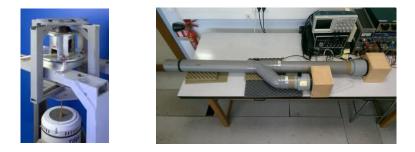
Adaptive Youla-Kucera Parametrization for Active Noise and Vibration Attenuation

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In collaboration with :

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ICSTCC, Sinaia, October 8-10, 2020





Attenuation of vibrations and noise has become a major objective!

Why?

- Technological reasons
- Environmental constraints

•Planes

•Spatial platforms

•Cars, trucks

•Pipes

•Ships

•Buildings

•Manufacturing devices

•Memory devices

•





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Techniques for noise and vibration attenuation:

•Passive – use of materials having absorbtion properties

•Semi-active – changing the properties of the absorbing materials and devices

•Active – use of an auxiliary source for compensating vibrations and noise

•Adaptive – to take into account in real time the unknown and time varying characteristics of vibrations and noise.

This presentation is concerned by *« adaptive active noise and vibration attenuation »*

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- Presence of resonant modes in physical structures can be a source of generation/amplification of noise and vibrations
- Active damping (active absorbers) is a basic control problem where by feedback one tries to damp the low damped modes taking into account the Bode Integral limitations (frequency shaping of the sensitivity function)
- Active damping may be adaptive in order to take into account large variabilities of the plant parameters.





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- Active damping may be adaptive in order to take into account large variabilities of the plant parameters.

This is not the subject of this presentation!





ANC – Active Noise Control AVC – Active Vibration Control

- The challenging problem is to attenuate the effect of the propagation of vibrations and noise
- Noise or vibration are disturbances which have to be attenuated
- Passive and semi-active approaches do not allow to achieve enough attenuation and an active approach has to be used
- ANC and AVC require the introduction of a secondary source (an *actuator* in control terms)





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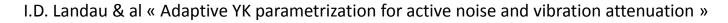
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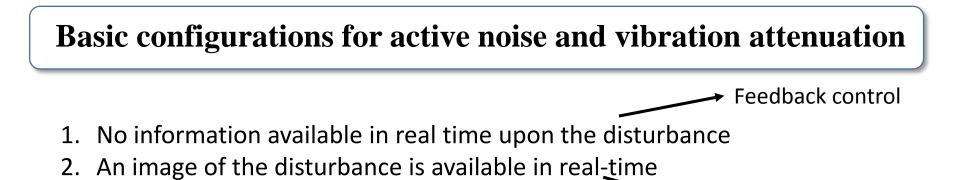
Objective : Strong attenuation of the disturbance (noise or vibration)

From a control point of view we have to solve a « regulation » problem

ANC and AVC raise the same control problems (and of course use the same solutions)



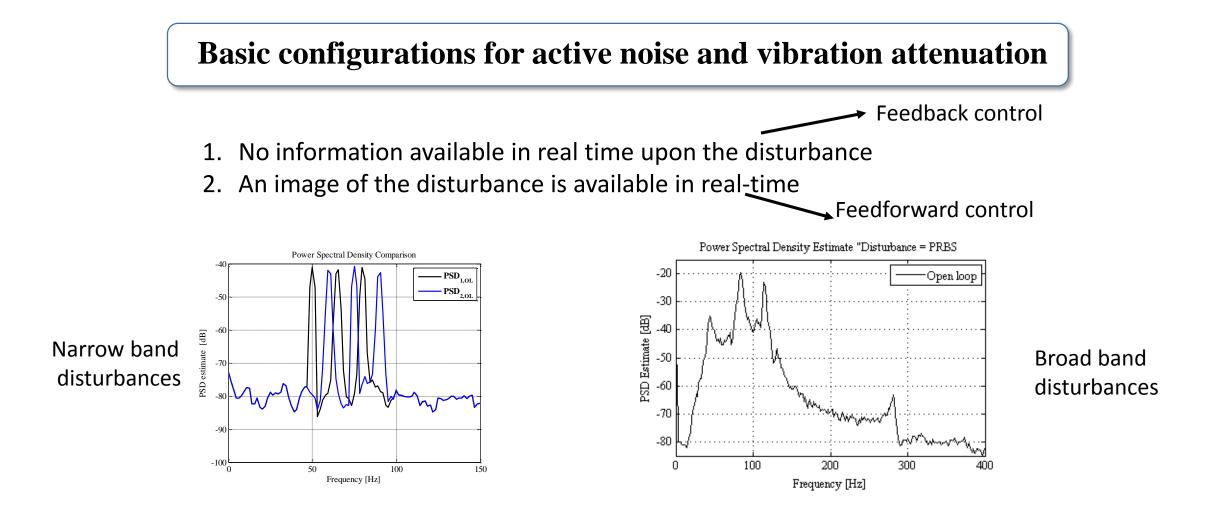




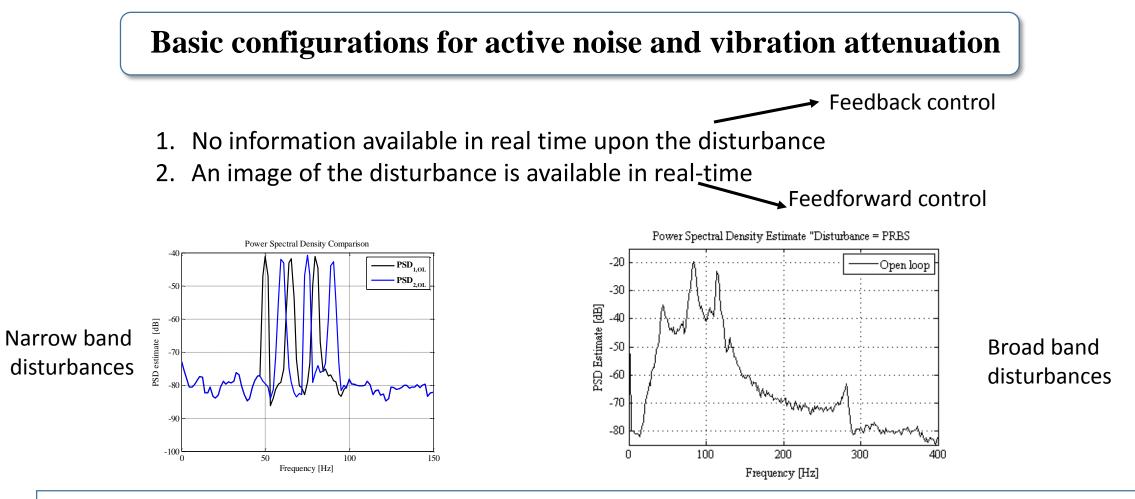


Feedforward control









• The disturbances are unknown and time varying

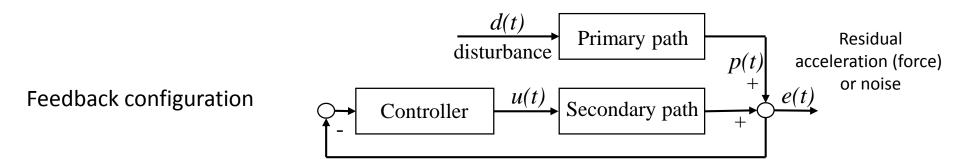
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• The « plant » has almost constant dynamic characteristics (in most of the applications)

Adaptive Regulation Problem

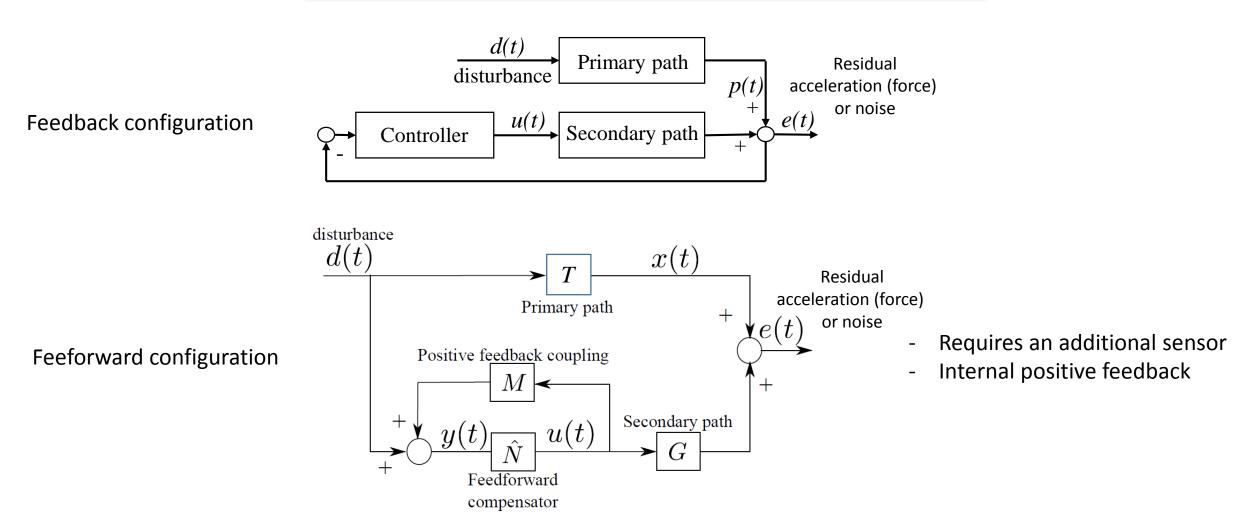
Basic Control Configurations in ANC and AVC







