{C}_{\Theta} = \mathbf{D}_s^{\mathrm{alg}}\mathbf{C}_s(\Theta)(\mathbf{D}_s^{\mathrm{alg}})^H = (\mathbf{S}^H \mathbf{C}_s^{-1}(\Theta)\mathbf{S})^{-1}$.

In practice, it is difficult to optimize the nonlinear function (??) where it involves the computation of $\mathbf{C}_s^{-1}(\alpha)$. Porat and Friedlander proved for the real case in [?], that the lowest bound (??) is also obtained if an arbitrary consistent estimate $\mathbf{C}_{s,T}$ of $\mathbf{C}_s(\alpha)$ is used in (??). This property extends to the complex non-circular case and to any Hermitian positive definite weighting matrix. And we prove:

Theorem 3 The covariance of the asymptotic distribution of Θ_T given by an arbitrary nonlinear least square algorithm

$$\Theta_T = \arg\min_{\alpha \in \mathcal{R}^L} [\mathbf{s}_T - \mathbf{s}(\alpha)]^H \mathbf{W}(\alpha) [\mathbf{s}_T - \mathbf{s}(\alpha)], \tag{2.4}$$

is preserved if the Hermitian positive definite weighting matrix $\mathbf{W}(\alpha)$ is replaced by an arbitrary consistent estimate \mathbf{W}_T that satisfies $\mathbf{W}_T = \mathbf{W}(\Theta) + O(\mathbf{s}_T - \mathbf{s}(\Theta))$.

Proof: Following a perturbation analysis similar to those of the proof of theorem ??, it is straightforward to show that the differential $\mathbf{D}_s^{\mathrm{alg}} = \left(\mathbf{S}^H \mathbf{W}(\Theta) \mathbf{S}\right)^{-1} \mathbf{S}^H \mathbf{W}(\Theta)$ of the mapping alg(.) involved by (??) is preserved.

So the minimization (??) can be preferably replaced by the following

$$\Theta_T = \arg\min_{\alpha \in \mathcal{R}^L} [\mathbf{s}_T - \mathbf{s}(\alpha)]^H \mathbf{C}_{s,T}^{-1} [\mathbf{s}_T - \mathbf{s}(\alpha)]. \tag{2.5}$$

3 Application to estimation of DOA

In the following, we will be concerned with the signal model

$$\mathbf{y}_t = \mathbf{A}\mathbf{x}_t + \mathbf{n}_t, \qquad t = 1, \dots, T$$

where $(\mathbf{y}_t)_{t=1,...,T}$ represents the independent identically distributed M-vectors of observed complex envelope at the sensor output. $\mathbf{A} = [\mathbf{a}_1, \ldots, \mathbf{a}_K]$ is the steering matrix where each vector \mathbf{a}_k is parameterized by the real scalar parameter θ_k to avoid unnecessary notational complexity. But the results presented here apply to a general parameterization. $\mathbf{x}_t = (x_{t,1}, \ldots, x_{t,K})^T$ and \mathbf{n}_t model signals transmitted by K sources and additive measurement noise respectively. \mathbf{x}_t and \mathbf{n}_t are multivariate independent, zero-mean, complex wide-sense stationary. \mathbf{n}_t is assumed Gaussian complex circular, spatially uncorrelated with $\mathbf{E}(\mathbf{n}_t\mathbf{n}_t^H) = \sigma_n^2\mathbf{I}_M$, while \mathbf{x}_t is complex circular or not, Gaussian or not and possibly spatially correlated or even coherent with $\mathbf{R}_x \stackrel{\text{def}}{=} \mathbf{E}(\mathbf{x}_t\mathbf{x}_t^H)$ and $\mathbf{R}_x' \stackrel{\text{def}}{=} \mathbf{E}(\mathbf{x}_t\mathbf{x}_t^T)$. Consequently this leads to the covariance matrices of \mathbf{y}_t :

$$\mathbf{R}(\Theta) = \mathbf{A}\mathbf{R}_x\mathbf{A}^H + \sigma_n^2\mathbf{I}_M$$
 and $\mathbf{R}'(\Theta) = \mathbf{A}\mathbf{R}_x'\mathbf{A}^T$.

 $(\mathbf{R}(\Theta), \mathbf{R}'(\Theta)) \text{ is generically parametrized by the } L = K + K^2 + K(K + 1) + 1 \text{ real parameters } \Theta = (\Theta_1, \Theta_2) \text{ with } \Theta_1 \stackrel{\text{def}}{=} (\theta_1, \dots, \theta_K)^T \text{ and } \Theta_2 \stackrel{\text{def}}{=} (\Re([\mathbf{R}_x]_{i,j}), \Re([\mathbf{R}_x']_{i,j}), \Re([\mathbf{R}_x']_{i,j}))_{1 \leq j < i \leq K}, ([\mathbf{R}_x]_{i,i}, \Re([\mathbf{R}_x']_{i,i}), \Re([\mathbf{R}_x']_{i,i}))_{i=1,\dots,K}, \sigma_n^2)^T.$

