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Reliable method for accurate measurements of the breakdown voltage in microgaps

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Abstract- The up-to-date tendency of compact design of low/medium voltage equipment implies micrometer distances between the active parts. In order to reduce losses and prevent operational failure of the devices under such conditions, knowledge on the breakdown voltage is essential. However, measurements of the breakdown voltage in microgaps remains quite delicate experiment, mainly because of difficulties related to the accurate definition of the distance between the electrodes. To overcome these limitations, we propose here a reliable method based on measurements of the current amplitude and phase between current and applied voltage for AC signals. It is shown that when contact between the two electrodes occurs the current amplitude increases substantially and the phase changes abruptly to zero. Thus, no arbitrary conditions are needed to define the contact point. The system is well-defined and the procedure accuracy improved.

I. INTRODUCTION

Due to the fast development of massive technology solutions, microelectronic industry and nanotechnology products increase very rapidly in numbers and varieties. The era of migration towards smaller dimensions of the outcome devices presents new challenges related among others to the reliability of the equipment. To meet these challenges, an in-depth understanding of the equipment failure on the physical side is required, so that the long-term reliability of products or processes can be evaluated. Accordingly, many research groups renewed their interest to the study of electrical breakdown phenomena in micro- and even nano-gaps [1-9].

Electrical breakdown in gases is well described by the Paschen's law that uniquely relates the breakdown voltage with the pressure times distance ($p \times d$) product. Contrary to the case of large scale electrode separations, the experiments performed in microgaps (< 10 µm) at atmospheric pressure show much lower breakdown voltages than the ones predicted by the Paschen's law [1-4, 6-8]. When reducing the inter-electrode distance, mechanisms like field electron emission start playing

role in the electron yield in addition to the secondary electron emission and the ionization in the gas, leading to lower breakdown voltages. In general, the electrical breakdown is defined by the combined effect of a large number of parameters: nature of the gas, pressure, inter-electrode distance, geometry of the electrodes, cathode material, and in particular its work function, surface roughness, protrusions, ambient temperature, *etc.* For microgaps the above parameters may present instable features due to gradient variations and interdependence, and thus make the analysis of the underlying mechanisms more difficult. Therefore, to reach a better understanding of the phenomena at play a reliable method for measurements of the breakdown voltage in micrometer gaps is essential.

The breakdown voltage criterion is often chosen to be the presence of measurable current (in the order of a few milliamps) in the circuit after applying voltage from the power supply [1-4, 7, 8]. Critical step in the breakdown voltage measurement procedure in microgaps is defining the contact point between the two electrodes. Once the contact point is found, one of the electrodes is stepped back to precisely define the inter-electrode distance. According to the reported in the literature studies there are different protocols for determining the contact point, electrode particularly in tip-to-plane configuration: (i) observation under optical microscope until achieving mechanical contact between the electrodes [7]. This procedure is acceptable in the general case, but does not consider any electrical properties of the environment. It might lead to a large uncertainty in the breakdown voltage value that is attributed to a given inter-electrode distance, especially for very small microgaps; (ii) more reliable criterion is obtaining electrical contact between the two electrodes [1-4, 7, 8]. To reach contact, the tip-electrode is slowly approached to the plane and the contact point is defined when measurable current is detected in the circuit after applying a DC bias of 1-2 V. This criterion meets all required conditions, but registering gradually increasing currents in the order of 1 nA might be delicate.

To precisely define the contact point our strategy relies on using the threshold behavior of some characteristics of the current and voltage waveforms by using AC signals. In this paper we focus on description of the performed experiments and the simulation approach that has been applied for development of the method for accurate measurements of breakdown voltages in microgaps. Control of the contact point, hence the inter-electrode distance, certainly adds value to the precise description of breakdown phenomena in microgaps.

II. EXPERIMENTAL SETUP, SIMULATION APPROACH AND MEASUREMENT PROCEDURE

A. Schematic Design and Apparatus

The electrode configuration is tip-to-plane; the tip electrode is a tungsten needle with well controlled radius ($R_c = 10 \mu m$) facing an Au-metallized plate (Fig. 1). This experimental setup is an improved version of a previously developed system in LAPLACE [9, 10]. The needle is connected to a voltage source through a 142 k Ω series resistance (R_m) and the plate electrode is grounded. An AC signal (0.1 – 10 MHz, 10 V) is applied to the needle for the contact point definition. Then the breakdown voltage is measured for 1 – 10 μm gaps at atmospheric pressure and room temperature. To that end the tip is connected to a Keithley 2410 DC voltage source supplying voltages up to 1100 V. The current is detected using a Tektronix CT-1 with P6041 current probe (5 mV/mA) connected to a Wavesurfer 3024 oscilloscope.