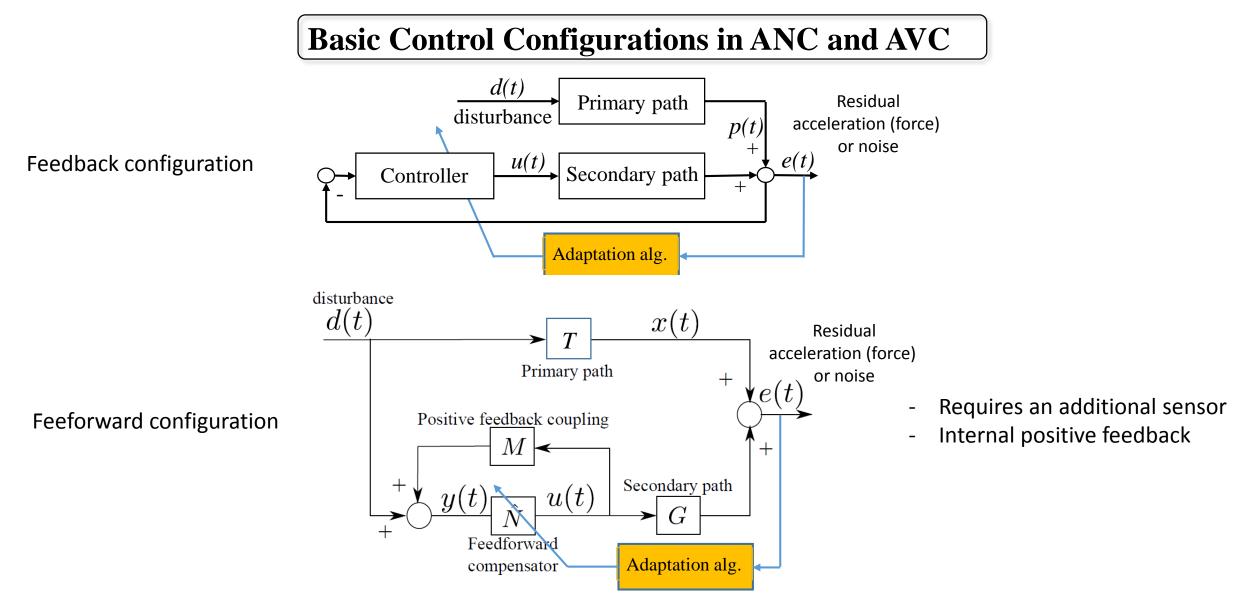
Basic Control Configurations in ANC and AVC



Do as much as possible by feedback (limitations due to Bode Integral) and then add feedforward compensation

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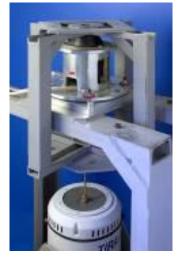
Variability of the disturbance chraracteristics requires the use of an adaptive approach

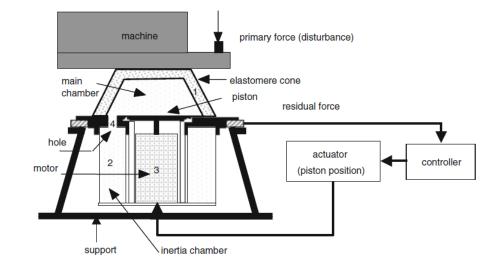


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Active suspension

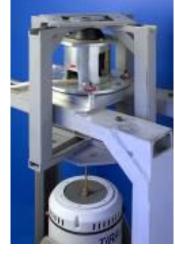








Active suspension





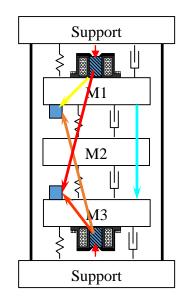
machine primary force (disturbance) elastomere cone main chamber piston residual force hole actuator controller motor (piston position) inertia chamber support primary force (disturbance) machine elastomere cone inertial actuator residual force steel housing coil winding power controller elastic link amplifier mobile part (magnet) support

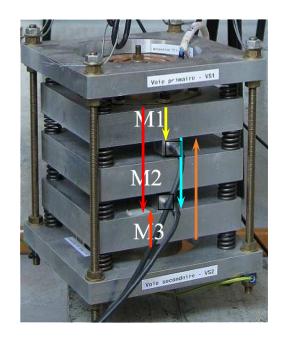
Inertial actuator



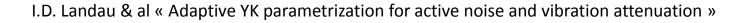
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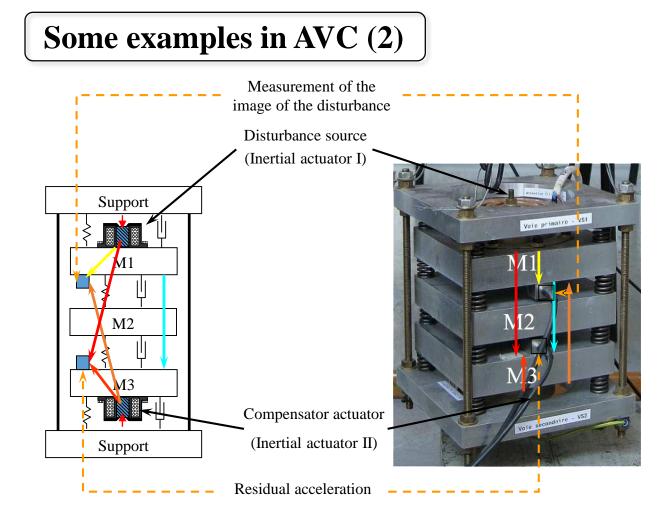