For performance analysis, some extra hypotheses are needed. The rank of \mathbf{R}_x is denoted \tilde{K} . Clearly $\tilde{K} \leq K$, and strict inequality implies linear dependence among the signal waveforms emanating from, e.g., specular multipath or smart jamming in communication applications. We suppose that the signal waveforms are linearly issued from \tilde{K} independent signals $(\tilde{x}_{t,k})_{k=1,\ldots,\tilde{K}}$, i.e., there exists a full colomn rank matrix \mathbf{B} such that $\mathbf{x}_t = \mathbf{B}\tilde{\mathbf{x}}_t$. The fourth-order cumulants of these \tilde{K} sources are denoted by $\kappa_{\tilde{x}_k} \stackrel{\text{def}}{=} \text{Cum}(\tilde{x}_{t,k}, \tilde{x}_{t,k}^*, \tilde{x$

We note that $\mathbf{s}(\Theta)$ is linear with respect to Θ_2 . Consequently (see e.g., [?]) there exists ¹ a known matrix $\Psi(\Theta_1)$ of the unknown DOA parameters Θ_1 :

$$\mathbf{s}(\Theta) = \mathbf{\Psi}(\Theta_1)\Theta_2.$$

Because, we suppose ² in this paper that Θ is identifiable from $(\mathbf{R}(\Theta), \mathbf{R}'(\Theta))$, Θ must be identifiable from $\mathbf{s}(\Theta)$, and necessarily $\Psi(\Theta_1)$ has column full rank [?]. In these conditions, the minimization (??) with respect to Θ_2 is immediate if Θ_2 is not restricted to be real. With a geometric procedure, we obtain:

$$\widehat{\Theta}_2 = [\mathbf{\Psi}^H(\Theta_1)\mathbf{W}\mathbf{\Psi}(\Theta_1)]^{-1}\mathbf{\Psi}^H(\Theta_1)\mathbf{W}\mathbf{s}_T$$
(3.1)

with $\mathbf{W} \stackrel{\text{def}}{=} \mathbf{C}_{s,T}^{-1}$. Because $\text{vec}(\mathbf{y}_t \mathbf{y}_t^H) = \mathbf{y}_t^* \otimes \mathbf{y}_t$ and $\text{v}(\mathbf{y}_t \mathbf{y}_t^T) = \mathbf{U}(\mathbf{y}_t \otimes \mathbf{y}_t)$, where \mathbf{U} is the $\frac{M(M+1)}{2} \times M^2$ selection matrix that satisfies $\text{v}(.) = \mathbf{U}\text{vec}(.)$ for all $M \times M$ matrices, \mathbf{s}_T can be written as

$$\mathbf{s}_T = rac{1}{T} \sum_{t=1}^T \mathbf{s}(t)$$
 with $\mathbf{s}(t) \stackrel{\mathrm{def}}{=} \left(egin{array}{c} \mathbf{y}_t^* \otimes \mathbf{y}_t \ \mathbf{U}(\mathbf{y}_t \otimes \mathbf{y}_t) \ \mathbf{U}(\mathbf{y}_t^* \otimes \mathbf{y}_t^*) \end{array}
ight).$

Consequently, \mathbf{s}_T is the mean of the T independent equidistributed random variables $\mathbf{s}(t)$. Therefore $\text{Cov}(\mathbf{s}_T) = \frac{1}{T}\text{Cov}(\mathbf{s}(t)) = \frac{1}{T}\text{E}\left[\left(\mathbf{s}(t) - \text{E}(\mathbf{s}(t))\right)\left(\mathbf{s}(t) - \text{E}(\mathbf{s}(t))\right)^H\right]$ and

$$\mathbf{C}_{s,T} = \frac{1}{T} \sum_{t=1}^{T} \left[\left(\mathbf{s}(t) - \frac{1}{T} \sum_{t=1}^{T} \mathbf{s}(t) \right) \left(\mathbf{s}(t) - \frac{1}{T} \sum_{t=1}^{T} \mathbf{s}(t) \right)^{H} \right]$$

 $[\]overline{{}^{1}}$ An explicit expression for $\Psi(\Theta_{1})$ will depend on the parameterization of $\mathbf{R}(\Theta)$ and $\mathbf{R}'(\Theta)$.

²We note that sufficient conditions for the identifiability will be application specific since they will depend on the structure of the array, the spatial correlation and the type of non-circularity of the sources.

is a consistent estimate of $\mathbf{C}_s(\Theta)$ structured as $\mathbf{s}_T \mathbf{s}_T^H$ for the real/imaginary part point of view. With arguments similar to that of COMET [?], we prove that $\widehat{\Theta}_2$ is real-valued.

Proof: If **J** denotes the linear invertible transformation that associates to \mathbf{s}_T , the real-valued vector γ_T comprised of the real and imaginary parts of \mathbf{s}_T , $\gamma_T = \mathbf{J}\mathbf{s}_T$ and $\widehat{\Theta}_2$ given by (??) assumes the form: $[(\mathbf{J}\Psi)^H(\mathbf{J}\mathbf{W}^{-1}\mathbf{J}^H)^{-1}(\mathbf{J}\Psi)^H(\mathbf{J}\mathbf{W}^{-1}\mathbf{J}^H)^{-1}\mathbf{J}\mathbf{s}_T$, where $\mathbf{J}\mathbf{s}_T$ is real and so is $\mathbf{J}\Psi$. It remains to examine $\mathbf{J}\mathbf{W}^{-1}\mathbf{J}^H$. Because $\mathbf{J}\mathbf{s}_T\mathbf{s}_T^H\mathbf{J}^H = \gamma_T\gamma_T^H$ is real-valued and because $\mathbf{C}_{s,T}$ is structured as $\mathbf{s}_T\mathbf{s}_T^H$, the matrix $\mathbf{J}\mathbf{W}^{-1}\mathbf{J}^H = \mathbf{J}\mathbf{C}_{s,T}\mathbf{J}^H$ is real-valued.

