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ISS and integral-ISS of switched systems with nonlinear supply functions

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Abstract The problem of input-to-state stability (ISS), and its integral version (iISS), are considered for switched nonlinear systems with inputs, resets and possibly unstable subsystems. For the dissipation inequalities associated with the Lyapunov function of each subsystem, it is assumed that the supply functions, which characterize the decay rate and ISS/iISS gains of the subsystems, are nonlinear. The change in the value of Lyapunov functions at switching instants is described by a sum of growth and gain functions, which are also nonlinear. Using the notion of average dwell-time (ADT) to limit the number of switching instants on an interval, and the notion of average activation time (AAT) to limit the activation time for unstable systems, a formula relating ADT and AAT is derived to guarantee ISS/iISS of the switched system. Case studies of switched systems with saturating dynamics and switched bilinear systems are included for illustration of the results.

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1 Introduction

Switched systems - comprising a family of dynamical subsystems orchestrated by a piecewise constant switching signal – provide a mathematical framework for modeling systems where the state trajectories exhibit sudden transitions, either due to instant change in the vector field or due to jumps in some of the state variables [13,14]. A canonical approach for stability analysis of such systems is to develop characterizations based on the stability properties of the individual subsystems, and the effects observed at switching times. When all subsystems are stable, the undesired instability effects may occur at switching times. In such cases, if one restricts the frequency of switching times over an interval, then the decay in the norm of state trajectories can overcome the growth at switching times. Such slow switching results are usually obtained by imposing lower bounds on the *dwell-time* or *average dwell-time* (ADT) between switching instances [26, 11]. In the case when some subsystems are unstable, one needs to restrict the average activation time (AAT) of these unstable subsystems along with sufficiently large ADT bounds in order for the resulting switched system to be stable, and this has been studied in [17, 29, 31]. For switched linear systems, stability conditions based on matrix measure can also be used to determine stability when the switching signals are periodic [19], and this criterion is later extended to switching signals satisfying asymptotic AAT constraints [28]. Meanwhile, it is also possible to consider cases where the continuous dynamics of all the individual subsystems are unstable, but the impulsive effects from switching between these subsystems stabilize the system [25,30]. Conditions describing how fast such stabilizing impulsive effects must occur are captured under the notion of reverse average dwell-time (r-ADT), introduced in [9].

An additional important element that we incorporate in our stability analysis of switched systems is the presence of inputs or disturbances. For nonlinear systems, in general, quantitative relationships between the norm of the state trajectories and the magnitude of the inputs are appropriately captured by the notion of *input-to-state stability* (ISS), or its integral variant, *integral input-to-state stability* (iISS). These notions have been pioneered in [21] and [22], and since their inception, these tools have been used for analysis and design purposes in many control-related problems. In the context of switched systems under arbitrary switching, converse results regarding the existence of ISS and iISS Lyapunov functions first appeared in [16], and some implications relating ISS and iISS for switched systems have been developed in [8]. Switching dynamics with reset maps can also be captured by the framework of hybrid systems [7] and ISS characterizations via Lyapunov functions for hybrid systems have been developed in [3,4]. Relaxing certain assumptions in these works, along with some developments based on converse results, iISS characterizations via Lyapunov functions for hybrid systems appear in [18]. However, these results do not explicitly work out the stability conditions for switched systems in terms of the data associated with individual subsystems.

Combining the aforementioned two directions of research, it is rather natural to ask whether the Lyapunov characterizations for individual subsystems can be combined with ADT and AAT notions to obtain sufficient conditions for ISS/iISS of switched systems with inputs. This question has indeed been addressed in the literature in different settings. In the work [9], ISS/iISS is studied for impulsive systems, which can be viewed as switched systems with a single mode. The paper [24] provides lower bounds on ADT in terms of the data associated with individual ISS subsystems so that the switched system is ISS. Later the paper [17] relaxes the assumption of all ISS subsystems and allows possible unstable subsystems. A result on ISS switched system is concluded in that work when the switching signal satisfies both ADT and AAT conditions. Nevertheless, a major restriction of these works is that most results require the assumption that both the decay rates in the inequalities associated with the derivative of Lyapunov functions or the growth rate in the value of the Lyapunov functions at switching instants are linear. This assumption becomes overly restrictive when studying iISS. An exception is the result [9, Theorem 6], which provides conditions for the impulsive system to be iISS under arbitrary switching even if the decay rate is nonlinear. However, growth in the Lyapunov function when impulses occur is not allowed in that result.

Compared to the aforementioned references, the major contribution of this work is to study ISS and iISS of switched systems with *nonlinear* decay rate in the dissipation inequalities associated with the Lyapunov functions of individual subsystems. In other words, using the terminology from [23], we consider Lyapunov functions for each subsystem with nonlinear supply functions. Moreover, the change in the value of Lyapunov function at switching times is also described by *nonlinear* growth function. This setup was, in particular, adopted in the conference version of this paper [15], and the previous independent work of the authors [32, 20]. While [32] provides the construction of an ISS Lyapunov function to find the lower bounds on ADT for stability of cascade interconnections, the present work uses a conceptually similar approach for a broader class of switched systems and extends the study to iISS. In contrast to [20], where the analysis is carried out using trajectory-based methods with nested comparison functions, this paper makes transition from dwell-time conditions to a more quantitative ADT estimate. In our conference paper [15], we only study switched systems whose subsystems are all (i)ISS and the growth of the Lyapunov functions at the switching times is not allowed to depend on the input. Such assumptions have been relaxed in this work and consequently, a generalization of the results in [17] is obtained with a Lyapunov-functionbased approach that allows us to handle additional nonlinearities in the supply functions.

The rest of the paper is organized as follows. After some necessary preliminaries about switched systems given in Section 2, our main result is stated in Section 3 with some discussions about the validity of its assumptions. Section 4 contains the technical tools used to prove our main result. Case studies of switched systems with saturating dynamics and switched bilinear systems are provided in Section 5 for illustration of the results, followed by the conclusion in Section 6.

2 Preliminaries

In this section, we introduce the basic notation of comparison functions, the definition of switched systems and switching signals, and the definitions of ISS and iISS of switched systems.

Comparison functions. A function $\alpha : \mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}_{\geq 0}$ is said to be positive definite if it is continuous, $\alpha(0) = 0$ and $\alpha(s) > 0$ for all s > 0. If α is also strictly increasing, then it is said to be of class \mathcal{K} . In addition if α is also unbounded, then it is said to be of class \mathcal{K}_{∞} . A function $\beta : \mathbb{R}_{\geq 0} \times \mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}_{\geq 0}$ is said to be of class $\mathcal{K}_{\mathcal{L}}$ if $\beta(\cdot, t)$ is of class \mathcal{K} for each $t \in \mathbb{R}_{\geq 0}$, $\beta(s, \cdot)$ is continuously decreasing and $\beta(s, t) \longrightarrow 0$ as $t \longrightarrow \infty$ for each fixed $s \in \mathbb{R}_{\geq 0}$; see [12, Chapter 4] for their use in formulation of common stability notions. In addition, we require class \mathcal{KLL} functions: A function $\beta : \mathbb{R}_{\geq 0} \times \mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}_{\geq 0}$ is a class \mathcal{KLL} function if $\beta(\cdot, \cdot, j)$ is a class \mathcal{KL} function for each $j \geq 0$ and $\beta(\cdot, t, \cdot)$ is a class \mathcal{KL} function for each $s \geq 0$.

Switched systems. Let $\mathcal{P} \subset \mathbb{N}$ be a set of either finite or infinite cardinality. For each $p \in \mathcal{P}$, there is a locally Lipschitz vector field $f_p : \mathbb{R}^n \times \mathbb{R}^m \longrightarrow \mathbb{R}^n$. The differential equations

$$\dot{x} = f_p(x,\omega), \quad p \in \mathcal{P}$$
 (1)

are the dynamics of the *subsystems* or *modes* of the switched system. For each pair $(p,q) \in \mathcal{P} \times \mathcal{P}$, there is also a continuous jump map $g_{q,p} : \mathbb{R}^n \times \mathbb{R}^m \longrightarrow \mathbb{R}^n$. Let $\overline{\Sigma}$ be the set of all right-continuous, piecewise constant mappings from $\mathbb{R}_{\geq 0}$ to \mathcal{P} with a locally finite number of discontinuities, called *switching signals*. For each switching signal $\sigma \in \overline{\Sigma}$, define

$$\mathcal{T}(\sigma) := \{t > 0 : \sigma(t) \neq \sigma(t^{-})\}$$
(2)

In other words, $\mathcal{T}(\sigma)$ is the collection of switching instants. With this data, we consider switched dynamical systems with inputs and resets, described by

$$\dot{x}(t) = f_{\sigma(t)}(x(t), \omega(t)) \qquad \text{if } t \notin \mathcal{T}(\sigma), \tag{3a}$$

$$x(t) = g_{\sigma(t^{-}),\sigma(t)}(x(t^{-}),\omega(t)) \quad \text{if } t \in \mathcal{T}(\sigma),$$
(3b)

where $\omega : \mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}^m$ is locally essentially bounded on $\mathbb{R}_{\geq 0}$ and bounded on $\mathcal{T}(\sigma)$. We call (3a) the *continuous flow* and (3b) the *discrete jumps*. We denote the solution of (3) with initial state x_0 , input ω and switching signal σ by $x(\cdot; x_0, \omega, \sigma)$. When x_0, ω, σ are clear from the context, the solution is also abbreviated by $x(\cdot)$.

Switching signals. We now specify the class of switching signals for which we study the stability of the system (3). Conceptually, the class of switching signals with ADT constraints puts an upper bound on the average number of switches over any time interval. According to [11], we say that a switching signal σ satisfies average dwell-time (ADT) condition with parameters $\tau_a > 0$ (average dwell-time) and $N_0 \ge 1$ if

$$\forall t_2 \ge t_1 \ge 0: \quad N_{\sigma}(t_1, t_2) \le N_0 + \frac{t_2 - t_1}{\tau_a},\tag{4}$$

where $N_{\sigma}(t_1, t_2)$ is the number of switches occurring in the interval (t_1, t_2) : $N_{\sigma}(t_1, t_2) := |(t_1, t_2] \cap \mathcal{T}(\sigma)|$. We denote $\Sigma_{\text{ADT}}(\tau_a, N_0)$ to be the collection of all ADT switching signals with parameters τ_a, N_0 .

Following the definition in [17], we take a disjoint union $\mathcal{P} = \mathcal{P}_{s} \cup \mathcal{P}_{u}$ and say that a switching signal σ satisfies *average activation time* (AAT) condition with parameters $\eta \in [0, 1]$ (percentage activation time), $T_0 \ge 0$, if

$$\forall t_2 \ge t_1 \ge 0: \quad T_{\sigma, \mathcal{P}_u}(t_1, t_2) \le T_0 + \eta(t_2 - t_1), \tag{5}$$

where $T_{\sigma,\mathcal{P}_{u}}(t_{1},t_{2})$ is the activation time of modes in \mathcal{P}_{u} over the interval $(t_{1},t_{2}]$. More precisely, if we consider the function $\mathbf{1}_{\mathcal{P}_{u}}: \mathcal{P} \longrightarrow \{0,1\}$, defined as

$$\mathbf{L}_{\mathcal{P}_{\mathbf{u}}}(\sigma) = \begin{cases} 1, & \text{if } \sigma \in \mathcal{P}_{\mathbf{u}}, \\ 0, & \text{otherwise,} \end{cases}$$
(6)

then $T_{\sigma,\mathcal{P}_{\mathbf{u}}}(t_1,t_2) := \int_{t_1}^{t_2} \mathbf{1}_{\mathcal{P}_{\mathbf{u}}}(\sigma(\tau)) d\tau$. We denote $\Sigma_{\text{AAT}}(\mathcal{P}_{\mathbf{u}},\eta,T_0)$ to be the collection of all AAT switching signals of modes in $\mathcal{P}_{\mathbf{u}}$ with parameters η, T_0 . Notice that the AAT condition is always with respect to the set $\mathcal{P}_{\mathbf{u}}$, which we later associate with the collection of subsystems with not necessarily stable dynamics. In addition, in the special case when $\eta = 1$, (5) imposes no constraints on the activation time and hence the switching signal can be arbitrary. On the other hand, when $T_0 = 0$ and $\eta < 1$, (5) implies that modes in $\mathcal{P}_{\mathbf{u}}$ are not activated at all.

Stability definitions. Let $\Sigma \subset \overline{\Sigma}$ be a set of switching signals. In this paper, we will focus on the following two uniform external stability properties of switched systems with respect to all switching signals in Σ . In what follows, the essential supremum norm of ω over a set I (that is, the supremum of ω on I except for a set of measure zero) is denoted by $\operatorname{ess\,sup}_{s\in I} |\omega(s)|$.

Definition 1 A switched system (3) is uniformly *input-to-state stable* (ISS) over Σ if there exist $\beta \in \mathcal{KL}, \gamma_1, \gamma_2 \in \mathcal{K}$ such that

$$|x(t;x_0,\omega,\sigma)| \leq \beta(|x_0|,t) + \gamma_1(\underset{s\in[0,t]\setminus\mathcal{T}(\sigma)}{\operatorname{ess\,sup}} |\omega(s)|) + \gamma_2(\underset{s\in[0,t]\cap\mathcal{T}(\sigma)}{\operatorname{sup}} |\omega(s)|)$$
(7)

for all $t \ge 0, x_0 \in \mathbb{R}^n$ and $\sigma \in \Sigma$.

