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RESEARCH PAPER

Design Criteria of X-Wave Launchers for Millimeter-Wave Applications

WALTER FUSCALDO¹, SANTI C. PAVONE², DAVIDE COMITE¹, GUIDO VALERIO³, MATTEO ALBANI², MAURO ETTORRE⁴, AND ALESSANDRO GALLI¹

Bessel-beam launchers are promising and established technologies for focusing applications at microwaves. Their use in time-domain leads to the definition of a new class of devices, namely, the X-wave launchers. In this work, we discuss the focusing features of such devices with a specific interest at millimeter waves. The spatial resolutions of such systems are described under a rigorous mathematical framework to derive novel operating conditions for designing X-wave launchers. These criteria might be particularly appealing for specific millimeter-wave applications. In particular, it is shown that an electrically-large aperture is not strictly required, as it seemed from previous works. However, the use of an electrically-small aperture would demand a considerably wideband capability. The various discussions presented here provide useful information for the design of X-wave launchers. This aspect is finally shown with reference to the practical design of two different X-wave launchers.

Keywords: Bessel beams, X-waves, nondiffracting waves, localized waves, millimeter waves.

I. INTRODUCTION

In applications such as security screening, medical imaging, and radiometry, to name but a few, it is important to focus the energy in the millimeter-wave range to benefit of both the non-ionizing character of millimeter waves and their millimeter resolution [1].

In this context, nondiffracting waves [2], especially Bessel beams and their polychromatic version, i.e., X-waves, are gaining much attention in the millimeter-wave community due to their remarkable focusing features [3]. Indeed, nondiffracting waves do not undergo diffraction within a certain depth of field. This is a striking difference with respect to more conventional beams and pulses such as Gaussian beams [4] and pulses [5, 6], or complex-source beams [7–9] and complex-source pulsed-beams [10, 11], that gradually spread their energy as they propagate. It is worth noting that Gaussian beams and pulses are *equivalent* to a *paraxial* complex spherical wave, whereas nondiffracting waves are *generalizations* of complex-source wavefields retaining bidirectional spectra [9, 12, 13].

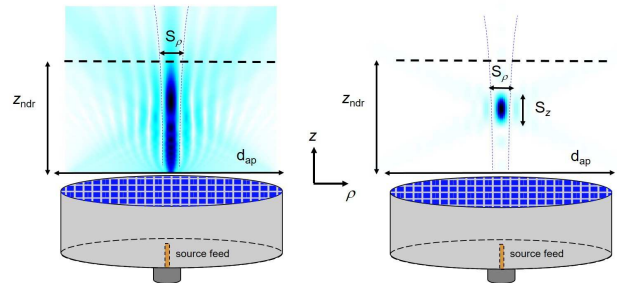


Fig. 1. Comparison between Bessel-beam launchers and X-wave launchers. On the left-side, a Bessel-beam launcher generating a transversely focused Bessel beam with transverse spot size S_ρ . The transverse profile is maintained along the longitudinal axis up to the nondiffractive range z_{ndr} . On the right side, the same device acts as an X-wave launcher in the time domain. The propagation of a focused X-wave is captured at a given time. Differently from the Bessel beam, the X-wave is confined also along the longitudinal z -axis where it exhibits a longitudinal spot size S_z considerably narrower than z_{ndr} .

Bessel beams [14] as well as X-waves [15] are able to maintain their nondiffractive character along the axis of propagation for a large distance (commonly known as *nondiffractive range*), as experimentally demonstrated in optics [16, 17] and acoustics [18, 19]. Interestingly, while Bessel beams are focused only along the transverse axis, X-waves are focused along both the transverse and the longitudinal axis (see Fig. 1).

Although Bessel beams have been experimentally generated in the microwave and millimeter-wave range (see, e.g., [20–24]), experimental results for X-waves are very recent [25, 26]. The reason for this lack of experimental results was motivated in [15]. There, the development of

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an original theoretical framework allowed the derivation of several physical insights on the problem. The proposed analysis revealed that the design of an *efficient* X-wave launcher (i.e., able to focus energy along both the transverse and the longitudinal axis) usually requires a moderate wavenumber dispersion, a wide fractional bandwidth, and an electrically-large aperture. These requirements actually limit the class of suitable planar devices at millimeter waves to those using holographic approaches [15, 27] and leaky-wave techniques [26, 28]. In those works, the design process takes advantage of the concept of metric of confinement introduced in [15]. The design criteria adopted in [15, 28] will be clearly outlined in this manuscript.

On this basis, we provide here a more general criterion for designing X-wave launchers. It is shown that an electrically-large aperture is not theoretically required to focus energy along both directions. Using the results recently outlined in [31], we show that the aperture size reduction dictates a larger fractional bandwidth.

The paper is organized as follows. In Section I, the focusing features of X-waves are briefly outlined. In Section II, we review the concept of metric of confinement and discuss some original aspects relevant for specific millimeter-wave applications. In Section III the criteria for designing electrically-small apertures are discussed. Section IV reports some practical design examples of X-wave launcher based on holographic approaches and leaky-wave techniques. Finally, conclusions are drawn in Section V.

