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1 ILD Silicon Tungsten Electromagnetic Calorimeter 2 First Full Scale Electronic Prototype

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9 *Introduction.* The SiW-Ecal [1] is a sampling electromagnetic calorimeter devel-
10 oped for the ILD detector [2] of the future International Linear Collider (ILC).
11 Its detector elements are called slabs and are based on high-resistivity silicon
12 diodes of several hundred microns thickness, divided in pixels of $5 \times 5 \text{ mm}^2$,
13 alternated with tungsten absorbers. The previous prototypes tested a tower of
14 “short slabs” featuring an active region of $18 \times 18 \text{ cm}^2$ for each single ASU, and
15 fully described in [3]. The long slab is a new prototype that has been designed
16 to demonstrate that it is electronically viable to operate a long electronic board
17 chain. It consists of 8 electronic front-end boards (named ASU for Active Sensor
18 Unit), assembled together to create a $144 \times 18 \text{ cm}^2$ detector which is the typical
19 size of the slabs in the ILD barrel (Figure 1).

20 For this prototype, we have relaxed the geometrical constraints imposed by
21 the ILD to focus on the electronics and the performance of the detector along
22 the length. Each board has been equipped with a $320 \mu\text{m}$ thick small sensor
23 ($2 \times 2 \text{ cm}^2$ for 4×4 pixels) produced by the Hamamatsu Photonics company.

24 *Mechanics.* For use in beam tests, a mechanical structure of 3 meters long,
25 shown in Figure 2, has been built, allowing to support the slab and to incline
26 it in the beam with a precision of 1° . It ensures the mechanical rigidity (max
27 1 mm of bend over a 3 m length) to ensure reliable operation in all conditions.



Figure 1: Assembled long slab with 8 ASUs.

28 Shielding is added to avoid electro-magnetic induced noise. The directional
29 blockable wheels allow placing the detector in the beam easily.

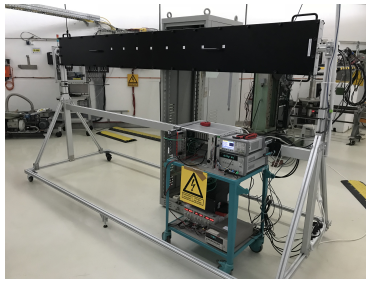


Figure 2: Long slab setup at DESY

30 *Electronics.* The building block of the detector is the ASU. It is a multi-layered
31 PCB with two very different external sides. The lower side is covered with
32 pads with the size of the silicon pixels. The silicon sensors are glued on this
33 side with a conductive glue. On the upper side sits the reading ASICs, named
34 Skiroc 2 and designed by OMEGA group [4]. The ASUs are chainable to obtain
35 long slabs with an almost continuous active surface. At the end of the chain,
36 a front-end board reads the ASICs and transmits data to the Data Acquisition

37 System (DAQ) described in [5]. The long slab prototype is the first attempt
38 to chain the ASUs to test their communication and performance. For testing
39 purpose, flat cables have been chosen for data and clocks to ease dismounting
40 without degrading transmission performance. The impedances and adaptive
41 systems (called drivers) have been tuned for every differential line by termination
42 resistors. For bias voltage, shielded cables has been used with lemo connectors,
43 allowing to reduce the electromagnetic noise and ease the connection.

44 The length of the prototype induces difficulties for clock and signal propa-
45 gation and data integrity. In particular, reflections appear on the clock lines.
46 Indeed, the clock line has a comb shape (in order to be evenly distributed), and
47 the signal bounces in the stubs, creating reflection. As a consequence, the clock
48 signal is not square but include glitches in the transition region. This creates
49 extra clock beats and does not allow to transfer the configuration bitstream
50 correctly to the ASICs. This problem was solved by adjusting the level of the
51 glitches by adding an RC filter in front. The filter parameters were optimized by
52 simulating an isolated clock line using the Cadence Sigrity tool [6], integrating
53 the PCB electrical specification and the driver characteristics. Even if the RC
54 filter permits the use of the actual prototype, a new design of clock lines has to
55 be developed in future versions of ASU to avoid the comb shape, as it has been
56 done for data lines.