Fig. 1. Schematic design of experimental set-up.

The position of the plane (cathode) electrode can be adjusted manually through a x-y linear stage. The tip (anode) electrode settings are made in x-axis by using a micro positioner: piezo-actuator, PI Hera 621.UD with E.709-CRG controller, connected to a PC. The piezo-electric actuator enables control of the distance between the two electrodes with accuracy of \pm 10 nm. To obtain this high precision the electrodes are placed on a marble plate. Observation of the tip-to-plane electrode configuration is performed by using a Keyence VHX-1000 digital microscope.

B. Simulation approach to consider effects induced by protrusions on the equivalent capacitance in the gap

As mentioned above irregularities on the electrode surface (roughness, protrusions) modify the breakdown voltage for a given electrode configuration due to a strong enhancement of the electrical field in the gap [9, 10].



Fig. 2. The simulated electrode configuration without protrusion in (a) and with protrusions in (b).

To account for effects induced by protrusions on the equivalent capacitance variation in the gap we have performed numerical simulations by using the FEMM software (Ver.4.2). The simulation applies finite element modeling (FEM) to extract the equivalent capacitance in the gap as function of the inter-electrode distance and compare it to the one in presence of protrusions on the electrode surface. Basically, the concept of FEM analysis involves four steps, namely discretizing the solution area into a number of sub-regions or finite elements, then obtaining an equation for a typical element followed by assembling all the elements in the solution region, and the last step is solving the system of the obtained equations [11].

The considered in the simulation tip-to-plane electrode configuration is shown in Fig. 2(a). The surface irregularities are considered by an isosceles triangle protrusion with a maximum height of 10 nm (comparable to the measured by profilometry highest peak on the Au-metalized plate electrode, results not shown here) placed on the surface of the plane electrode in front of the tip, as illustrated in Fig. 2(b). Such configuration represents the worst scenario. The other parameters involved in the FEM simulation are: needle (tip) electrode made of tungsten with radius $R_c = 1 \ \mu m$ and length of 50 μm , and gap between the electrodes varied in the range $10 - 200 \ nm$. The applied on the tip (anode) voltage is $V_{anode} = 300 \ V$, while the plane electrode is grounded ($V_{cathode} = 0 \ V$).

C. Procedure for measurement of contact between electrodes

The measurement procedure, applied to define contact between the electrodes, follows two protocols for accuracy. The first protocol concerns the needle approach to the plate at nanometer scale (flowchart is shown in Fig. 3). The needle electrode is moved in small steps by using the piezo-actuator. In non-contact conditions, current flow occurs with significant phase difference with respect to the input source. When the contact point is achieved (d = 0) the phase difference between the input source and the output current becomes close to zero.

The second protocol is performed after each step of the first one. It consists in a simple measurement of the current amplitude. The contact occurs when the current amplitude increases drastically. Thus the contact point between the two electrodes is defined when both protocols are validated.



Fig. 3. Flowchart of the procedure to define contact point between needle and plate electrodes (d = 0 position).

III. RESULTS AND DISCUSSION

The developed method is presented for a frequency of 0.2 MHz and applied voltage of 10 V between the electrodes (Fig. 4). The variation of the current amplitude and phase is followed when the tip-electrode approaches the plane-electrode until contact. We note hereafter the amplitude of current with contact I_C and without contact I_{NC}, respectively. We also define the phase between the voltage and the current when the electrodes are not in contact by φ_{NC} and the phase when the electrodes are in contact by φ_{C} .



Fig. 4. Current and voltage waveforms. Applied voltage 10V (black curve) in both cases. In blue, the current when the electrodes are not in contact. In red, the current when contact occurs.