Thus $\widehat{\Theta}_2$ given by (??) is the real value that minimizes (??). $\Theta_{1,T}$ is obtained by subtituting $\widehat{\Theta}_2$ in (??):

$$\Theta_{1,T} = \arg \max_{\alpha_1 \in \mathcal{R}^K} V'(\alpha_1) \tag{3.2}$$

with

$$V'(\alpha_1) \stackrel{\text{def}}{=} \mathbf{s}_T^H \mathbf{W} \mathbf{\Psi}(\alpha_1) [\mathbf{\Psi}^H(\alpha_1) \mathbf{W} \mathbf{\Psi}(\alpha_1)]^{-1} \mathbf{\Psi}^H(\alpha_1) \mathbf{W} \mathbf{s}_T.$$

This COMET estimate is in general obtained by maximizing a multidimensional non-linear cost function. The reader interested in some implementational aspects (scoring technique, initialization of the multidimensional search, regularization of the sample covariance matrices...) may refer to [?].

To evaluate the improvement provided by the use of the covariance matrix \mathbf{R}_T' compared to the case in which only \mathbf{R}_T is considered, we first consider AMV second-order algorithms based on \mathbf{R}_T only.

4 Performance analysis

4.1 AMV estimator based on R_T only

We suppose here that Θ is identifiable from $\mathbf{R}(\Theta)$ only. In this case, the asymptotic minimum variance of the estimated parameters relies on the following standard central limit theorem applied to the independent equidistributed complex non-circular random variables $\mathbf{y}_t^* \otimes \mathbf{y}_t$. Thanks to simple algebraic manipulations of $\mathbf{C}_r = \mathbf{E}\left((\mathbf{y}_t^* \otimes \mathbf{y}_t - \text{vec}(\mathbf{R}(\Theta)))(\mathbf{y}_t^* \otimes \mathbf{y}_t - \text{vec}(\mathbf{R}(\Theta)))^H\right)$ and $\mathbf{C}_r' = \mathbf{E}\left((\mathbf{y}_t^* \otimes \mathbf{y}_t - \text{vec}(\mathbf{R}(\Theta)))(\mathbf{y}_t^* \otimes \mathbf{y}_t - \text{vec}(\mathbf{R}(\Theta)))^T\right)$, we straightforwardly prove:

Lemma 1 \sqrt{T} $(\text{vec}(\mathbf{R}_T) - \text{vec}(\mathbf{R}(\Theta)))$ converges in distribution to the zero-mean complex non-circular Gaussian

distribution of covariances \mathbf{C}_r and $\mathbf{C}'_r = \mathbf{C}_r \mathbf{K}$, where ³

$$\mathbf{C}_r = (\mathbf{A}^* \otimes \mathbf{A})\mathbf{C}_{r_x}(\mathbf{A}^T \otimes \mathbf{A}^H) + \sigma_n^4 \mathbf{I}_{M^2} + \sigma_n^2 \mathbf{I}_M \otimes \mathbf{A} \mathbf{R}_x \mathbf{A}^H + \mathbf{A}^* \mathbf{R}_x^* \mathbf{A}^T \otimes \sigma_n^2 \mathbf{I}_M$$
(4.3)

with 4

$$\mathbf{C}_{r_x} = \mathbf{R}_x^* \otimes \mathbf{R}_x + \mathbf{K}(\mathbf{R}_x' \otimes \mathbf{R}_x'^*) + \mathbf{Q}_x \quad and \quad \mathbf{Q}_x = (\mathbf{B}^* \otimes \mathbf{B}) \left(\sum_{k=1}^{\tilde{K}} \kappa_{\tilde{x}_k} (\mathbf{e}_{\tilde{K},k} \otimes \mathbf{e}_{\tilde{K},k}) (\mathbf{e}_{\tilde{K},k}^T \otimes \mathbf{e}_{\tilde{K},k}^T) \right) (\mathbf{B}^T \otimes \mathbf{B}^H).$$

First, we note that theorems 1-3 apply to the statistics \mathbf{r}_T , because a second-order algorithm based on \mathbf{R}_T only is a mapping $\mathbf{R}_T \to \Theta_T = \operatorname{alg}(\mathbf{R}_T)$ which is complex differentiable w.r.t. \mathbf{R}_T at the point $\mathbf{R}(\Theta)$ and the covariance $\mathbf{C}_r(\Theta)$ of the asymptotic distribution of \mathbf{R}_T is regular. By application of theorem ?? applied to the statistics \mathbf{r}_T , the covariance of the asymptotic distribution of the minimum variance second-order DOA estimator (??) based on \mathbf{R}_T only is given by the top left $K \times K$ "DOA corner" of $(\mathbf{S}^H\mathbf{C}_r^{-1}(\Theta)\mathbf{S})^{-1}$ where $\mathbf{C}_r(\Theta)$ is given by (??). If we note here that $\mathbf{S} \stackrel{\text{def}}{=} \frac{d\mathbf{r}}{d\Theta} = [\mathbf{S}_1, \boldsymbol{\Psi}]$ with $\mathbf{S}_1 \stackrel{\text{def}}{=} \frac{\partial \mathbf{r}}{\partial \Theta_1}$ and $\boldsymbol{\Psi}$ given by $\mathbf{r} = \boldsymbol{\Psi}(\Theta_1)\Theta_2$, the partitioned matrix inversion lemma gives

