

Interval observers design for continuous-time linear switched systems

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Interval observers design for continuous-time linear switched systems Haifa Ethabet [∗] Tarek Raissi [∗] Messaoud Amairi [∗] Haifa Ethabet [∗] Tarek Raissi ∗∗ Messaoud Amairi [∗] Interval observers design for continuous-time linear switched systems Interval observers design for continuous-time linear switched systems

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systems. The considered systems are subject to disturbances which are assumed to be unknown systems. The considered systems are subject to disturbances which are assumed to be unknown
but bounded. First, observer gains are computed to ensure the stability of the estimation error. Then, under some changes of coordinates an interval observer is designed. Efficiency of the proposed method is demonstrated through a numerical example. Abstract: This paper is devoted to investigate interval observers design for linear switched proposed memory is demonstrated an ought a numerical example.

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© 2017, IFAC (International Federation of Automatic Control) Hosting by Elsevier Ltd. All rights reserved. \sim 2011, it is continuous that is depended to recommend solution produce systems for the region sub- \oslash 2017, IFAC (International Federation of Automatic Control) Hosting b

Keywords: Interval observer, stability, continuous-time switched systems, hybrid systems, cooperativity cooperativity Keywords: Interval observer, stability, continuous-time switched systems, hybrid systems, Keywords: Interval observer, stability, continuous-time switched systems, hybrid systems, cooperativity

1. INTRODUCTION 1. INTRODUCTION 1. **INTRODUCTION**

Switched systems are one of the most important classes of Hybrid Dynamical Systems (HDS). They consist of contintry order by harmon systems (HDs). They consider contain
uous or discrete-time subsystems where only one subsystem is active at each time. A switching rule orchestrating among them. Several works treated switched systems such anong them. Several world treated swretched systems such
as (Djemai and Defoort (2015); Liberzon (2012); Branicky (1998)). Among these studies, control and state estimation (1998)). were crucial and fundamental problems. were crucial and fundamental problems. (1998)). Among these studies, control and state estimation In order to stability the stability of systems, specifically systems, were crucial and fundamental problems.

In order to study the stability of switched systems, specific results have been developed. For example, in (Hu et al. results have been developed. For example, in (Hu et al. (2002)) a common Lyapunov function yields sufficient conditions for the global asymptotic stability. However, it $\frac{1}{2}$ conditions for the global asymptotic stability. However, it must be possible to get this common function. Therefore, multiple Lyapunov functions were proposed for instance in (Liberzon and Morse (1999)). Approaches Therefore, multiple Lyapunov functions were proposed may not always be possible to get this common function. for instance in (Liberzon and Morse (1999)). Approaches based on average dwell time have also been studied to for instance in (Liberzon and Morse (1999)). Approaches Therefore, multiple Lyapunov functions were proposed for instance in (Liberzon and Morse (1999)). Approaches ensure the stability of switched systems (see Hetel (2007); Serres et al. (2011) and the references therein). Serres et al. (2011) and the references therein). Serres et al. (2011) and the references therein).

As far as the stability problem is widely concerned, it is The far as the stability problem is widely concerned, it is worth pointing out that the state is not always already measured but may be estimated from the input and
the output of the process. State estimation for switched systems remains a challenging problem by reason of the combined discrete and continuous states (see Hocine et al. combined discrete and continuous states (see Hocine et al. systems remains a challenging problem by reason of the combined discrete and continuous states (see Hocine et al. (2005); Tian et al. (2009); Birouche et al. (2006); Arichi et al. (2015)). et al. (2015)). $\frac{1000}{\pi}$ (2000) ; Tian et al. (2000) ; Birouche et al. (2000) ; Arieni $\frac{1}{2010}$.

In practice, measurements are usually subject to noises. To compute robust estimates, interval observers assume that noises and disturbances are bounded without any stochas-noises and disturbances are bounded without any stochas-compute robust estimates, interval observers assume that noises and disturbances are bounded without any stochastic assumptions. This approach assumes that the measurement and process errors are unknown but bounded. the assumptions. This approach assumes that the measure-
ment and process errors are unknown but bounded. ment and process errors are unknown but bounded.

In the last decade, several works have investigated es-In the last decade, several works have investigated estems in an unknown but bounded error
(UBBE) framework for different classes of continuous sys- $\frac{1}{\sqrt{2L}}$ and $\frac{1}{\sqrt{2L}}$ and $\frac{1}{\sqrt{2L}}$ from $\frac{1}{\sqrt{2L}}$ from $\frac{1}{\sqrt{2L}}$ tems such as Linear Time-Invariant (LTI), Linear Time- $\frac{1}{2}$ Varying (LTV) and Linear Parameter-Varying (LPV) systems (see Ra¨ıssi et al. (2012); Mazenc and Bernard (2011); Videau et al.; Ra¨ıssi et al. (2010); Mazenc and Bernard tems (see Ra¨ıssi et al. (2012); Mazenc and Bernard (2011); Varying (LTV) and Linear Parameter-Varying (LPV) sys-Videau et al.; Raïssi et al. (2010); Mazenc and Bernard (2010) ; Gouzé et al. (2000) ; Lamouchi et al. (2016) ; Yousfi et al. $(2016b,a)$ and the references therein). (2016), Gouze et al. (2016 b,a) and the references therein). et al. (2016b,a) and the references therein).