57 Another difficulty induced by the length of the prototype is the increased
58 level of noise on the sensor's bias voltage. The SiW-Ecal readout is very sensitive
59 to this noise which must be reduced to the minimum. To achieve this goal, we
60 insert RC filters on the bias line between the ASUs. We use third order filters,
61 adapted to reduce high frequency noises above 100 kHz. In the final version,
62 for compacity, it is planned to use a Kapton sheet instead of cables; a change
63 of the design will be required to include these RC filters.

64 Finally, it has been observed that the bandwidth of the DAQ is not entirely
65 sufficient when the long slab has all its memories full. For example, in case of a
66 pedestal calibration, it is necessary to read all channels of all chips at the same
67 time. In the future, we need to use a parallel reading of partitions of ASICs

68 inside the front-end board.

69 *Testing.* The long slab performance has been tested with cosmics, radioactive
70 sources and with 3 GeV electrons in the beam tests at DESY, Hamburg. In
71 particular, radioactive sources have been used at every step of mounting a new
72 ASU to test the behaviour of the new board and also to check the noise induced
73 by the card on the others. This methodology allowed us to detect the problems
74 before they mingle with another one.

75 During the test beam at DESY, we have accumulated data for both normal
76 and inclined incidence of the beam to the silicon sensor surface. The positioning
77 of the mechanical structure was performed with a laser to ensure full exposure of
78 every pixel. We took data in all the pixels at three different angles of incidence
79 to the silicon sensor surface (0, 45 and 60 degrees).

80 *MIP fit.* The Minimum Ionizing Particle (MIP) energy deposition is determined
81 by fitting the pedestal subtracted analog-to-digital converter (ADC) spectra.
82 This pedestal varies among the ASICs memories and channels and thus has to
83 be evaluated for every memory associated with every channel. Once the pedestal
84 subtracted, we fit the histogram of the energy deposition in every pixel. This
85 is fitted by a sum of two Landau distributions convolved with a Gaussian. The
86 first Landau distribution corresponds to a single MIP hitting the pixel during
87 the integration time. The second Landau distribution models 2 MIPs (pile-up).
88 The contribution of a third Landau distribution can be neglected.

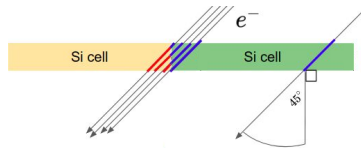


Figure 3: When the prototype is inclined, the incident particle can cross two pixels, modifying the energy deposition spectrum.

89 As shown in Figure 3, with the inclined beam, the particle can sometimes
90 go across two pixels. For a sufficiently uniform beam and in the simplest model

91 when electron-hole pairs in the silicon always drift perpendicularly to its surface,
 92 the one-pixel signal is reduced in this case such that it has equal probabilities
 93 of being anywhere between zero and the total energy deposition in the silicon
 94 shared by two pixels. The energy deposited distribution is then modified. In
 95 particular, it starts not from zero on the left but from a plateau which re-
 96 mains flat until the point where the original Landau distribution starts growing
 97 significantly. A simulation of this situation has been performed with Geant4,
 98 presented in Figure 4, showing clearly the plateau on the left of the energy
 99 deposition histogram.

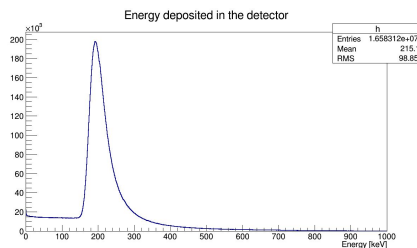


Figure 4: On the GEANT4 simulation of 2 regions, the two-pixels energy sharing adds a low energy plateau to the single-pixel energy deposition histogram.