II. NONDIFFRACTIVE FEATURES OF X-WAVES

X-waves can be obtained as a spectral superposition of Bessel beams with the same *axicon angle* [2] over a certain frequency bandwidth $\Delta\omega$. Throughout the paper, we consider only *spectrally-flat* X-waves, i.e., we tacitly assume a uniform frequency spectrum. Under this assumption, an X-wave can be expressed as

$$\chi(\rho, z, t) = \int_{\Delta\omega} J_0(k_\rho(\omega)\rho) e^{-jk_z(\omega)z} e^{j\omega t} d\omega, \quad (1)$$

where $J_0(\cdot)$ is a zeroth-order Bessel function of the first kind, ω is the angular frequency, t is time, k_ρ and k_z are the wavenumbers along the radial ρ and longitudinal z directions, respectively (related each other through the separation relation $k_0^2 = k_\rho^2 + k_z^2$, where k_0 is the free-space wavenumber), and $\theta = \arctan[k_\rho(\omega)/k_z(\omega)]$ is the axicon angle. Therefore, any Bessel-beam launcher characterized by a negligible wavenumber dispersion over a certain frequency bandwidth can act as an X-wave launcher in time domain. Moreover, since X-waves are obtained from Bessel beams, they inherit their focusing features.

Specifically, X-waves are localized along the transverse axis and maintain their transverse spot size S_ρ up to the nondiffractive range $z_{\text{ndr}} = \rho_{\text{ap}} \cot \theta$ [14], where ρ_{ap} is

the aperture radius. In addition, X-waves are also localized along the longitudinal axis. However, the longitudinal spot size S_z is narrow as long as a considerable fractional bandwidth is guaranteed [15]. This results from the definition of a suitable metric of confinement for X-waves [15, 31], as we will review in the following paragraphs.

We note here that throughout the paper we always refer to zeroth-order X-waves, although higher-order X-waves are of extreme interest for their capability to carry orbital angular momentum [29, 30].

A) Metric of confinement

The definition of metric of confinement states that X-waves are localized if and only if both their transverse S_ρ and longitudinal S_z spot sizes are smaller than the aperture diameter $d_{\text{ap}} = 2\rho_{\text{ap}}$ and the nondiffractive range z_{ndr} , respectively [15, 31]. Hence, by introducing the following confinement ratios

$$C_\rho = S_\rho/d_{\text{ap}}, \quad (2)$$

$$C_z = S_z/z_{\text{ndr}} \quad (3)$$

the definition of the following metric of confinement is straightforward:

$$C_{\rho,z} = \begin{cases} 1 & \text{if } \max(C_\rho, C_z) > 1, \\ C_\rho C_z & \text{elsewhere,} \end{cases} \quad (4)$$

Clearly, when $C_{\rho,z} < 1$, the resulting X-wave will be confined along *both* the radial *and* the longitudinal direction. Therefore, the value of $C_{\rho,z}$ gives a well-defined measure of the localization of the energy. Otherwise, when $C_{\rho,z} = 1$ the X-wave would not be confined along the radial *or* the longitudinal direction.

B) Resolutions along the radial and longitudinal axis

For *spectrally-flat* X-waves, the expressions for S_ρ and S_z and in turn C_ρ and C_z can be found in analytical closed form [15]. In the case of a *nondispersive* X-wave characterized by an axicon angle θ and generated at the carrier angular frequency ω_0 around a fractional bandwidth $\text{FBW} = \Delta\omega/\omega_0$, S_ρ and S_z take the following expressions:

$$S_\rho = \frac{j_{0,1} \lambda_0}{\pi \sin \theta}, \quad (5)$$

$$S_z = \frac{2\lambda_0}{\text{FBW} \cos \theta}, \quad (6)$$

where $j_{0,1} \simeq 2.405$ is the first null of the zeroth-order Bessel function of the first kind, and $\lambda_0 = 2\pi/k_0$, with $k_0 = \omega_0/c_0$ c_0 being the light speed in vacuum. From these expressions, we note that the transverse resolution S_ρ is lower-bounded by $S_{\rho,a} \simeq \lambda_0/1.3$ (which is close to the Abbe diffraction limit for optical systems [32]). Conversely, the longitudinal resolution S_z could be reduced

