THE POLE BEHAVIOUR OF THE PHASE DERIVATIVE OF THE SHORT-TIME FOURIER TRANSFORM

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ABSTRACT. The short-time Fourier transform (STFT) is a timefrequency representation widely used in applications, for example in audio signal processing. Recently it has been shown that not only the amplitude, but also the phase of this representation can be successfully exploited for improved analysis and processing. In this paper we describe a rather peculiar pole phenomenon in the phase derivative, a recurring pattern that appears in a characteristic way in the neighborhood around any of the zeros of the STFT, a negative peak followed by a positive one. We describe this phenomenon numerically and provide a complete analytical explanation.

1. INTRODUCTION

The short-time Fourier transform (STFT) [5, 13] is a time-frequency representation widely used in audio signal processing. A common definition of the STFT^1 is

(1)
$$V(f,g)(x,\omega) = \int f(t)\overline{g(t-x)}e^{-2\pi i\omega t} dt.$$

The STFT $V(f,g)(x,\omega)$ provides information about the frequency content of the signal f at time x and frequency ω . The analyzing window g determines the resolution in time and frequency.

The interpretation of the modulus of the STFT is relatively easy, considering the fact that the spectrogram (defined as the square absolute value of the STFT) can be interpreted as a time-frequency distribution of the signal energy. This interpretation led to the important success of the STFT in signal processing. In particular, it has been widely used

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¹This is the *frequency-invariant* STFT.

for applications in speech processing and acoustics as a graphical tool for signal analysis [21].

But the interpretation of the phase of the STFT is less obvious, and was thus hardly considered in applications for some time.

The phase can be of particular interest for certain applications, as illustrated by important applications such as phase vocoder [12, 7] or reassignment [18, 2]. In digital image processing it is well known that the phase information of the discrete Fourier transform is at least as important as the amplitude information. In [19] it is shown that as long as the phase of the discrete Fourier transform of an image is retained and the amplitude is set to 1, the image can still be recognized. Similar effects can also be shown for acoustic signal depending on the parameters of the STFT [4].

For applications modifying the STFT coefficients, phase information is essential again. For these types of applications, in particular for the applications using Gabor frame multipliers [10, 3] which motivated the present study, better understanding of the structure of the phase is necessary to improve the processing possibilities.

The phase of the STFT is usually not considered directly. In fact, it is more interesting to consider the phase derivative over time or frequency. Indeed, these quantities appear naturally in the context of reassignment [2] and manipulations of phase derivative over time is the idea behind the phase vocoder [7]. Their interpretation is easier, as the derivative of phase over time can be interpreted as local instantaneous frequency while the derivative of the phase over frequency can be interpreted as a local group delay.

To numerically compute the local instantaneous frequency, an unwrapping of the phase is needed to avoid discontinuities. This is the classical method used in [7, 18]. Another method was found in [2]:

(2)
$$\frac{\partial}{\partial x} \arg(V(f,g)(x,\omega)) = \operatorname{Im}\left(\frac{V(f,g')(x,\omega)\overline{V(f,g)(x,\omega)}}{|V(f,g)(x,\omega)|^2}\right),$$

with $g'(t) = \frac{dg}{dt}(t)$. The benefit of this method is that is does not require unwrapping, instead the phase derivative is computed by pointwise operations using a second STFT based on the derivative of the window.

To understand the phase of the STFT more thoroughly, in particular for applications dealing with multipliers, see for example [20, 22, 23], we conducted related extensive numerical experiments. In the process we observed a rather peculiar phenomenon in the phase derivative, a recurring pattern that appears in a similar way in the neighborhood around any of the zeros of the STFT. The behaviour of the phase derivative close to the singularity always shows the same characteristic shape, i.e., a negative peak followed by a positive one. We describe this phenomenon and provide a complete analytical explanation.

This paper is organized as follows: In Section 2 we report the numerical results. In Section 3 we give a short, instructive, analytical example for this behaviour. In Section 4 we give the full analytical results.