100 This plateau and the left part of the Landau distribution can be modelled by
 101 assuming a uniform energy sharing of the deposited energy between two pixels.
 102 Let's denote dE/dx Landau distribution by $L(x)$ and the probability that the
 103 particle passes through two pixels by c . Then, the normalized distribution of
 104 energies deposited in one pixel is:

$$f(x) = (1 - c) \frac{L(x)}{\int_0^{+\infty} L(t) dt} + c \frac{\int_x^{+\infty} \frac{L(t)}{t} dt}{\int_0^{+\infty} \int_x^{+\infty} \frac{L(t)}{t} dt dx} \quad (1)$$

105 where the second term describes the energy sharing. Namely, the total
 106 energy $L(t)$ in the numerator $\int_x^{+\infty} L(t)/t dt$ is uniformly distributed in the
 107 range $[0...L(t)]$ and creates the density $L(t)/t$ which is then integrated over $L(t)$
 108 for all $t \geq x$. This plateau is ideally suited for measuring the position and the
 109 shape of the trigger threshold, modelled by an error function denoted in the

110 following by erf. The full fit can be performed with the following function:

$$\frac{\text{erf}(x) + 1}{2} \int_{-\infty}^{+\infty} f(t) \cdot \text{Gauss}(x - t, \sigma) dt \quad (2)$$

111 where $\text{Gauss}(x, \sigma)$ denotes the Gaussian with the center x and the sigma σ .

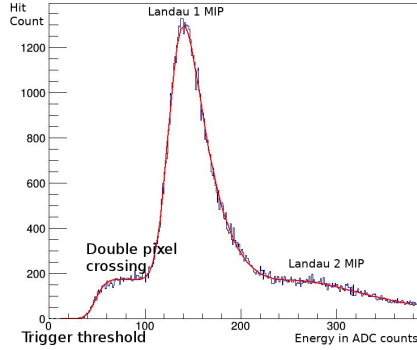


Figure 5: Full fit of the pedestal-subtracted signal histogram : two Landau distributions representing 1 and 2 MIPs respectively are convolved with a Gaussian. On the left, the pixel-sharing plateau is truncated by an erf function.

112 The full fit is shown in Figure 5. The Most Probable Value (MPV) of the first
 113 Landau distribution gives the value of the MIP in ADC counts. The inflexion
 114 point of the erf function provides the value of the trigger threshold. Figure
 115 5, therefore clearly demonstrates that the threshold was set to sufficiently low
 116 value far below the MIP signal. With these two pieces of information, we can
 117 infer the value of this threshold in term of a MIP percentage which is very
 118 valuable to perform accurate calibration of the system.

119 *MIP uniformity along the slab.* With this MIP fit function, we can get accurate
 120 values of the MIP on all the sensors, depending on the angle. Figure 6 sum-
 121 marises the extracted MIP values for the 8 ASUs for the three considered angles.
 122 A $1/\cos(\alpha)$ relates the three series as expected from the proportionality of the
 123 MIP value and the thickness of silicon crossed by the particle. On Figure 6, the
 124 MIP values are corrected. Points corresponding to the last ASU under 45 and
 125 60-degree angle are missing because the corresponding sensor was inaccessible
 126 due to the size of the beam room.

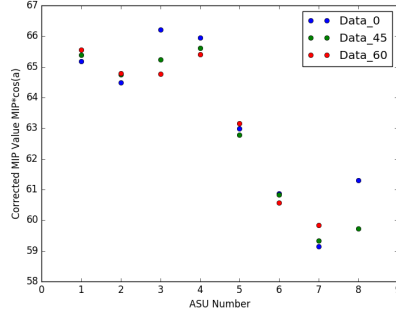


Figure 6: MIP values corrected for incidence angle for all the ASUs of the prototype, at the incidence angles (0, 45 and 60 degrees)

127 In an ideal prototype, the three series should be completely flat but they are
 128 not. They show a strange shape including a slope of about 10 % in the value
 129 of the MIP. This kind of variations may come from different sources. We have
 130 investigated the ASIC's power supply voltage along the length of the prototype
 131 and measured a 150 mV linear decrease over the total length. This might explain
 132 a part of the variation. We have also investigated a reference voltage, so-called
 133 bandgap, an internal reference inside the readout ASIC which can change the
 134 amplification and thus the MIP value. Figure 7 shows the variation of this
 135 reference on the different ASU.