As can be observed on Fig. 4, the current amplitude doubles and the phase is greatly reduced when the electrodes are in contact compared to the no contact conditions. When the electrodes are not in contact, the current is limited by the equivalent capacitance due to the gap (C_g) between the electrodes and the resistance of measurement (R_m) as presented in Fig. 5. Basically, when the contact occurs there is no gap. The capacitance is short circuited and the current is only limited by the resistance. Therefore, the phase ϕ_C is close to zero and the current I_C increases.



Fig. 5. Equivalent circuit of the system: R_m is the measurement resistance, C_g is the equivalent capacitance of the gap.

The gap capacitance is the main difference between the measurement with contact and without contact. However, it is very difficult to estimate experimentally the value of this capacitance from the current and voltage measurements. Indeed, a very small error on the estimation of a parasitic impedance (as the wire inductance) affects strongly the value of the capacitance. In consequence, we have calculated this capacitance by simulation, as shown in Fig. 6. The obtained results show that the Cg-value is not significantly affected by the inter-electrode distance (d) or by the presence of protrusions on the electrode surface. The Cg-value remains between 7 and 8 femto Farad. As long as there is no contact, the displacement of the electrode does not affect the equivalent circuit (R_m and C_g). This explains as why experimentally the current (I_{NC} and ϕ_{NC}) is not modified by the gap d as long as there is no contact between the electrodes. It also strengthens the reliability level of the method because modification on the current (phase and amplitude) can only occur due to contact between the electrodes.



Fig. 6. Comparison of the equivalent capacitance values against interelectrode gaps with and without protrusions based on the simulation results.

The method, that we propose here, to detect the contact is accurate as long as the impedance of the capacitance is higher or equivalent to the resistance. As defined by the two protocols presented above, there are two conditions to fulfil in order to get high accuracy: the current amplitude ratio I_C/I_{NC} and the current phase difference $\phi_{NC} - \phi_C$ should be large. In addition, the change of the current phase occurs suddenly when the contact condition is achieved. Fig. 7 shows the frequency range

of applicability of the method. For frequencies below 0.1 MHz, the current amplitude is not significant to be measured with accuracy and the method cannot be applied. For 1 MHz the current amplitude ratio I_C/I_{NC} is close to 1, as shown in Fig. 7. There is no difference in the current amplitude, yet the phase difference is still up to 25°. For frequencies above 1 MHz, the current phase difference $\phi_{NC} - \phi_C$ is close to zero because the value of the capacitance impedance is quite low compared to the series resistance. Therefore, in our configuration, this method is practical for applied frequencies in the range 0.1 MHz.



Fig. 7. Evolution with the frequency of the current ratio (black) and the phase difference (red). The current ratio is defined as I_C/I_{NC} . The current phase difference is defined as $\phi_{NC}-\phi_C$.

The developed in this work method allows to position accurately the needle against the plate electrode, *i.e.* the corresponding breakdown voltage can accurately be attributed to the inter-electrode distance. The uncertainty in the inter-electrode distance is defined only by the piezo-actuator. Compared to the other methods described earlier, the double information acquired here (phase shift and growth of the current amplitude) ensures the accuracy of the contact. Moreover, the conditions for contact are not arbitrary. They are imposed by the system. Then the step back of the needle electrode with the piezo-electric actuator defines exactly the inter-electrode distance *d*.



Fig. 8. Breakdown voltage versus the inter-electrode distance in air at atmospheric pressure and room temperature.

Application of the method is demonstrated in Fig. 8. A continuous voltage is applied and the breakdown voltage (V_B) is determined for different inter-electrode distances in the range $1 - 10 \mu m$. The three characteristic regions for microgaps are clearly observable: the classical curve of breakdown voltage following the Paschen law (for gaps higher than 9 μm), the plateau (between 5.5 and 9 μm) and the deviation from Paschen region (for gaps below 5.5 μm).

IV. CONCLUSION

The method developed in this work allows accurate measurements of the breakdown voltage in microgaps. It relies on the threshold behavior of some characteristics of the current and voltage waveforms by using AC signals to define contact between the two electrodes. Two conditions are required to be fulfilled in order to get high accuracy: the current amplitude ratio I_C/I_{NC} and the current phase difference $\phi_{NC} - \phi_C$ should be large. Thus the criterion for contact between the electrodes is not arbitrary. It is entirely imposed by the system. Future work will be directed to systematic studies of the breakdown voltage in microgaps as function of the gas nature, pressure, cathode material, surface roughness, and other relevant factors.

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