Definition 2 A switched system (3) is uniformly *integral input-to-state stable* (iISS) over Σ if there exist $\alpha_0 \in \mathcal{K}_{\infty}, \beta \in \mathcal{KL}, \gamma_1, \gamma_2 \in \mathcal{K}$ such that

$$\alpha_0(|x(t;x_0,\omega,\sigma)|) \leqslant \beta(|x_0|,t) + \int_0^t \gamma_1(|\omega(s)|)ds + \sum_{s \in [0,t] \cap \mathcal{T}(\sigma)} \gamma_2(|\omega(s)|) \quad (8)$$

for all $t \ge 0, x_0 \in \mathbb{R}^n$, and $\sigma \in \Sigma$.

Our definitions of ISS and iISS are adopted from [3] and [18]. Notice that the last terms in (7), and in (8) do not appear in standard definitions of ISS and iISS for non-switched systems; these additional terms are needed in our framework to capture the growth of state trajectories at switching instants due to the presence of inputs in the jump dynamics (3b).

3 Assumptions and main results

We now state some assumptions on the data of system (3) required for the statement of the main result.

Assumption 1 There exist C^1 Lyapunov functions $V_p : \mathbb{R}^n \longrightarrow \mathbb{R}_{\geq 0}, p \in \mathcal{P}$, satisfying the conditions:

(L1) There exist $\underline{\alpha}, \overline{\alpha} \in \mathcal{K}_{\infty}$ such that

$$\underline{\alpha}(|x|) \leqslant V_p(x) \leqslant \overline{\alpha}(|x|), \quad \forall x \in \mathbb{R}^n, p \in \mathcal{P}.$$
(9)

(L2) There exist a disjoint partition $\mathcal{P} = \mathcal{P}_{s} \cup \mathcal{P}_{u}$, two positive definite functions α_{s}, α_{u} , and $\gamma \in \mathcal{K}$ such that

$$\left\langle \frac{\partial}{\partial x} V_p(x), f_p(x,\omega) \right\rangle \leqslant -\alpha_{\rm s}(V_p(x)) + \gamma(|\omega|) \quad \forall x \in \mathbb{R}^n, \omega \in \mathbb{R}^m, p \in \mathcal{P}_{\rm s},$$
(10a)

and

$$\left\langle \frac{\partial}{\partial x} V_p(x), f_p(x,\omega) \right\rangle \leqslant \alpha_{\mathrm{u}}(V_p(x)) + \gamma(|\omega|) \quad \forall x \in \mathbb{R}^n, \omega \in \mathbb{R}^m, p \in \mathcal{P}_{\mathrm{u}}.$$
(10b)

(L3) There exist $\chi \in \mathcal{K}_{\infty}, \rho \in \mathcal{K}$ such that

$$V_p(g_{q,p}(x,\omega)) \leqslant \chi(V_q(x)) + \rho(|\omega|) \quad \forall x \in \mathbb{R}^n, \omega \in \mathbb{R}^m, p, q \in \mathcal{P}.$$
(11)

Remark 1 For each subsystem in \mathcal{P}_{s} , (L1) and the condition (10a) in (L2) provide a necessary and sufficient condition for it to be iISS [2]. In addition if it is assumed that α_{s} in (10a) is \mathcal{K}_{∞} , then each subsystem in \mathcal{P}_{s} is ISS.

Remark 2 The partition \mathcal{P}_{s} , \mathcal{P}_{u} is not necessarily an indicator whether a subsystem is iISS or not. Indeed, when a subsystem is not iISS, (10a) will not hold for any choice of V_{p} but (10b) may still hold. However, even if a subsystem is iISS, it may still lead to the inequality (10b) due to the improper choice of V_{p} . In this case this subsystem needs to be categorized to the set \mathcal{P}_{u} .

Based on Assumption 1, we introduce the function $\psi:\mathbb{R}_{\geqslant 0}\longrightarrow\mathbb{R}_{\geqslant 0}$ as

$$\psi(t) := \min_{s \in [0,t]} \{ \alpha_{s}(s) + c(t-s) \}$$
(12)

where c > 0 is some constant. The next assumption bounds the divergence rate for the subsystems in \mathcal{P}_{u} :

Assumption 2 The supremum

$$\kappa := \sup_{s>0} \frac{\alpha_{\mathbf{u}}(s)}{\psi(s)} \tag{13}$$

is finite, where α_u comes from (L2) in Assumption 1 and ψ is defined in (12).

Our last assumption will be instrumental in deriving a lower bound for ADT in a similar fashion as in [32]:

Assumption 3 The supremum

$$\zeta^* := \sup_{s>0} \int_s^{\chi(s)} \frac{1}{\psi(r)} dr \tag{14}$$

is finite, where χ comes from (L3) in Assumption 1 and ψ is defined in (12).

We are now ready to state the main theorem of this paper:

Theorem 1 Consider the switched system (3) and suppose that Assumptions 1, 2, and 3 hold with the function ψ constructed via (12) for some c > 0. If $\tau_a > \zeta^*$, then for every $\eta \in [0, 1]$ satisfying the inequality

$$(1+\kappa)\eta + \frac{\zeta^*}{\tau_a} < 1,\tag{15}$$

the system (3) is uniformly iISS over $\Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$ for any $N_0 \ge 1$ and $T_0 \ge 0$. Moreover, if in (L2) $\alpha_s \in \mathcal{K}_{\infty}$, and $\tau_a > \zeta^*$, then system (3) is uniformly ISS over $\Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$ for every $\eta \in [0, 1]$ satisfying (15) and any $N_0 \ge 1, T_0 \ge 0$.

To provide some insight about Theorem 1, let us consider its application to some special types of switched systems. We first consider the case when all the subsystems are (i)ISS. In this case, we can always find V_p satisfying (10a) in (L2) for all $p \in \mathcal{P}$ so that $\mathcal{P}_u = \emptyset$. Hence the AAT condition vanishes and we could simply set $\eta = 0$. As a result, (15) reduces to $\tau_a > \zeta^*$. In other words, in this case the switched system (3) is uniformly (i)ISS over $\Sigma_{ADT}(\tau_a, N_0)$ for any $\tau_a > \zeta^*, N_0 \ge 1$.

We next consider the case when the switching does not introduce jumps in the Lyapunov functions evaluated along the solutions of the unforced system, so that $\chi = \text{id}$. We then conclude from (14) that $\zeta^* = 0$, so by Theorem 1, the switched system is (i)ISS with arbitrarily small values of the average dwell time τ_a . In addition, (15) results in $\eta < \frac{1}{1+\kappa}$ and the switched system (3) is uniformly (i)ISS over $(\cup_{\tau_a>0} \Sigma_{\text{ADT}}(\tau_a, N_0)) \cap \Sigma_{\text{AAT}}(\mathcal{P}_u, \eta, T_0)$.

Lastly, we consider the case when the decaying rates and growth rates in (L2), (L3) of Assumption 1 are all linear; that is, there exist $\lambda_s, \lambda_u, \mu > 0$ such that $\alpha_s(s) = \lambda_s s, \alpha_u(s) = \lambda_u s, \chi(s) = \mu s$ for all $s \ge 0$. In this case, the formula (14) yields

$$\zeta^* = \int_s^{\mu s} \frac{1}{\lambda_{\rm s} r} dr = \frac{\ln \mu}{\lambda_{\rm s}}$$

and consequently the inequality (15) becomes

$$\left(1+\frac{\lambda_{\rm u}}{\lambda_{\rm s}}\right)\eta+\frac{\ln\mu}{\lambda_{\rm s}\tau_a}<1.$$

After rearranging the terms, we recover the conditions on τ_a and η which guarantee ISS switched system as stated in [17, Theorem 2]. Thus our work turns out to be a generalization from linear to nonlinear setting of the known results in the literature.

Discussion of the assumptions. Once we impose some characterizations of the stability or instability of individual subsystems and the reset maps in Assumption 1, it is seen that the stability condition in (15) primarily depends on finiteness of κ in Assumption 2 and ζ^* in Assumption 3. In both the assumptions, the function ψ constructed in (12) plays a key role.

- 1. Note that (12) immediately implies that $\psi(t) \leq \overline{\psi}(t) := \min\{\alpha_{\rm s}(t), ct\}$. In [15], we proposed stability conditions using the function $\overline{\psi}$ while assuming that it is globally one-sided Lipschitz. That assumption is necessary in [15]; for example, take $\alpha_{\rm s}(s) = se^{-s} + 1 \cos(s^2)$, then $\overline{\psi}(t)$ is not one-sided Lipschitz for any c > 0. In this work, the Lipschitzness assumption is no longer required because the function ψ in (12) inherits this property by construction, which will be shown later by Lemma 8.
- 2. The function ψ defined by (12) is essentially a global one-sided Lipschitz modification of α_s . In case α_s is already globally Lipschitz, ψ can be chosen equal to α_s by setting c larger than its Lipschitz constant. Functions like $\alpha_s(s) = s, \alpha_s(s) = \ln(1+s)$ and $\alpha_s(s) = se^{-s}$ are examples of such a case.

- 3. Combining (12) and (13), we conclude from (10b) that for all $p \in \mathcal{P}_{u}$, $\left\langle \frac{\partial}{\partial x} V_p(x), f_p(x,\omega) \right\rangle \leqslant \kappa c V_p(x) + \gamma(|\omega|)$. According to [1, Theorem 1], this property in turn implies that these subsystems are forward complete, which is a very reasonable assumption on the subsystems as one cannot expect the switched system to be iISS if an active subsystem has finite escape time.
- 4. In order to get a qualitative answer to the question of when the switched system is iISS/ISS under slow switching, we need to better understand Assumption 3. The next lemma states that, in order to show that ζ^* is finite, we only need to examine the values of $\psi(s), \chi(s)$ for small and large s without computing the integral in (14). To this end, we say a positive definite function $\alpha : \mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}_{\geq 0}$ is *initially increasing* if there exists r > 0such that $\alpha(s) \leq \alpha(t)$ for all $0 \leq s \leq t \leq r$, and eventually increasing (resp. eventually decreasing) if there exists R > 0 such that $\alpha(s) \leq \alpha(t)$ (resp. $\alpha(s) \ge \alpha(t)$ for all $R \le s \le t$.

Lemma 2 The value ζ^* defined in (14) is finite when both conditions (a) and (b) hold:

- (a) ψ is initially increasing and $\limsup_{s \longrightarrow 0^+} \frac{\chi(s)-s}{\psi(s)} < \infty$, and (b) ψ is eventually increasing and $\limsup_{s \longrightarrow \infty} \frac{\chi(s)-s}{\psi(s)} < \infty$, or ψ is eventually decreasing and $\limsup_{s \longrightarrow \infty} \frac{\chi(s)-s}{\psi(\chi(s))} < \infty$.

Proof. We only study the non-trivial case when $\chi(s) > s$ for all s > 0. By inspecting the definition of ζ^* in (14), we see that it suffices to show that $\limsup_s \int_s^{\chi(s)} \frac{1}{\psi(r)} dr$ is finite when s approaches both 0^+ and infinity in order for ζ^* to be finite. When ψ is initially increasing, we have $\overline{r} > 0$ such that for all s > 0 with $\chi(s) \leq \overline{r}$,

$$\int_{s}^{\chi(s)} \frac{1}{\psi(r)} dr \leqslant \int_{s}^{\chi(s)} \frac{1}{\psi(s)} dr = \frac{\chi(s) - s}{\psi(s)}.$$

Take the limit as $s \longrightarrow 0^+$, and it follows from the first condition in Lemma 2 that $\limsup_{s \to 0^+} \int_s^{\chi(s)} \frac{1}{\psi(r)} dr$ is bounded. Similarly, the second condition results in $\limsup_{s \to \infty} \int_s^{\chi(s)} \frac{1}{\psi(r)} dr$ being bounded. \Box

It is not hard to see that ζ^* is finite for the linear functions when $\psi(s) =$ $\lambda s, \chi(s) = \mu s$. In addition, using Lemma 2, it is straightforward to verify that the nonlinear pairs such as $(\psi(s), \chi(s)) = (1 - e^{-s}, s + 1 - e^{-s}),$ $(e^{-s}(1-e^{-s}), \ln(e^s+1))$ also ensure finite ζ^* .

4 Technical Tools and Proofs

Our proof of Theorem 1 relies on rewriting the switched system into a hybrid system in the framework of [7] and then showing ISS/iISS of that hybrid system. In this section we provide a brief description of hybrid systems, followed with some supporting lemmas and the proof of Theorem 1.