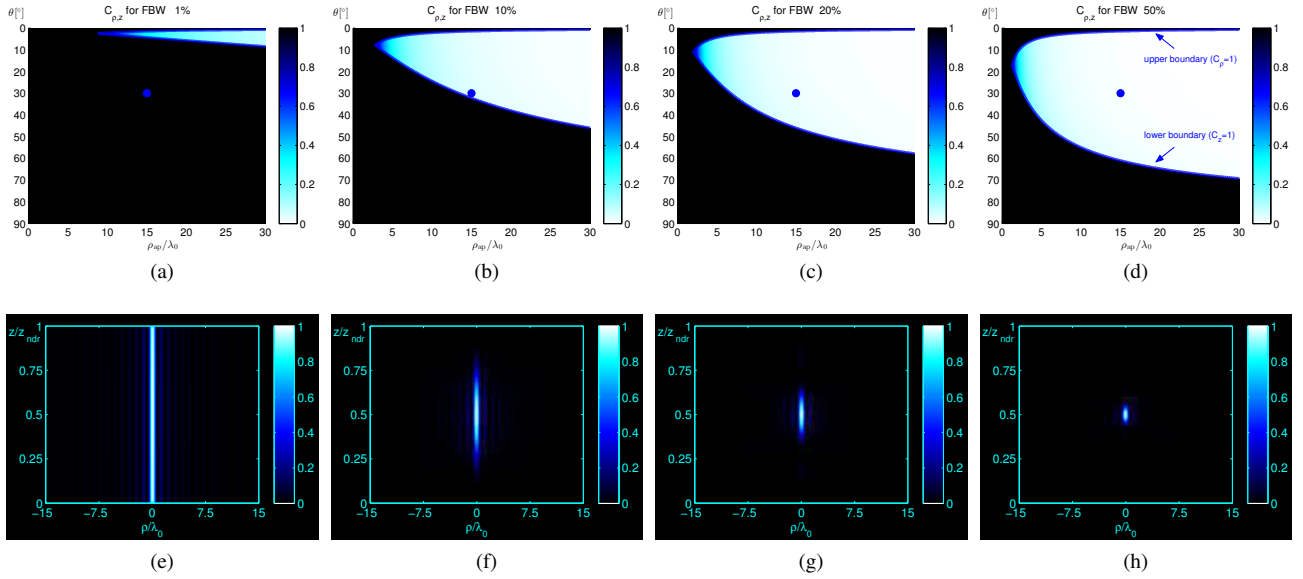


Fig. 2. (a)–(d) The metric of confinement $C_{\rho,z}$ vs. ρ_{ap}/λ_0 and θ is shown for (a) FBW = 1%, (b) FBW = 10%, (c) FBW = 20%, and (d) FBW = 50%. In the black region $C_\rho > 1$ or $C_z > 1$, thus the resulting X-wave is not confined along *both* axes. The solid blue lines represent the boundaries for which $C_\rho = 1$ (upper boundary) and $C_z = 1$ (lower boundary). The two boundaries cross at $\theta = \theta'_{eq}$ and ρ'_{eq} . (e)–(h) Normalized field intensities $|\chi(\rho, z, t_0)|^2 / \max |\chi|^2$ vs. ρ/λ_0 and z/z_{ndr} for an X-wave generated with $\theta = 30^\circ$ and $\rho_{ap} = 15\lambda_0$ (see the blue dots in Figs. 2(a)–(d)). The time t is set at the instant t_0 for which the X-wave peak has reached $z = z_{ndr}$.

arbitrarily, in principle, by increasing the fractional bandwidth. Nonetheless, planar radiating structures operating at millimeter waves are generally limited to 10–20% fractional bandwidths, thus S_z is also practically lower-bounded by $2\lambda_0$.

It is worth noting here that the axicon angle has an opposite effect over the longitudinal and the transverse resolutions, and thus it is not possible to narrow the resolution along both directions by solely changing the axicon angle.

From the expressions of S_ρ and S_z , the confinement ratios C_ρ and C_z are readily found

$$C_\rho = \frac{j_{0,1}}{(\rho_{ap}/\lambda_0)2\pi \sin \theta}, \quad (7)$$

$$C_z = \frac{2 \sin \theta}{(\rho_{ap}/\lambda_0)\text{FBW} \cos^2 \theta}. \quad (8)$$

As for S_ρ and S_z , it is manifest that the FBW is the only parameter that allows for improving the longitudinal localization without negatively affecting the transverse one.

A numerical demonstration of this aspect is shown in Fig. 2. In Figs. 2(a)–(d), the metric of confinement $C_{\rho,z}$ is reported as a contour-plot of the two variables ρ_{ap}/λ_0 and θ for values of FBW = 1%, 10%, 20%, 50%. The black region represents the pairs of values $(\rho_{ap}/\lambda_0, \theta)$ for which the X-wave is not confined along at least one direction for a given FBW. This region is determined by the intersection of the two boundaries (see blue solid lines) that represent the condition for which $C_\rho = 1$ (upper boundary) and $C_z = 1$ (lower boundary), respectively.

As is shown, the region within these two boundaries increases as the FBW increases. In other words, a pair of values $(\rho_{ap}/\lambda_0, \theta)$ for which the X-wave is not confined

for a certain FBW, might be confined for a higher FBW. This is corroborated by numerical results (see Figs. 2(e)–(f)) considering a non-dispersive X-wave characterized by the pair $(\rho_{ap}/\lambda_0 = 15, \theta = 30^\circ)$ (see blue dots in Figs. 2(a)–(d)). As is shown, for low fractional bandwidths the X-wave is loosely localized along the longitudinal axis (the main spot has a ‘needle-like’ shape), whereas it is tightly focused (the main spot has a ‘bullet-like’ shape) for high fractional bandwidths.

C) Resolution and confinement equalization

For some applications, such as medical imaging, it could be more important to have the same resolution along both axes (i.e., $S_\rho = S_z$). Then, it is interesting to find the operating conditions equalizing the transverse and longitudinal resolutions. By equating the expressions of S_ρ and S_z , it is readily found that the equalization condition for resolution is obtained for θ_{eq} :

$$\theta_{eq} = \arctan \left(\frac{j_{0,1}\text{FBW}}{2\pi} \right) \simeq \arctan(0.38\text{FBW}). \quad (9)$$

Inserting (9) into (5) we get an expression for the equalized spot-size

$$S_{eq} = \lambda_0 \sqrt{\left(\frac{j_{0,1}}{\pi} \right)^2 + \left(\frac{2}{\text{FBW}} \right)^2}, \quad (10)$$

which represents the diameter of the resulting spherical-like pulse generated by the considered aperture. As expected, the minimum equalized spot-size is achieved for a theoretically infinite bandwidth. More precisely, S_{eq}