$$\begin{split} \left(\mathbf{S}^H \mathbf{C}_r^{-1} \mathbf{S}\right)_{(1:K,1:K)}^{-1} &= & \left(\mathbf{S}_1^H \mathbf{C}_r^{-1} \mathbf{S}_1 - \mathbf{S}_1^H \mathbf{C}_r^{-1} \boldsymbol{\Psi} \left[\boldsymbol{\Psi}^H \mathbf{C}_r^{-1} \boldsymbol{\Psi}\right]^{-1} \boldsymbol{\Psi}^H \mathbf{C}_r^{-1} \mathbf{S}_1\right)^{-1} \\ &= & \left(\mathbf{S}_1^H \mathbf{C}_r^{-1/2} \mathbf{P}_{\mathbf{C}_r^{-1/2} \boldsymbol{\Psi}}^{\perp} \mathbf{C}_r^{-1/2} \mathbf{S}_1\right)^{-1}, \end{split}$$

where $\mathbf{P}_{\mathbf{C}_r^{-1/2}\Psi}^{\perp}$ denotes the projector onto the ortho-complement of the colomns of $\mathbf{C}_r^{-1/2}\Psi$. Consequently, we prove the following theorem:

Theorem 4 For Gaussian or non Gaussian and complex circular or non-circular sources, the covariance of the asymptotic distribution of the minimum variance second-order DOA estimator based on \mathbf{R}_T only has the common closed-form expression:

$$\mathbf{C}_{\Theta_1} = \left(\mathbf{S}_1^H \mathbf{C}_r^{-1/2} \mathbf{P}_{\mathbf{C}_r^{-1/2} \mathbf{\Psi}}^{\perp} \mathbf{C}_r^{-1/2} \mathbf{S}_1\right)^{-1}.$$

$$(4.4)$$

This expression (??) extends to non Gaussian and/or complex non-circular sources, the expression of the asymptotic covariance given in [?] for Gaussian complex circular sources. On the other hand, we note that this expression is no longer equal to the Cramer-Rao bound because this AMV second-order estimator based on \mathbf{R}_T Because, $\operatorname{vec}^T(\mathbf{y}_t\mathbf{y}_t^H - \mathbf{R}(\Theta)) = \operatorname{vec}^H(\mathbf{y}_t\mathbf{y}_t^H - \mathbf{R}(\Theta))\mathbf{K}$, $\mathbf{C}'_r = \mathbf{C}_r\mathbf{K}$ and the non-circular complex Gaussian asymptotic distribution of \mathbf{R}_T is characterized by \mathbf{C}_r only.

⁴If the K sources are independent, \mathbf{Q}_x is reduced to $\mathbf{Q}_x = \sum_{k=1}^K \kappa_{x_k} (\mathbf{e}_{K,k} \otimes \mathbf{e}_{K,k}) (\mathbf{e}_{K,k}^T \otimes \mathbf{e}_{K,k}^T)$.

only is no longer efficient for non Gaussian and/or complex non-circular sources.

Remark 1: The expression of \mathbf{C}_{Θ_1} is generally sensitive to the non-circularity and the distribution of the sources. Furthermore, we note that a parameterization of \mathbf{R}_x and \mathbf{R}'_x may be introduced to incorporate a priori knowledge on the spatial correlation of the sources. For example, if the sources are supposed to be spatially uncorrelated, \mathbf{R}_x will be parameterized by $([\mathbf{R}_x]_{i,i})_{i=1,\dots,K}$ and if, moreover, they are independent, \mathbf{R}_x and \mathbf{R}'_x will be parameterized by $([\mathbf{R}_x]_{i,i}, \Re([\mathbf{R}'_x]_{i,i}))_{i=1,\dots,K}$ only. Consequently the expression of \mathbf{C}_{Θ_1} is generally sensitive to these a priori information as well.

Remark 2: Note that the derivative $\mathbf{D}_r^{\mathrm{AMV_1}}$ of the mapping which associates to \mathbf{R}_T , the estimate $\Theta_{1,T}$ depends on the non-circularity and the distribution of the sources through the expression of the weighting matrix \mathbf{C}_r^{-1} (see (??)). Consequently, the lemma proved in [?] which states that the constraints $\mathbf{D}_r^{\mathrm{AMV_1}}(\mathbf{A}^* \otimes \mathbf{A}) = \mathbf{O}$ or $\mathbf{D}_r^{\mathrm{AMV_1}}(\mathbf{a}_k^* \otimes \mathbf{a}_k) = \mathbf{0}, k = 1, \ldots, K$ that satisfy the derivative $\mathbf{D}_r^{\mathrm{AMV_1}}$ if the sources are not supposed to be spatially uncorrelated or respectively supposed spatially uncorrelated does not allow us to conclude that the expression of \mathbf{C}_{Θ_1} is generally unsensitive to the non-circularity and the distribution of the sources.