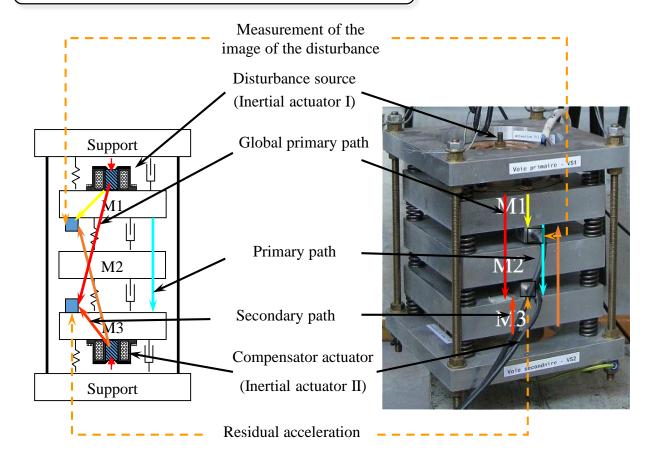






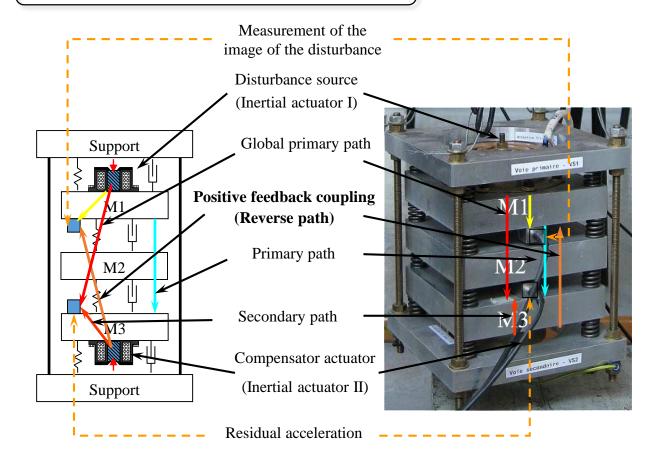






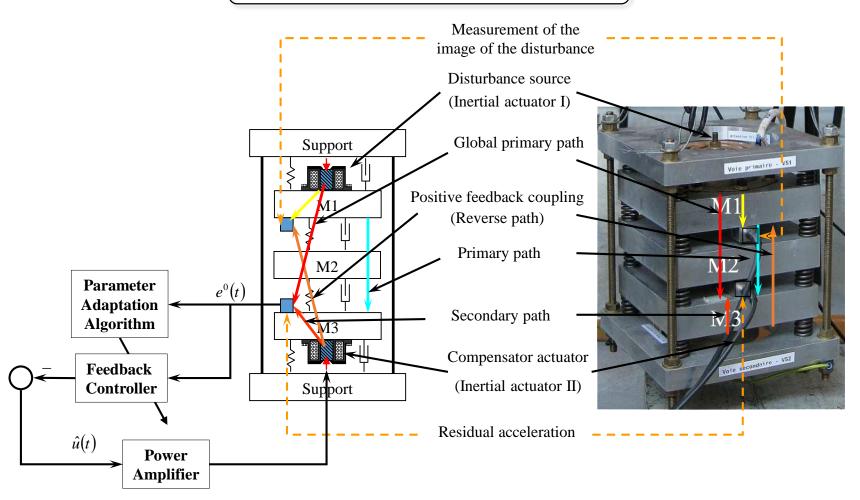










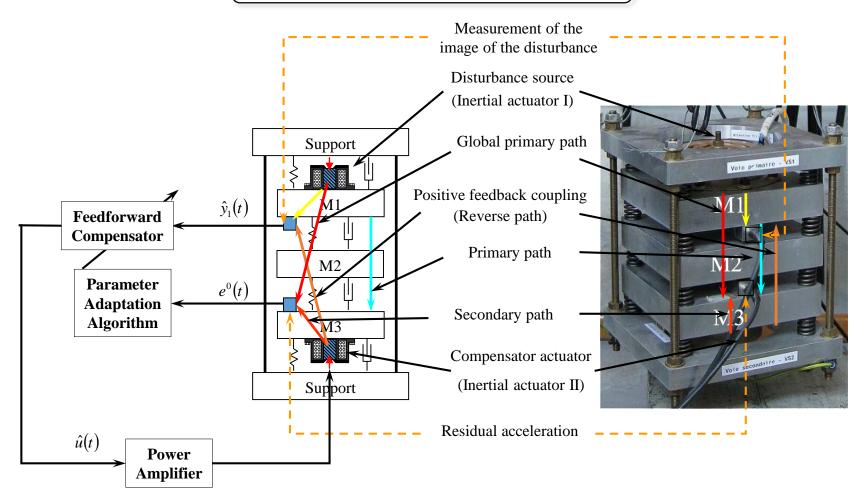


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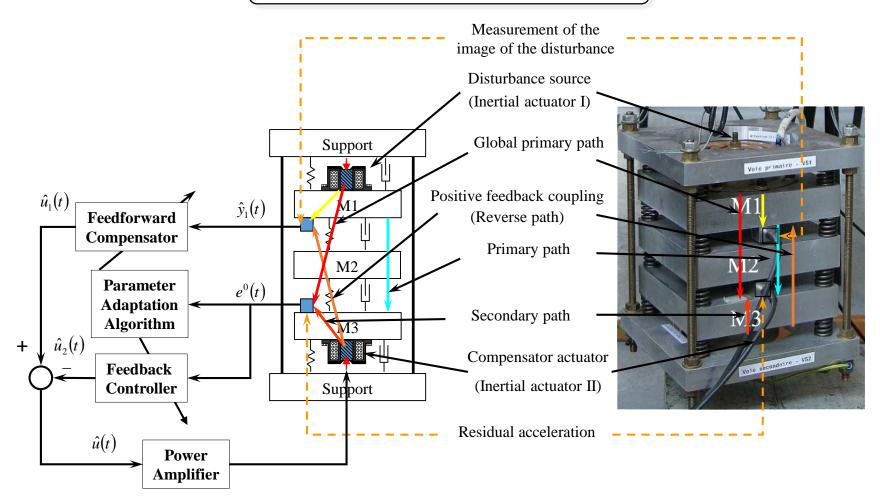






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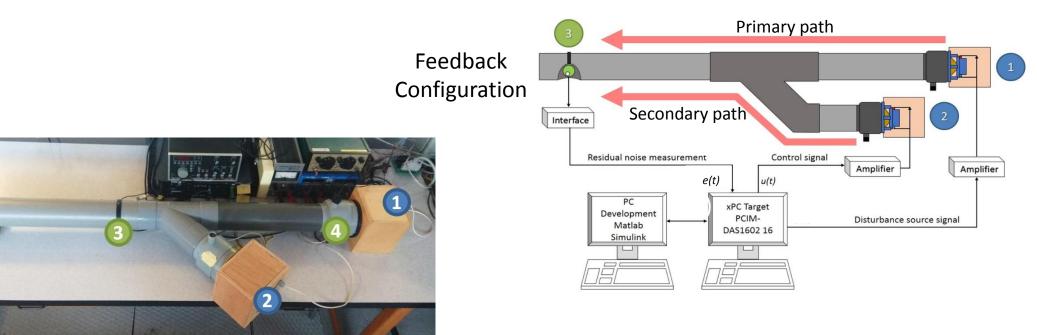








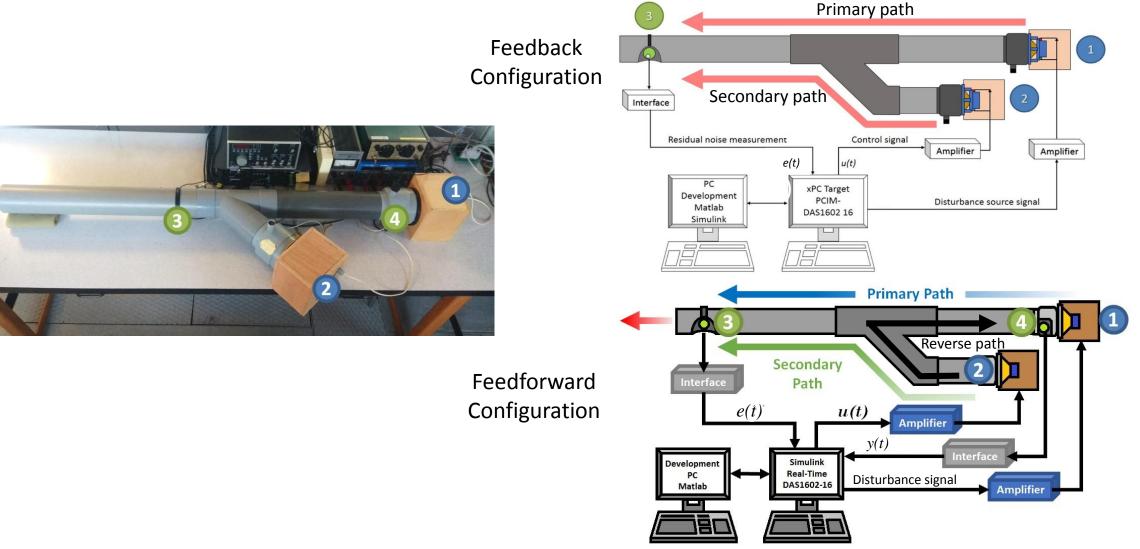
An example in ANC (silencer)







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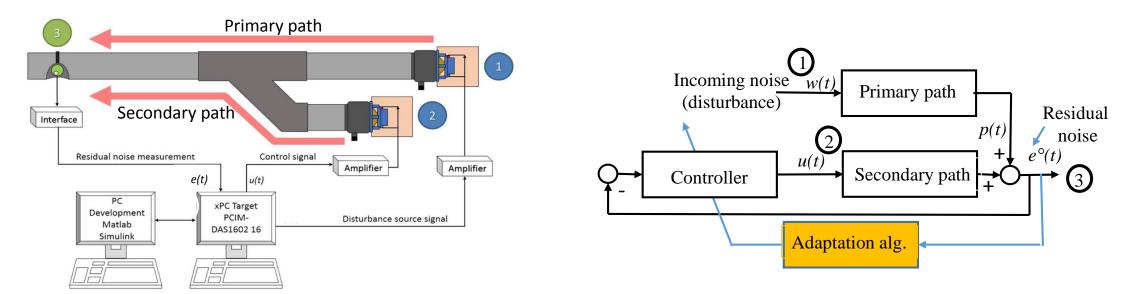




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Adaptive feedback noise attenuation

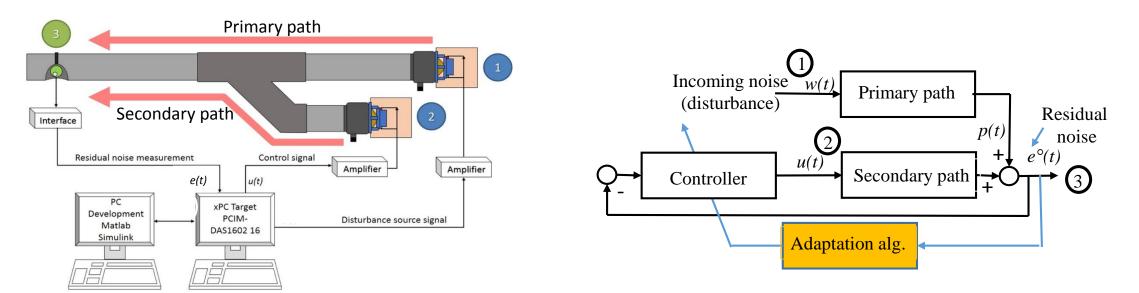


Type of disturbance: Multiple tonal (narrow band) noise with unknown and time varying characteristics