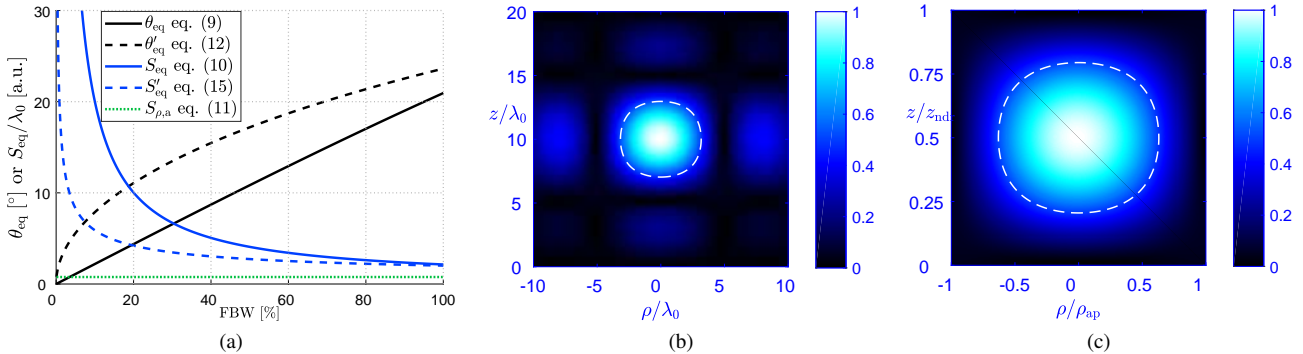


Fig. 3. (a) θ_{eq} , θ'_{eq} (in degrees) and S_{eq} , S'_{eq} (normalized to λ_0) vs. the fractional bandwidth in the range $0 \leq \text{FBW} \leq 1$. The green dotted line identifies the asymptotic limit $S_{\text{eq}} \sim S_{\rho,a}$. (b) Numerical demonstration of the resolution equalization condition (i.e., equation (9)). Normalized field amplitude $|\chi(\rho, z, t_0)| / \max |\chi|$ vs. ρ/λ_0 and z/λ_0 for an X-wave generated with $\text{FBW} = 20\%$ and $\theta_{\text{eq}} \simeq 4.38^\circ$. The aperture radius is fixed at $\rho_{\text{ap}} = 10\lambda_0$ without loss of generality. (c) Numerical demonstration of the confinement equalization condition (i.e., Eq. (12)). Normalized field amplitude $|\chi(\rho, z, t_0)| / \max |\chi|$ vs. ρ/ρ_{ap} and z/z_{ndr} for an X-wave generated with $\text{FBW} = 20\%$ and $\theta_{\text{eq}} \simeq 11.06^\circ$. The aperture radius is fixed at $\rho_{\text{ap}}/\lambda_0 = \rho'_{\text{eq}}$ to also show the validity of Eq. (14). Indeed, the field amplitude vanishes as $\rho \rightarrow \rho_{\text{ap}}$ along the transverse axis, and $z \rightarrow z_{\text{ndr}}$ along the propagation axis. In both (b) and (c) the time t is set at the instant t_0 for which the X-wave peak has reached $z = z_{\text{ndr}}/2$. The white dashed line highlights the iso-line of the half-amplitude level.

asymptotically approaches the limit $\lambda_0/1.3$ (represented as a green dotted line in Fig. 3(a)) as pointed out in Section II.B, i.e.,

$$\lim_{\text{FBW} \rightarrow \infty} S_{\text{eq}} = S_{\rho,a} \simeq \lambda_0/1.3. \quad (11)$$

In Fig. 3(a) we have reported Eqs. (9) and (10) on the same plot to show the variation of θ_{eq} and S_{eq} , respectively, as functions of the fractional bandwidth in the range $0 \leq \text{FBW} \leq 1$.

In order to numerically validate the equalization condition we generate an X-wave with a fractional bandwidth $\text{FBW} = 20\%$ at which corresponds $\theta_{\text{eq}} \simeq 4^\circ$ (see Fig. 3(a)). The contour plot of the generated field distribution over the ρz -plane for this numerical example is reported in Fig. 3(b).

This equalization criterion gives the design rule to equalize the *absolute* resolutions (that is dimensionally a length). However, for some other applications, such as wireless power transfer, it could be more interesting to have the same *relative* resolutions (adimensional) with respect to the aperture diameter and the nondiffractive range, for the radial and longitudinal direction, respectively. This can be accomplished by equalizing the confinement ratios along both axes (i.e., $C_\rho = C_z$) in place of the resolutions (i.e., $S_\rho = S_z$). In that case, Eq. (9) should be replaced by

$$\theta'_{\text{eq}} = \arctan \sqrt{\frac{j_{0,1} \text{FBW}}{4\pi}} \simeq \arctan (0.44 \sqrt{\text{FBW}}) \quad (12)$$

where θ'_{eq} gives the condition for equalizing the confinement ratios rather than the resolutions. Inserting (12) into (7) we get an expression for the equalized confinement

$$C_{\text{eq}} = \frac{\lambda_0}{\rho_{\text{ap}}} \frac{2\pi}{j_{0,1}} \sqrt{\frac{j_{0,1}}{2\pi} + \frac{2}{\text{FBW}}}, \quad (13)$$

which depends on both the normalized aperture radius and the fractional bandwidth. However, from Figs. 2(a)–(d) we

note that a point exists for which the upper and lower boundaries of the metric of confinement (see blue lines in Figs. 2(a)–(d)) meet at $\theta = \theta'_{\text{eq}}$ and $\rho_{\text{ap}}/\lambda_0 = \rho'_{\text{eq}}$ with

$$\rho'_{\text{eq}} = \frac{j_{0,1}}{2\pi \sin \theta'_{\text{eq}}}. \quad (14)$$

At this point (viz., $(\theta'_{\text{eq}}, \rho'_{\text{eq}})$) we clearly have $C_\rho = C_z = 1$. This point represents the limit condition of confinement equalization. In other words, for any aperture radius such that $\rho_{\text{ap}}/\lambda_0 > \rho'_{\text{eq}}$, the resulting X-wave will be efficiently confined and equalized (with respect to the relative resolutions), provided that $\theta = \theta'_{\text{eq}}$. The absolute resolution achieved under the confinement equalization condition S'_{eq} is obtained from Eq. (5), by replacing C_ρ with C_{eq} (see Eq. (13)), and reads

$$S'_{\text{eq}} = \frac{\pi \lambda_0}{j_{0,1}} \sqrt{\frac{j_{0,1}}{2\pi} + \frac{2}{\text{FBW}}}, \quad (15)$$

which clearly does not depend on ρ_{ap} . The behavior of S'_{eq} as a function of FBW is reported in Fig. 3(a) in dashed blue lines.