Remark 3: Note that in the particular case of one source, the numerical value of \mathbf{C}_{Θ} is block diagonal $\begin{bmatrix} C_{\Theta_1} & \mathbf{0}^T \\ \mathbf{0} & \mathbf{C}_{\Theta_2} \end{bmatrix}$ where C_{Θ_1} does not depend on the non-circularity and the distribution of the source, but we have not succeeded in proving these properties analytically.

4.2 AMV estimator based on (R_T, R_T)

To extend lemma $\ref{eq:total_total$

Lemma 2
$$\sqrt{T}$$
 $\left(\begin{array}{c} \operatorname{vec}(\mathbf{R}_T) - \operatorname{vec}(\mathbf{R}(\Theta)) \\ \operatorname{vec}(\mathbf{R}_T') - \operatorname{vec}(\mathbf{R}'(\Theta)) \end{array}\right)$ converges in distribution to the zero-mean complex non-circular

Gaussian distribution of covariances $\mathbf{C}_{r'} = \begin{pmatrix} \mathbf{C}_{r} & \mathbf{C}_{r,r'} \\ \mathbf{C}_{r,r'}^{H} & \mathbf{C}_{r'} \end{pmatrix}$ and $\mathbf{C'}_{r'} = \begin{pmatrix} \mathbf{C}_{r}\mathbf{K} & \mathbf{K}\mathbf{C}_{r,r'}^{*} \\ \mathbf{C}_{r,r'}^{H}\mathbf{K} & \mathbf{C}_{r'} \end{pmatrix}$ where \mathbf{C}_{r} is given by (??) and

$$\mathbf{C}_{r'} = (\mathbf{A} \otimes \mathbf{A}) \mathbf{C}_{r'_x} (\mathbf{A}^H \otimes \mathbf{A}^H) + \sigma_n^4 (\mathbf{I}_{M^2} + \mathbf{K})$$

$$+ (\mathbf{I}_{M^2} + \mathbf{K}) (\sigma_n^2 \mathbf{I}_M \otimes \mathbf{A} \mathbf{R}_x \mathbf{A}^H + \mathbf{A} \mathbf{R}_x \mathbf{A}^H \otimes \sigma_n^2 \mathbf{I}_M)$$

$$\mathbf{C}'_{r'} = (\mathbf{A} \otimes \mathbf{A}) \mathbf{C}'_{r'_x} (\mathbf{A}^T \otimes \mathbf{A}^T)$$

$$\mathbf{C}_{r,r'} = (\mathbf{A}^* \otimes \mathbf{A}) \mathbf{C}_{r_x,r'_x} (\mathbf{A}^H \otimes \mathbf{A}^H)$$

with

$$\mathbf{C}_{r_x'} = \mathbf{R}_x \otimes \mathbf{R}_x + \mathbf{K}(\mathbf{R}_x \otimes \mathbf{R}_x) + \mathbf{Q}_x$$

$$\mathbf{C}_{r_x'}' = \mathbf{R}_x' \otimes \mathbf{R}_x' + \mathbf{K}(\mathbf{R}_x' \otimes \mathbf{R}_x') + \mathbf{Q}_x'$$

$$\mathbf{C}_{r_x,r_x'} = \mathbf{R}_x'^* \otimes \mathbf{R}_x + \mathbf{K}(\mathbf{R}_x \otimes \mathbf{R}_x'^*) + \mathbf{Q}_x''$$

where \mathbf{Q}_x is given in lemma ?? and⁵

$$\mathbf{Q}_{x}' = (\mathbf{B} \otimes \mathbf{B}) \left(\sum_{k=1}^{\tilde{K}} \kappa_{\tilde{x}_{k}}'(\mathbf{e}_{\tilde{K},k} \otimes \mathbf{e}_{\tilde{K},k})(\mathbf{e}_{\tilde{K},k}^{T} \otimes \mathbf{e}_{\tilde{K},k}^{T}) \right) (\mathbf{B}^{T} \otimes \mathbf{B}^{T})$$

$$\mathbf{Q}_{x}'' = (\mathbf{B}^{*} \otimes \mathbf{B}) \left(\sum_{k=1}^{\tilde{K}} \kappa_{\tilde{x}_{k}}''(\mathbf{e}_{\tilde{K},k} \otimes \mathbf{e}_{\tilde{K},k})(\mathbf{e}_{\tilde{K},k}^{T} \otimes \mathbf{e}_{\tilde{K},k}^{T}) \right) (\mathbf{B}^{H} \otimes \mathbf{B}^{H}).$$

And thanks to the standard continuity theorem, the asymptotic behavior of \mathbf{s}_T and $(\mathbf{R}_T, \mathbf{R}_T')$ are directly related. Therefore lemma ?? extends to the statistic \mathbf{s}_T :

$$\sqrt{T} (\mathbf{s}_T - \mathbf{s}(\Theta)) \stackrel{\mathcal{L}}{\rightarrow} \mathcal{N}_c (\mathbf{0}; \mathbf{C}_s(\Theta), \mathbf{C}_s'(\Theta))$$

with

$$\mathbf{C}_{s}(\Theta) = \begin{pmatrix} \mathbf{C}_{r} & \mathbf{C}_{r,r'}\mathbf{U}^{T} & \mathbf{K}\mathbf{C}_{r,r'}^{*}\mathbf{U}^{T} \\ \mathbf{U}\mathbf{C}_{r,r'}^{H} & \mathbf{U}\mathbf{C}_{r'}\mathbf{U}^{T} & \mathbf{U}\mathbf{C}_{r'}^{'}\mathbf{U}^{T} \\ \mathbf{U}\mathbf{C}_{r,r'}^{T}\mathbf{K} & \mathbf{U}\mathbf{C}_{r'}^{H}\mathbf{U}^{T} & \mathbf{U}\mathbf{C}_{r'}^{*}\mathbf{U}^{T} \end{pmatrix}$$
 and
$$\mathbf{C}_{s}'(\Theta) = \mathbf{C}_{s}(\Theta)\mathbf{P}.$$
 (4.5)