Results in this paper have partly been reported at a conference [17], and a preprint of this paper has already been cited in [1].

2. NUMERICAL OBSERVATIONS

For noise, naturally only statistical properties of the phase are accessible. Some interesting results for the phase derivative have been shown in the context of reassignment. In [8], the following result is given: We consider a zero-mean Gaussian analytic white noise f such that

(3)
$$\operatorname{E}[\operatorname{Re}(f(t)) \cdot \operatorname{Re}(f(s))] = \operatorname{E}[\operatorname{Im}(f(t)) \cdot \operatorname{Im}(f(s))] = \frac{\sigma^2}{2}\delta(t-s)$$

and E[f(t)f(s)] = 0 for any $(t,s) \in \mathbb{R}^2$, with its real and imaginary parts a Hilbert transform pair. Using a Gaussian window given by $g(t) = e^{-\pi \frac{t^2}{2\sigma^2}}$, the phase derivative over time of V(f,g) is a random variable with distribution of the form:

(4)
$$\rho(v) = \frac{1}{2(1+v^2)^{\frac{3}{2}}}$$

This distribution is shown in Figure 1. As can be seen, it is a quite "peaky" distribution, indicating that the values of the phase derivative are mainly values close to zero, with some rare values with higher absolute values.

The spatial distribution of the phase derivative seems to be difficult to be solved analytically. Therefore we conducted systematic numerical experiments to study this spatial distribution.

For this, we need to compute the derivative of the phase in discrete settings. We used the expression (2) to compute the phase derivative.

We see on this formula that we will face numerical difficulties when the denominator $V(f, g)(x, \omega)$ is close to zero. But using double precision, these problems only appear for really small values of the modulus



FIGURE 1. Distribution of the values of the phase derivative over time of the STFT for a white Gaussian noise.

(on the order of 10^{-13}), which allows us to reliably observe the values of the phase derivative even close to the zeros of the STFT. In the figures of this paper, the phase derivative values are ignored and represented as white at the points where the value of the modulus is too small.

The results of our experiments are illustrated by Figure 2. The timefrequency distribution of the values appears to be highly structure, as e.g. noted in [15]. The values of the phase derivative with high absolute values are concentrated around several time-frequency points, which can be identified as the zeros of the transform when looking at the modulus. Furthermore, the shape of the phase derivative seems to be very similar in the neighbourhood of the zeros, with a typical pattern repeating at each zero, see Figure 2. When going from low to high frequencies, it presents a negative peak followed by a positive one.

This phenomenon is related to the fact that the STFT of white noise is a correlated process, with a correlation determined by the window through the reproducing kernel of the transform (see part 6.2.1 of [5]). It is thus interesting to study the influence of the window choice on the observed structure of the phase derivative, as illustrated in Figure 3.

We can observe that narrowing the window results in similar patterns around the zeros, but with a scaled shape: the resulting pattern is narrower over time, but wider over frequency. Figure 3 also shows the influence of the window type. The structure is more complicated for windows with bad time-frequency concentration. On the representation using a Hamming window, we still observe repeating patterns at the zeros of the transform, but the variability of the shape of this pattern seems higher, and the pattern orientation slightly varies, whereas it is fixed in the case of a Gaussian window. For the case of the rectangular window, the zeros of the STFT form a more complicated, extended structures. This leads to much more variable patterns. Yet, we still, interestingly, observe that the values of the phase derivative with high



FIGURE 2. Observation for a Gaussian white noise, using a Gaussian window. Top: modulus of the STFT. Bottom-left: derivative over time of the phase of the STFT using the definition (1). Bottom-right: mesh plot of the derivative over time of the phase in the neighbourhood of a zero of the STFT.

absolute values concentrate around the zeros of the transform, whereas the phase derivative is close to zero in the regions of the STFT where the modulus is high.