Adaptive feedback noise attenuation



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Challenges

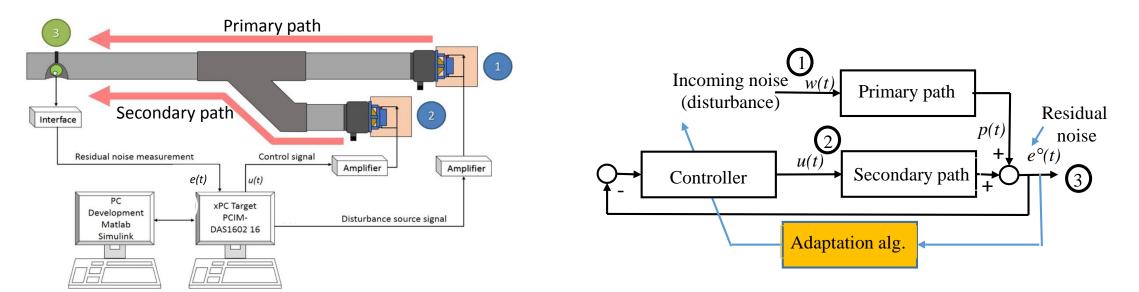
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- Removing the need for an aditional microphone
- What control strategy for cancelling this type of disturbance?
- Could we use a « direct » adaptation scheme? (to avoid explicit identification of the disturbance model and computation of the controller at each sampling)



Adaptive feedback noise attenuation



Type of disturbance: Multiple tonal (narrow band) noise with unknown and time varying characteristics

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- 1. Removing the need for an aditional microphone
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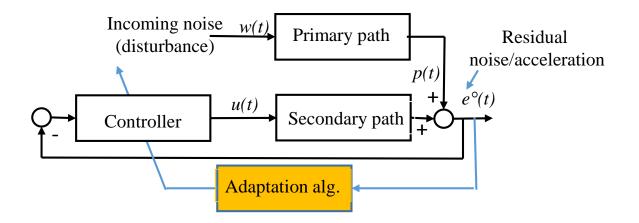
Answers:

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1) Feedback 2) Internal model principle 3) Youla-Kucera paramerization of the controller

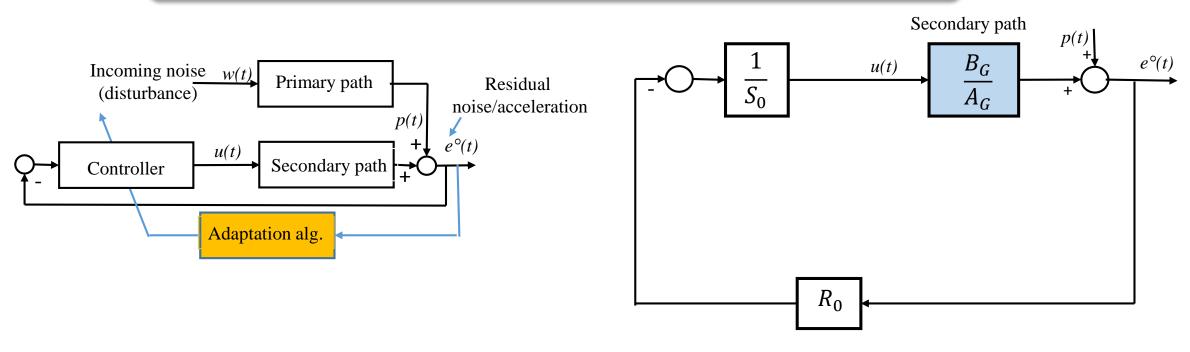
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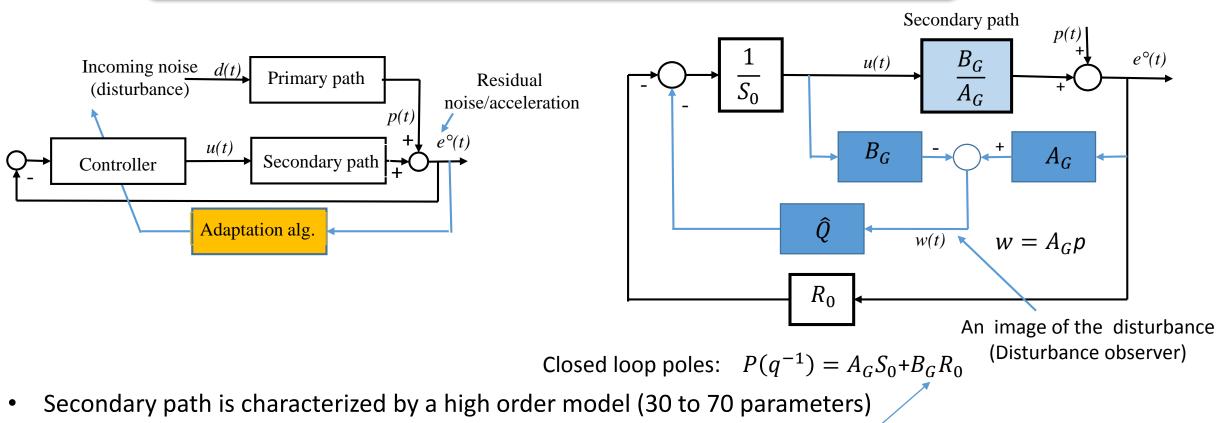




Closed loop poles: $P(q^{-1}) = A_G S_0 + B_G R_0$

• Secondary path is characterized by a high order model (30 to 70 parameters)



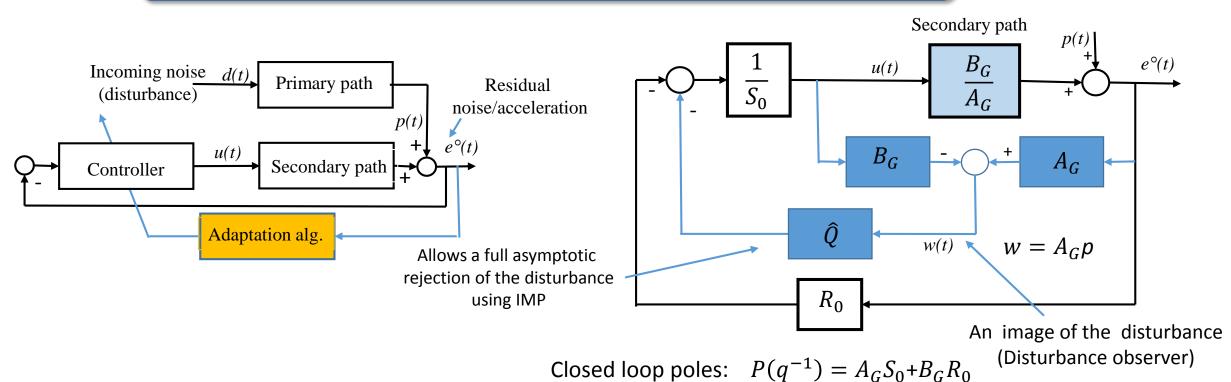


• Q is a FIR filter

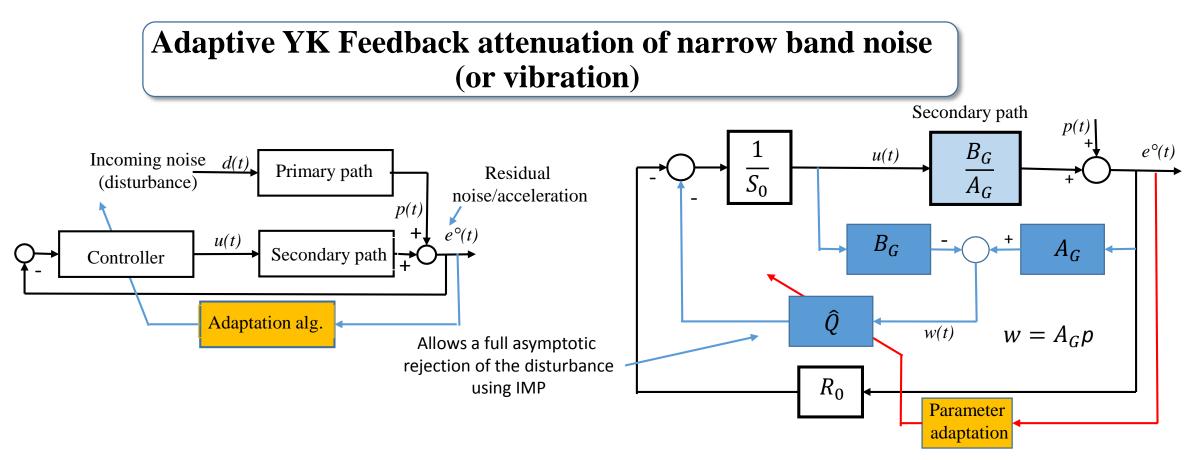
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Poles of the closed loop remain unchanged independently of the parameters of Q