⁵If the K sources are independent, \mathbf{Q}_x' and \mathbf{Q}_x'' are reduced to $\mathbf{Q}_x' = \sum_{k=1}^K \kappa_{x_k}' (\mathbf{e}_{K,k} \otimes \mathbf{e}_{K,k}) (\mathbf{e}_{K,k}^T \otimes \mathbf{e}_{K,k}^T)$ and $\mathbf{Q}_x'' = \sum_{k=1}^K \kappa_{x_k}'' (\mathbf{e}_{K,k} \otimes \mathbf{e}_{K,k}) (\mathbf{e}_{K,k}^T \otimes \mathbf{e}_{K,k}^T)$ respectively.

Consequently, theorem ?? extends to the minimum variance second-order DOA estimator (??) based on $(\mathbf{R}_T, \mathbf{R}_T')$ by direct application of theorem ??. Following the same procedure used to prove theorem ?? where here $\mathbf{S}_1 \stackrel{\text{def}}{=} \frac{\partial \mathbf{s}}{\partial \Theta_1}$, Ψ given by $\mathbf{s} = \Psi(\Theta_1)\Theta_2$ and $\mathbf{C}_r(\Theta)$ is replaced by $\mathbf{C}_s(\Theta)$ given in (??), we prove:

Theorem 5 For Gaussian or non Gaussian and complex circular or non-circular sources, the covariance of the asymptotic distribution of the minimum variance second-order DOA estimator based on \mathbf{R}_T and \mathbf{R}_T' has the common closed-form expression:

$$\mathbf{C}_{\Theta_1} = \left(\mathbf{S}_1^H \mathbf{C}_s^{-1/2} \mathbf{P}_{\mathbf{C}_s^{-1/2} \mathbf{\Psi}}^{\perp} \mathbf{C}_s^{-1/2} \mathbf{S}_1 \right)^{-1}. \tag{4.6}$$

Remark 1: If the sources are Gaussian complex non-circular, the stochastic maximum likelihood estimator is a second-order algorithm based on \mathbf{R}_T and \mathbf{R}_T' . Because it is asymptotically efficient, the closed-form expression (??) where the fourth order terms \mathbf{Q}_x , \mathbf{Q}_x' and \mathbf{Q}_x'' are canceled in $\mathbf{C}_s(\Theta)$ equals the Cramer-Rao bound on the DOA parameters alone in these conditions.

Remark 2: If the sources are complex circular up to the fourth order, $\mathbf{R}'_x = \mathbf{O}$, $\mathbf{Q}'_x = \mathbf{Q}''_x = \mathbf{O}$, and consequently

$$\mathbf{C}_{r,r'} = \mathbf{O} \text{ and } \mathbf{C}'_{r'} = \mathbf{O}.$$
 Therefore \mathbf{C}_s is block diagonal: $\mathbf{C}_s = \begin{pmatrix} \mathbf{C}_r & \mathbf{O} & \mathbf{O} \\ \mathbf{O} & \times & \mathbf{O} \\ \mathbf{O} & \mathbf{O} & \times \end{pmatrix}$. Consequently, the AMV of a

second-order algorithm based on $(\mathbf{R}_T, \mathbf{R}_T')$ given by theorem ?? reduces to

$$\mathbf{C}_{\Theta} = \left(\begin{pmatrix} d\mathbf{r}^{H} \\ d\Theta \end{pmatrix}, \mathbf{0}^{T} \right) \begin{pmatrix} \mathbf{C}_{r}^{-1} & \mathbf{O} \\ \mathbf{O} & \times \end{pmatrix} \begin{pmatrix} \frac{d\mathbf{r}}{d\Theta} \\ \mathbf{0} \end{pmatrix} \right)^{-1} = \left(\frac{d\mathbf{r}^{H}}{d\Theta} \mathbf{C}_{r}^{-1}(\Theta) \frac{d\mathbf{r}}{d\Theta} \right)^{-1},$$

which is the AMV given by a second-order algorithm based on \mathbf{R}_T only.

5 Simulations

In this section, numerical comparisons and Monte Carlo simulations are made between the AMV estimator based on \mathbf{R}_T only and the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$. This will give an indication of the information contributed by the second covariance matrix. The sources emit equipowered unfiltered BPSK modulated signals. We consider a uniform linear array of M=6 sensors separated by a half-wavelength for which $\mathbf{a}_k=(1,e^{i\theta_k},\ldots,e^{i(M-1)\theta_k})^T$ where $\theta_k=\pi\sin(\alpha_k)$ with α_k , the DOAs relative to the normal of the array.