¿From the experiments above, we expect those properties to be highly correlated with co-orbit properties regarding the STFT, i.e. inclusion in certain modulation spaces [9]. In particular, choosing windows in the Feichtinger algebra S_0 [11] should result in a phase derivative behavior comparable to the Gaussian window. We also expect results to be valid as in Section 4, for windows in S_0 . The systematic investigation of the phase derivative behaviour for modulation spaces is beyond the scope of this paper, and will be investigate in future work.

The behaviour that we observe is not specific to noise signals. Indeed, further experiments on other synthesized and recorded complex sounds showed that the same characteristics can be observed for all signals: the values of the phase derivative of high absolute value are concentrated in



FIGURE 3. Influence of the window when analyzing a frozen Gaussian white noise. For three different windows, on the left, modulus of the STFT, on the right, derivative over time of the phase of the STFT using the definition (1). From top to bottom, the windows are: a narrower Gaussian window, a Hamming window, a rectangular window.

the neighbourhood of the zeros of the STFT, and for "nice" windows, a specific pattern appears in this neighbourhood.

3. A Simple Explicit Analytic Example

In this section we give a simple analytical example for which we can explicitly compute the phase derivative. Considering the signal given by

(5)
$$f(t) = e^{2\pi i\omega_1 t} + e^{2\pi i\omega_2 t}$$

and using a Gaussian window $g(t) = e^{-\pi \frac{t^2}{2\sigma^2}}$, we can explicitly compute the expression of the STFT, which results in the formula: (6)

$$V(f,g)(x,\omega) = e^{-2\pi i x(\omega-\omega_1)} e^{-2\pi\sigma^2(\omega-\omega_1)^2} + e^{-2\pi i x(\omega-\omega_2)} e^{-2\pi\sigma^2(\omega-\omega_2)^2}.$$

The zeros of this STFT are the points of coordinates (x_k, ω_{mid}) in the time-frequency plane, with $\omega_{mid} = \frac{\omega_1 + \omega_2}{2}$ and $x_k = \frac{1+2k}{2(\omega_1 - \omega_2)}$ for $k \in \mathbb{Z}$.

The expression of the phase derivative for this signal, given in part VI-12 of [6], is:

(7)

$$\frac{\partial}{\partial x} \arg(V(f,g)(x,\omega)) = 2\pi \left(\omega_{mid} - \omega + \delta \tanh(s) \frac{1 + \tan^2(2\pi\delta x)}{1 + \tan^2(2\pi\delta x) \tanh^2(s)}\right)$$
with $\delta = \frac{\omega_2 - \omega_1}{2}$ and $s = 4\pi\sigma^2(\omega - \omega_{mid})\delta$.

The plot of this function around one of the zeros is visible in Figure 4. We see the pattern that was already observed in the previous section.



FIGURE 4. Observation for the signal defined in (5). Top: modulus of the STFT. Bottom: derivative over time of the phase of the STFT according to (7) represented as an image (left) and as a mesh (right).

For an extended treatment of this analytic example, see [1], where the authors are already referring to a preprint of our present work.

4. Analytical Results

In this section, we denote by M_h the modulation operator M_h : $L^2(\mathbb{R}) \to L^2(\mathbb{R}), f(t) \mapsto M_h f(t) := e^{-2\pi i h t} f(t)$, and by T_h the translation operator $T_h : L^2(\mathbb{R}) \to L^2(\mathbb{R}), f(t) \mapsto T_h f(t) := f(t-h)$ (with $h \in \mathbb{R}$). The set $\mathcal{S}(\mathbb{R})$ is the Schwartz class of rapidly decaying functions. The Fourier transform is denoted by \mathcal{F} .