4.1 Hybrid systems with inputs

Consider the hybrid dynamical system with inputs, which is described as:

$$\begin{cases} \dot{\xi} \in \mathcal{F}(\xi, d), & (\xi, d) \in \mathcal{C}, \\ \xi^+ \in \mathcal{G}(\xi, d), & (\xi, d) \in \mathcal{D}, \end{cases}$$
(16)

where the state trajectory ξ evolves in $\mathcal{X} \subset \mathbb{R}^n$, and the disturbance d takes values in \mathbb{R}^m . The sets \mathcal{C}, \mathcal{D} are subsets of $\mathcal{X} \times \mathbb{R}^m$ and are called flow and jump sets, respectively. The evolution of the state is thus described by \mathcal{F} (during flows) and by \mathcal{G} (at jump instants), which are set-valued mappings from $\mathcal{X} \times \mathbb{R}^m$ to \mathcal{X} . The solutions of the hybrid system (16) are defined on hybrid time domains. A set $I \subset \mathbb{R}_{\geq 0} \times \mathbb{Z}_{\geq 0}$ is called a compact hybrid time domain if $I = \bigcup_{j=0}^{J-1}([t_j, t_{j+1}], j)$ for some finite sequence of times $0 = t_0 \leqslant t_1 \cdots \leqslant t_J$; and I is a hybrid time domain if for all $(T, J) \in I$, $I \cap ([0, T] \times \{0, 1, \dots, J\})$ is a compact hybrid time domain. The domain of the hybrid input d in (16) is a hybrid time domain, and $d: \operatorname{dom} d \longrightarrow \mathbb{R}^m$ is such that $d(\cdot, j)$ is Lebesgue measurable and locally essentially bounded for each $j \in \mathbb{Z}_{\geq 0}$. It is assumed that when the system data $(\mathcal{F}, \mathcal{G}, \mathcal{C}, \mathcal{D})$ satisfies certain basic assumptions [4, 7], then for each input $d : \operatorname{dom} d \longrightarrow \mathbb{R}^m$, the system (16) admits a local solution, called hybrid arc, $\xi : \operatorname{dom} \xi \longrightarrow \mathcal{X}$. A hybrid arc $\xi : \operatorname{dom} \xi \longrightarrow \mathcal{X}$ and a hybrid input $d: \operatorname{dom} d \longrightarrow \mathbb{R}^m$ are a solution pair to the hybrid model (16) if:

- $-\operatorname{dom}\xi = \operatorname{dom}d;$
- for all $j \in \mathbb{Z}_{\geq 0}$ and almost all $t \in \mathbb{R}_{\geq 0}$ such that $(t, j) \in \text{dom }\xi$, we have $(\xi(t, j), d(t, j)) \in \mathcal{C}$ and $\dot{\xi}(t, j) \in \mathcal{F}(\xi(t, j), d(t, j))$; and
- for $(t, j) \in \text{dom } \xi$ such that $(t, j + 1) \in \text{dom } \xi$ we have $(\xi(t, j), d(t, j)) \in \mathcal{D}$ and $x(t, j + 1) \in \mathcal{G}(\xi(t, j), d(t, j)).$

For an initial state ξ_0 and an input d, a hybrid arc of (16) is denoted by $\xi(\cdot, \cdot; \xi_0, d)$. In what follows, we denote the distance between a vector $\xi \in \mathcal{X}$ and a compact set $\mathcal{A} \subset \mathcal{X}$ by $|\xi|_{\mathcal{A}}$, that is, $|\xi| := \inf_{\zeta \in \mathcal{A}} |\xi - \zeta|$. Following [3], for a hybrid signal d, we use the notation $||d||_{(t,j)}$ to denote the maximum between $\underset{(\hat{t},\hat{j}+1)\notin \operatorname{dom} d, \hat{t}+\hat{j}\leqslant t+j}{\operatorname{ess\,sup}} |d(\hat{t},\hat{j})|$ and $\underset{(\hat{t},\hat{j}+1)\in \operatorname{dom} d, \hat{t}+\hat{j}\leqslant t+j}{\operatorname{sup}} |d(\hat{t},\hat{j})|$.

Definition 3 A hybrid system (16) is said to be *input-to-state stable* (ISS) with respect to \mathcal{A} if there exist $\beta_h \in \mathcal{KLL}$ and $\gamma_h \in \mathcal{K}$ such that for all $\xi_0 \in \mathcal{X}$ and all $(t, j) \in \text{dom } \xi$, each solution pair (ξ, d) satisfies

$$|\xi(t,j;\xi_0,d)|_{\mathcal{A}} \leqslant \beta_h(|\xi_0|_{\mathcal{A}},t,j) + \gamma_h(||d||_{(t,j)}).$$
(17)

Definition 4 A hybrid system (16) is said to be *integral input-to-state stable* (iISS) with respect to \mathcal{A} if there exist $\alpha_h \in \mathcal{K}_{\infty}, \beta_h \in \mathcal{KLL}$ and $\gamma_h \in \mathcal{K}$ such that for all $\xi_0 \in \mathcal{X}$ and all $(t, j) \in \text{dom } \xi$, each solution pair (ξ, d) satisfies

$$\alpha_{h}(|\xi(t,j;\xi_{0},d)|_{\mathcal{A}}) \leqslant \beta_{h}(|\xi_{0}|_{\mathcal{A}},t,j) + \int_{0}^{t} \gamma_{h}(|d(s,j_{d,s})|)ds + \sum_{\substack{(\hat{t},\hat{j}+1)\in\mathrm{dom}\,d\\\hat{t}+\hat{j}\leqslant t+j}} \gamma_{h}(|d(\hat{t},\hat{j})|)$$

$$(18)$$

where $j_{d,s} := \max\{i \in \mathbb{N} \mid (s,i) \in \operatorname{dom} d\}.$

We can also characterize ISS and iISS using Lyapunov functions, as stated in the following two lemmas [3,18]:

Lemma 3 A hybrid system (16) is ISS with respect to \mathcal{A} if there exists a smooth function $V : \mathcal{X} \longrightarrow \mathbb{R}_{\geq 0}$ and functions $\alpha_1, \alpha_2, \alpha_c, \alpha_d \in \mathcal{K}_{\infty}$ and $\gamma_c \in \mathcal{K}$ such that

$$\alpha_1(|\xi|_{\mathcal{A}}) \leqslant V(\xi) \leqslant \alpha_2(|\xi|_{\mathcal{A}}) \tag{19}$$

for all $\xi \in \mathcal{X}$,

$$\left\langle \frac{\partial}{\partial \xi} V(\xi), \overline{f} \right\rangle \leqslant -\alpha_c(|\xi|_{\mathcal{A}}) + \gamma_c(|d|)$$
 (20)

for all $(\xi, d) \in C, \overline{f} \in \mathcal{F}(\xi, d)$, and

$$V(\overline{g}) - V(\xi) \leqslant -\alpha_d(|\xi|_{\mathcal{A}}) + \gamma_d(|d|)$$
(21)

for all $(\xi, d) \in \mathcal{D}, \overline{g} \in \mathcal{G}(\xi, d)$.

Lemma 4 A hybrid system is iISS with respect to \mathcal{A} if there exists a smooth function $V : \mathcal{X} \longrightarrow \mathbb{R}_{\geq 0}$ and functions $\alpha_1, \alpha_2 \in \mathcal{K}_{\infty}$, positive definite functions α_c, α_d and $\gamma_c \in \mathcal{K}$ such that (19) holds for all $\xi \in \mathcal{X}$, (20) holds for all $(\xi, d) \in \mathcal{C}, \overline{f} \in \mathcal{F}(\xi, d)$ and (21) holds for all $(\xi, d) \in \mathcal{D}, \overline{g} \in \mathcal{G}(\xi, d)$.

We call a function V satisfying the inequalities in Lemma 3 (resp. Lemma 4) a hybrid ISS (resp. iISS) Lyapunov function.

Remark 3 Note that Lemma 4 differs from Lemma 3 in the sense that α_c, α_d are only assumed to be positive definite instead of class \mathcal{K}_{∞} , which also reflects the fact that iISS is strictly weaker than ISS.

4.2 Hybrid models for ADT and AAT

We now model the switched systems with ADT and AAT constraints as a hybrid system (16). It is assumed that $\mathcal{P} = \mathcal{P}_u \cup \mathcal{P}_s$, and we recall that $\mathbf{1}_{\mathcal{P}_u}(\cdot)$, defined in (6), is the indicator function associated with unstable modes indexed

by \mathcal{P}_{u} . Recall that the definitions of ADT and AAT constrained switching signals in (4) and (5) depend on the positive integer N_{0} and the scalar T_{0} . We consider an autonomous hybrid system evolving on $\mathcal{X} := \mathcal{P} \times [0, N_{0}] \times [0, T_{0}]$ with state denoted by (p, τ, τ_{u}) . The dynamics of this hybrid system is given by

$$\begin{cases} \dot{p} = 0, \\ \dot{\tau} \in [0, \frac{1}{\tau_a}], \\ \dot{\tau}_u \in [\mathbf{1}_{\mathcal{P}_u}(p) - \eta, \mathbf{1}_{\mathcal{P}_u}(p)], \end{cases} \quad \xi \in \mathcal{C}, \quad (22a)$$

$$\begin{cases} p^+ \in \mathcal{P}, \\ \tau^+ = \tau - 1, \\ \tau_u^+ = \tau_u, \end{cases} \quad \xi \in \mathcal{D} \quad (22b)$$

where the parameters τ_a , η come from (4), (5) and the flow and jump sets are given by $\mathcal{C} := \mathcal{P} \times [0, N_0] \times [0, T_0]$, $\mathcal{D} := \mathcal{P} \times [1, N_0] \times [0, T_0]$, respectively. Note that the domain of $\mathbf{1}_{\mathcal{P}_u}(\cdot)$ is discrete, which makes it continuous with discrete topology. Consequently, the right-hand side of (22a) is outer semi-continuous, closed and convex-valued so it satisfies the regularity assumption on the hybrid system. Essentially, in our model (22), we are introducing two timers: τ and τ_u . The underlying intuition is that τ keeps track of ADT constraint, and τ_u models the AAT constraint. Recall the definition of $\mathcal{T}(\sigma)$ in (2) and denote $\mathcal{T}(\sigma) =: \{t_1, t_2, \cdots\}$ with $t_1 < t_2 < \cdots$ and denote $t_0 := 0$. In addition we abuse the notation that when $|\mathcal{T}(\sigma)| < \infty$, $[t_{|\mathcal{T}(\sigma)|}, t_{|\mathcal{T}(\sigma)+1|}] := [t_{|\mathcal{T}(\sigma)|}, \infty)$. The next proposition formalizes the relation between the complete solutions of the hybrid system (22) and the switching signals satisfying both ADT and AAT conditions.

Proposition 5 For every switching signal $\sigma \in \Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$, there exists a complete solution $\xi = (p, \tau, \tau_u)$ to the hybrid system (22) with the hybrid domain dom $\xi = \bigcup_{j=0}^{|\mathcal{T}(\sigma)|} [t_j, t_{j+1}] \times \{j\}$ and it satisfies that

$$p(t,j) = \sigma(t_j) \quad \forall (t,j) \in \operatorname{dom} \xi.$$
(23)

On the other hand, for every complete solution $\xi = (p, \tau, \tau_u)$ to the hybrid system (22), denote its hybrid domain by dom $\xi := \bigcup_{j=0}^{J} [t_j, t_{j+1}] \times \{j\}$, where $[t_J, t_{J+1}] = [t_J, \infty)$ if $J < \infty$. Define a switching signal

$$\sigma(t) := p(t,j) \quad \forall t \in [t_j, t_{j+1}), \ j = 1, \cdots, J$$

$$(24)$$

then $\sigma \in \Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0).$

Proof. As shown in [7, Example 2.15], the dynamics of p and τ guarantees the equivalence between complete solutions of the hybrid system (22) and switching signals $\sigma \in \Sigma_{ADT}(\tau_a, N_0)$ such that $p(t, j) = \sigma(t)$. It remains to show that the dynamics on τ_u confines p(t, j) to be equivalent to an AAT switching signal. To show the first part of Proposition 5, let $\sigma \in \Sigma_{ADT}(\tau_a, N_0)$ $\Sigma_{\text{AAT}}(\mathcal{P}_u, \eta, T_0)$ be arbitrary. If $\eta = 1$, then there is no AAT constraint on the switching signal and a hybrid arc $\xi = (p, \tau, \tau_u)$ with $p(0, 0) = \sigma(0)$, $\dot{p} = 0$ if $t \notin \mathcal{T}(\sigma)$, $p^+ = \sigma(t)$ if $t \in \mathcal{T}(\sigma)$ and $\tau_u \equiv 0$ is complete and satisfies (23). If $\eta < 1$, we construct this hybrid arc ξ with the same p dynamics as earlier. Also, we take $\tau_u(0, 0) = 0$ and in the set-valued dynamics of $\dot{\tau}_u$ for this solution, we choose

$$\dot{\tau}_u(t,j) = \begin{cases} -\eta & \text{if } \sigma(t) \in \mathcal{P}_{\mathrm{s}}, \ \tau_u(t,j) > 0, \\ 0 & \text{if } \sigma(t) \in \mathcal{P}_{\mathrm{s}}, \ \tau_u(t,j) = 0, \\ 1 - \eta & \text{if } \sigma(t) \in \mathcal{P}_{\mathrm{u}}. \end{cases}$$
(25)