In the first experiment, the two sources are independent and matrices \mathbf{R}_x and \mathbf{R}'_x are parameterized by their diagonal terms. Fig.1 exhibits the theoretical and empirical (averaged on 1000 independent Monte Carlo runs)

 $Var(\theta_{1,T})$ given by

- the AMV estimator based on \mathbf{R}_T only,
- $\bullet \;$ the AMV estimator based on $(\mathbf{R}_T,\mathbf{R}_T'),$
- $\bullet\,$ the MUSIC-like algorithm introduced in [?] 6
- the standard MUSIC algorithm,

versus the signal noise ratio for $\theta_2 - \theta_1 = 0.2$ rd and T = 500. This figure shows a good agreement between the theoretical and empirical curves and we notice that the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$ outperforms the AMV estimator based on \mathbf{R}_T only, for all values of the signal noise ratio. Naturally, the AMV estimators based on \mathbf{R}_T only and $(\mathbf{R}_T, \mathbf{R}_T')$ perform better than the MUSIC algorithms based on respectively \mathbf{R}_T only and $(\mathbf{R}_T, \mathbf{R}_T')$. Fig.2 exhibits the theoretical normalized asymptotic variance $[\mathbf{C}_{\Theta_1}]_{1,1}$ given by the AMV estimator based on \mathbf{R}_T only and the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$, versus the DOA separation for a SNR of 10dB. The AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$ clearly outperforms the AMV estimator based on \mathbf{R}_T only, and the difference is particularly prominent when the sources are very close.

⁶Because no performance study is available in the literature, only the empirical $Var(\theta_{1,T})$ is plotted for this algorithm.

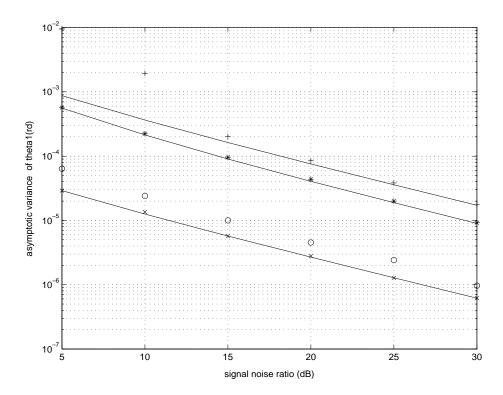


Fig.1 Theoretical and empirical $Var(\theta_{1,T})$ given by the AMV estimator based on \mathbf{R}_T only (\star) , by the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$ (\times) , by the MUSIC-like algorithm given in [?] (o) and by the standard MUSIC algorithm (+) versus the signal noise ratio.

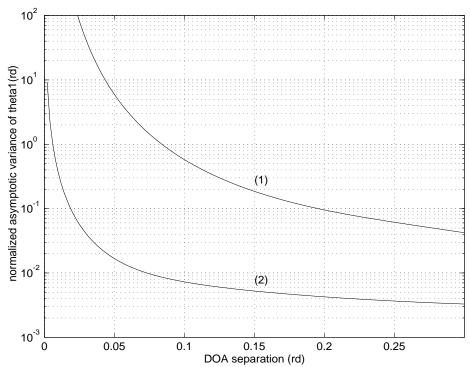


Fig.2 Theoretical normalized asymptotic variance of $\theta_{1,T}$ ($[\mathbf{C}_{\Theta_1}]_{1,1}$) given by the AMV estimator based on $(\mathbf{R}_T \text{ only } (1)$ and the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$ (2), versus the DOA separation.

In the second experiment, we select a scenario where the second covariance matrix contributes almost no additional information beyond the information in the first covariance matrix. We consider two spatially correlated waveforms including coherence. The matrices \mathbf{R}_x and \mathbf{R}'_x are parameterized by the real and imaginary parts of their entries (i.e., by $\Re([\mathbf{R}_x]_{2,1}), \Im([\mathbf{R}_x]_{2,1}), \Im([\mathbf{R}'_x]_{2,1}), \Im([\mathbf{R}'_x]_{2,1}), ([\mathbf{R}_x]_{i,i}, \Re([\mathbf{R}'_x]_{i,i}), \Im([\mathbf{R}'_x]_{i,i}))_{i=1,2})$. We suppose the signals consist of two equipowered multipaths issued from the DOAs θ_1 and θ_2 . Referenced on the first sensor and from the DOA θ_1 , we have equivalently: $x_{t,1} = \tilde{x}_{t,1}$ and $x_{t,2} = \cos(\alpha)\tilde{x}_{t,1} + \sin(\alpha)\tilde{x}_{t,2}$ with $\mathbf{R}_{\tilde{x}} = \sigma_1^2\mathbf{I}_2$ and $\mathbf{R}'_{\tilde{x}} = \sigma_1^2\left(\begin{array}{cc} e^{i2\phi_1} & 0 \\ 0 & e^{i2\phi_2} \end{array}\right)$. Consequently

$$\mathbf{R}_{x} = \sigma_{1}^{2} \begin{pmatrix} 1 & \cos(\alpha) \\ \cos(\alpha) & 1 \end{pmatrix} \quad \text{and} \quad \mathbf{R}_{x}' = \sigma_{1}^{2} \begin{pmatrix} e^{2i\phi_{1}} & \cos(\alpha)e^{2i\phi_{1}} \\ \cos(\alpha)e^{2i\phi_{1}} & \cos^{2}(\alpha)e^{2i\phi_{1}} + \sin^{2}(\alpha)e^{2i\phi_{2}} \end{pmatrix}.$$

