

ISS Lyapunov functions for cascade switched systems and sampled-data control

Guangxue Zhang, Aneel Tanwani

▶ To cite this version:

Guangxue Zhang, Aneel Tanwani. ISS Lyapunov functions for cascade switched systems and sampled-data control. Automatica, 2019, 105, pp.216-227. 10.1016/j.automatica.2019.03.028. hal-02144470

HAL Id: hal-02144470 https://hal.science/hal-02144470v1

Submitted on 30 May 2019

HAL is a multi-disciplinary open access archive for the deposit and dissemination of scientific research documents, whether they are published or not. The documents may come from teaching and research institutions in France or abroad, or from public or private research centers. L'archive ouverte pluridisciplinaire **HAL**, est destinée au dépôt et à la diffusion de documents scientifiques de niveau recherche, publiés ou non, émanant des établissements d'enseignement et de recherche français ou étrangers, des laboratoires publics ou privés.

ISS Lyapunov Functions for Cascade Switched Systems and Event-Triggered Output Feedback Stabilization

GuangXue Zhang^a, Aneel Tanwani^{b,*}

^aDepartment of Aerospace Engineering, University of California at Irvine, USA

Abstract

Input-to-state stability (ISS) of switched systems is studied where the individual subsystems are connected in a serial cascade configuration, and the states are allowed to reset at switching times. An ISS Lyapunov function is associated to each of the two blocks connected in cascade, and these functions are used as building blocks for constructing ISS Lyapunov function for the interconnected system. The derivative of individual Lyapunov functions may be bounded by nonlinear decay functions, and the growth in the value of Lyapunov function at switching times may also be a nonlinear function of the value of other Lyapunov functions. The stability of overall hybrid system is analyzed by constructing a newly constructed ISS-Lyapunov function and deriving lower bounds on the average dwell-time. The particular case of linear subsystems and quadratic Lyapunov functions is also studied. The tools are also used for studying the observer-based feedback stabilization of a nonlinear switched system with event-based sampling of the output and control inputs. We design dynamic sampling algorithms based on the proposed Lyapunov functions and analyze the stability of the resulting closed-loop system. This material presented here is a reprint of (Zhang and Tanwani, 2019).

Keywords: Switched systems, input-to-state stability, cascade connection, multiple Lyapunov functions, average dwell-time, output feedback, event-based control

1. Introduction

Switched systems, or in general, hybrid dynamical systems provide a framework for modeling a large class of physical phenomenon and engineering systems which combine discrete and continuous dynamics. Due to their wide utility, such systems have been extensively studied in the control community over the past two decades; see the books by Liberzon (2003) and Goebel et al. (2012) for comprehensive overview.

This article addresses a robust stability problem for systems with switching vector fields and jump maps. In our setup, each subsystem has a two-stage serial cascade structure where the output of first block acts as an input to the second block, and the disturbances we consider are an exogenous input to the first block, see Figure 1. By proposing a novel construction for multiple Lyapunov functions for such configurations, we analyze the stability of the interconnected switched system by deriving lower bounds on average dwell-time between switching times. It is seen that

such a configuration arises in the context of output feedback stabilization of switched systems with known switching signal where the inputs and outputs are time-sampled. The theoretical tools developed in the earlier part of this paper are then used to design sampling algorithms and analyze stability of the resulting sampled-data system. A preliminary version of the sampled-data problem, studied in the later part of this paper, has appeared in (Zhang and Tanwani, 2018).

Stability of switched systems has been a topic of interest in control community for past two decades now. Depending on the class of switching signals, or the assumptions imposed on the continuous dynamics, different approaches have been adopted in the literature to study the convergence of the state trajectories. The book (Liberzon, 2003) provides an overview on this subject. For our purposes, the approach based on slow switching is more relevant. In this direction, the pioneering contribution comes from (Hespanha and Morse, 1999) where the lower bounds on average dwell-time are computed using multiple Lyapunov functions. Another result on slow switching, but with state-dependent average dwell-time, has appeared in (Persis et al., 2003). The second fundamental tool, that we build on, relates to the robustness with respect to external disturbances, formalized by the notion of input-tostate stability (ISS) introduced in (Sontag, 1989). Using these classical works as foundation, our article provides a certain construction of the ISS-Lyapunov functions for