- Secondary path is characterized by a high order model (30 to 70 parameters)
- Q is a FIR filter
- Poles of the closed loop remain unchanged independently of the parameters of Q
- Based on the Internal Model Principle (the controller should contain the model of the disturbance)



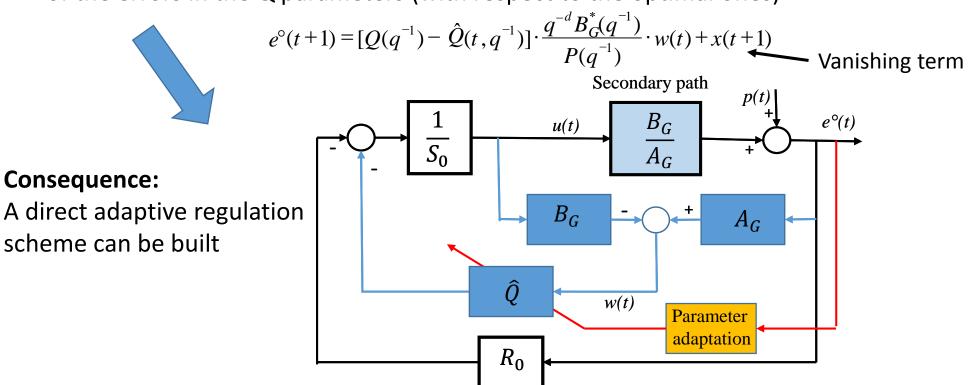
- Q is a FIR filter •
- Poles of the closed loop remain unchanged independently of the parameters of Q •
- Based on the internal model principle (the controller should contain the model of the disturbance) ullet
- Extremely fast adaptation,

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Most important:

YK parametrization allows to express the measured residual error (e°(t)) as a function of the errors in the Q parameters (with respect to the optimal ones)

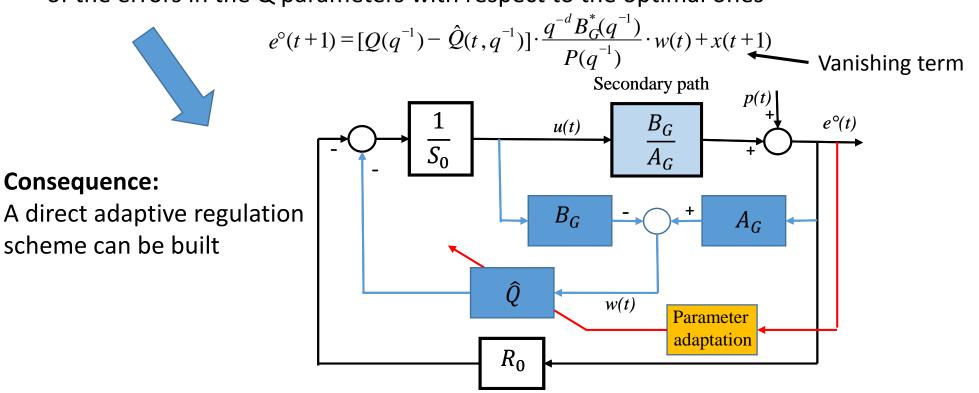






Most important:

YK parametrization allows to express the measured residual error (e[°](t)) as a function of the errors in the Q parameters with respect to the optimal ones



- The order of the Q polynomial depends upon the order of the disturbance model denominator (D_P) and not upon the complexity of the plant model
- Particularly dedicated to multiple narrow band disturbances or limited band disturbances (Bode limitations) gipsa-lab

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Most important:

YK parametrization allows to express the measured residual error (e°(t)) as a function of the errors in the Q parameters with respect to the optimal ones

Consequence:

A direct adaptive regulation scheme can be built

$$e^{\circ}(t+1) = [Q(q^{-1}) - \hat{Q}(t, q^{-1})] \cdot \frac{q^{-d} B_{G}^{*}(q^{-1})}{P(q^{-1})} \cdot w(t) + x(t+1)$$
Vanishing term
$$\theta = [q_{1}, q_{2}, ...]; \hat{\theta}(t) = [\hat{q}_{1}(t), \hat{q}_{2}(t) ...]$$

$$\phi(t) = [w(t), w(t-1)]$$

$$L(q^{-1}) = q^{-d} B_{G}^{*}/P$$

$$\psi(t) = L(q^{-1})\phi(t)$$

$$e^{0}(t+1) = [\theta - \hat{\theta}(t)]^{T}\psi(t)$$





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A direct adaptive regulation scheme can be built

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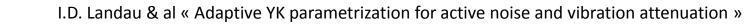
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Standard form for the synthesis of stable adaptive schemes

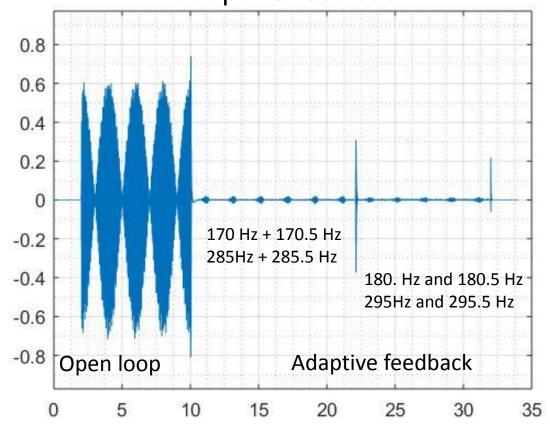
Performance will depend of course also upon the design of the parameter adaptation algorithms





Adaptive Attenuation of Acoustical Interference

Nominal frequencies: 1) 170 Hz and 170.5 Hz + 2) 285 Hz and 285.5 Hz Variations of + 10 Hz for the nominal frequencies YK adaptive feedback



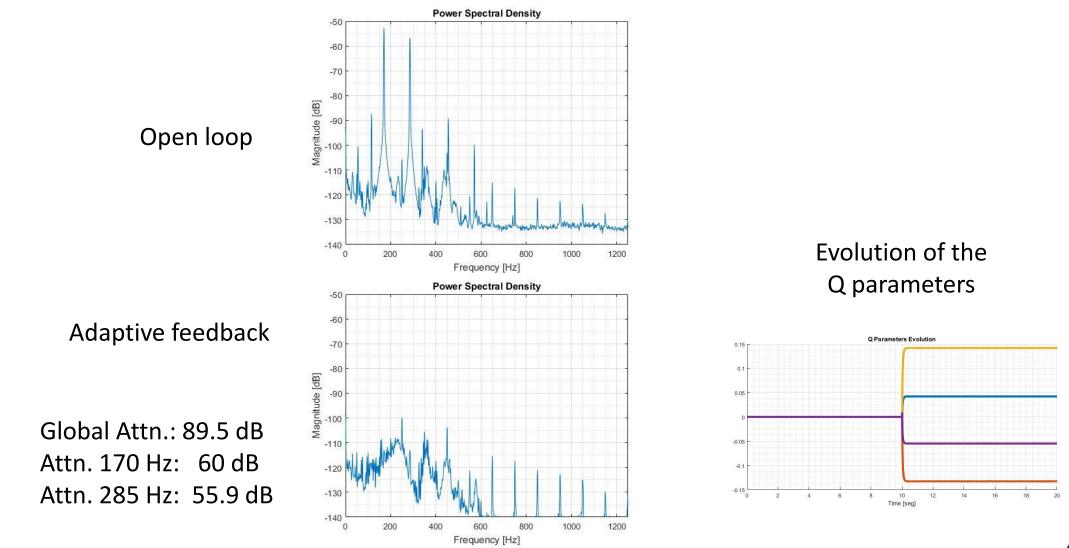


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Attenuation of 2 simultaneous tonal noises