Fig.3 exhibits the theoretical normalized asymptotic variance $[\mathbf{C}_{\Theta_1}]_{1,1}$ given by the AMV estimator based on \mathbf{R}_T only and the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}'_T)$, versus the DOA separation for uncorrelated $(\alpha = \frac{\pi}{2})$ and coherent $(\alpha = 0)$ sources, for a SNR of 10dB. We see that the AMV estimators based on \mathbf{R}_T and on $(\mathbf{R}_T, \mathbf{R}'_T)$ have the same performance with coherent signals, whereas the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}'_T)$ slightly outperforms the AMV estimator based on \mathbf{R}_T for uncorrelated sources. Compared with Fig.1, we see the crucial role of the parameterization of \mathbf{R}_x and \mathbf{R}'_x . If the sources are known to be uncorrelated, we must parameterize these matrices by their diagonal only to benefit from the second covariance matrix.

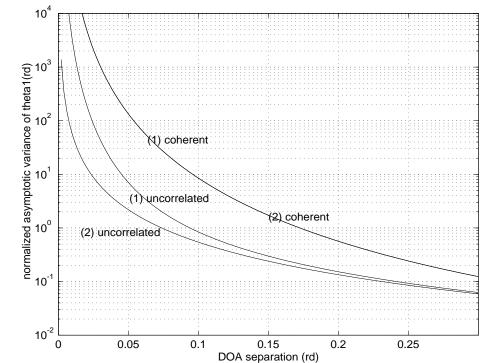


Fig.3 Theoretical normalized asymptotic variance of $\theta_{1,T}$ ($[\mathbf{C}_{\Theta_1}]_{1,1}$) given by the AMV estimator based on \mathbf{R}_T only (1) and the AMV estimator based on $(\mathbf{R}_T, \mathbf{R}_T')$ (2) for uncorrelated or coherent sources, versus the DOA separation.

6 Conclusion

This paper has introduced asymptotically minimum variance algorithms in the class of algorithms based on secondorder statistics for estimating DOA parameters of possibly spatially correlated even coherent narrowband noncircular sources impinging on arbitrary array structures. The performance of the proposed algorithms were evaluated by closed-form expressions of the asymptotic covariance of the DOA estimates which can be used as a lower
bound for assessing the performance of any suboptimal second-order algorithms. These asymptotic covariances
were numerically compared with that obtained by AMV algorithms based on the first covariance matrix only. We
have then realized that the expected benefits due to the non-circular property mainly happens for uncorrelated
sources and furthermore if the parameterization takes this information into account. Naturally, these conclusions
must be mitigated because a thorough comparison between these two AMV algorithms would need a large quantity
of scenarios (various geometry arrays, number of sources, non-circularity, correlation and SNR).

An issue which was not addressed in this paper is the sufficient conditions that guarantee the identifiability of the DOA parameters from the two covariance matrices for non-circular signals. This crucial question is not trivial and it is in fact application specific since it depends on the structure of the array, the spatial covariance and the type of non-circularity of the sources. A study to deal with this issue is underway.

References

- [1] L.C. Godara, "Application of antenna arrays to mobile communications, Part II: beamforming and direction of arrival considerations," *Proceedings of the IEEE*, vol. 85, pp. 1193-1245, August 1997.
- [2] B. Picinbono, "On circularity," IEEE Trans. Signal Processing, vol. 42, no. 12, pp. 3473-3482, December 1994.
- [3] P. Ciblat, P. Loubaton, E. Serpedin and G. Giannakis, "Performance analysis of blind carrier frequency offset estimators for non-circular transmissions through frequency-selective channels," *IEEE Trans. on Signal Processing*, vol. 50, no. 1, pp. 130-140, January 2001.

- [4] O. Grellier, P. Comon, B. Mourrain and P. Trebuchet, "Analytical blind channel identification," *IEEE Trans. on Signal Processing*, vol. 50, no. 9, pp. 2196-2207, September 2002.
- [5] P. Gounon, C. Adnet and J. Galy, "Angular location for non-circular signals," Traitement du Signal, vol. 15, no. 1 pp. 17-23, 1998.
- [6] P. Chargé, Y. Wang and J. Saillard, "A non-circular sources direction finding method using polynomial rooting," Signal Processing, vol. 81, pp. 1765-1770, 2001.
- [7] B. Porat and B. Friedlander, "Performance analysis of parameter estimation algorithms based on high-order moments," *International journal of adaptive control and signal processing*, vol. 3, pp. 191-229, 1989.
- [8] B. Ottersten, P. Stoica and R. Roy, "Covariance matching estimation techniques for array signal processing," Digital Signal Processing, vol. 8, pp. 185-210, July 1998.
- [9] B. Friedlander and B. Porat, "Asymptotically optimal estimation of MA and ARMA parameters of Non-Gaussian processes from high-order moments," IEEE Trans. Automatic Control, vol. 35, no. 1, pp. 27-35, January 1990.
- [10] J.P. Delmas, "Asymptotic performance of second-order algorithms," IEEE Transactions on Signal Processing, vol. 50, no. 1, pp. 49-57, January 2002.