Proceeding by contradiction, assume that this solution to the hybrid system is not complete. This happens when the solution reaches the boundary of $\mathcal{C} \cup \mathcal{D}$ and the continuous dynamics keeps "pushing" the solution to go beyond the boundary of $\mathcal{C} \cup \mathcal{D}$. By inspecting the dynamics of τ_u as defined in (25), we see that the solution is not complete if and only if there exists (\bar{t}, \bar{j}) such that $\tau_u(\bar{t}, \bar{j}) = T_0$ and $\sigma(\bar{t}) \in \mathcal{P}_u$. Note that the complete solutions whose hybrid domain only consists of discrete jumps after hybrid time (\bar{t}, \bar{j}) are ruled out by the ADT condition. Now, let

$$(\underline{t},\underline{j}) := \max\left\{(t,j) \in \operatorname{dom} \xi : (t,j) \leqslant (\overline{t},\overline{j}), \tau_u(t,j) = 0\right\}.$$

It is noted that $(\underline{t}, \underline{j})$ exists as $\tau_u(0, 0) = 0, \tau_u(\overline{t}, \overline{j}) = T_0$ and the function τ_u is continuous. Use the notation

$$j_s := \max\{j \in \mathbb{N} : (s, j) \in \operatorname{dom} \xi\}$$

$$(26)$$

and because the switching signal is piecewise constant, we have the observation that

$$\int_{\underline{t}}^{\overline{t}} \mathbf{1}_{\mathcal{P}_{\mathrm{s}}}(p(s, j_s)) ds = \int_{\underline{t}}^{\overline{t}} \mathbf{1}_{\mathcal{P}_{\mathrm{s}}}(\sigma(s_{j_s})) ds = \int_{\underline{t}}^{\overline{t}} \mathbf{1}_{\mathcal{P}_{\mathrm{s}}}(\sigma(s)) ds$$

and the relation also holds if replacing \mathcal{P}_s by \mathcal{P}_u . Hence we have

$$\begin{split} T_0 &= \tau_u(\bar{t},\bar{j}) - \tau_u(\underline{t},\underline{j}) = \int_{\underline{t}}^t \dot{\tau}_u(s,j_s) \, ds \\ &= \int_{\underline{t}}^{\overline{t}} \dot{\tau}_u(s,j_s) \mathbf{1}_{\mathcal{P}_u}(p(s,j_s)) \, ds + \int_{\underline{t}}^{\overline{t}} \dot{\tau}_u(s,j_s) \mathbf{1}_{\mathcal{P}_s}(p(s,j_s)) \, ds \\ &= \int_{\underline{t}}^{\overline{t}} (1-\eta) \mathbf{1}_{\mathcal{P}_u}(p(s,j_s)) \, ds + \int_{\underline{t}}^{\overline{t}} (-\eta) \mathbf{1}_{\mathcal{P}_s}(p(s,j_s)) \, ds \\ &= \int_{\underline{t}}^{\overline{t}} \mathbf{1}_{\mathcal{P}_u}(\sigma(s)) \, ds - \int_{\underline{t}}^{\overline{t}} \eta \, ds \\ &= T_{\sigma,\mathcal{P}_u}(\underline{t},\overline{t}) - \eta(\overline{t}-\underline{t}). \end{split}$$

This shows that the given switching signal σ has reached the maximum activation time over $[\underline{t}, \overline{t})$. Therefore, the switched system has to switch to a stable mode so that $\sigma(\bar{t}) \in \mathcal{P}_s$, which is a contradiction. This proves that there exists a complete solution to the hybrid system (22) which satisfies (23).

To prove the second part, we need to verify that the switching signal given by (24) satisfies the AAT condition. We first observe that (24) implies that $\sigma(s) = (s, j_s)$, where j_s is defined in (26). Since $\xi \in \mathcal{C} \cup \mathcal{D}$, we see that $\tau_u(t, j) \in [0, T]$ for all $(t, j) \in \text{dom } \xi$. Thus for any $(t_1, j_1), (t_2, j_2) \in \text{dom } \xi$ with $t_2 > t_1$,

$$\begin{split} T_0 &\ge \tau_u(t_2, j_2) - \tau_u(t_1, j_1) = \int_{t_1}^{t_2} \dot{\tau}_u(s, j_s) \, ds \\ &= \int_{t_1}^{t_2} \dot{\tau}_u(s, j_s) \mathbf{1}_{\mathcal{P}_u}(p(s, j_s)) \, ds + \int_{t_1}^{t_2} \dot{\tau}_u(s, j_s) \mathbf{1}_{\mathcal{P}_s}(p(s, j_s)) \, ds \\ &= \int_{t_1}^{t_2} (1 - \eta) \mathbf{1}_{\mathcal{P}_u}(\sigma(s)) \, ds + \int_{t_1}^{t_2} (-\eta) \mathbf{1}_{\mathcal{P}_s}(\sigma(s)) \, ds \\ &= \int_{t_1}^{t_2} \mathbf{1}_{\mathcal{P}_u}(\sigma(s)) \, ds - \int_{t_1}^{t_2} \eta \, ds \\ &= T_{\sigma, \mathcal{P}_u}(t_1, t_2) - \eta(t_1 - t_2). \end{split}$$

This shows that the switching signal σ defined by (24) satisfies the AAT condition and this completes the proof.

Remark 4 We make a comparison between our model and the one proposed in [27]. While we use two timers to track ADT and AAT constraints individually, a single timer is developed in the context of a small-gain theorem in [27], which imitates a switching signal under both ADT and AAT constraints. We acknowledge here that although the models are different, Proposition 1 in [27] is analogous to Proposition 5 in our paper. Nevertheless, it turns out that for our purpose (see Section 4.4), using two individual timers makes the argument fairly intuitive and straightforward.

Using the result of Proposition 5, one can write the switched system (3), along with the ADT and AAT constraints on the switching signal, more naturally in the form of (16). To do so, we let

$$\xi := \begin{pmatrix} y \\ p \\ \tau \\ \tau_u \end{pmatrix} \in \mathbb{R}^n \times \mathcal{P} \times [0, N_0] \times [0, T_0] =: \mathcal{X}.$$
(27)

The hybrid model capturing the dynamics of the switched system, driven by an \mathbb{R}^m -valued disturbance d and a switching signal $\sigma \in \Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$, is compactly written as

$$\dot{\xi} \in \mathcal{F}(\xi, d) : \begin{cases} \dot{y} = f_p(y, d), \\ \dot{p} = 0, \\ \dot{\tau} \in [0, \frac{1}{\tau_a}], \\ \dot{\tau}_u \in [\mathbf{1}_{\mathcal{P}_u}(p) - \eta, \mathbf{1}_{\mathcal{P}_u}(p)], \end{cases}$$
(28a)

$$\xi^{+} \in \mathcal{G}(\xi, d) : \begin{cases} y^{+} \in G_{p}(y, d), \\ p^{+} \in \mathcal{P}, \\ \tau^{+} = \tau - 1, \\ \tau^{+}_{u} = \tau_{u}, \end{cases} \quad (\xi, d) \in \mathcal{D},$$
(28b)

where $\mathcal{C} := \mathcal{X} \times \mathbb{R}^m$, $\mathcal{D} := \mathbb{R}^n \times \mathcal{P} \times [1, N_0] \times [0, T_0] \times \mathbb{R}^m$, and $G_p(y, d) := \bigcup_{q \in \mathcal{P}} g_{p,q}(y, d)$. For the system (28), it is of interest to study whether the compact set

$$\mathcal{A} := \{0\}^n \times \mathcal{P} \times [0, N_0] \times [0, T_0] \tag{29}$$

is an attractor. Comparing (17) and (18) with (7) and (8) respectively, we have the following straightforward result:

Corollary 6 The switched system (3) is uniformly ISS (resp. iISS) over the class of switching signals $\Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$ if the hybrid system (28), with augmented state variable defined in (27), is ISS (resp. iISS) with respect to \mathcal{A} .

Proof. We show that a solution of the switched system (3) satisfies the estimate (7) if the hybrid system (28) is ISS with respect to \mathcal{A} . The proof for iISS follows by similar arguments. Let $x(\cdot; x_0, \omega, \sigma)$ be a solution of the switched system. From Proposition 5, we see that x is equivalent to a complete solution of the hybrid system (28) with a hybrid input signal d defined by $d(0,0) := \omega(0)$, $d(t,j) := \omega(t)$ for $t \in [t_j, t_{j+1})$, where we recall that $\{t_1, t_2, \cdots\} = \mathcal{T}(\sigma)$ is the set of switching instants, in the sense that (23) holds and $y(0,0) = x_0, y(t,j) = x(t; x_0, \omega, \sigma)$ for all $t \in [t_j, t_{j+1})$. With the hybrid system (28) being ISS, dom $\xi = \text{dom } d$ and the pair (ξ, d) satisfies the estimate (17). Therefore, by observing that $|x(t; x_0, \omega, \sigma)| = |y(t, j)| = |\xi(t, j)|_{\mathcal{A}}$ for all $t \in [t_j, t_{j+1})$, and

$$\|d\|_{(t,j)} = \max\left\{ \operatorname{ess\,sup}_{s \in [0,t] \setminus \mathcal{T}(\sigma)} |\omega(s)|, \sup_{s \in [0,t] \cap \mathcal{T}(\sigma)} |\omega(s)| \right\},\$$

the estimate in (7) is now obtained by taking $\beta(s,t) = \beta_h(s,t,0)$ and $\gamma_1 = \gamma_2 = \gamma_h$.

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4.3 Supporting lemmas

In this subsection we provide a couple of supporting lemmas which will be used in our proof of Theorem 1. We start with an obvious observation on ψ :

Lemma 7 The function ψ constructed in (12) is positive definite. It is also of class \mathcal{K}_{∞} if α_s is of class \mathcal{K}_{∞} .

Proof. By construction it is clear that ψ is positive definite. To show that $\psi \in \mathcal{K}_{\infty}$ when $\alpha_{s} \in \mathcal{K}_{\infty}$, notice that ψ is continuous, $\psi(0) = 0$ so we are left to show that ψ is strictly increasing and unbounded. To show ψ is strictly increasing, we take arbitrary $t_{2} > t_{1} \ge 0$. Suppose that $\alpha_{s}(s) + c(t_{2} - s)$ achieves minimum at some $t^{*} \in [0, t_{2}]$, that is, $\psi(t_{2}) = \alpha_{s}(t^{*}) + c(t_{2} - t^{*})$. We consider two cases by comparing t^{*} and t_{1} . If $t^{*} \le t_{1}$, then

$$\psi(t_2) = \alpha_{\rm s}(t^*) + c(t_1 - t^*) + c(t_2 - t_1) > \alpha_{\rm s}(t^*) + c(t_1 - t^*)$$

$$\geqslant \min_{s \in [0, t_1]} \{\alpha_{\rm s}(s) + c(t_1 - s)\} = \psi(t_1).$$

Otherwise, if $t^* > t_1$, then

$$\psi(t_2) = \alpha_{\rm s}(t_1) + (\alpha_{\rm s}(t^*) - \alpha_{\rm s}(t_1)) + c(t_2 - t^*) > \alpha_{\rm s}(t_1) \\ \ge \min_{s \in [0, t_1]} \{\alpha_{\rm s}(s) + c(t_1 - s)\} = \psi(t_1).$$

Therefore, we conclude that $\psi(t_2) > \psi(t_1)$. Next we show that $\lim_{t \to \infty} \psi(t) = \infty$. If this is not true, there exists K > 0 such that $\psi(t) \leq K$ for all $t \geq 0$. Because $\alpha_{\rm s} \in \mathcal{K}_{\infty}$, there exists r > 0 such that $\alpha_{\rm s}(r) = 2K$. Then, for any $s \in [0, r + \frac{2K}{c}]$, either $s \in [0, r]$ so that $c(r + \frac{2K}{c} - s) \geq 2K$, or $s \in (r, r + \frac{2K}{c}]$ so that $\alpha_{\rm s}(s) > 2K$. Thus,

$$\psi\left(r+\frac{2K}{c}\right) = \min_{s \in [0,r+\frac{2K}{c}]} \left\{ \alpha_{s}(s) + c\left(r+\frac{2K}{c}-s\right) \right\} \ge 2K > K,$$

which is a contradiction. This completes the proof.

The next lemma shows that ψ is globally one-sided Lipschitz. This property often appears in the study of differential inclusions (see, e.g., [6]).

Lemma 8 The function ψ constructed in (12) is a globally one-sided Lipschitz function on $\mathbb{R}_{\geq 0}$ with constant c; that is, for any $t_2 \geq t_1 \geq 0$, $\psi(t_2) - \psi(t_1) \leq c(t_2 - t_1)$.

Proof. For any $t_2 \ge t_1 \ge 0$,

$$\psi(t_2) = \min_{s \in [0, t_2]} \{ \alpha_s(s) + c(t_2 - s) \}$$

$$\leqslant \min_{s \in [0, t_1]} \{ \alpha_s(s) + c(t_2 - s) \}$$

$$= \min_{s \in [0, t_1]} \{ \alpha_s(s) + c(t_1 - s) \} + c(t_2 - t_1)$$

$$= \psi(t_1) + c(t_2 - t_1)$$

which is the desired inequality.

Next we define a function which will be used in the construction of the (i)ISS Lyapunov function for the hybrid system. Define the function φ : $\mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}_{\geq 0}$ by

$$\varphi(s) := \begin{cases} \exp\left(\int_1^s \frac{2c}{\psi(r)} dr\right), & \text{if } s > 0, \\ 0, & \text{if } s = 0. \end{cases}$$
(30)

where we recall c and ψ are introduced in (12). We state two properties associated with φ .

Lemma 9 It holds that $\varphi \in \mathcal{K}_{\infty}$.