The main restriction of interval observer theory is the cooperativity feature. Considering that most of systems are not cooperative, it has been shown in the literature that not cooperative, it has been shown in the heriature that any linear invariant-time system could be changed into any linear invariant-time system could be changed into
a cooperative form. The technique proposed in (Mazenc and Bernard (2011)) considers that it is possible to design and Bernard (2011)) considers that it is possible to design
an interval observer via a time-varying change of coordinates based on the Jordan representation. In (Raïssi et al. (2012)) an invariant change of coordinates is used to ensure the cooperativity property. the cooperativity property. $\frac{1}{2}$ (2012)) an invariant change of coordinates is used to ensure the cooperativity property.

The extension of interval observers to switched systems has not been fully considered in the literature. To the has not been fully considered in the literature. To the The extension of interval observers to switched systems has not been fully considered in the literature. To the
best of the authors knowledge, only a preliminary work has been developed in (He and Xie (2016, 2015)) with
the strong assumption that there exist gains L_q such the strong assumption that there exist gains L_q such that $(A_q - L_q C_q)$, $q \in \overline{1, N}$, are Metzler (the off-diagonal elements are nonnegative). Unfortunately, this assumption elements are nonnegative). is rarely verified. is rarely verified. $\sum_{i=1}^{n}$ to design and $\sum_{i=1}^{n}$ to design and $\sum_{i=1}^{n}$ a elements are nonnegative). Unfortunately, this assumption $\frac{1}{2}$ rately verified.

In this paper, the main contribution is to design an interval observer for switched linear systems subject to disturbances. The measurement noise and the state dis-disturbances. The measurement noise and the state dis-interval observer for switched linear systems subject to disturbances. The measurement noise and the state dis-

turbance are unknown but bounded with known bounds. State estimation conditions are given in terms of Linear Matrix Inequalities (LMIs). The suggested methodology allows one to overcome the strong limitations in (He and Xie (2016, 2015)). It will be shown that the constructive methodology can be applied for a large class of linear switched systems.

This paper is structured as follows. Some preliminaries are described in Section 2. Interval state estimation for switched linear systems is stated in Section 3. Section 4 is dedicated to show efficiency of the proposed method via a numerical example. Section 5 concludes the paper.

2. PRELIMINARIES

The sets of real and natural numbers are denoted by R and N respectively. $|x|$ is the elementwise absolute value of a vector $x \in \mathbb{R}^n$. The sequence of integers $1, ..., N$ is denoted by $\overline{1, N}$. E_p is a $(p \times 1)$ vector whose elements are equal to 1. I is the identity matrix of proper dimension. For a matrix $P = P^T$, $P \prec 0$ $(P \succ 0)$ means that the matrix $P \in \mathbb{R}^n$ is negative (positive) definite. Lower and upper bounds \underline{x} and \overline{x} of x satisfy $\underline{x} \leq x \leq \overline{x}$, where the relation \leq is interpreted elementwise for vectors and matrices. For a matrix $A \in \mathbb{R}^{m \times n}$, let $A^+ = \max\{0, A\}$ and $A^- = A^+ - A$.

Lemma 1. (Chebotarev et al. (2015)) Let $x \in \mathbb{R}^n$ be a vector satisfying $\underline{x} \leq x \leq \overline{x}$ and $A \in \mathbb{R}^{m \times n}$ be a constant matrix, then

$$
A^{+}\underline{x} - A^{-}\overline{x} \leq Ax \leq A^{+}\overline{x} - A^{-}\underline{x}.
$$
 (1)

A matrix $A = \{a_{ij}\}\in \mathbb{R}^{n \times n}$ is called Metzler if all its offdiagonal elements are nonnegative, i.e. $a_{ij} \geq 0$, $\forall i \neq j$.

Lemma 2. (Gouzé et al. (2000)) Let
\n
$$
\dot{x}(t) = Ax + u, \quad x(0) = x_0
$$
\nwhere A is a Metzler matrix and $u \ge 0$. If $x_0 \ge 0$, then

$$
x(t) \ge 0, \quad \forall t \ge 0. \tag{2}
$$

A continuous-time linear system $\dot{x}(t) = Ax$ is said to be cooperative if A is a Metzler matrix.