4.1. Regularity Properties of the STFT.

Definition 4.1. Define the (unbounded) operator P on $L^2(\mathbb{R})$ as the multiplication operator

$$Pf(t) := 2\pi i t \cdot f(t)$$

with domain

$$Dom(P) := \{ f \in L^2(\mathbb{R}) : \int_{\mathbb{R}} |t f(t)|^2 dt < \infty \} \subset L^2(\mathbb{R}).$$

Further, define the (unbounded) operator

 $Q := \mathcal{F}^{-1} P \mathcal{F}$

(where \mathcal{F} denotes the Fourier transform) with domain

$$Dom(Q) := \{ f \in L^2(\mathbb{R}) : \mathcal{F}f \in Dom(P) \} \subset L^2(\mathbb{R}).$$

In quantum mechanics, these operators are essentially the momentum and position operator, respectively. The operator P (and thus also Q) are clearly densely-defined, since $\mathcal{S}(\mathbb{R}) \subset \text{Dom}(P)$ (and $\mathcal{F}^{-1}\mathcal{S}(\mathbb{R}) =$ $\mathcal{S}(\mathbb{R}) \subset \text{Dom}(Q)$). It can be shown that P and Q are closed unbounded operators and that iP and iQ are self-adjoint.

We collect basic properties of these operators in the following lemma.

Lemma 4.2. The operators P and Q have the following properties:

- (i) $\mathcal{F}Q = P\mathcal{F}$ on Dom(Q), $Q\mathcal{F} = -\mathcal{F}P$ on Dom(P);
- (ii) Q is a (maximal extension of a) differential operator, more precisely: if $f \in \mathcal{S}(\mathbb{R})$, then $Qf(t) = \frac{d}{dt}f(t)$.

The next lemma is in essence a well-known result from the theory of operator (semi-)groups; it gives the infinitesimal generators of the modulation and translation group, respectively, see e.g. [14, 16]. The version below needed in this manuscript can be proved in a straightforward way. **Lemma 4.3.** Let $f \in Dom(P)$, then

$$\|\frac{1}{h}(M_h - Id)f - Pf\|_{L^2} \to 0$$

for $h \to 0$. Let $f \in Dom(Q)$, then

$$\|\frac{1}{h}(T_h - Id)f + Qf\|_{L^2} \to 0$$

for $h \to 0$.

We can now prove a regularity result for the short-time Fourier transform.

Proposition 4.4. Let $f, g \in L^2(\mathbb{R})$.

 (i) If f belongs to Dom(P), then V(f, g) has a continuous partial derivative with respect to the second argument ω, and we have

$$\frac{\partial}{\partial \omega} V(f,g)(x,\omega) = -V(Pf,g)(x,\omega).$$

(ii) If g belongs to Dom(Q), then V(f,g) has a continuous partial derivative with respect to the first argument x, and we have

$$\frac{\partial}{\partial x}V(f,g)(x,\omega) = -V(f,Qg)(x,\omega).$$

Proof. Assume $f \in \text{Dom}(P)$. Then, by the preceding lemma,

$$\begin{aligned} \frac{1}{h}(V(f,g)(x,\omega+h) - V(f,g)(x,\omega)) &= \langle f, \frac{M_h - Id}{h} M_\omega T_x g \rangle \\ &= \langle \frac{M_{-h} - Id}{h} f, M_\omega T_x g \rangle \\ &\xrightarrow{h \to 0} \langle -Pf, M_\omega T_x g \rangle = -V(Pf,g)(x,\omega), \end{aligned}$$

which is a continuous function on \mathbb{R}^2 . If $g \in \text{Dom}(Q)$, then, analogously,

$$\frac{1}{h}(V(f,g)(x+h,\omega) - V(f,g)(x,\omega)) = \langle f, M_{\omega}T_x \frac{T_h - Id}{h}g \rangle$$
$$\xrightarrow{h \to 0} \langle f, -M_{\omega}T_x Qg \rangle = -V(f, Qg)(x,\omega).$$

Using $V(f,g)(x,\omega) = e^{-2\pi i x \omega} \overline{V(g,f)(-x,-\omega)}$, the partial derivatives of V(f,g) exist if and only if those of V(g,f) exist. We may thus change the roles of f and g.

Corollary 4.5. Let $f, g \in L^2(\mathbb{R})$.