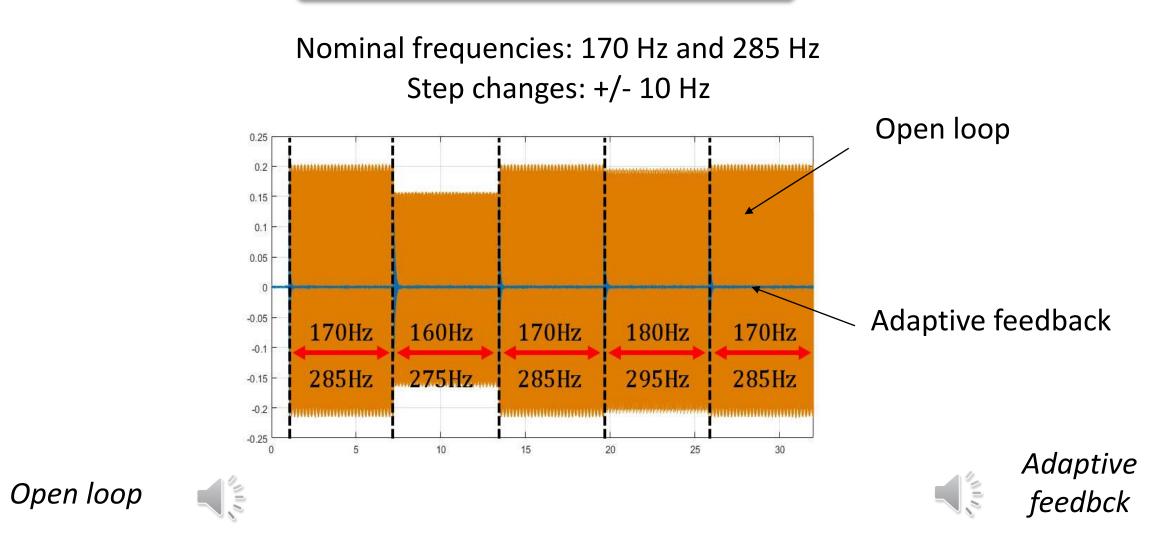
Noise Frequencies : 170 Hz and 285 Hz



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Step Changes in Noise Frequencies



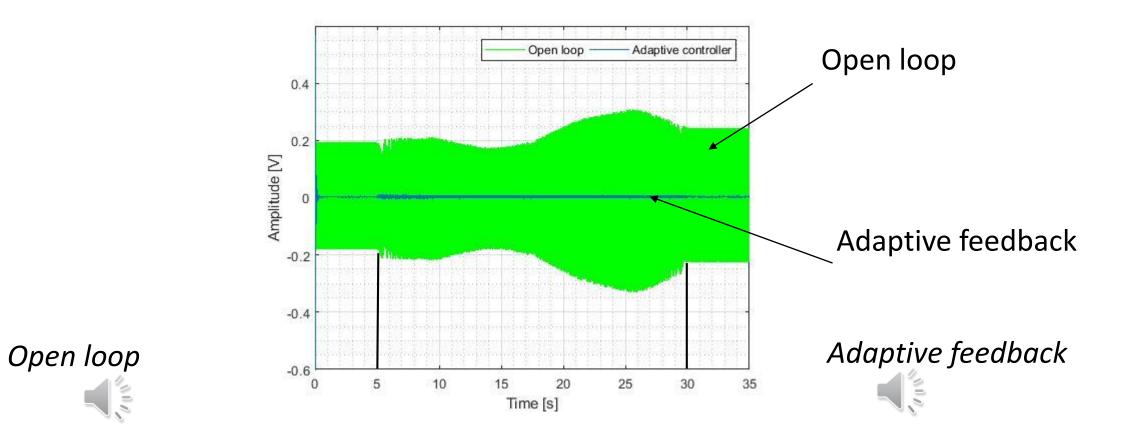


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Chirp frequency variation of 2 simulatenous tonal disturbances

1) From 100 Hz to 200 Hz + 2) From 200 Hz to 300 Hz (from 5s to 30s)

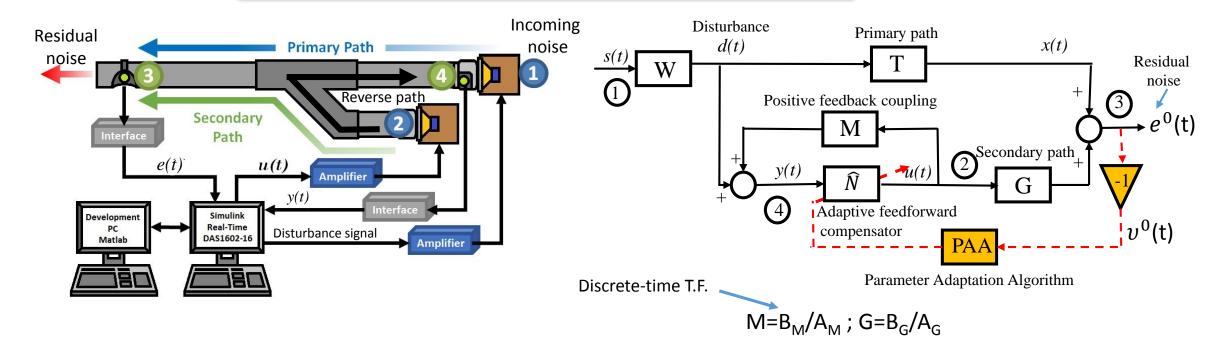




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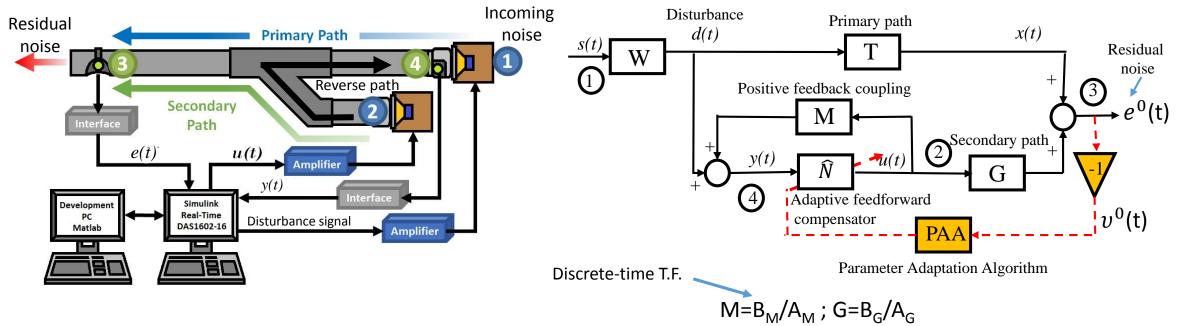
Adaptive Feedforward Noise Attenuation







Adaptive Feedforward Noise Attenuation

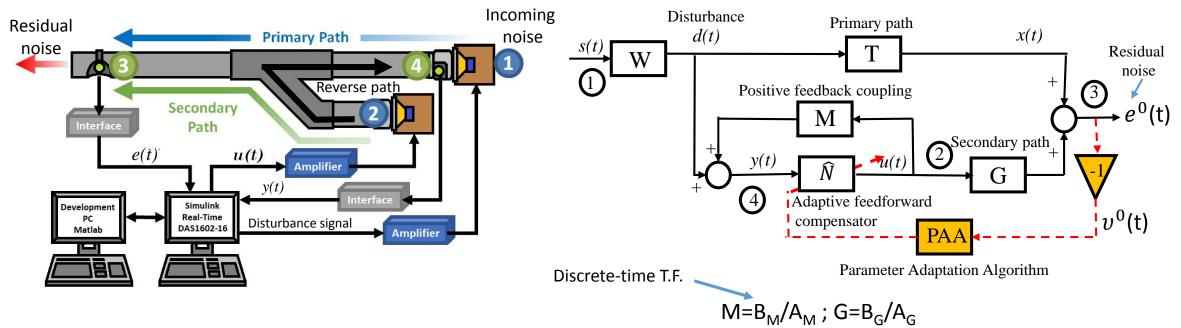


State of the art (before 2011)

- Stability problems caused by the internal positive coupling
- Use of refined LMS (XLMS, FULMS...) adaptation algorithms but without a stability analysis



Adaptive Feedforward Noise Attenuation



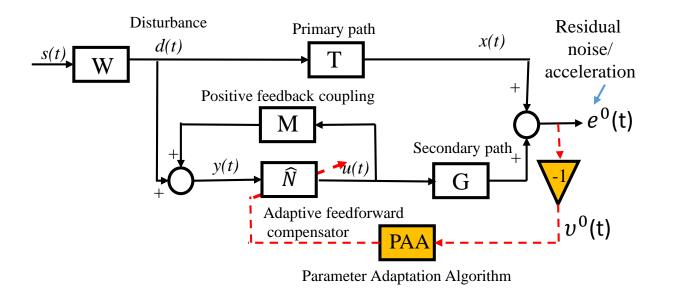
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New solutions

- Introduction of stability based adaptive algorithms
- Inherent stable structures despite internal positive feedback using Youla-Kucera parametrization I.D. Landau & al « Adaptive YK parametrization for active noise and vibration attenuation »