Proof. By definition, $\varphi(s)$ is strictly increasing and continuous for all s > 0. It remains to show that φ is continuous at 0 and $\lim_{s \longrightarrow \infty} \varphi(s) = \infty$. It follows from (12) that $\psi(r) \leq cr$ for all $r \geq 0$ so $\frac{2c}{\psi(r)} \geq \frac{2}{r}$. Therefore when s > 1, $\int_{1}^{s} \frac{2c}{\psi(r)} dr \geq \int_{1}^{s} \frac{2c}{cr} dr = 2 \ln s$ so $\lim_{s \longrightarrow \infty} \varphi(s) = \infty$. On the other hand when s < 1, $\int_{1}^{s} \frac{2c}{\psi(r)} dr \leq \int_{1}^{s} \frac{2c}{cr} dr = 2 \ln s$ so $\lim_{s \longrightarrow 0^{+}} \varphi(s) = 0$ and thus φ is continuous at 0.

Lemma 10 Let $c_0 > 0$ and $s \ge t > 0$ such that $\varphi(s) \le c_0 \varphi(t)$. It follows that $\frac{\psi(s)}{\psi(t)} \le \sqrt{c_0}$.

Proof. Fix $s \ge t > 0$. Then, the inequality $\varphi(s) \le c_0 \varphi(t)$ implies that

$$\exp\left(\int_{1}^{s} \frac{2c}{\psi(r)} dr\right) \leqslant c_{0} \exp\left(\int_{1}^{t} \frac{2c}{\psi(r)} dr\right).$$

Take logarithm on both sides and subtract the integral on the right from the left, to get

$$\int_{t}^{s} \frac{2c}{\psi(r)} dr \leq \ln(c_0).$$
(31)

Define $\widetilde{\psi}_t : [t, \infty) \longrightarrow \mathbb{R}_+$ by $\widetilde{\psi}_t(r) := \psi(t) + c(r-t)$. Recall from Lemma 8 that c is also the global one-sided Lipschitz constant of ψ , so we have $\widetilde{\psi}_t(r) \ge \psi(r)$ for all $r \ge t$. Hence, continuing from (31), we have

$$\ln(c_0) \ge \int_t^s \frac{2c}{\psi(r)} dr \ge \int_t^s \frac{2c}{\widetilde{\psi}_t(r)} dr = \int_t^s \frac{2c}{\psi(t) + c(r-t)} dr$$
$$= 2\ln\left(\psi(t) + c(r-t)\right)\Big|_{r=t}^{r=s} = 2\ln\frac{\psi(t) + c(s-t)}{\psi(t)} \ge 2\ln\frac{\psi(s)}{\psi(t)},$$

which results in $\frac{\psi(s)}{\psi(t)} \leqslant \sqrt{c_0}$.

The next Lemma tells that a special transformation of function φ inherits the global one-sided Lispchitzness from ψ .

Lemma 11 The function $\theta : \mathbb{R}_{\geq 0} \longrightarrow \mathbb{R}_{\geq 0}$ defined by $\theta(s) := \varphi^{-1}(c_0\varphi(s))$ with any $c_0 \geq 1$ is globally one-sided Lipschitz with constant $\sqrt{c_0}$, where φ is defined by (30).

Proof. Before we start the proof, we note that φ^{-1} in the definition of θ is well-defined because $\varphi \in \mathcal{K}_{\infty}$, stated by Lemma 9. To show θ is globally one-sided Lipschitz with constant $\sqrt{c_0}$, it suffices to show that

$$\theta(t_2) - \theta(t_1) \leqslant \sqrt{c_0}(t_2 - t_1)$$

for all $t_2 \ge t_1 \ge 0$; in other words, we need to show that $\theta(t) - t\sqrt{c_0}$ is a non-increasing function on $\mathbb{R}_{\ge 0}$. To this end, we first observe that it follows from the definition of φ in (30) that

$$\varphi'(s) = \frac{2c\varphi(s)}{\psi(s)} \tag{32}$$

and

$$(\varphi^{-1})'(s) = \frac{1}{\varphi'(\varphi^{-1}(s))} = \frac{\psi(\varphi^{-1}(s))}{2cs}.$$
(33)

Hence, we conclude that for any $t \in \mathbb{R}_{\geq 0}$,

$$\frac{d}{dt}\theta(t) = \frac{d}{dt} \Big(\varphi^{-1}\big(c_0\varphi(t)\big)\Big) = \frac{\psi\big(\varphi^{-1}(c_0\varphi(t))\big)}{2cc_0\varphi(t)} \frac{2cc_0\varphi(t)}{\psi(t)} = \frac{\psi(\theta(t))}{\psi(t)}$$

Let $s := \theta(t)$, so that $\varphi(s) = c_0 \varphi(t)$. Since $c_0 \ge 1$, $s \ge t \ge 0$, Lemma 10 yields $\frac{\psi(s)}{\psi(t)} \le \sqrt{c_0}$, and

$$\frac{d}{dt} \big(\theta(t) - t\sqrt{c_0} \big) = \frac{\psi(s)}{\psi(t)} - \sqrt{c_0} \leqslant 0.$$

Therefore, $\theta(t) - t\sqrt{c_0}$ is non-increasing and this completes the proof.

4.4 Proof of the main result

We now use the developments carried out in Sections 4.1, 4.2, and 4.3 to prove Theorem 1.

Proof of Theorem 1. Since $\tau_a > \zeta^*$ and (15) holds, we pick $\zeta \in (\zeta^*, \tau_a)$ which is sufficiently close to ζ^* so that

$$(\kappa+1)\eta + \frac{\zeta}{\tau_a} < 1. \tag{34}$$

We prove the theorem by showing that

$$V(\xi) := \varphi^{-1} \left(e^{2c \left(\zeta \tau + (\kappa + 1)(T_0 - \tau_u) \right)} \varphi \left(V_p(x) \right) \right)$$
(35)

is an (i)ISS Lyapunov function for the hybrid system (28) corresponding to the switched system (3) with ADT and AAT constrained switching signals.

Firstly, to verify the condition (19), we set

$$\alpha_1(s) = \underline{\alpha}(s), \text{ and } \alpha_2(s) = \varphi^{-1} \left(e^{2c \left(\zeta N_0 + (\kappa+1)T_0 \right)} \varphi(\overline{\alpha}(s)) \right)$$

and the bounds in (19) clearly hold.

Secondly, to verify the condition (20) for the continuous flow, we define

$$W(\xi) := e^{2c\left(\zeta\tau + (\kappa+1)(T_0 - \tau_u)\right)}\varphi(V_p(x)).$$
(36)

Picking $(\xi, d) \in \mathcal{C}$ and $\overline{f} \in \mathcal{F}(\xi, d)$, we have

$$\left\langle \frac{\partial}{\partial \xi} W(\xi), \overline{f} \right\rangle = 2c\zeta W(\xi)\dot{\tau} - 2c(\kappa+1)W(\xi)\dot{\tau}_u + \left\langle e^{2c(\zeta\tau + (\kappa+1)(T_0 - \tau_u))}\varphi'(V_p(x))\frac{\partial}{\partial x}V_p(x), f_p(x,d) \right\rangle$$

$$= 2c\left(\zeta\dot{\tau} - (\kappa+1)\dot{\tau}_u + \frac{1}{\psi(V_p(x))}\left\langle \frac{\partial}{\partial x}V_p(x), f_p(x,d) \right\rangle \right)W(\xi)$$

where we have used (32) for φ' . When $p \in \mathcal{P}_s$, it follows from (10a) and (12) that $\left\langle \frac{\partial}{\partial x} V_p(x), f_p(x, d) \right\rangle \leq -\psi(V_p(x)) + \gamma(|d|)$. In addition $\dot{\tau} \in [0, \frac{1}{\tau_a}]$ and $\dot{\tau}_u \in [-\eta, 0]$, so

$$\left\langle \frac{\partial}{\partial \xi} W(\xi), \overline{f} \right\rangle \leqslant 2c \left(\frac{\zeta}{\tau_a} + (\kappa + 1)\eta - 1 \right) W(\xi) + \frac{2c W(\xi)}{\psi(V_p(x))} \gamma(|d|).$$

When $p \in \mathcal{P}_{u}$, it follows from (10b) and (13) that $\left\langle \frac{\partial}{\partial x} V_{p}(x), f_{p} \right\rangle \leq \kappa \psi(V_{p}(x)) + \gamma(|d|)$. In addition $\dot{\tau} \in [0, \frac{1}{\tau_{a}}]$ and $\dot{\tau}_{u} \in [1 - \eta, 1]$, so

$$\left\langle \frac{\partial}{\partial \xi} W(\xi), \overline{f} \right\rangle \leqslant 2c \left(\frac{\zeta}{\tau_a} + (\kappa + 1)(\eta - 1) + \kappa \right) W(\xi) + \frac{2cW(\xi)}{\psi(V_p(x))} \gamma(|d|)$$

and we reach the same upper bound as in the first case. Appealing to (34), we conclude that

$$\left\langle \frac{\partial}{\partial \xi} W(\xi), \overline{f} \right\rangle \leqslant -2 c a W(\xi) + \frac{2 c W(\xi)}{\psi(V_p(x))} \gamma(|d|)$$

where

$$a := 1 - \frac{\zeta}{\tau_a} - (\kappa + 1)\eta > 0.$$
 (37)

Consequently it can be computed that along the differential inclusion (28a),

$$\left\langle \frac{\partial}{\partial \xi} V(\xi), \overline{f} \right\rangle = (\varphi^{-1})'(W(\xi)) \left\langle \frac{\partial}{\partial \xi} W(\xi), \overline{f} \right\rangle$$

$$\leqslant \left(\frac{\psi(\varphi^{-1}(W(\xi)))}{2cW(\xi)}\right) \left(-2caW(\xi) + \frac{2cW(\xi)}{\psi(V_p(x))}\gamma(|d|)\right)$$
$$= -a\psi(\varphi^{-1}(W(\xi))) + \frac{\psi(\varphi^{-1}(W(\xi)))}{\psi(V_p(x))}\gamma(|d|)$$
$$= -a\psi(V(\xi)) + \frac{\psi(V(\xi))}{\psi(V_p(x))}\gamma(|d|).$$

To further simplify the right-hand side, let $v := V_p(x)$, $w := V(\xi)$. Then we have $w \ge v \ge 0$ and $\varphi(w) = W(\xi) = e^{2c \left(\zeta \tau + (\kappa+1)(T_0 - \tau_u)\right)} \varphi(V_p(x)) \le e^{2c \left(\zeta N_0 + (\kappa+1)T_0\right)} \varphi(v)$. Hence from Lemma 10,

$$\left\langle \frac{\partial}{\partial \xi} V(\xi), \overline{f} \right\rangle \leqslant -a\psi(V(\xi)) + e^{c\left(\zeta N_0 + (\kappa+1)T_0\right)} \gamma(|d|)$$
$$\leqslant -a\psi \circ \alpha_1(|\xi|_{\mathcal{A}}) + e^{c\left(\zeta N_0 + (\kappa+1)T_0\right)} \gamma(|d|)$$

and (20) holds with

$$\alpha_c(s) = a\psi \circ \alpha_1(s), \quad \gamma_c(s) = e^{c\left(\zeta N_0 + (\kappa+1)T_0\right)}\gamma(s).$$
(38)

It follows from Lemma 7 that if α_s is positive definite (resp. $\alpha_s \in \mathcal{K}_{\infty}$), then α_c is also positive definite (resp. $\alpha_c \in \mathcal{K}_{\infty}$). In addition, $\gamma_c \in \mathcal{K}$.