 (i) If g belongs to Dom(P), then V(f, g) has a continuous partial derivative with respect to the second argument ω, and we have

$$\frac{\partial}{\partial \omega} V(f,g)(x,\omega) = V(f,Pg)(x,\omega) - 2\pi i x V(f,g)(x,\omega).$$

(ii) If f belongs to Dom(Q), then V(f,g) has a continuous partial derivative with respect to the first argument x, and we have

$$\frac{\partial}{\partial x}V(f,g)(x,\omega) = V(Qf,g)(x,\omega) - 2\pi i\omega V(f,g)(x,\omega).$$

If f (resp. g) belongs to Schwartz class $\mathcal{S}(\mathbb{R})$, then Pf, Qf (resp. $Pg, Qg) \in \mathcal{S}(\mathbb{R})$, as well. Iterated application of Proposition 4.4 or Corollary 4.5 gives the following smoothness result for the STFT.

Theorem 4.6. Let $f, g \in L^2(\mathbb{R})$, and at least one of them in Schwartz class $\mathcal{S}(\mathbb{R})$. Then $V(f, g)(x, \omega)$ is infinitely partially differentiable in both variables x and ω .

Although this result may be considered mathematical folklore, to our knowledge it has not been stated and proved in the literature so far. Note that this proves in particular that the STFT with Gaussian window is smooth.

4.2. The Derivative of the Phase Around the Zeros of the STFT. In this section we present an analytic explanation of the peculiar behaviour of the phase derivatives of the STFT for a large class of window functions. It turns out that the phenomenon is connected to the smoothness and continuous differentiability of the STFT which, as we have seen in the previous paragraph, is in turn connected to the smoothness of the window.

Consider first the partial derivative of the phase of the STFT with respect to the first variable (i.e., the 'time' variable). For convergence along a vertical path, we have

Theorem 4.7 (Phase derivatives of the STFT, part I). Let $f, g \in L^2(\mathbb{R})$. Assume that

- $V(f,g) = V = U + i \cdot W \in C^2(\mathbb{R}^2)$
- $V(x_0,\omega_0)=0$

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• det $J_V(x_0, \omega_0) < 0$, where

$$J_{V}(x_{0},\omega_{0}) = \begin{pmatrix} \frac{\partial U}{\partial x}(x_{0},\omega_{0}) & \frac{\partial U}{\partial \omega}(x_{0},\omega_{0}) \\ \frac{\partial W}{\partial x}(x_{0},\omega_{0}) & \frac{\partial W}{\partial \omega}(x_{0},\omega_{0}) \end{pmatrix}_{10}$$

Then the phase $\psi(x,\omega)$ of $V(f,g)(x,\omega)$ satisfies

$$\frac{\partial \psi}{\partial x}(x_0,\omega) \longrightarrow \begin{cases} +\infty, & \text{if } \omega \uparrow \omega_0 \text{ from below} \\ -\infty, & \text{if } \omega \downarrow \omega_0 \text{ from above.} \end{cases}$$

Proof. We have

$$\frac{\partial \psi}{\partial x}(x_0,\omega) = \frac{U(x_0,\omega) \cdot W_x(x_0,\omega) - W(x_0,\omega) \cdot U_x(x_0,\omega)}{U^2(x_0,\omega) + W^2(x_0,\omega)}.$$

Since W_x and U_x are continuous and thus remain bounded in a neighbourhood of (x_0, ω_0) , both numerator and denominator tend to zero for $\omega \to \omega_0$. However, both functions are differentiable, since $V \in C^2(\mathbb{R}^2, \mathbb{R}^2)$. So L'Hospital's Rule is applicable and yields