Lemma 3. (Jiang et al. (2002)) Let $\delta > 0$ be a scalar and $P \in \mathbb{R}^n$ be a symmetric positive definite matrix, then

$$
2x^T y \le \frac{1}{\delta} x^T P x + \delta y^T P^{-1} y \quad x, y \in \mathbb{R}^n. \tag{3}
$$

Consider the system described by:

$$
\begin{cases} \n\dot{x} = Ax + \phi(t) \\ \ny = Cx \n\end{cases} \n\tag{4}
$$

where ϕ is a continuous function and assume that there exist two known functions $\overline{\phi}$ and $\underline{\phi} : \mathbb{R} \to \mathbb{R}^n$ Lipschitz continuous such that $\phi(t) \leq \phi(t) \leq \overline{\phi}(t)$ for all $t \geq 0$.

Theorem 1. (Gouzé et al. (2000)) If there exists a gain K such that $(A - KC)$ is Metzler and if $\underline{x}_0 \leq x_0 \leq \overline{x}_0$, then the system

$$
\begin{cases} \n\dot{\overline{x}} = A\overline{x} + \overline{\phi} + K\left(y - C\overline{x}\right) \\
\underline{\dot{x}} = A\underline{x} + \underline{\phi} + K\left(y - C\underline{x}\right)\n\end{cases} \tag{5}
$$

is a framer for the system (4) such that

$$
\underline{x}(t) \leq x(t) \leq \overline{x}(t), \quad \forall t \geq 0
$$

Theorem 2. (Gouzé et al. (2000)) The system (5) is called an interval observer for system (4) if the lower $(x - x)$ and upper $(\overline{x} - x)$ estimation errors are asymptotically stable.

3. INTERVAL STATE ESTIMATION FOR SWITCHED LINEAR SYSTEMS

Consider a Switched Linear System (SLS) described by:

$$
\begin{cases} \n\dot{x}(t) = A_q x(t) + B_q u(t) + w(t) \\
y_m(t) = C_q x(t) + v(t)\n\end{cases}, q \in \overline{1, N}, N \in \mathbb{N} \tag{6}
$$

where $x \in \mathbb{R}^n$, $u \in \mathbb{R}^m$, $y_m \in \mathbb{R}^p$, $w \in \mathbb{R}^n$, $v \in \mathbb{R}^p$ are respectively the state vector, the input, the output, the disturbance and the measurement noise. A_q , B_q and C_q are constant matrices of proper dimensions. \hat{q} is the index of the active subsystem and N is the number of subsystems. The measurement noise and the state disturbance are assumed to be unknown but bounded with *a priori* known bounds such that

$$
|w(t)| \leq \overline{w}, \quad |v(t)| \leq \overline{V}E_p, \quad \forall t \geq 0 \tag{7}
$$

where $\overline{w} \in \mathbb{R}^n$ and \overline{V} is a scalar.

The aim is to derive two trajectories $x(t)$ and $\overline{x}(t)$ where $x(t) \leq x(t) \leq \overline{x}(t), \forall t \geq 0$, despite the disturbances, starting from the initial condition x_0 which is assumed to be bounded by two known bounds $x_0 \leq x_0 \leq \overline{x}_0$.

To design an interval observer for (6), a necessary condition is given in the following assumption.

Assumption 1. There exist gains L_q such that the matrices $(A_q - L_qC_q)$ are Metzler for $q \in \overline{1, N}$.

The matrices L_q $(q \in \overline{1,N})$ denote the observer gains associated with each subsystem q.

A candidate interval observer structure for the estimation of \overline{x} , \underline{x} is described by:

$$
\begin{cases}\n\dot{\overline{x}} = (A_q - L_q C_q) \overline{x} + B_q u + \overline{w} + L_q y_m + |L_q| \overline{V} E_p \\
\dot{\underline{x}} = (A_q - L_q C_q) \underline{x} + B_q u - \overline{w} + L_q y_m - |L_q| \overline{V} E_p\n\end{cases}
$$
\n(8)

Similarly to *Theorem* 1, the following theorem gives the conditions for achieving partially the desired design goal. *Theorem 3.* Consider the system described by (6). Let *Assumption 1* be satisfied. For any initial condition $x_0 \leq$

 $x_0 \leq \overline{x}_0$, if there exist observer gains L_q such that

$$
(A_q - L_q C_q)
$$
 are Metzler, $\forall q \in \overline{1, N}$ (9)

then the system (8) is a framer for the system (6) with $x(t) \leq x(t) \leq \overline{x}(t)$, $\forall t \geq 0$. (10)

Proof. Let $\overline{e}(t) = \overline{x} - x$ be the upper observation error and $\underline{e}(t) = x - \underline{x}$ be the lower observation error. Let us show that \overline{e} and \underline{e} are positive.