^bLaboratory for Analysis and Architecture of Systems, Toulouse, CNRS, France

^{*}Corresponding author

Email addresses: gxzhang7@uci.edu (GuangXue Zhang), aneel.tanwani@laas.fr (Aneel Tanwani)

¹The work of G. Zhang was supported under "IDEX grant for nouveaux entrants" as a part of her Masters internship at LAAS. The work of A. Tanwani is partially supported by ANR JCJC project ConVan with grant number ANR-17-CE40-0019-01.

$$\frac{d}{e^{+}} = g_{o}(e, d)$$

$$e \xrightarrow{\dot{x}} = f_{c,\sigma}(x, e)$$

$$x^{+} = g_{c}(x, e)$$

Figure 1: Two-stage serial cascade system with switching dynamics.

the switched systems in cascade configuration and develops lower bounds on the dwell-time that guarantee ISS property for the switched system.

One of the first results on input-to-state stability of switched systems appears in (Vu et al., 2007), where the authors associate an ISS-Lyapunov function to each subsystem with linear decay rate, and assume that the Lyapunov function for each subsystem can be linearly dominated by the Lyapunov function of another subsystem at switching times. Other relevant papers studying ISS property for systems with jump maps using dwell-time conditions are (Hespanha et al., 2008), (Dashkovskiy and Mironchenko, 2013). Using the notion of ISS, tools such as small gain theorems (Jiang et al., 1994), or cascade principles (Sontag and Teel, 1995) are developed to study different applications. The small gain theorems have in particular found utility in the stability analysis of interconnected systems (Ito, 2006), (Dashkovskiy et al., 2010). For hybrid systems, in general, the ISS results using Lyapunov functions appear in (Cai and Teel, 2009), (Cai and Teel, 2013). Their utility is seen in analyzing stability of two interconnected hybrid systems in (Sanfelice, 2014) and (Liberzon et al., 2014), where the later in particular focuses on small-gain theorems and their application in control over networks. The stability of interconnected switched systems based on small gain theorems also appears in (Yang and Liberzon, 2015). The more recent article then generalizes the results on interconnections (Yang et al., 2016) while allowing for potentially unstable subsystems and jump dynamics.

The first part of this article is also built on analyzing the stability of interconnected subsystems with continuous and discrete dynamics. However, we are interested in studying systems where the interconnection is described by a cascade configuration, see Fig. 1. Using the Lyapunov function construction in (Tanwani et al., 2015), we construct the Lyapunov functions for this cascade interconnection. We then use the framework of hybrid systems to describe the overall system with jump maps, and switching signal with average dwell-time constraints. A novel Lyapunov function is constructed for this hybrid system and the corresponding analysis provides the lower bounds on average dwell-time which yield global asymptotic stability of a certain set. In our approach, we do not require the decay rates of the individual Lyapunov functions to be linear, and the upper bounds on the value of individual Lyapunov function at jump instants may be nonlinear functions of other Lyapunov functions. When studying linear systems as an example, even though we associate quadratic Lyapunov functions to individual subsystems, the Lyapunov function for the overall hybrid system involves a product of the exponential function with a non-quadratic function, which to the best of our knowledge is a novel observation.

We then use these constructions to study the feedback stabilization of switched nonlinear systems when the output measurements and control inputs are time-sampled. Using an observer-based controller, where the estimation error dynamics and the closed-loop system (with static control) are ISS with respect to measurement errors, we rewrite the whole system in cascade configuration where the estimation error drives the state of the controlled plant. The measurement errors are introduced because we only send time-sampled outputs to the controller, and the controller only sends sampled control inputs to the plant. In both cases, the sampled measurements are subjected to a zero-order hold, and thus remain constant until the next sampling instant. Our goal is to derive algorithms to compute sampling algorithms which result in global asymptotic stability of the closed-loop system under the average dwell-time assumptions derived earlier. The event-based sampling strategy that we use is inspired from (Tanwani et al., 2015), where the dynamic filters are introduced. The next sampling instant occurs when the difference between the current value of the output (resp. input) and its last sample is comparatively larger than the value of the dynamic filter's state. Beyond the realm of periodic sampling, stabilization of dynamical systems has been studied subject to various sampling techniques, see for example (Heemels et al., 2012) and (Tanwani et al., 2018) for recent surveys. Among these methods, event-based control has received attention as an effective means of sampling and various variants of this problem have been studied over the past few years. However, this technique has not yet been studied for switched systems which is the second main contribution of this article.