$$\begin{split} \lim_{\omega \to \omega_0} \frac{U(x_0, \omega) \cdot W_x(x_0, \omega) - W(x_0, \omega) \cdot U_x(x_0, \omega)}{U^2(x_0, \omega) + W^2(x_0, \omega)} \\ \stackrel{\text{L'Hosp.}}{=} \lim_{\omega \to \omega_0} \frac{\frac{d}{d\omega} \left(U(x_0, \omega) \cdot W_x(x_0, \omega) - W(x_0, \omega) \cdot U_x(x_0, \omega) \right)}{\frac{d}{d\omega} \left(U^2(x_0, \omega) + W^2(x_0, \omega) \right)} \\ = \lim_{\omega \to \omega_0} \frac{\left(U_\omega W_x + U W_{x\omega} - W_\omega U_x - W U_{x\omega} \right) (x_0, \omega)}{\left(2U U_\omega + 2W W_\omega \right) (x_0, \omega)} \\ = \lim_{\omega \to \omega_0} \frac{\left(U W_{x\omega} - W U_{x\omega} \right) (x_0, \omega) + \left(U_\omega W_x - W_\omega U_x \right) (x_0, \omega)}{\left(2U U_\omega + 2W W_\omega \right) (x_0, \omega)}. \end{split}$$

Concerning this limit, we clearly have

$$\left(UW_{x\omega} - WU_{x\omega}\right)(x_0,\omega) \to 0,$$

since $U(x_0, \omega) \to U(x_0, \omega_0) = 0$, $W(x_0, \omega) \to W(x_0, \omega_0) = 0$, and $W_{x\omega}$ and $U_{x\omega}$ are continuous and thus remain bounded in a neighborhood of (x_0, ω_0) . Furthermore,

$$(U_{\omega}W_x - W_{\omega}U_x) (x_0, \omega) \to (U_{\omega}W_x - W_{\omega}U_x) (x_0, \omega_0)$$

= $-\det J_V(x_0, \omega_0) \neq 0,$

by assumption. Hence the numerator tends to a nonzero number, in this case (det $J_V(x_0, \omega_0) < 0$) a positive one. For the denominator, we find

$$(2UU_{\omega} + 2WW_{\omega})(x_0, \omega) = \frac{d}{d\omega} (U^2 + W^2)(x_0, \omega)$$
$$= \begin{cases} < 0, & \text{if } \omega < \omega_0 \\ > 0, & \text{if } \omega > \omega_0 \end{cases}$$

since the function $\omega \mapsto (2UU_{\omega} + 2WW_{\omega})(x_0, \omega)$ has a strict local minimum in ω_0 . At the same time,

$$(2UU_{\omega} + 2WW_{\omega})(x_0, \omega) \to 0$$
¹¹

for $\omega \to \omega_0$, hence the denominator goes to zero from below for $\omega \uparrow \omega_0$ and from above for $\omega \downarrow \omega_0$. This concludes the proof.

Note that for simplicity we have only considered the case that the Jacobian determinant det $J_V(x_0, \omega_0)$ is negative; this case corresponds to the examples we presented above. For positive Jacobian determinant, the situation is completely analogous, although reversed in the sense that the positive and negative singularities switch roles. Apart from this, the general behaviour remains the same.

For convergence along a horizontal path, we need slightly more regularity:

Theorem 4.8 (Phase derivatives of the STFT, part II). Let $f, g \in L^2(\mathbb{R})$. Assume that

- $V(f,g) = V = U + i \cdot W \in C^3(\mathbb{R}^2)$
- $V(x_0, \omega_0) = 0$
- det $J_V(x_0, \omega_0) < 0$, where J is the Jacobian as in Theorem 4.7

Then there exists a number $c \in \mathbb{R}$ such that the phase $\psi(x,\omega)$ of $V(f,g)(x,\omega)$ satisfies