From (6) and (8) the dynamics of the interval estimation errors are given by:

$$
\dot{\overline{e}}(t) = \dot{\overline{x}} - \dot{x} = (A_q - L_q C_q) \overline{e}(t) + \overline{\Gamma}_q \tag{11}
$$

and
$$
\underline{\dot{e}}(t) = \dot{x} - \underline{\dot{x}} = (A_q - L_q C_q) \underline{e}(t) + \underline{\Gamma}_q
$$
 (12)

where

$$
\overline{\Gamma}_q = \overline{w} - w + L_q v + |L_q| \overline{V} E_p \tag{13}
$$

$$
\underline{\Gamma}_q = w + \overline{w} - L_q v + |L_q| V E_p \tag{14}
$$

As $|v(t)| \le \overline{V}E_p$, $\forall t \ge 0$ and by construction $(\overline{w} - w)$ and $(w + \overline{w})$ are positive then $\overline{\Gamma}_q$ and $\underline{\Gamma}_q$ are positive for all $q \in \overline{1, N}$. In addition, since L_q are computed to verify (9) and x_0 and \overline{x}_0 are chosen such that

$$
\begin{cases} \overline{e}(0) = \overline{x}_0 - x_0 \ge 0\\ \underline{e}(0) = x_0 - \underline{x}_0 \ge 0 \end{cases}
$$

then, according to *Lemma* 2, $\bar{e}(t)$ and $\underline{e}(t)$ are positive $\forall t \geq 0$. Thus, $\underline{x}(t) \leq \overline{x}(t) \leq \overline{x}(t)$ for all $t \geq 0$.

Remark 1. The framer (8) is initialized with the initial conditions \underline{x}_0 and \overline{x}_0 for the active subsystem $(q = 1)$. At the switching time instant, the output of the previous active subsystem $(q = i)$ is used to initialize (8) with the subsystem $(q = i + 1)$.

In addition, *Theorem* 3 ensures only the inclusion relation $x(t) \leq x(t) \leq \overline{x}(t)$. However, the errors $(x - \underline{x})$ and $(\overline{x} - x)$ are not guaranteed to be bounded.

For the stability analysis of (11) and (12), let us introduce the following lemma.

Lemma 4. (Liberzon and Morse (1999)) Let

$$
\dot{x}(t) = A_q x(t), \quad q \in \overline{1, N} \tag{15}
$$

The switched system (15) is globally asymptotically stable if there exists a matrix $S = \overline{S}^T > 0$ such that

 $\dot{V}(x) = x^T (A_q^T S + S A_q) x < 0, \quad q \in \overline{1, N}$ (16) where $V(x)$ is the common Lyapunov function given by:

$$
V\left(x\right) = x^T S x. \tag{17}
$$

Theorem 4. Let *Assumption 1* be satisfied. Given scalars $\delta_q > 0$, if there exists a symmetric positive definite matrix $S \in \mathbb{R}^n$ for all $q \in \overline{1, N}$ such that

$$
A_q^T S + S A_q - C_q^T W_q^T - W_q C_q + \alpha_q S \prec 0 \qquad (18)
$$

where $\alpha_q = \frac{2}{\delta_q}$ and $W_q = SL_q$, then the framer (8) is an asymptotically stable interval observer for (6).

Proof. As mentioned in *Theorem* 2, the global asymptotic stability of the interval observer is guaranteed by applying a common Lyapunov function to the estimation errors. Consider the following Lyapunov function:

$$
V(\overline{e}) = \overline{e}(t)^T S \overline{e}(t), \quad S \succ 0.
$$
 (19)

$$
\dot{V}(\overline{e}) = \overline{e}^T S \overline{e} + \overline{e}^T S \overline{e}
$$
\n
$$
= \overline{e}^T \left[\left(A_q - L_q C_q \right)^T S + S(A_q - L_q C_q) \right] \overline{e} - 2 \overline{e}^T S w
$$
\n
$$
+ 2 \overline{e}^T S L_q v + 2 \overline{e}^T S \overline{w} + 2 \overline{e}^T S |L_q| \overline{V} E_p \tag{20}
$$

$$
2c D_q e + 2c D_w + 2c D |L_q| V L_p
$$

om *Lemma 3* it is clear that

From Lemma 3, it is clear that

$$
\dot{V}\left(\overline{e}\right) \leq \overline{e}^T \mathcal{B}_1 \overline{e} + \mathcal{C}_1 \tag{21}
$$

where

$$
\mathcal{B}_1 = A_q{}^T S + S A_q - C_q{}^T W_q{}^T - W_q C_q + \frac{2}{\delta_q} S \quad (22)
$$

$$
C_1 = w^T \left[-\delta_q S \right] w + \overline{w}^T \left[\delta_q S \right] \overline{w} + v^T \left[\delta_q L_q^T S L_q \right] v + E_p^T \left[\delta_q \overline{V} L_q^T S L_q \overline{V} \right] E_p
$$
(23)

From (18), it is assumed that $\mathcal{B}_1 \prec 0$. In addition, the noises and disturbances are bounded it follows that C_1

is bounded. Therefore the error \bar{e} is bounded. The same arguments show that the error ϵ is also bounded.

The methodology described above, although simple, is not always constructive. Indeed, it is not always possible to find gains L_q such that *Assumption 1* is satisfied. Hence, the key point of the idea is to find a change of coordinates that transforms the observation errors into cooperative forms. The changes of coordinates proposed for instance in (Raïssi et al. (2012) ; Mazenc and Bernard (2011)) for continuous systems can be used to transform the matrices $(A_q - L_qC_q)$ into a Metzler form.