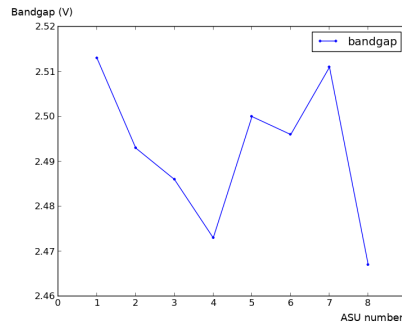


Figure 7: The measured bandgap of the 8 ASIC of the long slab.

136 A more detailed measurement on all 128 ASICs has shown that this bandgap
 137 varies from ASIC to ASIC. There is no systematic effect and the distribution

138 of the bandgap over the ASICs seems to follow a normal distribution with a
 139 standard deviation of 19.2 mV and a maximum peak-to-peak difference of 200
 140 mV.

141 Based on the measured ASIC power and bandgap voltages, an attempt has
 142 been made to reproduce the MIP variation across ASUs using a simple model
 143 of the form:

$$\text{MIP}(\text{ASU}) = a * \text{ASU} + b - c * \text{bandgap}(\text{ASU}). \quad (3)$$

144 The best fit on 0 degree incidence MIP values is shown in Figure 8. Further
 145 studies are needed to understand fully the observed variation of the MIP signal
 146 across ASUs.

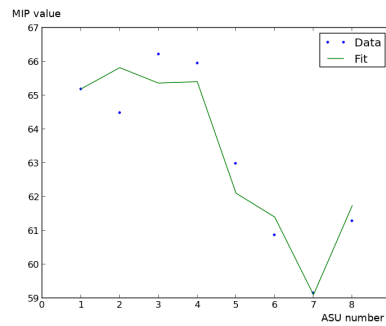


Figure 8: Fit of the MIP values along the length of the prototype (in ASU)

147 *Conclusion and Perspectives.* The electronic long slab prototype has proved
 148 that the ASUs are almost suitable for chained operation. After solving the
 149 different electronics problems, we found out that the long slab has almost the
 150 same performance at low energy (MIP) as the short slabs tested previously [3].

151 Nevertheless, many improvements could be envisaged for the next version of
 152 the ASU:

- 153 • Clock line adaptation: the shape of this line has to be optimised to avoid
 154 reflections and limit the possibility of bitstream corruption

- 155 • RC filtering of the sensor’s bias voltage: the current design uses a long
156 Kapton covering all ASUs. It should be replaced by a single Kapton per
157 ASU design to integrate a low-noise filter.
- 158 • The power supply of the ASICs has to be stabilised along the length of
159 the slab, for example by equalising the length of all the power lines.
- 160 • The analysis has proved that the bandgap disparity between ASICs is
161 not neglectable. Improvement requires the refactoring of the ASIC design
162 (which has already been partly done with the new version Skiroc 2a). In
163 case the bandgap spread between ASICs remains too large, they will have
164 to be individually measured and compensated by software.
- 165 • The sequential reading of the ASIC limits the bandwidth and can cause
166 problems during pedestal or charge injection calibrations. A solution with
167 readout partitions has to be considered.

168 The successful operation of a long slab is a milestone toward a proof of
169 the feasibility of the ILD SiW-Ecal. A prototype including all the previously
170 mentioned improvements, the use of broader and thicker sensors and including
171 all the mechanical constraints of the ILD experiment still needs to be built and
172 tested with high energy electrons to fully assess the design.

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