The remainder of the article is organized as follows: In Section 2, we provide an overview of basic stability notions and existing results which will be used in this article. The system class of interest is introduced in Section 3, where we develop the main theoretical results on construction of Lyapunov functions, and developing bounds for average dwell-time. These results are applied in Section 4 to study dynamic feedback stabilization of switched nonlinear systems with sampled-data, and the second main results concerning the design of sampling algorithms and stability analysis of closed-loop system is developed in this section. As an illustration, we provide simulation results for an academic example in Section 5, followed by some concluding remarks in Section 6.

2. Preliminaries

In this section, we recall some basic notions of interest which relate to the stability of a hybrid system. For our purposes, it is useful to consider hybrid systems with inputs studied in (Cai and Teel, 2013), which are described by following inclusions:

$$\begin{cases} \dot{\xi} \in \mathcal{F}(\xi, d), & (\xi, d) \in \mathcal{C}, \\ \xi^{+} \in \mathcal{G}(\xi, d), & (\xi, d) \in \mathcal{D}, \end{cases}$$
 (1)

where $\xi \in \mathcal{X}$ is the state and $d \in \mathbb{R}^{\overline{d}}$ is the external disturbance. The flow set $\mathcal{C} \subseteq \mathcal{X} \times \mathbb{R}^{\overline{d}}$, and the jump set $\mathcal{D} \subseteq \mathcal{X} \times \mathbb{R}^{\overline{d}}$ are assumed to be relatively closed in $\mathcal{X} \times \mathbb{R}^{\overline{d}}$. The set-valued map $\mathcal{F} : \mathcal{C} \rightrightarrows \mathcal{X}$ describes the continuous dynamics when ξ belongs to the flow set \mathcal{C} . The mapping $\mathcal{G} : \mathcal{D} \rightrightarrows \mathcal{X}$ defines the state reset map, when ξ belongs to the jump set \mathcal{D} .

The solution of the hybrid system (1) is defined on a hybrid time domain. A set $E \subseteq \mathbb{R}_{\geq 0} \times \mathbb{Z}_{\geq 0}$ is called a compact hybrid time domain if $E = \bigcup_{j=0}^{J} ([t_j, t_{j+1}], j)$ for some finite sequence $0 = t_0 \le t_1 \le \cdots \le t_{J+1}$. We say that E is a hybrid time domain if for each (T, J) in E, the set $E \cap [0,T] \times \{0,1,\ldots,J\}$ is a compact hybrid time domain. A function defined on a hybrid time domain is called a hybrid signal. In (1), the disturbance d is a hybrid signal, so that $d(\cdot, j)$ is locally essentially bounded. A hybrid arc ξ is a hybrid signal for which $\xi(\cdot,j)$ is locally absolutely continuous for each $j \in \mathbb{Z}_{>0}$, and we use the notation $\xi(t,j)$ to denote the value of ξ at time t after j jumps. For a given initial condition $\xi(0,0) \in \mathcal{C} \cup \mathcal{D}$, and a hybrid signal d, the solution ξ to (1) is a hybrid arc if $\operatorname{dom} \xi \subseteq \operatorname{dom} d$, and it holds that i) for every $j \in \mathbb{Z}_{>0}$ and almost every $t \in \mathbb{R}_{>0}$ such that $(t,j) \in \text{dom } \xi$, we have $(\xi(t,j), d(t,j)) \in \mathcal{C}$ and $\dot{\xi}(t,j) \in \mathcal{F}(\xi(t,j), d(t,j))$; ii) for $(t,j) \in \text{dom } \xi$ such that $(t,j+1) \in \text{dom } \xi$, we have $(\xi(t,j),d(t,j)) \in \mathcal{D}$ and $\xi(t,j+1) \in \mathcal{G}(\xi(t,j),d(t,j))$. It is assumed that the quadruple $(\mathcal{F}, \mathcal{G}, \mathcal{C}, \mathcal{D})$ satisfies the basic assumptions listed in (Goebel et al., 2012, Assumption 6.5), so that system (1) has a well-defined solution ξ in the space of hybrid arcs, for each hybrid signal d, but not necessarily unique.

