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# Comparative experimental study of junctionless and inversionmode nanowire transistors for analog applications

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Abstract— We fabricated JunctionLess (JL) and Inversion-Mode (IM) monocrystalline nanowire NMOSFETs down to L=18nm gate length and W=20nm width. We demonstrate record performance of nanowire junctionless transistors for analog applications:  $A_{VT}$ =1.4mV,µm matching, Av0=62dB gain (L=200nm),  $f_{T}$ =126GHz cut-off frequency and  $f_{MAX}$ =182GHz maximum operating frequency. Junctionless transistor performances even exceed those of IM-FETs in terms of back-bias capability, low-frequency noise, hot-carrier degradation and  $f_{MAX}$ . This is explained by JL physics combining characterization and TCAD simulations: channel length modulation, bulk conduction and high channel depth sensitivity to back bias.

#### I. INTRODUCTION

Recent years have seen renewed interest for JL CMOS transistors because of the relative simplicity of its process integration and its potential fabrication at low temperature as a Back-End-Of-Line transistor for a 3D-sequential integration [1]. On the one hand, polycrystalline silicon or germanium channels, whose doping is activated by a laser anneal, are a promising solution for low-cost and extremelylow-temperature fabrication [2, 3]. On the other hand, monocrystalline Si junctionless channels have already been demonstrated on either planar Silicon-On-Insulator (SOI) [4], FinFET [5] or nanowire architectures [6,7]. Furthermore, JL devices are attractive for analog applications since they are less prone to Hot Carrier Injection (HCI).

#### II. DEVICE PROCESS FLOW

We fabricated IM and JL nanowire nMOS down to W=20nm channel width and L=18nm gate length. The channel thickness is 11nm. The junctionless devices are made by epitaxially growing a 7nm thick *in-situ* phosphorous (P) doped Si film on a 4nm undoped SOI layer. Excellent crystalline quality is obtained (Fig.1). After full process integration the final channel doping level is uniform and equal to  $10^{19}$  at/cm<sup>3</sup> Such channel doping and thickness are optimized for planar devices. The fabrication process is outlined in Fig.2. All the devices feature the same gate stack with HfO<sub>2</sub> dielectrics (equivalent oxide thickness EOT=1nm), TiN + poly-Si and identical 8nm thick spacer.

In order to assess the impact of the channel doping and the source/drain resistance, we added the same 17nm thick Raised Souce/Drain (RSD) and HDD+LDD doping processes for so-called Junctionless Accumulation Mode (JAM) transistors as for the IM reference (final anneal at 1050°C). Purely Junctionless transistors (named JL) have also been fabricated without doping under the spacer. Instead, a 5keV P implantation in the 17nm thick RSD followed by a Solid Phase Epitaxy Regrowth (SPER) annealing at 525°C 30 min was carried out, in order to only dope the JL-RSD (see Kinetic Monte-Carlo profile on Fig.2) and avoid lateral doping diffusion. This process module is suitable for a 525°C 3D-sequential integration process [8].

#### **III. DEVICE ELECTRICAL CHARACTERIZATION**

**Digital** I<sub>ON</sub>-I<sub>OFF</sub> figure of merit (Fig 3) show no significant difference between JAM and IM FET. Also, similar parasitic gate to SD capacitance  $C_{GDS}$  suggests a similar direct overlap controlled by the SD implant (Fig. 4). On the other hand, the JL FET has a 0.06fF/µm lower  $C_{GDS}$  vs JAM/IM, which cannot be explained by the fringe components but rather by a depletion region extending below the spacers in JL transistors (Fig.5). Nevertheless JL devices suffers from higher access resistance (R<sub>0</sub>) due to non-optimized junctions (Fig 3).

**Local variability:** matching coefficient  $A_{VT}$  is higher for JL/JAM devices than IM (1.7/1.4 *vs.* 1.0 mV.µm, Fig. 6). This degradation with the channel doping is attributed to Random Dopant Fluctuation [9].

**Analog**: we consider a nominal transistor width of W=0.24µm, (planar SOI configuration instead of a trigate nanowire structure). The JL/JAM subthreshold slope (SS) is SS=64mV/dec *vs.* 61mV/dec for the IM device (Fig. 7). However, **back-bias** can be leveraged in order to adjust the threshold voltage and tune performance [10]. Indeed, back-bias is more effective for wider than for narrower devices and it is more effective on JL/JAM than on IM transistors (Fig.8). Markedly, a negative back-bias applied on JAM moves the bulk conduction channel upwards towards the gate, which results in an improvement of the electrostatic control (SS, g<sub>d</sub>) and, therefore, improving g<sub>m</sub>/I<sub>d</sub> and g<sub>m</sub>/g<sub>d</sub>, as well as the Early voltage (V<sub>EA</sub>) (Figs 9-10). As a result, JAM FETs reach analog performances that are slightly better than IM devices, up to an A<sub>v0</sub>=68dB gain.