Lastly, to verify the condition (21) for the discrete jumps, we pick $(\xi, d) \in \mathcal{D}$ and $\bar{g} \in \mathcal{G}(\xi, d)$. Notice that

$$\begin{split} \varphi \left(\chi(V_p(x)) \right) &= \exp \left(\int_1^{\chi(V_p(x))} \frac{2}{\psi(r)} dr \right) \\ &= \exp \left(\int_{V_p(x)}^{\chi(V_p(x))} \frac{2c}{\psi(r)} dr \right) \exp \left(\int_1^{V_p(x)} \frac{2c}{\psi(r)} dr \right) \leqslant e^{2c\zeta^*} \varphi(V_p(x)) \end{split}$$

where we recall that ζ^* is defined in (14). Hence, if we define

$$\theta(s) := \varphi^{-1} \left(e^{2c(\zeta(\tau-1) + (\kappa+1)(T_0 - \tau_u))} \varphi(s) \right),$$

and recall the definition of W from (36), we have the estimate that $\theta(\chi(V_p(x))) \leq \varphi^{-1}(e^{2c(\zeta(\tau-1)+(\kappa+1)(T_0-\tau_u))+2c\zeta^*}\varphi(V_p(x))) = \varphi^{-1}(e^{2c(\zeta^*-\zeta)}W(\xi))$. Therefore, using the assumption (L3) and Lemma 11, we get

$$V(\bar{g}) = \varphi^{-1} \left(e^{2c(\zeta\tau^{+} + (\kappa+1)(T_0 - \tau_u^{+}))} \varphi(V_{p^+}(x^{+})) \right)$$

$$\leq \varphi^{-1} \left(e^{2c(\zeta(\tau-1) + (\kappa+1)(T_0 - \tau_u))} \varphi(\chi(V_p(x)) + \rho(|d|)) \right)$$

$$= \theta(\chi(V_p(x)) + \rho(|d|))$$

$$\leq \theta(\chi(V_p(x)) + e^{c(\zeta(\tau-1) + (\kappa+1)(T_0 - \tau_u))} \rho(|d|)$$

$$\leq \varphi^{-1} \left(e^{2c(\zeta^* - \zeta)} W(\xi) \right) + e^{c(\zeta(N_0 - 1) + (\kappa + 1)T_0)} \rho(|d|)$$

= $\varphi^{-1} \left(e^{2c(\zeta^* - \zeta)} \varphi(V(\xi)) \right) + e^{c(\zeta(N_0 - 1) + (\kappa + 1)T_0)} \rho(|d|).$

Define

$$\kappa(s) := s - \varphi^{-1} \left(e^{2c(\zeta^* - \zeta)} \varphi(s) \right), \tag{39}$$

then the inequality (21) is seen to hold with

$$\alpha_d(s) = \kappa \circ \alpha_1(s), \quad \gamma_d(s) = e^{c(\zeta(N_0 - 1) + (\kappa + 1)T_0)}\rho(s).$$

$$\tag{40}$$

Recall that $\zeta > \zeta^*$ and $\varphi \in \mathcal{K}_{\infty}$; thus by construction κ is positive definite, so is α_d . In addition, $\gamma_d \in \mathcal{K}$. Lastly we show that $\alpha_d \in \mathcal{K}_{\infty}$ when $\alpha_s \in \mathcal{K}_{\infty}$ and it suffices to show that $\kappa \in \mathcal{K}_{\infty}$. It follows from (39) that $\varphi((s - \kappa(s)) = e^{2c(\zeta^* - \zeta)}\varphi(s)$. Using the definition of φ in (30), and taking logarithm, it follows that

$$\int_1^{s-\kappa(s)} \frac{2c}{\psi(r)} dr = 2c(\zeta^* - \zeta) + \int_1^s \frac{2c}{\psi(r)} dr.$$

Hence, we have

$$\int_{s-\kappa(s)}^{s} \frac{1}{\psi(r)} dr = \zeta - \zeta^* > 0.$$

$$\tag{41}$$

By Lemma 7, when $\alpha_{s} \in \mathcal{K}_{\infty}$, $\psi \in \mathcal{K}_{\infty}$ as well and thus $\frac{1}{\psi(r)}$ decreases to 0 as r increases to infinity. Therefore, (41) holds for all s > 0 only when the length of the interval of integration grows to infinity as $s \longrightarrow \infty$. In other words, we conclude that $\kappa \in \mathcal{K}_{\infty}$.

With all the three conditions (19), (20) and (21) verified, we appeal to Lemma 3 (resp. Lemma 4) and conclude that the hybrid system (28) is ISS (resp. iISS) when $\alpha_s \in \mathcal{K}_{\infty}$ (resp. α_s is only positive definite). Then by Corollary 6, the switched system (3) is uniformly ISS (resp. uniformly iISS) over $\Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$.

Remark 5 It is observed that the formulas for γ_c, γ_d in (38), (40) only depend on the parameters N_0, T_0 but not τ_a, η . Since it is known in [3,18] that the (i)ISS gain is related to γ_c, γ_d , we conclude that the average dwell-time or percentage activation time of the unstable modes do not affect the (i)ISS gain. On the other hand, it is seen that the functions α_c, α_d depend on τ_a, η . Thus we conclude that the convergence rate of the unforced switched system depends on the average dwell-time and percentage activation time of the unstable modes. In particular when the left-hand side of (15) is close to 1, the convergence rate of the unforced switched system may become extremely slow and when it is larger than 1, the unforced switched system may become unstable and thus the switched system with inputs is not (i)ISS. This observation is consistent with the known results in [10].

5 Case Studies

In this section we will demonstrate how our results can be applied to determining stability of several switched nonlinear systems. The subsystems of these switched systems will be constructed using the matrices:

$$A_1 = \begin{pmatrix} -0.1 & -1\\ 2 & -0.1 \end{pmatrix}, \ A_2 = \begin{pmatrix} -0.1 & -2\\ 1 & -0.1 \end{pmatrix}, \ A_3 = \begin{pmatrix} 0.1 & -1\\ 2 & 0.1 \end{pmatrix}.$$
(42)

Notice that A_1 and A_2 are both Hurwitz and they have the same eigenvalues; A_3 is not Hurwitz. In addition, it is discussed in [13, Section 3.2] that the switched linear system with modes $\mathcal{P} = \{1, 2\}$

$$\dot{x} = A_{\sigma} x \text{ if } t \notin \mathcal{T}(\sigma), \tag{43a}$$

$$x = x^{-} \quad \text{if } t \in \mathcal{T}(\sigma). \tag{43b}$$

has no common Lyapunov function and for some particular switching signals the solution will diverge.

5.1 Switched system with saturation-like scaling

x

Consider the switched system with modes $\mathcal{P} = \{1, 2\}$

$$\dot{x} = \frac{1}{1+|x|} A_{\sigma} x + \omega \quad \text{if } t \notin \mathcal{T}(\sigma), \tag{44a}$$

$$=x^{-}$$
 if $t \in \mathcal{T}(\sigma)$ (44b)

where $\omega \in \mathbb{R}^2$ is the input. The coefficient $\frac{1}{1+|x|}$ mimics the effect of saturation such that $\dot{x} \approx A_{\sigma}x + \omega$ when |x| is small and $\dot{x} \approx A_{\sigma}\frac{x}{|x|} + \omega$ when |x| is large. The subsystems of (44) are both iISS, which can be verified by picking the iISS-Lyapunov functions

$$V_p(x) = \sqrt{1 + x^{\top} M_p x} - 1$$
 (45)

where M_p are the solutions to the Lyapunov equations

$$M_p A_p + A_p^{\top} M_p + I = 0. (46)$$

However, this switched system is not iISS under slow switching. To see this, notice that when $u \equiv 0$, the solutions of the switched system (44) are the same as the solutions of (43) after a time re-parameterization. By similar arguments as discussed in [20], for any $\tau_a > 0$ and $N_0 \ge 1$, we can pick an initial state far away enough from the origin and a switching signal from $\Sigma_{ADT}(\tau_a, N_0)$, yet the solution diverges. Hence the switched system (44) is not uniformly iISS over $\Sigma_{ADT}(\tau_a, N_0)$ for any $\tau_a > 0$. Nevertheless, we show that the switched system (44) is iISS over AAT switching signals. To do this, let V_1 be defined

as in (45) with p = 1 and set $V_2 = V_1$. We first see that (9) is verified with $\underline{\alpha}(s) := \sqrt{1 + \underline{\sigma}(M_1)s^2} - 1$ and $\bar{\alpha}(s) := \sqrt{1 + \overline{\sigma}(M_1)s^2} - 1$, where $\bar{\sigma}(\cdot), \underline{\sigma}(\cdot)$ denote the largest and smallest singular values, respectively. We further denote $Q_1 := -I, Q_2 := M_1 A_2 + A_2^{\top} M_1$. Notice that because V_2 is not a Lyapunov function for the subsystem p = 2, Q_2 defined in this way is sign indefinite. It can be computed that

$$\left\langle \frac{\partial}{\partial x} V_p(x), f_p(x, \omega) \right\rangle = \frac{1}{\sqrt{1 + x^\top M_1 x}} \left(\dot{x}^\top M_1 x + x^\top M_1 \dot{x} \right)$$

$$= \frac{x^\top \left(A_p^\top M_1 + M_1 A_p \right) x}{(1 + |x|)\sqrt{1 + x^\top M_1 x}} + \frac{2x^\top M_p}{\sqrt{1 + x^\top M_p x}} \omega$$

$$= \frac{x^\top Q_p x}{(1 + |x|)\sqrt{1 + x^\top M_1 x}} + \frac{2x^\top M_1}{\sqrt{1 + x^\top M_1 x}} \omega$$

$$= \frac{x^\top Q_p x}{(1 + |x|)(V_p(x) + 1)} + \frac{2x^\top M_1}{\sqrt{1 + x^\top M_1 x}} \omega.$$

It is observed that the function $h(s) = \frac{s^2}{1+s}$ is increasing on $\mathbb{R}_{\geq 0}$; in addition, from (45) we have $\sqrt{\frac{V_p^2 + 2V_p}{\bar{\sigma}(M_1)}} \leq |x| \leq \sqrt{\frac{V_p^2 + 2V_p}{\underline{\sigma}(M_1)}}$. Hence we conclude that

$$\left\langle \frac{\partial}{\partial x} V_1(x), f_1(x,\omega) \right\rangle \leqslant -\frac{V_1^2 + 2V_1}{\left(\bar{\sigma}(M_1) + \sqrt{\bar{\sigma}(M_1)(V_1^2 + 2V_1)}\right)(V_1 + 1)} + \frac{2\bar{\sigma}(M_1)}{\sqrt{\underline{\sigma}(M_1)}} |\omega|$$

which fits the form (10a) with $\alpha_r(s) := \frac{s^2 + 2s}{\sqrt{\underline{\sigma}(M_1)}}$ and $\gamma(s) := \frac{s^2 + 2s}{\sqrt{\underline{\sigma}(M_1)}}$

In fits the form (10a) with $\alpha_{s}(s) := \frac{1}{\left(\overline{\sigma}(M_{1}) + \sqrt{\overline{\sigma}(M_{1})(s^{2}+2s)}\right)(s+1)}$ and $\gamma(s)$. $\frac{2\bar{\sigma}(M_1)}{\sqrt{\underline{\sigma}(M_1)}}s.$ It can be verified that α_s is globally Lipschitz so as discussed in Sec-

tion 3, we can set $\psi = \alpha_s$. For p = 2,

$$\left\langle \frac{\partial}{\partial x} V_2(x), f_2(x,\omega) \right\rangle \leqslant \frac{\bar{\sigma}(Q_2)(V_2^2 + 2V_2)}{\left(\underline{\sigma}(M_1) + \sqrt{\underline{\sigma}(M_1)(V_2^2 + 2V_2)}\right)(V_2 + 1)} + \frac{2\bar{\sigma}(M_1)}{\sqrt{\underline{\sigma}(M_1)}} |\omega|$$

which fits the form (10b) with $\alpha_{\rm u}(s) := \frac{\bar{\sigma}(Q_2)(s^2+2s)}{\left(\underline{\sigma}(M_1) + \sqrt{\underline{\sigma}(M_1)(s^2+2s)}\right)(s+1)}$. Plugging the numerical values (42) for A_1, A_2 into the formulas for $\psi, \alpha_{\rm u}$, it is computed that $\kappa = \sup_{s \ge 0} \frac{\alpha_{\rm u}(s)}{\psi(s)} = \bar{\sigma}(Q_2) \frac{\bar{\sigma}(M_1)}{\underline{\sigma}(M_1)} = 24.9436$ and $\frac{1}{1+\kappa} = 0.0385$. In addition, because we have chosen $V_1 = V_2$ and the switched system has no resets at switches, $\zeta^* = 0$ and hence by Theorem 1 and the discussion after the theorem statement, we conclude that the switched system (44) is uniformly ISS over $\Sigma_{ADT}(\tau_a, N_0) \cap \Sigma_{AAT}(\{2\}, \eta, T_0)$ for any $\tau_a > 0, N_0 \ge 1, \eta < 0.0385, T_0 \ge 0.$

5.2 Switched bilinear system with resets and stable modes

Consider the switched system with modes $\mathcal{P} = \{1, 2\}$ $\dot{x} = A_{\sigma}x + \omega Bx \qquad \text{if } t \notin \mathcal{T}(\sigma),$ (47a)

$$x = Dx^{-} + \omega Ex^{-} \quad \text{if } t \in \mathcal{T}(\sigma), \tag{47b}$$

where $\omega \in \mathbb{R}$ is the input, A_p are given in (42) and $B, D, E \in \mathbb{R}^{2 \times 2}$. Each subsystem is shown to be iISS (see, e.g., [5]) by picking the iISS-Lyapunov function $V_p(x)$ as

$$V_p(x) := \ln(1 + x^{\top} M_p x)$$
 (48)

where again M_p are the solutions of (46). We have

$$\underline{\alpha}(|x|) := \ln(1 + \min_{p \in \mathcal{P}} \underline{\sigma}(M_p)|x|^2) \leqslant V_p(x) \leqslant \ln(1 + \max_{p \in \mathcal{P}} \overline{\sigma}(M_p)|x|^2) =: \overline{\alpha}(|x|).$$
(49)

For the continuous flow, it can be computed that

$$\begin{split} \left\langle \frac{\partial}{\partial x} V_p(x), f_p(x, \omega) \right\rangle &= \frac{1}{1 + x^\top M_p x} \left(\dot{x}^\top M_p x + x^\top M_p \dot{x} \right) \\ &= \frac{1}{1 + x^\top M_p x} x^\top (A_p^\top M_p + M_p A_p) x + \frac{2x^\top M_p B x}{1 + x^\top M_p x} \omega \\ &= -\frac{|x|^2}{1 + x^\top M_p x} + \frac{2x^\top M_p B x}{1 + x^\top M_p x} \omega \\ &\leqslant -\frac{(e^{V_p} - 1)}{\bar{\sigma}(M_p) e^{V_p}} + \frac{\bar{\sigma}(M_p B)}{\underline{\sigma}(M_p)} |\omega| \end{split}$$