Let us assume that there exists a non singular transformation matrix P such that, with the new coordinates $z = Px$, the system (6) is transformed into the form

$$
\begin{cases}\n\dot{z} = PA_q P^{-1}z + PB_q u + Pw \\
y = C_q P^{-1}z + v\n\end{cases}
$$
\n(24)

A Luenberger based candidate observer for the system (24) can be written in the new coordinates z as:

$$
\begin{cases}\n\dot{\overline{z}} = P(A_q - L_q C_q) P^{-1} \overline{z} + P B_q u + P^+ \overline{w} - P^- \underline{w} \\
+ P L_q y_m + |P L_q| \overline{V} E_p \\
\dot{z} = P(A_q - L_q C_q) P^{-1} \underline{z} + P B_q u + P^+ \underline{w} - P^- \overline{w} \\
+ P L_q y_m - |P L_q| \overline{V} E_p \\
\forall q \in \overline{1, N}\n\end{cases}
$$
\n(25)

where

$$
\overline{z}(0) = P^{+}\overline{x}_{0} - P^{-}\underline{x}_{0}
$$
 (26)

$$
\underline{z}(0) = P^+ \underline{x}_0 - P^- \overline{x}_0 \tag{27}
$$

P is the solution of the Sylvester equation given by

$$
PA_q - R_qP = Q_qC_q, \quad Q_q = PL_q \tag{28}
$$

and

$$
R_q = P\left(A_q - L_q C_q\right)P^{-1}
$$

Let $\overline{e_z}(t) = \overline{z} - z$ be the upper observation error and $\underline{e}_z(t) = z - \underline{z}$ be the lower one.

From systems (24) and (25), the dynamics of the interval estimation errors are given by:

$$
\dot{\overline{e}}_z(t) = \dot{\overline{z}} - \dot{z} = \left[\left(P^+ \overline{w} - P^- \underline{w} \right) - P w \right] + |PL_q| \overline{V} E_p + P \left(A_q - L_q C_q \right) P^{-1} \overline{e_z} + P L_q v = R_q \overline{e_z} + \overline{\Upsilon}_q (29)
$$

$$
\underline{\dot{e}}_z(t) = \dot{z} - \underline{\dot{z}} = [Pw - (P^+\underline{w} - P^-\overline{w})] + |PL_q|\overline{V}E_p
$$

+ $P(A_q - L_qC_q)P^{-1}\underline{e}_z - PL_qv = R_q\underline{e}_z + \underline{\Upsilon}_q(30)$

where

$$
\overline{\Upsilon}_q = \left[\left(P^+ \overline{w} - P^- \underline{w} \right) - P w \right] + \left| PL_q \right| \overline{V} E_p + PL_q v \quad (31)
$$

$$
\underline{\Upsilon}_q = \left[P w - \left(P^+ \underline{w} - P^- \overline{w} \right) \right] + \left| PL_q \right| \overline{V} E_p - PL_q v \quad (32)
$$

Similarly to the proof of $Theorem 4$, the asymptotic stability of the observer (25) is ensured by applying a common Lyapunov function to the observation errors as follows:

$$
V\left(\overline{e}_z\right) = \overline{e}_z^T M \overline{e}_z, \ V\left(\underline{e}_z\right) = \underline{e}_z^T M \underline{e}_z \tag{33}
$$

In the following, only the dynamics of the upper observation error are considered. The derivative of $V(\overline{e}_z)$ is:

 \blacksquare

 \blacksquare

$$
\dot{V}(\overline{e}_z) = \dot{\overline{e}}_z^T M \overline{e}_z + \overline{e}_z^T M \dot{\overline{e}}_z \leq \overline{e}_z^T \mathcal{B}_2 \overline{e}_z + \mathcal{C}_3 \tag{34}
$$

where

$$
\mathcal{C}_3 = \overline{w}^T \left[\delta_q P^{+^T} M P^+ \right] \overline{w} - \underline{w}^T \left[\delta_q P^{-^T} M P^- \right] \underline{w}
$$

$$
- w^T \left[\delta_q P^T M P \right] w + v^T \left[\delta_q L_q^T P^T M P L_q \right] v
$$

$$
+ E_p^T \left[\delta_q \overline{V} |PL_q|^T M |PL_q| \overline{V} \right] E_p \tag{35}
$$

and

$$
\mathcal{B}_2 = P(A_q - L_q C_q) P^{-1} M + M \left[P(A_q - L_q C_q) P^{-1} \right]^T
$$

+
$$
\frac{1}{\delta_q} M = P^{-1} A_q^T P^T M + M P A_q P^{-1} + \frac{1}{\delta_q} M
$$

-
$$
P^{-1} C_q^T L_q^T P^T M - M P L_q C_q P^{-1} \prec 0
$$
 (36)

The existence of a common transformation matrix P such that $P(A_q - L_qC_q)P^{-1}$ for all $q \in \overline{1,N}$ are Metzler is difficult since (36) is a nonlinear inequality. Therefore, the stability of the observer (25) can not be easily ensured.