In order to numerically validate the equalization condition for the relative resolutions we generate an X-wave with a fractional bandwidth $\text{FBW} = 20\%$ at which corresponds $\theta'_{\text{eq}} \simeq 11^\circ$ (see Fig. 3(a)). We then fixed $\rho_{\text{ap}}/\lambda_0 = \rho'_{\text{eq}}$ to validate also Eq. (14). As a consequence, the generated X-wave should have a transverse spot size equal to the aperture diameter and a longitudinal spot size equal to the nondiffractive range (recall the definitions of C_ρ and C_z in Eqs. (7)–(8)). This can be easily inferred from Fig. 3(c), where the contour plots over the ρz -plane have been reported over the range $0 < \rho < \rho_{\text{ap}}$ and $0 < z < z_{\text{ndr}}$.

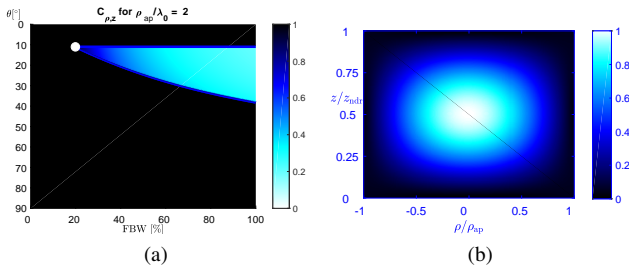


Fig. 4. (a) The metric of confinement $C_{\rho,z}$ vs. FBW and θ is shown for $\rho_{\text{ap}} = 2\lambda_0$. The blue solid lines represent the boundaries for which $C_\rho = 1$ and $C_z = 1$. The two boundaries cross at $\text{FBW} = \text{FBW}_{\text{min}} \simeq 20\%$ and $\theta = \theta_{\text{min}} \simeq 11^\circ$ (see the white dot). (b) Numerical demonstration of the minimum bandwidth condition for electrically-small apertures (i.e., Eqs. (16)–(17)). Normalized field amplitude $|\chi(\rho, z, t_0)| / \max |\chi|$ vs. ρ/ρ_{ap} and z/z_{ndr} for an X-wave generated with FBW_{min} and θ_{min} . The aperture radius is fixed at $\rho_{\text{ap}} = 2\lambda_0$. The time t is set at the instant t_0 for which the X-wave peak has reached $z = z_{\text{ndr}}/2$.

III. ALTERNATIVE CRITERIA FOR ELECTRICALLY-SMALL APERTURES

In the previous Section II, we focused on the equalization of the resolutions. Also, from the definitions of the confinement ratios we demonstrated that X-wave launchers generally require an electrically-large aperture, a considerable fractional bandwidth, and a low axicon angle [15]. In fact, from Figs. 2(a)–(b), it appears that for low fractional bandwidths and electrically-small apertures, an efficient confinement is possible only for few values of the axicon angle around 10° . However, the situation is rather different for higher fractional bandwidths (see Figs. 2(c)–(d)), for which efficient confinement is possible for axicon angles up to 40° .

All these aspects can be recast in a more systematic way, by inspecting Eqs. (7) and (8). In fact, from Eq. (7) one finds that the condition for having a transversely localized X-wave (i.e., $C_\rho < 1$) gives an aperture radius $\rho_{\text{ap}}/\lambda_0 > (j_{0,1}/2\pi) \csc \theta$, which is lower-bounded by the value $j_{0,1}/2\pi \simeq 0.38$. This means that, even for such electrically-small apertures, a minimum axicon angle θ_{min} would exist beyond which the X-wave will be transversely confined

$$\theta > \theta_{\text{min}} = \arcsin \left(\frac{j_{0,1}}{2\pi \rho_{\text{ap}}/\lambda_0} \right). \quad (16)$$

However, since the metric of confinement is lower than 1 if and only if $\max\{C_\rho, C_z\} < 1$ (see Eq. (4)), the X-wave has to be also longitudinally confined (i.e., $C_z < 1$) at θ_{min} . Then, from Eq. (7), it results that the fractional bandwidth has to be greater than a minimum value FBW_{min}

$$\text{FBW} > \text{FBW}_{\text{min}} = \frac{2 \tan \theta_{\text{min}}}{(\rho_{\text{ap}}/\lambda_0) \cos \theta_{\text{min}}}. \quad (17)$$

In this regard, in Fig. 4(a), we have reported the metric of confinement as a contour plot of the two variables θ and FBW for $\rho_{\text{ap}}/\lambda_0 = 2$. As is shown, even for

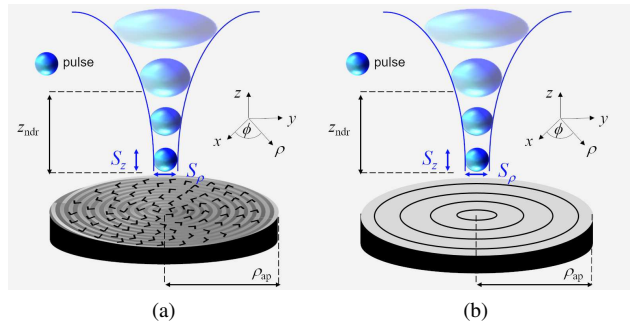


Fig. 5. (a)–(b) An X-wave is launched from (a) an RLSA antenna and (b) an LPRW. As the X-wave pulse propagates beyond the nondiffractive range z_{ndr} , the spot size increases and the intensity correspondingly decreases.

such an electrically-small aperture, the conditions given by Eqs. (16) and (17) identify a region where X-waves can be efficiently confined along both directions. In this numerical example, the limiting condition for the confinement is obtained for $\theta_{\text{min}} \simeq 11^\circ$ and $\text{FBW}_{\text{min}} \simeq 20\%$ (highlighted with a white dot in Fig. 4(a)) and corresponding to the intersection of the boundaries $C_\rho = 1$ and $C_z = 1$ (see both blue solid lines in Fig. 4(a)).