Hence (10a) is satisfied for all $p \in \mathcal{P}_s = \mathcal{P}$ with $\alpha_s(s) := \frac{1-e^{-s}}{\sigma_{\max}}, \gamma(s) := c_1 s$ where $\sigma_{\max} := \max_{p \in \mathcal{P}} \bar{\sigma}(M_p), c_1 := \max_{p \in \mathcal{P}} \frac{\bar{\sigma}(M_pB)}{\underline{\sigma}(M_p)}$. Now consider the discrete jumps and denote $g(x, \omega) := Dx + \omega Ex$. Let $\mu > 0$ be such that $x^{\top} D^{\top} M_p Dx \leq \mu x^{\top} M_q x$ for all $x \in \mathbb{R}^n$ and $p, q \in \mathcal{P}$, then we have the inequality that

$$\begin{split} 1 + g(x,\omega)^{\top} Mg(x,\omega) &= 1 + (Dx + uEx)^{\top} M_p(Dx + uEx) \\ &= 1 + x^{\top} D^{\top} M_p Dx + x^{\top} (2uE^{\top} M_p D + \omega^2 E^{\top} M_p E) x \\ &\leqslant 1 + \mu x^{\top} M_q x + x^{\top} (2uE^{\top} M_p D + \omega^2 E^{\top} M_p E) x \\ &= (1 + \mu x^{\top} M_q x) \left(1 + \frac{x^{\top} (2uE^{\top} M_p D + \omega^2 E^{\top} M_p E) x}{1 + \mu x^{\top} M_q x} \right) \\ &\leqslant (1 + \mu x^{\top} M_q x) \left(1 + \frac{2\bar{\sigma}(E^{\top} M_p D) |\omega| + \bar{\sigma}(E^{\top} M_p E) |\omega|^2}{\mu \underline{\sigma}(M_q)} \right) \end{split}$$

Hence when a switch occurs,

$$V_p(g(x,\omega)) = \ln(1+g(x,\omega)^\top Mg(x,\omega))$$

$$\leq \ln\left((1+\mu x^\top M_q x)\left(1+\frac{2\bar{\sigma}(E^\top M_p D)|\omega|+\bar{\sigma}(E^\top M_p E)|\omega|^2}{\mu\underline{\sigma}(M_q)}\right)\right)$$

$$= \ln(1+\mu x^\top M_q x) + \ln\left(1+\frac{2\bar{\sigma}(E^\top M_p D)|\omega|+\bar{\sigma}(E^\top M_p E)|\omega|^2}{\mu\underline{\sigma}(M_q)}\right)$$

$$= \ln\left(1 + \mu\left(e^{V_q(x)} - 1\right)\right) + \ln\left(1 + \frac{2\bar{\sigma}(E^{\top}M_pD)|\omega| + \bar{\sigma}(E^{\top}M_pE)|\omega|^2}{\mu\underline{\sigma}(M_q)}\right)$$

Hence (11) is satisfied with $\chi(s) := \ln \left(1 + \mu \left(e^s - 1\right)\right)$ and $\rho(s) := \ln(1 + c_2s + c_3s^2)$, where $c_2 := \frac{2 \max_{p \in \mathcal{P}} \bar{\sigma}(E^\top M_p D)}{\mu \min_{q \in \mathcal{P}} \underline{\sigma}(M_q)}$, $c_3 := \frac{2 \max_{p \in \mathcal{P}} \bar{\sigma}(E^\top M_p E)}{\mu \min_{q \in \mathcal{P}} \underline{\sigma}(M_p)}$. Therefore Assumption 1 is verified and since both modes are stable, Assumption 2 is omitted. Again because α_s is globally Lipschitz, we let $\psi = \alpha_s$ and it can be computed that

$$\begin{split} \zeta^* &= \sup_{s>0} \int_s^{\chi(s)} \frac{1}{\psi(r)} dr = \sup_{s>0} \int_s^{\chi(s)} \frac{\sigma_{\max} e^r}{(e^r - 1)} dr \\ &= \sigma_{\max} \sup_{s>0} \ln(e^r - 1) \Big|_s^{\ln(\mu(e^s - 1) + 1)} \\ &= \sigma_{\max} \sup_{s>0} \ln \frac{(\mu(e^s - 1) + 1) - 1}{e^s - 1} \\ &= \sigma_{\max} \ln \mu \end{split}$$

and Assumption 3 is verified. Hence we conclude from the special case of Theorem 1 that the switched system (47) with modes $\mathcal{P} = \{1, 2\}$ is uniformly iISS over $\Sigma_{ADT}(\tau_a, N_0)$ for all $\tau_a > \sigma_{\max} \ln \mu, N_0 \ge 1$. In particular if D = I, it can be numerically computed that $\sigma_{\max} \ln \mu = 5.1804$.

The same arguments also work for switched bilinear systems with resets and stable modes in general, in the form

$$\dot{x} = A_{\sigma}x + \sum_{j=1}^{m_c} B_{\sigma,j}x\omega_j + C_{\sigma}\omega, \qquad \text{if } t \in \mathcal{T}(\sigma), \qquad (50a)$$

$$x = D_{(\sigma^-,\sigma)}x^- + \sum_{k=1}^{m_d} E_{(\sigma^-,\sigma),k}x^-\omega_k + F_{(\sigma^-,\sigma)}\omega \quad \text{if } t \in \mathcal{T}(\sigma),$$
(50b)

where $x \in \mathbb{R}^n$ is the state, $\omega \in \mathbb{R}^m$ is the input and $A_p, B_{p,j}, D_{(p,q)}, E_{(p,q),k} \in \mathbb{R}^{n \times n}, C_p, F_{(p,q)} \in \mathbb{R}^{n \times m}$ for all $p, q \in \mathcal{P}, j = 1, \cdots, m_c, k = 1, \cdots, m_d$.

Proposition 12 Consider the switched system (50) with finitely many modes and assume that A_p are Hurwitz for all $p \in \mathcal{P}$ so that there exist positive definite symmetric matrices $M_p, Q_p \in \mathbb{R}^{n \times n}$ for which $A_p^{\top} M_p + M_p A_p + Q_p = 0$ hold. Let $\lambda > 0$ be such that $x^{\top} Q_p x \ge \lambda x^{\top} M_p x$ and $\mu > 0$ be such that $x^{\top} D_{(p,q)}^{\top} M_p D_{(p,q)} x \le \mu x^{\top} M_q x$ for all $x \in \mathbb{R}^n$ and $p, q \in \mathcal{P}$. Then (50) is uniformly iISS over $\Sigma_{\text{ADT}}(\tau_a, N_0)$ for all $\tau_a > \frac{\ln \mu}{\lambda}$ and $N_0 \ge 1$.

The proof is similar to the one carried out for system (47) and is provided in the Appendix. Remark 6 Notice that λ, μ defined in Proposition 12 are exactly the linear decay rate in the continuous flow and linear growth rate in the discrete jumps, respectively for the unforced switched system (50). It is known that in this case, $\frac{\ln \mu}{\lambda}$ is the lower bound on ADT such that the unforced switched linear system is globally asymptotically stable (see [11]). It follows that the bilinear input does not break the stability properties when the switched system is under slow switching. In addition, when there exists a common Lyapunov function, all M_p 's are the same and further when D = I, we conclude $\zeta^* = 0$ so τ_a can be as small as possible. This is indeed the case because such switched system is iISS with arbitrary switching [16].

5.3 Switched bilinear system with unstable modes

For the last example we still consider the switched system with dynamics given by (47) but now the set of modes is $\mathcal{P} = \{1, 3\}$. Recall that A_3 given in (42) is non-Hurwitz so this switched system is not uniformly iISS over $\Sigma_{ADT}(\tau_a, N_0)$ for any $\tau_a > 0$. We also let D = 2I, that is, whenever a switch occurs the state is scaled by 2 and hence this switched system is also not uniformly iISS over $\Sigma_{AAT}(\mathcal{P}_u, \eta, T_0)$ for any $\eta \in [0, 1]$. However, if we pick $V_3 = V_1 = \ln(1 + x^{\top} M_1 x)$, it can be computed that

$$\left\langle \frac{\partial}{\partial x} V_3(x), f_3(x,\omega) \right\rangle = \frac{1}{1+x^\top M_1 x} x^\top (A_3^\top M_1 + M_1 A_3) x + \frac{2x^\top M_1 B x}{1+x^\top M_1 x} \omega$$

$$\leqslant \frac{\bar{\sigma} (A_3^\top M_1 + M_1 A_3) |x|^2}{1+x^\top M_1 x} + \frac{2x^\top M_1 B x}{1+x^\top M_1 x} \omega$$

$$\leqslant \frac{\bar{\sigma} (A_3^\top M_1 + M_1 A_3)}{\underline{\sigma}(M_1)} \left(1 - e^{V_p(x)}\right) + \frac{\bar{\sigma}(M_1 B)}{\underline{\sigma}(M_1)} \omega$$

So we have $\mathcal{P}_{\mathbf{u}} = \{3\}$ with $\alpha_{\mathbf{u}}(s) := \frac{\overline{\sigma}(A_3^{\top}M_1 + M_1A_3)}{\underline{\sigma}(M_1)} (1-e^s)$. Thus Assumption 2 holds with $\kappa = \frac{\alpha_{\mathbf{u}}}{\psi} = \overline{\sigma}(A_3^{\top}M_1 + M_1A_3)\frac{\overline{\sigma}(M_1)}{\underline{\sigma}(M_1)} = 3.9868$. On the other hand, V_p being the same and D = 2I imply $\mu = 4$. Thus Assumption 3 holds with $\zeta^* = \sigma_{\max} \ln \mu = 10.3857$. Therefore by Theorem 1 we conclude that (47) is uniformly iISS over $\Sigma_{\mathrm{ADT}}(\tau_a, N_0) \cap \Sigma_{\mathrm{AAT}}(\{3\}, \eta, T_0)$ for any $N_0 \ge 1, T_0 \ge 0$ and all $\tau_a > 0, \eta \in [0, 1]$ satisfying $3.9868\eta + \frac{10.3857}{\tau_a} < 1$.

6 Conclusion

Within the context of literature on stability of switched systems, this paper provided conditions on the switching signals such that a switched nonlinear system with inputs, resets and unstable modes is ISS or iISS. In particular, we considered the case where the decay or growth rates of all subsystems are allowed to be nonlinear, and the changes of Lyapunov function values at switching instants are also assumed to be nonlinear. Under some mild assumptions on the switched system, and using the hybrid system framework, we derived a mixed ADT and AAT condition which guarantees ISS or iISS of the switched system. Using the main result of our work, switched systems with saturating dynamics and switched bilinear systems were shown to be uniformly iISS over switching signals satisfying this mixed ADT and AAT condition.

Appendix

Proof of Proposition 12. First of all, each subsystem is shown to be iISS (see, e.g., [5]) by picking the iISS-Lyapunov function $V_p(x)$ as

$$V_p(x) := \ln(1 + x^{\top} M_p x).$$
 (51)

We have

$$\underline{\alpha}(|x|) := \ln(1 + \min_{p \in \mathcal{P}} \underline{\sigma}(M_p)|x|^2) \leqslant V_p(x) \leqslant \ln(1 + \max_{p \in \mathcal{P}} \bar{\sigma}(M_p)|x|^2) =: \bar{\alpha}(|x|).$$
(52)

So the condition (L1) is verified. For the continuous flow, it can be computed that

$$\begin{split} \left\langle \frac{\partial}{\partial x} V_p(x), f_p(x, \omega) \right\rangle &= \frac{1}{1 + x^\top M_p x} \left(\dot{x}^\top M_p x + x^\top M_p \dot{x} \right) \\ &= \frac{1}{1 + x^\top M_p x} x^\top (A_p^\top M_p + M_p A_p) x + \sum_{j=1}^{m_c} \frac{2x^\top M_p B_{p,j} x}{1 + x^\top M_p x} \omega_j + \frac{2x^\top M_p C_p}{1 + x^\top M_p x} \\ &\leqslant -\frac{x^\top Q_p x}{1 + x^\top M_p x} + \sqrt{\sum_{j=1}^{m_c} \left(\frac{2x^\top M_p B_{p,j} x}{1 + x^\top M_p x} \right)^2} |\omega| + \frac{2|x^\top M_p C_p|}{1 + x^\top M_p x} |\omega| \\ &\leqslant -\frac{\lambda x^\top M_p x}{1 + x^\top M_p x} + 2\sqrt{n} \frac{\max_j \bar{\sigma}(M_p B_{p,j})}{\underline{\sigma}(M_p)} |\omega| + \frac{\bar{\sigma}(M_p C_p)}{\sqrt{\underline{\sigma}(M_p)}} |\omega| \\ &= -\lambda (1 - e^{-V_p(x)}) + \left(2\sqrt{n} \frac{\max_j \bar{\sigma}(M_p B_{p,j})}{\underline{\sigma}(M_p)} + \frac{\bar{\sigma}(M_p C_p)}{\sqrt{\underline{\sigma}(M_p)}} \right) |\omega|. \end{split}$$