However, it is rare, even impossible, to determine a non singular transformation matrix P to transform the system (6) into a cooperative form such that $P(A_q - L_qC_q)P^{-1}$ $(q \in \overline{1, N})$ are Metzler. As a solution to this problem, a second method is proposed. The main idea consists in redesigning two conventional observers in the original base " x ". Then, stability conditions will be given in terms of LMIs by applying a common Lyapunov function to the estimation errors. Consider the SLS (6) and two point observers described by

$$
\begin{cases}\n\dot{\hat{x}}^{+} = (A_{q} - L_{q}C_{q})\,\hat{x}^{+} + B_{q}u + P_{q}^{-1}\left(P_{q}^{+}\overline{w} + P_{q}^{-}\overline{w}\right) \\
+ L_{q}y_{m} + P_{q}^{-1}\left|P_{q}L_{q}\right|\overline{V}E_{p} \\
\dot{\hat{x}}^{-} = (A_{q} - L_{q}C_{q})\,\hat{x}^{-} + B_{q}u + P_{q}^{-1}\left(-P_{q}^{+}\overline{w} - P_{q}^{-}\overline{w}\right) \\
+ L_{q}y_{m} - P_{q}^{-1}\left|P_{q}L_{q}\right|\overline{V}E_{p}\n\end{cases}
$$
\n(37)

with $P_q, q \in \mathcal{I}$, are chosen as in the following theorem and

$$
\begin{cases}\n\hat{x}_0^+ = Q_q \left(P_q^+ \overline{x}_0 - P_q^- \underline{x}_0 \right) \\
\hat{x}_0^- = Q_q \left(P_q^+ \underline{x}_0 - P_q^- \overline{x}_0 \right)\n\end{cases} \tag{38}
$$

with

$$
Q_q = P_q^{\ -1}
$$

The observer (37) is not an interval observer for (6) and its structure is similar to the one proposed in Dinh et al. (2014) for the case of non switched systems. However, the estimates computed by (37) are used in *Theorem 5* to deduce an interval estimation.

Theorem 5. Consider matrices P_q ($q \in \mathcal{I}$) such that $F_q =$ P_q ($A_q - L_q C_q$) P_q^{-1} are Metzler. If the initial condition x_0 verifies $x_0(t) \leq x_0(t) \leq \overline{x}_0(t)$, then an interval estimation for (6) is given by:

$$
\begin{cases} \underline{x} = Q_q^+ P_q \hat{x}^- - Q_q^- P_q \hat{x}^+\\ \overline{x} = Q_q^+ P_q \hat{x}^+ - Q_q^- P_q \hat{x}^- \end{cases}
$$
\n(39)

satisfying

$$
\underline{x}(t) \le x(t) \le \overline{x}(t) \tag{40}
$$

In addition, if there exists a symmetric definite positive matrix M such that

$$
A_q^T M + M A_q - C_q^T W_q^T - W_q C_q + \sigma_q M \prec 0 \qquad (41)
$$

where

$$
\sigma_q = \frac{3}{\delta_q} \text{ and } W_q = ML_q
$$

then (37) is asymptotically stable and \overline{x} , \underline{x} are bounded.

Proof. Consider the errors $E_q^+ = P_q \hat{x}^+ - P_q x$ and $E_q^ P_qx - P_q\hat{x}^-$.

Let us show that $\overline{x} - x \geq 0$ and $x - x \geq 0$ where \overline{x} and x are computed by (39).

From (6) and (37) the dynamics of the errors E_q^+ and $E_q^$ are given by:

$$
\dot{E}_{q}^{+} = P_{q}\dot{\hat{x}}^{+} - P_{q}\dot{x} = P_{q} (A_{q} - L_{q}C_{q}) \hat{x}^{+} + |P_{q}L_{q}|\overline{V}E_{p} \n- P_{q} (A_{q} - L_{q}C_{q}) x + [(P_{q}^{+}\overline{w} + P_{q}^{-}\overline{w}) - P_{q}w] + P_{q}L_{q}v \n= P_{q} (A_{q} - L_{q}C_{q}) P_{q}^{-1} (P_{q}\hat{x}^{+} - P_{q}x) + \gamma_{q}^{+} \n= F_{q}E_{q}^{+} + \gamma_{q}^{+}
$$
\n(42)

where

 $\gamma_q^+ = \left[\left(P_q^+\overline{w} + P_q^-\overline{w}\right) - P_qw\right] + P_qL_qv + |P_qL_q|\,\overline{V}E_p$ Similarly to E_q^+ , the dynamics of E_q^- are given by:

$$
\dot{E}_q^- = P_q \dot{x} - P_q \dot{\hat{x}}^- = R_q E_q^- + \gamma_q^- \tag{43}
$$

where

 $\gamma_q^- = \left[P_q w - \left(-P_q^+ \overline{w} - P_q^- \overline{w} \right) \right] + |P_q L_q| \overline{V} E_p - P_q L_q v$ According to Lemma 1 we have

$$
-P_q^+\overline w-P_q^-\overline w\leq P_q w\leq P_q^+\overline w+P_q^-\overline w
$$