As for the previous equalization conditions, the numerical validation of Eqs. (16)–(17) is performed by generating an X-wave with $\rho_{\text{ap}} = 2\lambda_0$, $\text{FBW} = 20\%$, and $\theta = 11^\circ$, corresponding to the operating conditions highlighted with a white dot in Fig. 4(a). The contour plot over the ρz -plane for this numerical example is reported in Fig. 4(b). As is seen, the normalized field amplitude approaches the first null exactly at $\rho = \rho_{\text{ap}}$ and at $z = z_{\text{ndr}}$. We note that the field localization along the transverse plane improves as θ is increased (see Eq. (7)). This increase in the value of θ , would in turn adversely affect the longitudinal confinement which, however, can be improved by increasing FBW (see Eq. (8)). More generally, for the considered fixed aperture radius, viz., $\rho_{\text{ap}} = 2\lambda_0$, the pairs of θ and FBW values that grant an efficient confinement are represented by the bright region in Fig. 4(a).

IV. DESIGN OF X-WAVE LAUNCHERS

We consider here the design of two X-wave launchers both based on radial waveguides, but following two different approaches: a radial-line slot array (RLSA) antenna [33] (see Fig. 5(a)), adopting a holographic design principle [34], and a leaky periodic radial waveguide (LPRW) [28] (see Fig. 5(b)), exploiting leaky-wave theory [35].

Among the various planar realizations of microwave and millimeter-wave Bessel-beam launchers (see, e.g., [20–24] and references therein), the RLSA and the LPRW both possess wideband capabilities, and therefore they can also operate as X-wave launchers. The details for the accurate design of these devices as X-wave launchers can be found in [15] and in [28], for RLSA antennas and for LPRWs, respectively. Here we report the design of two X-wave

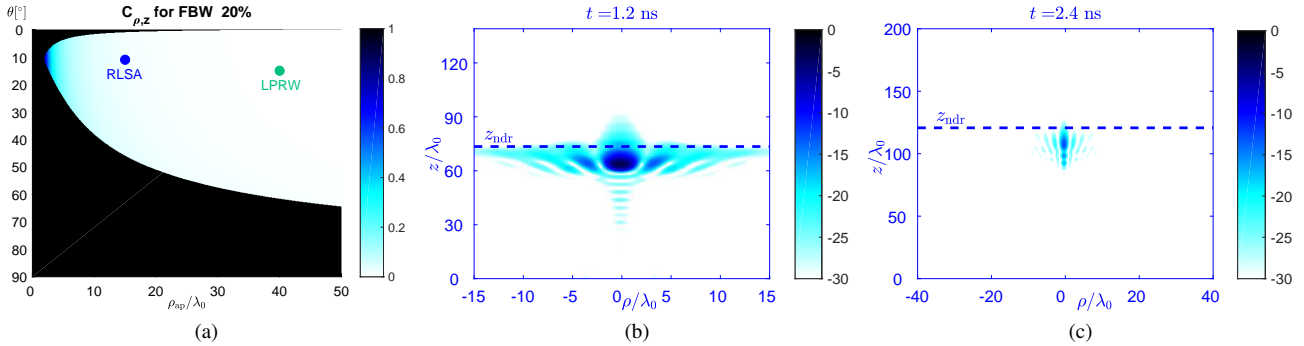


Fig. 6. (a) The confinement metric for a fractional bandwidth of 20%. A blue (green) dot represents the operating conditions of an RLSA(LRPW) design. (b)–(c) Contour plot of an X-wave launched by (b) an RLSA antenna and (c) by a LRPW with operating conditions as in (a). The X-wave is captured at a time instant for which the wavefront has traveled the distance $z = z_{ndr}/2$.

launchers to validate the analysis developed in the previous Sections.

A) RLSA antenna design

An RLSA antenna is a radial waveguide where slots are etched on the upper-plate in order to synthesize a prescribed aperture distribution for radiation (see Fig. 5(a)). Specifically, in [27, 33], the radiating slots are arranged in a spiral path to synthesize an inward cylindrical-wave aperture distribution creating a Bessel beam. Once the aperture radius, the operating frequency, and axicon angle are fixed, the position and the size of each slot are optimized according to an in-house method-of-moments (MoM) code [34]. Since the operating fractional bandwidth of an RLSA antenna easily achieves 10%–20%, the efficient generation of X-waves is granted for several combinations of the design parameters, i.e., ρ_{ap} and θ_0 , as shown in Fig. 6(a), reporting the metric of confinement for FBW = 20%. Indeed, all pairs (ρ_{ap}, θ) lying in the bright region of Fig. 6(a) can be used to launch efficiently-confined X-waves.

As an example, we report in Fig. 6(b) the contour plot of an X-wave generated at $f = 60$ GHz by an RLSA with $\rho_{ap} = 15\lambda_0$, and $\theta = 11^\circ$ corresponding to the operating conditions of the blue dot in Fig. 6(a). The resulting X-wave is obtained through the numerical evaluation of the radiation integral and the Fourier-Transform of the spectral field as extensively discussed in [15]. This result will be commented further on, in comparison with the result obtained through an LRPW design.