Hence (10a) is satisfied for all $p \in \mathcal{P}_{s} = \mathcal{P}$ with

$$\alpha_{s}(s) := \lambda(1 - e^{-s}), \gamma(s) := \left(\max_{p} \left(2\sqrt{n} \frac{\max_{j} \bar{\sigma}(M_{p}B_{p,j})}{\underline{\sigma}(M_{p})} + \frac{\bar{\sigma}(M_{p}C_{p})}{\sqrt{\underline{\sigma}(M_{p})}} \right) \right) s.$$

Now consider the discrete jumps. we start by defining the terms:

$$K_{DD} := (D_{(q,p)}x^{-})^{\top} M_p D_{(q,p)}x^{-},$$

$$K_{EE} := \Big(\sum_{k=1}^{m_d} E_{(q,p),k}x^{-}\omega_k\Big)^{\top} M_p\Big(\sum_{k=1}^{m_d} E_{(q,p),k}x^{-}\omega_k\Big),$$

$$K_{FF} := (F_{(q,p)}\omega)^{\top} M_p(F_{(q,p)}\omega),$$

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$$K_{DE} := (D_{(q,p)}x^{-})^{\top} M_p \Big(\sum_{k=1}^{m_d} E_{(q,p),k} x^{-} \omega_k \Big),$$

$$K_{DF} := (D_{(q,p)}x^{-})^{\top} M_p (F_{(q,p)} \omega),$$

$$K_{EF} := \Big(\sum_{k=1}^{m_d} E_{(q,p),k} x^{-} \omega_k \Big)^{\top} M_p (F_{(q,p)} \omega).$$

These are the terms when expanding the product

$$\left(D_{(q,p)}x^{-} + \sum_{k=1}^{m_d} E_{(q,p),k}x^{-}\omega_k + F_{(q,p)}\omega\right)^{\top} M_p \left(D_{(q,p)}x^{-} + \sum_{k=1}^{m_d} E_{(q,p),k}x^{-}\omega_k + F_{(q,p)}\omega\right).$$

From the hypothesis of Proposition (12) we have $K_{DD} \leq \mu(x^-)^\top M_q x^-$. Meanwhile, use the fact that $y^\top M y \leq \bar{\sigma}(M) |y|^2$ for any matrix $M \in \mathbb{R}^{n \times n}$ and any vectors $y \in \mathbb{R}^y$, in addition to the inequality that

$$\Big|\sum_{k=1}^{m_d} E_{(q,p),k} x \omega_k\Big| \leqslant \Big|\sum_{k=1}^{m_d} \omega_k E_{(q,p),k}\Big| |x| \leqslant \sqrt{\sum_{k=1}^{m_d} \bar{\sigma}(E_{(q,p),k})^2} |\omega| |x|,$$

we conclude that $K_{EE} \leq k_{EE} |x^-|^2 |\omega|^2$, $K_{FF} \leq k_{FF} |\omega|^2$, $K_{DE} \leq k_{DE} |x^-|^2 |\omega|$, $K_{DF} \leq k_{DF} |x^-||\omega|$ and $K_{EF} \leq k_{EF} |x^-||\omega|^2$, where

$$k_{EE} := \bar{\sigma}(M_p) \sum_{k=1}^{m_d} \bar{\sigma}(E_{(q,p),k})^2, \quad k_{FF} := \bar{\sigma}(F_{(q,p)}^\top M_p F_{(q,p)}),$$

$$k_{DE} := \bar{\sigma}(M_p D_{(q,p)}) \left(\sum_{k=1}^{m_d} \bar{\sigma}(E_{(q,p),k})^2\right)^{\frac{1}{2}}, \quad k_{DF} := \bar{\sigma}(D_{(q,p)}^\top M_p F_{(q,p)}),$$

$$k_{EF} := \bar{\sigma}(M_p F_{(q,p)}) \left(\sum_{k=1}^{m_d} \bar{\sigma}(E_{(q,p),k})^2\right)^{\frac{1}{2}}.$$

Therefore when a switch occurs,

$$\begin{split} V_{p}(x) &= \ln\left(1 + K_{DD} + K_{EE} + K_{FF} + 2K_{DE} + 2K_{DF} + 2K_{EF}\right) \\ &\leq \ln\left(1 + \mu(x^{-})^{\top}M_{q}x^{-} + (2k_{DE}|\omega| + k_{EE}|\omega|^{2})|x^{-}|^{2} + (2k_{DF}|\omega| + 2k_{EF}|\omega|^{2})|x^{-}| + k_{FF}|\omega|^{2}\right) \\ &= \ln\left((1 + \mu(x^{-})^{\top}M_{q}x^{-})\left(1 + \frac{(2k_{DE}|\omega| + k_{EE}|\omega|^{2})|x^{-}|^{2} + (2k_{DF}|\omega| + 2k_{EF}|\omega|^{2})|x^{-}| + k_{FF}|\omega|^{2}\right)\right) \\ &\ln(1 + \mu(x^{-})^{\top}M_{q}x^{-}) + \ln\left(1 + \frac{(2k_{DE}|\omega| + k_{EE}|\omega|^{2})|x^{-}|^{2} + (2k_{DF}|\omega| + 2k_{EF}|\omega|^{2})|x^{-}| + k_{FF}|\omega|^{2}}{1 + \mu(x^{-})^{\top}M_{q}x^{-}}\right) \\ &\leq \ln(1 + \mu(x^{-})^{\top}M_{q}x^{-}) + \ln\left(1 + \frac{(2k_{DE}|\omega| + k_{EE}|\omega|^{2})|x^{-}|^{2}}{\mu(x^{-})^{\top}M_{q}x^{-}} + \frac{(2k_{DF}|\omega| + 2k_{EF}|\omega|^{2})|x^{-}|}{1 + \mu(x^{-})^{\top}M_{q}x^{-}} + k_{FF}|\omega|^{2}\right) \\ &\leq \ln(1 + \mu(e^{V_{q}(x^{-})} - 1)) + \ln\left(1 + \frac{2k_{DE}|\omega| + k_{EE}|\omega|^{2}}{\mu\underline{\sigma}(M_{q})} + \frac{k_{DF}|\omega| + k_{EF}|\omega|^{2}}{\sqrt{\mu\underline{\sigma}(M_{q})}} + k_{FF}|\omega|^{2}\right) \end{split}$$

Hence (11) is satisfied with $\chi(s) := \ln \left(1 + \mu \left(e^s - 1\right)\right)$ and $\rho(s) := \ln \left(1 + \left(\frac{2k_{DE}}{\mu \underline{\sigma}(M_q)} + \frac{k_{DF}}{\sqrt{\mu \underline{\sigma}(M_q)}}\right)s + \left(\frac{k_{EE}}{\mu \underline{\sigma}(M_q)} + \frac{k_{EF}}{\sqrt{\mu \underline{\sigma}(M_q)}} + k_{FF}\right)s^2\right)$. Therefore Assumption 1 is verified. Because

 $\alpha_{\rm s}$ is globally Lipschitz, we let $\psi = \alpha_{\rm s}$ and it can be computed that

$$\begin{split} \zeta^* &= \sup_{s>0} \int_s^{\chi(s)} \frac{1}{\psi(r)} dr = \sup_{s>0} \int_s^{\chi(s)} \frac{e^r}{\lambda(e^r - 1)} dr \\ &= \frac{1}{\lambda} \sup_{s>0} \ln(e^r - 1) \Big|_s^{\ln(\mu(e^s - 1) + 1)} \\ &= \frac{1}{\lambda} \sup_{s>0} \ln \frac{(\mu(e^s - 1) + 1) - 1}{e^s - 1} \\ &= \frac{\ln \mu}{\lambda}. \end{split}$$

Thus Assumption 3 is verified and Proposition 12 is proven by Theorem 1.

References

- 1. D. Angeli and E. D. Sontag. Forward completeness, unboundedness observability, and their Lyapunov characterizations. Systems & Control Letters, 38(4):209 217, 1999.
- D. Angeli, E. D. Sontag, and Y. Wang. A characterization of integral input-to-state stability. *IEEE Trans. Automatic Control*, 45(6):1082–1097, 2000.
- C. Cai and A.R. Teel. Characterizations of input-to-state stability for hybrid systems. Systems & Control Letters, 58(1):47–53, 2009.
- C. Cai and A.R. Teel. Robust input-to-state stability for hybrid systems. SIAM J. Control Optim., 51(2):1651–1678, 2013.
- 5. A. Chaillet, D. Angeli, and H. Ito. Combining iISS and ISS with respect to small inputs: The strong iISS property. *IEEE Trans. Automatic Control*, 59(9):2518-2524, 2014.
- A. Dontchev and F. Lempio. Difference methods for differential inclusions: A survey. SIAM Review, 34(2):263–294, 1992.
- R. Goebel, R. G. Sanfelice, and A. R. Teel. Hybrid Dynamical Systems: Modeling, Stability, and Robustness. Princeton University Press, 2012.
- H. Haimovich and J.L. Mancilla-Aguilar. ISS implies iISS even for switched and timevarying systems (if you are careful enough). *Automatica*, 104:154 – 164, 2019.
- 9. J. P. Hespanha, D. Liberzon, and A. Teel. Lyapunov conditions for input-to-state stability of impulsive systems. *Automatica*, 44:2735 2744, 2008.
- J.P. Hespanha. L₂-induced gains of switched linear systems. In V.D. Blondel and A. Megretski, editors, Unsolved Problems in Mathematical Systems and Control Theory, chapter 4, pages 131–133. Princeton University Press, New Jersey, USA, 2004.
- J.P. Hespanha and A.S. Morse. Stability of switched systems with average dwell-time. In Proceedings of 38th IEEE Conf. on Decision and Control, pages 2655–2660, 1999.
- 12. H. K. Khalil. Nonlinear Systems. Prentice-Hall, Inc., 3rd edition, 2002.
- 13. D. Liberzon. Switching in Systems and Control. Birkhäuser, Boston, 2003.
- H. Lin and P.J. Antsaklis. Stability and stabilizability of switched linear systems: A survey of recent results. *IEEE Trans. Automatic Control*, 54(2):308 – 322, 2009.
- S. Liu, A. Tanwani, and D. Liberzon. Average dwell-time bounds for ISS and integral ISS of switched systems using Lyapunov functions. In *Proc. 59th IEEE Conf. on Decision* and Control, pages 6291–6296, 2020.
- J.L. Mancilla-Aguilar and R.A. García. On converse Lyapunov theorems for ISS and iISS switched nonlinear systems. Systems & Control Letters, 42(1):47 – 53, 2001.
- M. A. Müller and D. Liberzon. Input/output-to-state stability and state-norm estimators for switched nonlinear systems. *Automatica*, 48(9):2029 – 2039, 2012.
- N. Noroozi, A. Khayatian, and R. Geiselhart. A characterization of integral input-tostate stability for hybrid systems. *Math. Control, Signals, and Systems*, 29(3), 2017.
- M. Porfiri, D. G. Roberson, and D. J. Stilwell. Fast switching analysis of linear switched systems using exponential splitting. SIAM Journal on Control and Optimization, 47(5):2582-2597, 2008.

- A. Russo, S. Liu, D. Liberzon, and A. Cavallo. Quasi-integral input-to-state stability for switched nonlinear systems. In proceedings of 21st IFAC World Congress, 2020.
- E. D. Sontag. Smooth stabilization implies coprime factorization. *IEEE Trans. Automatic Control*, 34(4):435 – 443, 1989.
- E.D. Sontag. Comments on integral variants of ISS. Systems & Control Letters, 34(1):93–100, 1998.
- E.D. Sontag and A.R. Teel. Changing supply functions in input/state stable systems. IEEE Trans. Automatic Control, 40(8):1476–1478, 1995.
- 24. L. Vu, D. Chatterjee, and D. Liberzon. Input-to-state stability of switched systems and switching adaptive control. *Automatica*, 43(4):639 646, 2007.
- W. Xiang and J. Xiao. Stabilization of switched continuous-time systems with all modes unstable via dwell time switching. *Automatica*, 50(3):940 – 945, 2014.
- W. Xie, C. Wen, and Z. Li. Input-to-state stabilization of switched nonlinear systems. IEEE Trans. Automatic Control, 46(7):1111–1116, July 2001.
- G. Yang and D. Liberzon. A Lyapunov-based small-gain theorem for interconnected switched systems. Systems & Control Letters, 78:47–54, 2015.
- G. Yang, A. J. Schmidt, D. Liberzon, and J. P. Hespanha. Topological entropy of switched linear systems general matrices and matrices with commutation relations. *Math. Control, Signals, and Systems*, 32(3):411–453, 2020.
- 29. H. Yang, V. Cocquempot, and B. Jiang. On stabilization of switched nonlinear systems with unstable modes. Systems & Control Letters, 58(10):703 708, 2009.
- H. Yang, B. Jiang, V. Cocquempot, and H. Zhang. Stabilization of switched nonlinear systems with all unstable modes: Application to multi-agent systems. *IEEE Trans. Automatic Control*, 56(9):2230–2235, 2011.
- G. Zhai, B. Hu, K. Yasuda, and A.N. Michel. Stability analysis of switched systems with stable and unstable subsystems: An average dwell time approach. *Int. J. Systems Science*, 32(8):1055–1061, 2001.
- G.X. Zhang and A. Tanwani. ISS Lyapunov functions for cascade switched systems and sampled-data control. *Automatica*, 105:216–227, 2019.