Since $P_q (A_q - L_q C_q) P_q^{-1}$ are assumed to be Metzler, and by construction γ_q^+ and γ_q^- are positive for all $t \geq 0$. Then, if \overline{x}_0 and \underline{x}_0 are chosen such that $E_q^+(0)$ and $E_q^-(0)$ are positive, then the errors $E_q^+(t)$ and $E_q^-(t)$ stay positive $\forall t \geq 0$ such that Pqxˆ[−] ≤ Pqx ≤ Pqxˆ⁺

$$
P_q x \le P_q x \le P_q x
$$

As $Q_q = P_q^{-1}$, then $\underline{x} \le x \le \overline{x}$ where

$$
\begin{cases} \underline{x} = Q_q + P_q \hat{x} - Q_q - P_q \hat{x} + \overline{x} \\ \overline{x} = Q_q + P_q \hat{x} + Q_q - P_q \hat{x} \end{cases}
$$

For the stability analysis, Let us now show that E_q^+ and E_q^- are asymptotically stable or simply show that $(\hat{x}^{\vec{+}} - x)$ and $(x - \hat{x}^{-})$ are asymptotically stable.

Let $e^+ = (\hat{x}^+ - x)$ and $e^- = (x - \hat{x}^-)$ be the observation errors and consider the following Lyapunov function:

$$
V\left(e^{+}\right) = e^{+T}Me^{+} \tag{44}
$$

where M is a symmetric positive definite matrix. As in proof of $Theorem 4$, the derivative of the Lyapunov function can be given as follows:

$$
\dot{V}(e^{+}) = e^{+T}Me^{+} + e^{+T}M\dot{e}^{+}
$$
\n
$$
= e^{+T} \left[\left(A_{q} - L_{q}C_{q} \right)^{T} M + M \left(A_{q} - L_{q}C_{q} \right) \right] e^{+}
$$
\n
$$
+ 2e^{+T}MP_{q}^{-1} \left(P_{q}^{+}\overline{w} + P_{q}^{-}\overline{w} \right) - 2e^{+T}Mw
$$
\n
$$
+ 2e^{+T}ML_{q}v + 2e^{+T}M|L_{q}|\overline{V}E_{p} \tag{45}
$$

According to Lemma 3, we have

where

$$
\dot{V}\left(e^{+}\right) \leq e^{+T} \mathcal{B}_3 e^{+} + \mathcal{C}_5
$$

$$
\mathcal{B}_3 = (A_q - L_q C_q)^T M + M (A_q - L_q C_q) + \frac{3}{\delta_q} M
$$

= $A_q^T M + M A_q - C_q^T W_q^T - W_q C_q + \frac{3}{\delta_q} M$ (46)

$$
\mathcal{C}_{5} = \overline{w}^{T} \left[\delta_{q} P_{q}^{+T} P_{q}^{-1} M P_{q}^{-1} P_{q}^{+} \right] \overline{w} + v^{T} \left[\delta_{q} L_{q}^{T} M L_{q} \right] v
$$

$$
+ \overline{w}^{T} \left[\delta_{q} P_{q}^{-T} P_{q}^{-1} M P_{q}^{-1} P_{q}^{-} \right] \overline{w} - w^{T} \left[\delta_{q} M \right] w
$$

$$
+ E_{p}^{T} \left[\delta_{q} \overline{V} |L_{q}|^{T} M |L_{q} | \overline{V} \right] E_{p}
$$
(47)

The noise v and disturbance w are bounded, it follows that \mathcal{C}_5 is bounded. Therefore, if $\mathcal{B}_3 \prec 0$, the observation error e^+ is bounded. The same arguments allow one to show that the observation error e^- is also bounded. In addition, since P_q and Q_q are bounded for all $(q \in \overline{1,N})$ then E_q^+ and E_q^- are bounded.

In the next section, the performance of the suggested method is shown through a numerical example.