B) LRPW design

An LRPW is a radial waveguide where the upper plate is replaced by an annular periodic metal strip grating (MSG). As shown in [28], the period of the MSG is chosen so as to excite a *backward* leaky-wave in the frequency range of interest. We recall here that leaky waves are characterized by a complex propagation wavenumber $k_\rho = \beta - j\alpha$ where the imaginary part, the *leakage rate*, accounts for radiation losses [35]. Backward (forward) leaky waves are

those for which the group and phase velocities are in opposite (identical) directions [35]. As a result, the backward character of the excited leaky wave allows for recovering the inward character required for the generation of a Bessel beam (see [28] for a rigorous theoretical explanation). However, as opposed to RLSA antennas where the radiation efficiency is entirely controlled during the MoM optimization procedure [23], the radiation efficiency of an LRPW depends on the product between the leakage rate and the aperture radius of the device. As a consequence, a low value of the leakage rate (which is desirable to improve the focusing efficiency of the LRPW [28]) demands for a relatively large aperture radius, in order to obtain a satisfactory radiation efficiency (commonly above 90% [24]).

As an example, we report in Fig. 6(c) the contour plot of an X-wave generated at $f = 60$ GHz by an LRPW with $\rho_{ap} = 40\lambda_0$, and $\theta = 15^\circ$, corresponding to the operating conditions highlighted with a green dot in Fig. 6(a). At this point, it is worth to compare the results in Figs. 6(b)–(c) obtained through an RLSA and an LRPW. As is manifest, the X-wave launched by an RLSA is less confined, but more intense than the one launched through by an LRPW. These two aspects have a clear physical explanation.

With regard to the confinement efficiency, it could be predicted from Fig. 6(a) that the operating conditions of the LRPW antenna would lead to a more confined X-wave with respect to the RLSA design. The former lay in a region of higher confinement (the color is brighter) according to the metric $C_{\rho,z}$. However, it should be stressed that this improved confinement is only due to the larger aperture radius of the LRPW with respect to the RLSA. On the other hand, the confinement of the X-wave generated through the RLSA design can be improved by using a larger radiating aperture.

With regard to the different field intensities, we should comment on the role of the leakage constant in the LRPW. Indeed, the complex nature of leaky waves contribute to *taper* the aperture field distribution with an exponentially-decaying function of the type $\exp(-\alpha\rho)$. As a consequence, the resulting X-wave does not present tails in the

radial direction, but it experiences a higher field attenuation while propagating along the longitudinal direction (the interested reader may find a detailed analysis on this aspects in [28]), with respect to the one generated through an RLSA design.

As a final comment we should discuss the energy tails surrounding the main spot of both X-waves. Indeed, an ideal X-wave would have tails outlining an X-shape (precisely, the axicon angle determines the intersection angle of the two branches forming the X-shape). Conversely, the X-waves reported in Figs. 6(b)–(c) spread their energy even in other directions, partly masking the X-shape of the tails. This phenomenon is due to the dispersive features of an RLSA antenna and an LPRW, as shown in [27] and [28], respectively.

V. CONCLUSION

In this work, the design of efficient X-wave launchers has been discussed from an original perspective. By reviewing the concept of metric of confinement, we discussed alternative criteria for designing X-wave launchers that might be of interest for specific millimeter-wave applications. In particular, it has been shown that X-wave launchers can be designed to get either the same resolutions or the same confinement ratios along the longitudinal and transversal axes. Also, we discussed the possibility to design an efficient X-wave launcher without requiring an electrically-large aperture. This comes at the expense of a larger fractional bandwidth. The discussion outlined in this work can be beneficial for the design process of X-wave launchers. The design of two X-wave launchers based on holographic and leaky-wave approach has been outlined to validate the results of the paper. The possibility to extend these results to the relevant case of dispersive X-waves is envisaged for future works.