$$\lim_{x \to x_0} \frac{\partial \psi}{\partial x}(x, \omega_0) = c.$$

Proof. The assumptions allow us to apply L'Hospital's Rule twice, giving

$$\begin{split} \lim_{x \to x_0} \frac{\partial \psi}{\partial x}(x, \omega_0) \\ &= \lim_{x \to x_0} \frac{U(x, \omega_0) \cdot W_x(x, \omega_0) - W(x, \omega_0) \cdot U_x(x, \omega_0)}{U^2(x, \omega_0) + W^2(x, \omega_0)} \\ & \overset{\text{L'Hosp.}}{=} \lim_{x \to x_0} \frac{\frac{d}{dx} \left(U(x, \omega_0) \cdot W_x(x, \omega_0) - W(x, \omega_0) \cdot U_x(x, \omega_0) \right)}{\frac{d}{dx} \left(U^2(x, \omega_0) + W^2(x, \omega_0) \right)} \\ &= \lim_{x \to x_0} \frac{\left(U_x W_x + U W_{xx} - W_x U_x - W U_{xx} \right) (x, \omega_0)}{(2UU_x + 2WW_x) (x, \omega_0)} \\ &= \lim_{x \to x_0} \frac{\left(U W_{xx} - W U_{xx} \right) (x, \omega_0)}{(2UU_x + 2WW_x) (x, \omega_0)} \\ & \overset{\text{L'Hosp.}}{=} \lim_{x \to x_0} \frac{\frac{d}{dx} \left(U W_{xx} - W U_{xx} \right) (x, \omega_0)}{\frac{d}{dx} \left(2UU_x + 2W W_x \right) (x, \omega_0)} \\ &= \lim_{x \to x_0} \frac{\left(U_x W_{xx} + U W_{xxx} - W U_{xx} \right) (x, \omega_0)}{2 \left(U_x^2 + U U_{xx} + W_x^2 + W W_{xx} \right) (x, \omega_0)}. \end{split}$$

For the denominator,

$$UU_{xx} + WW_{xx}(x, \omega_0) \to (UU_{xx} + WW_{xx})(x_0, \omega_0) = 0$$

for $x \to x_0$, but

(

$$(U_x^2 + W_x^2)(x, \omega_0) \to (U_x^2 + W_x^2)(x_0, \omega_0) > 0$$

converges to a nonzero number, since not both $U_x(x_0, \omega_0)$ and $W_x(x_0, \omega_0)$ can be zero because of det $J_V(x_0, \omega_0) = (U_x W_\omega - U_\omega W_x) (x_0, \omega_0) \neq 0$. The numerator obviously converges:

$$(U_x W_{xx} + U W_{xxx} - W_x U_{xx} - W U_{xxx}) (x, \omega_0)$$

$$\rightarrow (U_x W_{xx} - W_x U_{xx}) (x_0, \omega_0) \in \mathbb{R},$$

thus

$$\lim_{x \to x_0} \frac{\partial \psi}{\partial x}(x, \omega_0) = \frac{\left(U_x W_{xx} - W_x U_{xx}\right)(x_0, \omega_0)}{2\left(U_x^2 + W_x^2\right)(x_0, \omega_0)} =: c \in \mathbb{R}.$$

Concerning the partial derivatives of the phase of the STFT with respect to the second variable (i.e., the 'frequency' variable), we can argue almost identically and thus find the following analogous results:

Theorem 4.9 (Phase derivatives of the STFT, part III). Let $f, g \in L^2(\mathbb{R})$. Assume that

•
$$V(x_0, \omega_0) = 0$$

• det $J_V(x_0, \omega_0) < 0$

Let $V(f,g) = V = U + i \cdot W \in C^2(\mathbb{R}^2)$. Then the phase $\psi(x,\omega)$ of $V(f,g)(x,\omega)$ satisfies

$$\lim_{x \to x_0} \frac{\partial \psi}{\partial \omega}(x, \omega_0) = \begin{cases} +\infty, & \text{if } x \to x_0 \text{ from the left} \\ -\infty, & \text{if } x_0 \leftarrow x \text{ from the right.} \end{cases}$$

Let $V(f,g) = V = U + i \cdot W \in C^3(\mathbb{R}^2)$, then

$$\lim_{\omega \to \omega_0} \frac{\partial \psi}{\partial \omega}(x_0, \omega) = c' \in \mathbb{R}, \quad \text{if } \omega \to \omega_0,$$

converges to some real number $c' \in \mathbb{R}$.