4. NUMERICAL EXAMPLE

Consider the linear switched system subject to disturbances described by:

$$
\begin{cases}\n\dot{x}(t) = A_q x(t) + B_q u(t) + w(t) \\
y(t) = C_q x(t) + v(t)\n\end{cases}, \forall q \in \overline{1,3} \tag{48}
$$

where

$$
A_1 = \begin{bmatrix} -1.5 & 0.262 \\ 0 & -1 \end{bmatrix}, A_2 = \begin{bmatrix} -0.5 & 2 \\ 0 & -1 \end{bmatrix}
$$

$$
A_3 = \begin{bmatrix} -0.6 & 1.5 \\ 0 & -1 \end{bmatrix}
$$

$$
B_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, B_2 = \begin{bmatrix} 1 \\ 0 \end{bmatrix}, B_3 = \begin{bmatrix} 1 \\ 1 \end{bmatrix}
$$

$$
C_1 = \begin{bmatrix} 1 & 0 \end{bmatrix}, C_2 = \begin{bmatrix} 1 & 1 & 1 \end{bmatrix}, C_3 = \begin{bmatrix} 1 & 1.5 \end{bmatrix}
$$

 $w(t)$ and $v(t)$ are uniformly distributed bounded signals such that $-\overline{w} \leq w(t) \leq \overline{w}$ with $\overline{w} = [0.03 \ 0.03]^T$ and $-V E_p \le v(t) \le VE_p$ with $V = 0.3$.

To verify the cooperativity property, a transformation of coordinates must be used such that $P(A_q - L_qC_q)P^{-1}$ are Metzler. However, it is not possible to compute this common transformation matrix P. Hence, non singular transformation matrices P_q are calculated. Consequently, a conventional observer (37) is constructed for the system (48).

Now, using the Matlab LMI toolbox, one can solve the LMI defined by (41). One feasible solution is given by:

$$
L_1 = \begin{bmatrix} -0.6555 \\ -0.1011 \end{bmatrix}, \ L_2 = \begin{bmatrix} 0.9397 \\ -0.2497 \end{bmatrix}, \ L_3 = \begin{bmatrix} 0.5734 \\ -0.2512 \end{bmatrix}
$$

$$
M = \begin{bmatrix} 81.6804 & 21.7133 \\ 21.7133 & 55.5904 \end{bmatrix}, \ \delta_q = 84.5569, \ \forall q \in \overline{1,3}
$$

Note that P_q are computed to verify that P_q $(A_q - L_qC_q) P_q^{-1}$ are Metzler for all $q \in \overline{1,3}$ and given by:

$$
P_1 = \begin{bmatrix} 0.2845 & -0.7262 \\ 0.4770 & 0.4841 \end{bmatrix}, \ P_2 = \begin{bmatrix} 0.7776 & -0.8577 \\ 0.1483 & 0.5718 \end{bmatrix}
$$

$$
P_3 = \begin{bmatrix} 0.7829 & -0.6580 \\ 0.2171 & 0.6580 \end{bmatrix}
$$

All conditions of *Theorem 5* are satisfied; it follows that the system (37) is asymptotically stable verifying

$$
\underline{x}(t) \le x(t) \le \overline{x}(t) \ \forall t \ge 0, \ \forall q \in 1, 3.
$$

 $\underline{x} = Q_q^{\dagger} P_q \hat{x}^{-} - Q_q^{-} P_q \hat{x}^{+}$

with

and

 \blacksquare

$$
\overline{x} = Q_q^{} P_q \hat{x}^+ - Q_q^{} P_q \hat{x}^-
$$
 where $Q_q = P_q^{-1}$ are given by:

$$
Q_1 = \begin{bmatrix} 1.0000 & 1.5000 \\ -0.9853 & 0.5876 \end{bmatrix}, \ Q_2 = \begin{bmatrix} 1.0000 & 1.5000 \\ -0.2593 & 1.3599 \end{bmatrix}
$$

$$
Q_3 = \begin{bmatrix} 1.0000 & 1.0000 \\ -0.3299 & 1.1899 \end{bmatrix}
$$

The results of simulation of the obtained observer are depicted in Fig. 1(a) for both coordinates where solid lines present the state and dashed lines present the estimated bounds.

The switching between the three subsystems is governed by the switching signal plotted in Fig. 1(b).

The initial state x_0 is assumed to be bounded such that

$$
\underline{x}_0 \le x_0 \le \overline{x}_0
$$

where $\overline{x}_0 = [1.5 \ 1.5]^T$, $\underline{x}_0 = [-1.5 \ -1.5]^T$

(a) State and estimated bounds

Fig. 1. Interval state estimation for the switched system with disturbances.

The results show that, despite the disturbances, the state is always inside the upper and the lower trajectories. The interval observer has exhibited approved stability properties. The inclusion

$$
\underline{x}(t) \le x(t) \le \overline{x}(t), \text{ for all } t \ge t_0
$$

is always verified.

As shown in Fig. 1(a), the interval is quite large at the beginning, although its width decreases despite the uncertainties on the measurements. Finally, the interval observer remains stable despite the switching instants.

5. CONCLUSION

This paper investigates state estimation for switched linear systems subject to disturbances. An interval observer is designed under some transformations where two conventional observers are reformulated in the base " x ". The stability and the cooperativity conditions are represented in terms of LMIs. Effectiveness of the proposed methodology is illustrated through a numerical example. In this work, the switching instants are assumed to be known nonetheless it is not the case for the most of switched systems. This concern will be the subject of further contributions.

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