REFERENCES

- [1] M. Ettore, S. C. Pavone, M. Casaletti, M. Albani, A. Mazzinghi, and A. Freni: Near-field focusing by non-diffracting Bessel beams, in *Aperture Antennas for Millimeter and Sub-Millimeter Wave Applications*, A. Boriskin and R. Sauleau, Springer, Cham, 2018.
- [2] H. E. Hernández-Figueroa, M. Zamboni-Rached, and E. Recami: *Nondiffracting Waves*, John Wiley & Sons, New York, 2013.
- [3] D. McGloin and K. Dholakia: Bessel beams: diffraction in a new light, *Contemporary Physics*, **46**(1) (2005), 15–28.
- [4] G.A. Deschamps: Gaussian beam as a bundle of complex rays, *Electronics Lett.*, **7**(23) (1971), 684–685.
- [5] A. N. Norris, B. S., White, and J. R. Schrieffer: Gaussian wave packets in inhomogeneous media with curved interfaces, *Proc. Roy. Soc. London. A: Math. Phys. Sci.*, **412**(1842) (1987), 93–123.
- [6] R. W. Ziolkowski: Exact solutions of the wave equation with complex source locations, *J. Math. Phys.*, **26**(4) (1985), 861–863.
- [7] N. L. Tsitsas, C. A. Valagiannopoulos, and A. I. Nosisch: Scattering and absorption of a complex source point beam by a grounded lossy dielectric slab with a superstrate, *J. Opt.*, **16**(10) (2014), 105712.
- [8] N. L. Tsitsas: Complex source point beam excitation of a lossless dielectric slab, *IEEE Int. Conf. Mathematical Methods in Electromagn. Theory (MMET 2016)*, Lviv, Ukraine, 2016.
- [9] A. M. Tagirdzhanov, A. S. Blagovestchenskii, and A. P. Kiselev: ‘Complex source’ wavefields: sources in real space, *J. Phys. A: Mathematical and Theoretical*, **44**(42) (2011), 425203.
- [10] E. Heyman and B. Z. Steinberg: Spectral analysis of complex-source pulsed beams, *J. Opt. Soc. Am. A*, **4**(3) (1987), 473–480.
- [11] E. Heyman and L. B. Felsen: Complex-source pulsed-beam fields, *J. Opt. Soc. Am. A*, **6**(6) (1989), 806–817.
- [12] E. Heyman: Focus wave modes: a dilemma with causality, *IEEE Trans. Antennas Propag.*, **37**(12) (1989), 1604–1608.
- [13] A. P. Kiselev: Localized light waves: Paraxial and exact solutions of the wave equation (a review), *Optics and Spectroscopy*, **102**(4) (2007), 603–622.
- [14] J. Durnin, J. J. Miceli Jr., and J. H. Eberly: Diffraction-free beams, *Phys. Rev. Lett.*, **58**(15) (1987), 1499–1501.
- [15] W. Fuscaldo, S. C. Pavone, G. Valerio, M. Albani, M. Ettore, and A. Galli: Analysis of limited-diffractive and limited-dispersive X-waves generated by finite radial waveguides, *J. App. Phys.*, **119**(19) (2016), 194903.
- [16] P. Saari and K. Reivelt: Evidence of X-shaped propagation-invariant localized light waves, *Phys. Rev. Lett.*, **79**(21) (1997), 4135.
- [17] H. Sónajal, M. R’atsep, and P. Saari: Demonstration of the Bessel-X pulse propagating with strong lateral and longitudinal localization in a dispersive medium, *Opt. Lett.*, **22**(5) (1997), 310–312.
- [18] R. W. Ziolkowski, D. K. Lewis, and B. D. Cook: Evidence of localized wave transmission, *Phys. Rev. Lett.*, **62**(2) (1989), 147.
- [19] J.-Y. Lu and J. F. Greenleaf: Experimental verification of non-diffracting X waves, *IEEE Trans. Ultrason., Ferroelectr. Freq. Control*, **39**(3) (1992), 441–446.
- [20] Z. Li, K. B. Alici, H. Caglayan, and E. Ozbay: Generation of an axially asymmetric Bessel-like beam from a metallic subwavelength aperture, *Phys. Rev. Lett.*, **102**(14) (2009), 143901.
- [21] M. Ettore, S. M. Rudolph, and A. Grbic: Generation of propagating Bessel beams using leaky-wave modes: experimental validation, *IEEE Trans. Antennas and Propag.*, **60**(6) (2012), 2645–2653.
- [22] W. Fuscaldo, G. Valerio, A. Galli, R. Sauleau, A. Grbic, and M. Ettore: Higher-order leaky-mode Bessel-beam launcher, *IEEE Trans. Antennas and Propag.*, **64**(3) (2016), 904–913.
- [23] S. C. Pavone, M. Ettore, and M. Albani: Analysis and design of Bessel beam launchers: Longitudinal polarization, *IEEE Trans. Antennas Propag.*, **64**(6) (2016), 2311–2318.
- [24] D. Comite, W. Fuscaldo, S. K. Podilchak, P. D. H. Re, V. Guillamón-Gómez Buendía, P. Burghignoli, P. Baccarelli, and A. Galli: Radially periodic leaky-wave antenna for Bessel beam generation over a wide-frequency range, *IEEE Trans. Antennas Propag.*, **66**(6) (2018), 2828–2843.
- [25] N. Chiotellis, V. Mendez, S. M. Rudolph, and A. Grbic: Experimental demonstration of highly localized pulses (X waves) at microwave frequencies, *Phys. Rev. B*, **97**(8) (2018), 085136.
- [26] D. Comite, W. Fuscaldo, S. K. Podilchak, V. Gómez-Guillamón Buendía, V., P. D. H. Re, P. D., P. Baccarelli, P. Burghignoli, and A. Galli: Microwave generation of X-waves by means of a planar leaky-wave antenna, *Appl. Phys. Lett.*, **113**(14) (2018), 144102.
- [27] S. C. Pavone, A. Mazzinghi, A. Freni, and M. Albani: Comparison between broadband Bessel beam launchers based on either Bessel or Hankel aperture distribution for millimeter wave short pulse generation, *Opt. Express*, **25**(16) (2017), 19548–19560.
- [28] W. Fuscaldo, D. Comite, A. Boesso, P. Baccarelli, P. Burghignoli, and A. Galli: Focusing Leaky Waves: A Class of Electromagnetic

Localized Waves with Complex Spectra, *Phys. Rev. App.*, **9**(5) (2018), 054005.

- [29] M. Ornigotti, C. Conti, and A. Szameit: Effect of Orbital Angular Momentum on Nondiffracting Ultrashort Optical Pulses, *Phys. Rev. Lett.*, **115**(10) (2015), 100401.
- [30] D. Comite, W. Fuscaldo, S. C. Pavone, G. Valerio, M. Ettore, M. Albani, and A. Galli: Propagation of nondiffracting pulses carrying orbital angular momentum at microwave frequencies, *App. Phys. Lett.*, **110**(11) (2017), 114102.
- [31] W. Fuscaldo, S. C. Pavone, G. Valerio, A. Galli, M. Albani, and M. Ettore: Parameterization of the nondiffractive features of electromagnetic localized pulses, *IEEE Int. Symp. Antennas Propag. (APS-URSI 2016)*, Fajardo, Puerto Rico, 2016.
- [32] M. Born and E. Wolf, *Principles of Optics*, Cambridge University Press, 1997.
- [33] S. C. Pavone, M. Ettore, M. Casaletti, and M. Albani: Transverse circular-polarized Bessel beam generation by inward cylindrical aperture distribution, *Opt. Express* **24**(10) (2016), 11103–11111.
- [34] M. Albani, S. C. Pavone, M. Casaletti, and M. Ettore: Generation of non-diffractive Bessel beams by inward cylindrical traveling wave aperture distributions, *Opt. Express*, **22**(15) (2014), 18354–18364.
- [35] D. R. Jackson and A. A. Oliner: Leaky-wave antennas, in *Modern Antenna Handbook*, C. A. Balanis, John Wiley & Sons, New York, 2011.



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