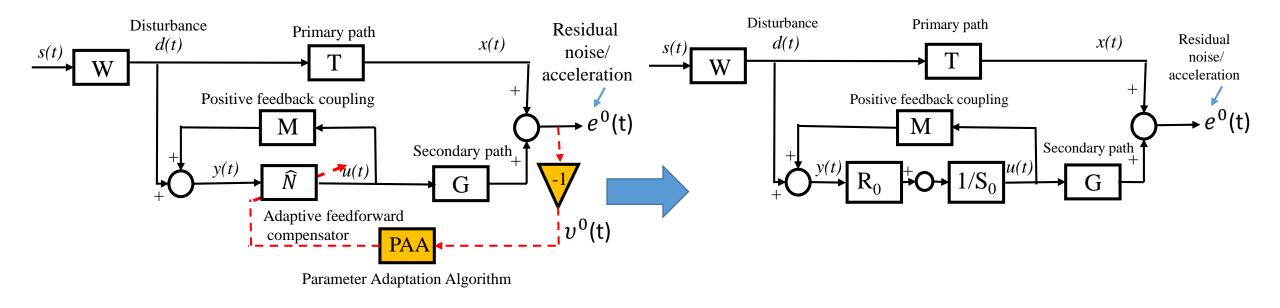




• **Problem**: To assure the stability of the internal coupling positive feedback loop independently of the parameters of the feedforward compensator which should minimize the residual noise



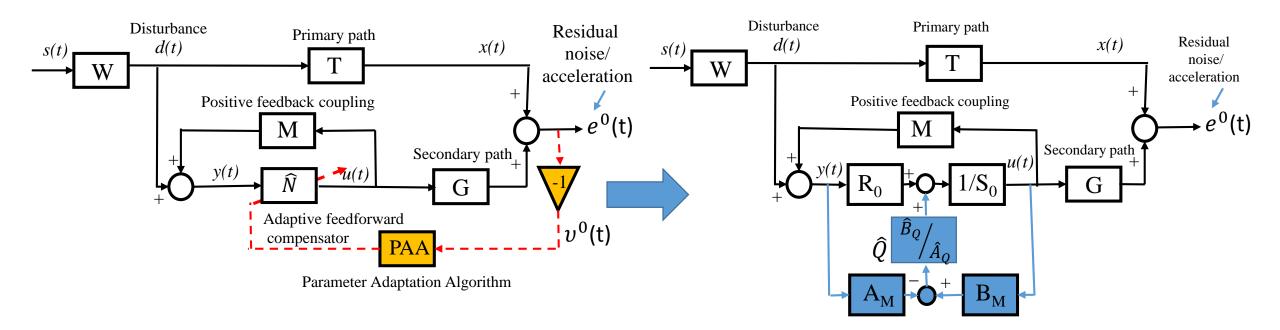




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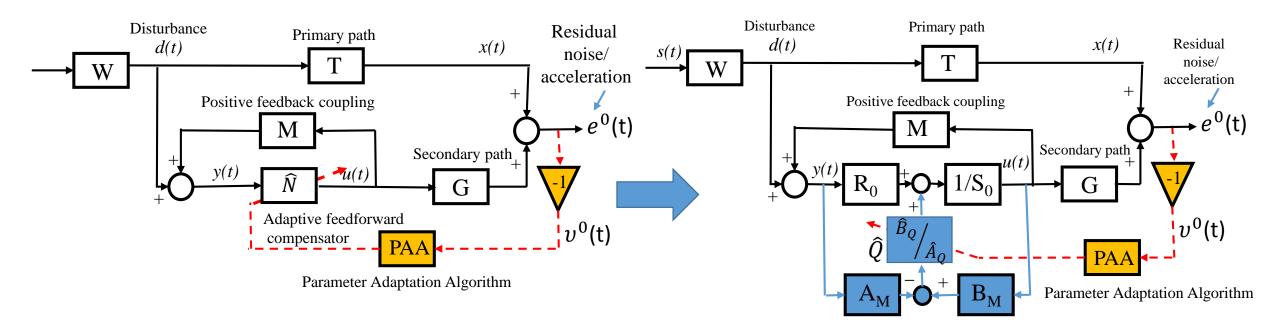






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- Youla Kucera parametrization assures the stability of the internal positive feedback loop (defined by the choice of Ro and So) independently of the parameters of the FIR feedforward compensator filter Q (A_Q=1)
- The stabilisation of the internal feedback loop and the optimisation of the feedforward compensator are dissociated gipsa-lab I.D. Landau & al « Adaptive YK parametrization for active noise and vibration attenuation »

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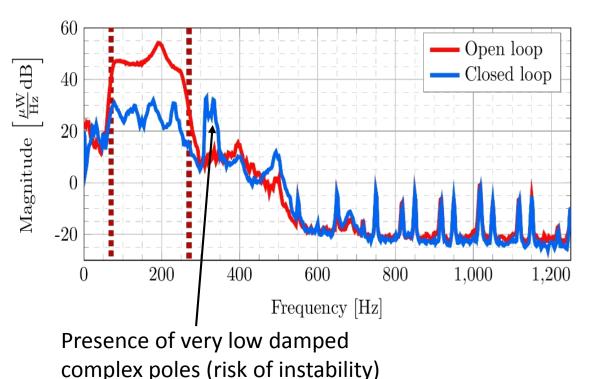
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Adaptive Feedforward Broad-band Noise Attenuation

Mainly dedicated to attenuation of broad-band disturbances

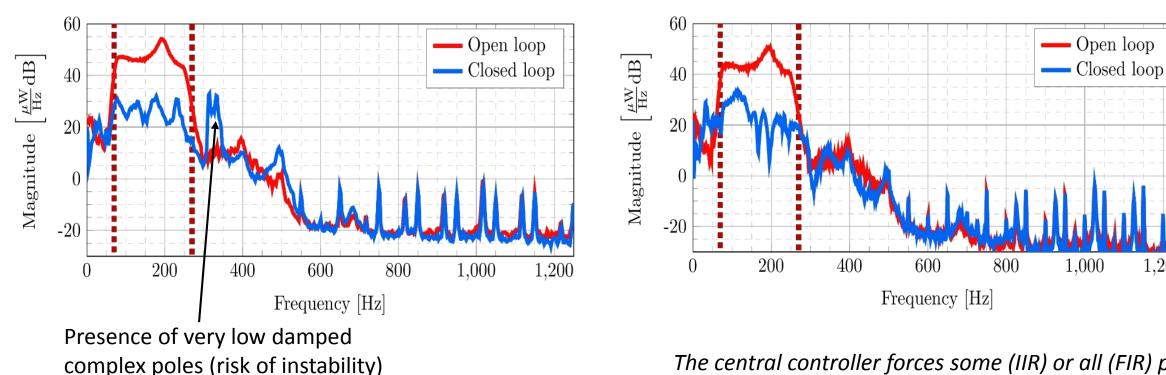


Stability based standard IIR compensator

The stability approach does not guarantee the damping of the poles of the internal loop (they may be very close to the unit circle)

Adaptive Feedforward Broad-band Noise Attenuation

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Stability based standard IIR compensator

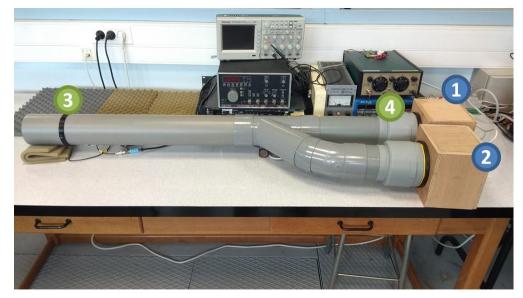
1,200

Stability based YK – IIR compensator

The central controller forces some (IIR) or all (FIR) poles to be enough far from the unit circle

The stability approach does not guarantee the damping of the poles of the internal loop (they may be very close to the unit circle)

Beyond the delay barrier using YK parametrization



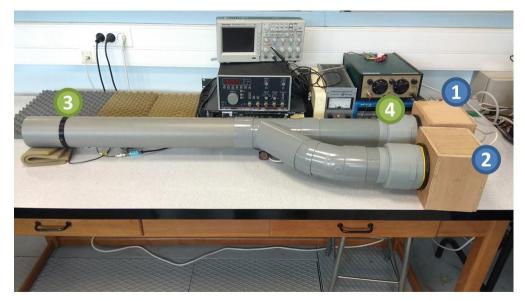
There are potential application areas where:Length secondary path $(2\rightarrow 3)$ > Length primary path $(4\rightarrow 3)$ (delay)(delay)

Beyond the theoretical limit for standard Feedforward Compensation (*Secondary path shorter than Primary path*)





Beyond the delay barrier using YK parametrization



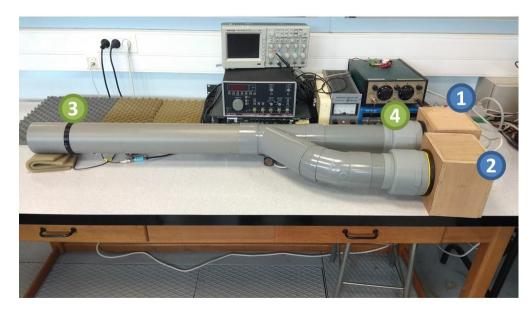
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- Standard FIR structures have very poor performance and tend to become unstable on a long horizon.