**Reliability:** we have performed Positive Bias Temperature Instability (PBTI) and Hot Carrier Injection (HCI) (Figs. 11-12). We demonstrate similar PBTI (88 years lifetime at  $V_{DD}$ =0.8V) and even better HCI for JL as compared to IM. It can be explain by a lower and shifted to the drain (not underneath the gate dielectric as for IM/JAM) peak electric field.

Low-frequency drain current noise measurements (Fig.13) show a 31-die average 1/f signature and a slightly lower input-referred gate voltage noise level (S<sub>Vg</sub>) for JAM. Using the Carrier number fluctuations with Correlated Mobility Fluctuations model [11] and taking into account the series resistance noise (S<sub>Rsd</sub>), we fitted the normalized drain current noise at f =10 Hz (Fig.28) to extract the volumetric oxide effective trap density  $N_T$ , and the remote Coulomb scattering coefficient  $\alpha_{sc}$  for all wafers. We extracted a value of  $N_T \approx$ 7.5 10<sup>17</sup> eV/cm<sup>3</sup> for all cases, reflecting a similar interface quality, independently of the conduction mode. This value is also very close to state-of-the-art N<sub>T</sub> values of high-k-metal-gate CMOS technologies [12]. Concerning  $\alpha_{sc}$ , a very similar value ( $\approx 4 \times 10^3$  Vs/C) is extracted for all wafers, showing that the remote Coulomb scattering is not affected by the different conduction modes. Finally, S<sub>Rsd</sub> has a significant impact only for JL, which can be linked to non-optimized source/drain doping [13].

**Mobility differences**, due to channel doping, are also translated into a cut-off frequency shift, measured at  $F_T$ =130 GHz for JAM vs. 136 GHz for IM. But JAM exceeds IM devices in terms of  $F_{MAX}$  because of lower gate capacitances  $C_{GDS}$ . We demonstrate a record  $F_{max}$ =182GHz for JAM nMOS (Fig 15).

#### **IV.** CONCLUSION

Finally, our devices feature record performance among junctionless nMOSFETs (Fig.16). Thanks to process technology variants, electrical characterizations and TCAD simulations, we demonstrated that such a technology offers a good tradeoff for mixed analog/digital applications.

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Fig.1: TEM cross section of JAM device. Excellent crystalline quality is observed.



Fig. 5 : Electron density (SD cut) highlighting three operation regimes



Fig.9: g<sub>d</sub> vs. L for V<sub>B</sub>=0 and V<sub>B</sub>=-10V. V<sub>B</sub><0 improves electrostatics.



power spectral density versus

frequency



Fig.2: Detailed Process flow for IM (N+-i-N+), JAM (N+-N-N+) and JL (N) devices.



Fig.6: Pelgrom plot (local variability)

Gain A <sub>vo</sub> (db)	V <sub>B</sub> (V)	L=100 nm	L=200 nm	L=400 nm	10
IMI	0	1	60	50	(s
	-10	18	65	53	ШЦ
MAL	0	1	12	51	⊢ <sub>10</sub>
	-10	64.5	68.8	55	
Л	0	59	47	38	
	-10	61.5	65	41	
					10

Fig.10: Gain A<sub>v0</sub> for different Gate length  $V_B < 0$  improves  $A_{v0}$ 





Fig.3: I<sub>ON</sub>-I<sub>OFF</sub> for L=35nm and W=20nm.



Fig.7:  $g_{\text{mMAX}}$  and SS as a function of

L for W=0.24µm

T=125°C

IM

AC

– model

L=0.1µm

W=10µm

3

0.92 V

AC land IM: V<sub>G-5y</sub>=0.92V

88y I @ V<sub>DD</sub>=0.8V

10

10<sup>-11</sup>

C<sub>ov</sub> (fF/µm): JAM: 0.48 <sub>G</sub>=-0.5 IM: 0.49 L=35nm JL: 0.42 (fF) ပ 0 0 8 12 4 W (µm)

Fig.4: C(W) and C<sub>GDS</sub> extraction at L=35nm. Inset: Schematic with parasitic capacitance contributions.



L (µm) Fig.8: Back bias efficiency for W=0.02µm and W=0.24µm.



Fig.11: Time-To-Failure for PBTI. The 5-year criterion is met



Fig.12: Time-To-Failure for HCI. The 5-year criterion is met.

Ref.	[14]	[6]	This Work
Techno	Low-Temp FDSOI	NW	NW/FDS OI
L <sub>G</sub> (nm)	80	48	35
V <sub>DD</sub> (V)	1.0	1.0	0.8
Α <sub>ντ</sub> (mV.μm)	-	4.4	1.4
A <sub>vo</sub> (dB)	55 for L=40nm	45 for L=80nm	68 for L=0.2μm
f <sub>T</sub> (GHz)	80	>70	126
f <sub>max</sub> (GHz)	117	~90	169

Fig.16: nMOS junctionless





 $V_{g}(V)^{2}$ 