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References

- F. Auger, r. Chassande-Mottin, and P. Flandrin. On phase-magnitude relationships in the short-time fourier transform. *IEEE Signal Process. Lett.*, 19(5):267-270, 2012.
- [2] F. Auger and P. Flandrin. Improving the readability of timefrequency and time-scale representations by the method of re-assignment. *IEEE Trans. Signal Process.*, 43(5):1068–1089, May 1995.
- [3] P. Balazs. Basic definition and properties of Bessel multipliers. Journal of Mathematical Analysis and Applications, 325(1):571–585, January 2007.
- [4] P. Balazs, H. Waubke, and W. A. Deutsch. Phasenanalyse mit akustischen Anwendungsbeispielen. In Proceedings DAGA 2003 - Fortschritte der Akustik, March 2003.
- [5] R. Carmona, W.-L. Hwang, and B. Torrésani. Practical Time-Frequency Analysis. Academic Press San Diego, 1998.
- [6] N. Delprat, B. Escudie, P. Guillemain, R. Kronland-Martinet, P. Tchamitchian, and B. Torrésani. Asymptotic wavelet and Gabor analysis: extraction of instantaneous frequencies. *IEEE Transactions on Information Theory*, 38(2):644–664, 1992.
- [7] M. Dolson. The phase vocoder: a tutorial. Computer Musical Journal, 10(4):11-27, 1986.
- [8] F. A. E. Chassande-Mottin and P. Flandrin. On the statistics of spectrogram reassignment vectors. *Multidim. Syst. and Signal Proc.*, 9(4):355–362, 1998.
- [9] H. G. Feichtinger. Modulation Spaces: Looking Back and Ahead. Sampl. Theory Signal Image Process., 5(2):109–140, 2006.
- [10] H. G. Feichtinger and K. Nowak. A first survey of Gabor multipliers, chapter 5, pages 99–128. Birkhäuser Boston, 2003.
- [11] H. G. Feichtinger and G. Zimmermann. A Banach space of test functions for Gabor analysis, chapter 3, pages 123–170. Birkhäuser Boston, 1998.
- [12] J. Flanagan and R. M. Golden. Phase vocoder. Bell Syst. Tech., 45:1493 1509, 1966.
- [13] P. Flandrin. *Time-Frequency/Time-Scale Analysis*. Academic Press, San Diego, 1999.
- [14] G. B. Folland and A. Sitaram. The uncertainty principle: A mathematical survey. J. Fourier Anal. Appl., 3(3):207–238, 1997.
- [15] T. Gardner and M. Magnasco. Sparse time-frequency representations. Proc. Natl. Acad. Sci. USA, 103(16):6094–6099, April 2006.
- [16] K. Gröchenig. Foundations of Time-Frequency Analysis. Appl. Numer. Harmon. Anal. Birkhäuser Boston, Boston, MA, 2001.
- [17] F. Jaillet, P. Balazs, M. Dörfler, and N. Engelputzeder. On the structure of the phase around the zeros of the short-time fourier transform. In M. M. Boone, editor, NAG/DAGA 2009, pages 1584–1587, Rotterdam, 2009.
- [18] K. Kodera, C. D. Villedary, and R. Gendrin. A new method for the numerical analysis of nonstationary signals(magnetospheric ULF). *Physics of the Earth* and Planetary Interiors, 12(2):142–150, 1976.

- [19] A. V. Oppenheim and J. S. Lim. The importance of phase in signals. Proceedings of the IEEE, 69(5):529–541, May 1981.
- [20] N. Perraudin, P. Balazs, and P. Soendergaard. A fast Griffin-Lim algorithm. In 2013 IEEE Workshop on Applications of Signal Processing to Audio and Acoustics, New Paltz, NY, USA, 2013., 2013.
- [21] T. Quatieri. Discrete-time speech signal processing: principles and practice. Prentice Hall Press, Upper Saddle River, NJ, USA, 2001.
- [22] D. T. Stoeva and P. Balazs. Invertibility of multipliers. Applied and Computational Harmonic Analysis, 33(2):292–299, 2012.
- [23] D. T. Stoeva and P. Balazs. Canonical forms of unconditionally convergent multipliers. *Journal of Mathematical Analysis and Applications*, 399:252–259, 2013.