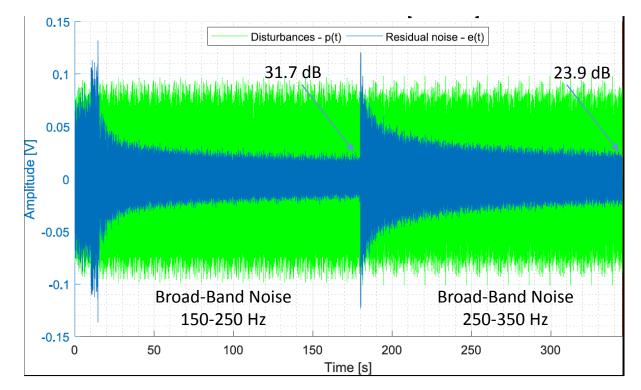
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Results using **FIR YK** adaptive feedforward compensator (60 parameters)





Design of adaptive feedforward attenuation schemes

• Establish an equation for the residual error (noise/acceleration) of the form:

$$v(t+1) = -e(t+1) = H_{\nu}(q^{-1}) \left[\theta - \hat{\theta}\right]^T \phi(t) \qquad \text{For } \hat{\theta} = \text{const.}$$

(not obvious at a first glance)

- θ = vector of optimal parameters of the compensator filter (unknown)
- $\hat{\theta}$ = vector of adjustable parameters of the compensator filter
- $\phi =$ vector of measurements (observations) inputs and outputs of the compensator filter



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- Introduce a filter L on the data (will results from stability considerations) $\psi(t) = \phi_f(t) = L(q^{-1})\phi(t)$
- Re-write equation of the residual error as

$$w(t+1) = H(q^{-1}) \left[\theta - \hat{\theta}\right]^T \psi(t)$$
 where $H(q^{-1}) = H_v(q^{-1}) \cdot \left[L(q^{-1})\right]^{-1}$



Design of adaptive feedforward attenuation schemes

Establish an equation for the residual error (noise/acceleration) of the form:

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 where $H(q^{-1}) = H_v(q^{-1}) \cdot \left[L(q^{-1})\right]^{-1}$

Replace $\hat{\theta}$ = const. by a time varying one

$$\mathbf{v}(t+1) = H\left(q^{-1}\right) \left[\theta - \hat{\theta}(t+1)\right]^T \psi(t)$$

This is a standard form for the synthesis of stable adaptive schemes.



 θ = vector of parameters ($\hat{\theta}(t)$ = estimated parameter vector) ϕ = vector of measurements (observation vector) ψ = regressor vector: $\psi = L\phi$; L = filter

- Adaptation error (*a priori*)
- Stability based adaptation algorithms

• Data Filter





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Measured residual error (noise/acc.)

• Adaptation error (*a priori*) $v^{0}(t)=-e^{0}(t)$ (feedforward schemes)

 $v^{0}(t)=e^{0}(t)$ (feedback schemes)

• Stability based adaptation algorithms

• Data Filter





 θ = vector of parameters ($\hat{\theta}(t)$ = estimated parameter vector) ϕ = vector of measurements ψ = observation (regressor) vector: $\psi = L\phi$; L = filterMeasured residual error (noise/acc.) Adaptation error (*a priori*) $v^{0}(t)=-e^{0}(t)$ (feedforward schemes) $v^{0}(t) = e^{0}(t)$ (feedback schemes) $\hat{\theta}(t+1) = \hat{\theta}(t) + \gamma \psi(t) \frac{\nu^0(t+1)}{1 + \gamma \psi^T(t)\psi(t)}$ $\hat{\theta}(t+1) = \hat{\theta}(t) + F(t)\psi(t) \frac{\nu^0(t+1)}{1 + \psi^T(t)F(t)\psi(t)}$ Scalar constant adaptation gain (γ) Stability based adaptation algorithms Matrix adaptation gain (F)

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Stability conditions for adaptive feedback/feedforward noise/vibration attenuation

 $\nu(t+1) = H(q^{-1}) \left[\theta - \hat{\theta}(t+1)\right]^T \psi(t)$

Stability condition:

 $H(z^{-1})$ (or $H(z^{-1}) - \lambda/2$ for time varying adaptation gain) should be Strictly Positive Real (SPR) (*)

 $H(q^{-1}) = H_{\nu} (q^{-1}) \cdot [L(q^{-1})]^{-1} \qquad H_{\nu}(q^{-1}) = f(G, M, R_0, S_0)$





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Best choice for *L* (trying to approach *H*=1): $L=f(\widehat{G}, \widehat{M}, Ro, So)$

When doing this, condition (*) becomes in most of the cases:

$$\frac{\widehat{G}}{G}$$
 and $\frac{\widehat{M}}{M}$ should be **strictly positive real** T.F.



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 $\frac{\widehat{G}}{G}$ and $\frac{\widehat{M}}{M}$ should be **strictly positive real** T.F.

Or equivalently:

The phase difference between \hat{G} and G and between \hat{M} and M should be $< |90^{\circ}|$ at all frequencies (can be relaxed using averaging arguments)

Consequence: one needs good models of G and M!

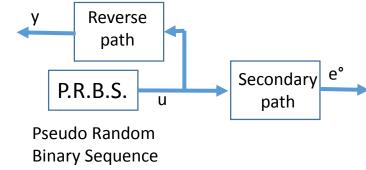
(system identification is a must)

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System identification – a key point for stability

Step 1 Data acquisition under experimental protocol (15 to 30s)



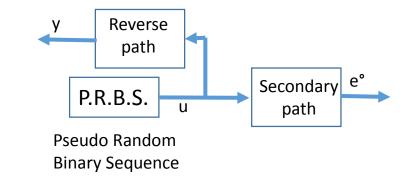
- Step 2 > Estimation of the model order from data (Duong-Landau method)
- Step 3 > Unbiased parameter estimation of an ARMAX model (XOLOE alg. gives the best results)
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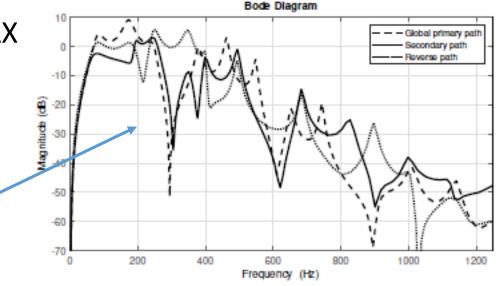


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- Step 3 > Unbiased parameter estimation of an ARMAX model (XOLOE alg. gives the best results)
- Step 4 > Model validation (whiteness test)
- High order models with many resonances and anti-resonances
- Double differentiator behaviour

((bsa-)))

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The methodology – A summary

Advances in Industrial Control

Ioan Doré Landau Tudor-Bogdan Airimitoaie Abraham Castellanos-Silva Aurelian Constantinescu



Methodology and Tests

AIC

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- This methodology has been developed in the field of active vibration control and applied in the field of active noise control
- R. Melendez et al. « Data driven design of tonal noise feedback cancellers », IFAC World Congr. 2017, Toulouse
- I. Landau et al. « Robust and Adaptive Feedback Noise Attenuation in Ducts", IEEE Trans. on CST, March 2019
- I. Landau et al. « Why one should use Youla Kucera parametrization in adaptive feeforward noise attenuation? » IEEE – CDC, Nice, 2019
- I. Landau et al. « Beyond the delay barrier in, active feedforward noise control using YK paramtrization » J. of Sound and Vibr., Sept. 2019
- T. Airimitoaie et al « Algorithms for adaptive feedforward noise • attenuation—A unified approach and experimental evaluation », IEEE Trans. on CST, to be pulished

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Some research subjects

- Combined feedback and feedforward attenuation of noise /vibrations
- Adaptive feedback and feedforward attenuation of noise and vibration in the presence of significant changes in the models of the secondary and reverse paths



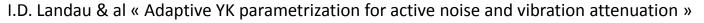


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Preliminary results indicates the usefulness of Youla – Kucera parametrization. However many problems remain to be solved!

- T. Airimitoaie I. Landau. « Combined adaptive feedback and feedforward compensation for active vibration control using Youla–Kucera parametrization", J. of Sound and Vibr., pp 422-441, 2018
- Ph. Mulhaupt et al. "Asymptotic rejection of nonvanishing disturbances despite plant–model mismatch » • International J. of Adaptive Control and Signal Porcessing, 2012, pp1090 -1110
- B. Vau I. Landau, «Youla-Kucera feedback disturbance rejection in the presence of plant uncertainties" gipsa-lab IEEE-CDC 2019, Nice, 2019





Thank you for your attention!

Do not hesitate to contact me: ioan-dore.landau@gipsa-lab.fr



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