To study the stability notions of interest for hybrid arcs, we need some notation. A function $\alpha: \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is said to be of class K if it is continuous, strictly increasing, and $\chi(0) = 0$. If α is also unbounded, then it is said to be of class \mathcal{K}_{∞} . A function $\beta: \mathbb{R}_{\geq 0} \times \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is said to be of class \mathcal{KL} if $\beta(\cdot,t)$ is of class \mathcal{K} for each $t \in \mathbb{R}_{\geq 0}$ and $\beta(r,t) \to 0$ as $t \to \infty$ for each fixed $r \in \mathbb{R}_{>0}$; see (Khalil, 2002, Chapter 4) for their use in formulation of common stability notions. In addition, we require class \mathcal{KLL} function: A function $\beta: \mathbb{R}_{\geq 0} \times \mathbb{R}_{\geq 0} \times \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is a class \mathcal{KLL} function if $\beta(\cdot,\cdot,j)$ is a class \mathcal{KL} function for each $j \geq 0$ and $\beta(\cdot, s, \cdot)$ is a class \mathcal{KL} function for each $s \geq 0$. For a compact set $\mathcal{A} \subset \mathcal{X}$ and $\xi \in \mathcal{X}, |\xi|_{\mathcal{A}} := \inf_{z \in \mathcal{A}} |\xi - z|, \text{ where } |\cdot| \text{ denotes the}$ usual Euclidean norm. Following (Cai and Teel, 2009), for a hybrid signal d, we use the notation $||d||_{(t,j)}$ to denote the maximum between ess-sup $|d(\hat{t}, \hat{j})|$ $(\hat{t}, \hat{j}\!+\!1) \not\in\! \mathrm{dom}\; d, \hat{t}\!+\!\hat{j}\!\leq\! t\!+\!j$

and $\sup_{\substack{(\hat{t},\hat{j}+1)\in \mathrm{dom}\,d,\hat{t}+\hat{j}\leq t+j\\\mathrm{tions}\;\alpha,\chi\;\mathrm{over}\;\mathbb{R}_{\geq 0},\;\mathrm{we}\;\mathrm{also}\;\mathrm{use}\;\mathrm{the}\;\mathrm{Landau-notation}\;\mathrm{to}} |d(\hat{t},\hat{j})|.$

write $\alpha(s) = o(\chi(s))$ as $s \to c$ if $\lim_{s \to c} \frac{\alpha(s)}{\chi(s)} = 0$; similarly, we write $\alpha(s) = O(\chi(s))$ as $s \to c$ if $\lim_{s \to c} \frac{\alpha(s)}{\chi(s)} \le M$ for some M > 0.

2.1. Input-to-State Stability

We recall basic definitions and Lyapunov characterizations of ISS for hybrid systems from (Cai and Teel, 2009).

Definition 1 (Input-to-state stability (ISS)). System (1) is *ISS* with respect to a compact set $\mathcal{A} \subset \mathcal{X}$ if there exist functions $\gamma \in \mathcal{K}$, and $\beta \in \mathcal{KLL}$ such that

$$|\xi(t,j)|_{\mathcal{A}} \le \beta(|\xi(0,0)|_{\mathcal{A}},t,j) + \gamma(||d||_{(t,j)}),$$
 (2)

for every $(t, j) \in \text{dom } \xi$.

Definition 2 (ISS Lyapunov function). A smooth function $V: \mathcal{X} \to \mathbb{R}_{\geq 0}$ is an ISS-Lyapunov function of the hybrid system (1) w.r.t. a compact set $\mathcal{A} \subset \mathcal{X}$ if the following hold:

• there exist $\underline{\alpha}, \overline{\alpha} \in \mathcal{K}_{\infty}$ such that

$$\underline{\alpha}(|\xi|_{\mathcal{A}}) \le V(\xi) \le \overline{\alpha}(|\xi|_{\mathcal{A}}), \quad \forall \, \xi \in \mathcal{C} \cup \mathcal{D} \cup \mathcal{G}(\mathcal{D}), \quad (3)$$

• there exist $\widehat{\alpha} \in \mathcal{K}_{\infty}$ and $\widehat{\gamma} \in \mathcal{K}$ such that

$$|\xi|_{\mathcal{A}} \ge \widehat{\gamma}(|d|) \Rightarrow \langle \nabla V(\xi), f \rangle \le -\widehat{\alpha}(|\xi|_{\mathcal{A}}),$$
 (4)

holds for every $(\xi, d) \in \mathcal{C}$ and $f \in \mathcal{F}(\xi, d)$.

• the functions $\widehat{\alpha}$ and $\widehat{\gamma}$ also satisfy

$$|\xi|_{\mathcal{A}} \ge \widehat{\gamma}(|d|) \Rightarrow V(g) - V(\xi) \le -\widehat{\alpha}(|\xi|_{\mathcal{A}}),$$
 (5)

for every $(\xi, d) \in \mathcal{D}$ and for every $g \in \mathcal{G}(\xi, d)$.

The following result provides an alternate characterization of ISS for system (1) by combining results given in (Cai and Teel, 2009, Proposition 1) and (Liberzon and Shim, 2015, Theorem 1).

Proposition 1. Consider system (1) and a compact set $A \subset \mathcal{X}$. A differentiable function $V : \mathcal{X} \to \mathbb{R}_{\geq 0}$ satisfying (3) with $\underline{\alpha}, \overline{\alpha} \in \mathcal{K}_{\infty}$ is an ISS-Lyapunov function w.r.t. A if and only if

• there exist $\alpha_{\mathcal{C}} \in \mathcal{K}_{\infty}$, $\gamma_{\mathcal{C}} \in \mathcal{K}$ and a continuous nonnegative function $\varrho : \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ such that

$$\langle \nabla V(\xi), f \rangle \le -\alpha_{\mathcal{C}}(|\xi|_{\mathcal{A}}) + \varrho(|\xi|_{\mathcal{A}}) \gamma_{\mathcal{C}}(|d|)$$
 (6a)

holds for each $(\xi, d) \in \mathcal{C}$ and $f \in \mathcal{F}(\xi, d)$,

• there exist $\alpha_{\mathcal{D}} \in \mathcal{K}_{\infty}$, $\gamma_{\mathcal{D}} \in \mathcal{K}$ that satisfy

$$V(g) - V(\xi) \le -\alpha_{\mathcal{D}}(|\xi|_{\mathcal{A}}) + \gamma_{\mathcal{D}}(|d|) \tag{6b}$$

for each $(\xi, d) \in \mathcal{D}$ and $g \in \mathcal{G}(\xi, d)$,

• the functions $\alpha_{\mathcal{C}}$, ϱ satisfy the asymptotic ratio condition

$$\limsup_{r \to \infty} \frac{\varrho(r)}{\alpha_{\mathcal{C}}(r)} = 0. \tag{7}$$

Remark 1. The inequality (6a) is different from the expression given in (Cai and Teel, 2009, Proposition 2.6). It can be shown that (6a) also implies (4). This implication is proved in a constructive manner, that is, the pair $(\widehat{\alpha}, \widehat{\gamma})$ is constructed from the triplet $(\alpha_{\mathcal{C}}, \varrho, \gamma_{\mathcal{C}})$, in (Liberzon and Shim, 2015, Theorem 1) using the condition (7), which appears in Remark 1 of that paper.

2.2. Cascade Switched Systems

The framework of (1) is useful for modeling switched systems. We are interested in studying switched systems in cascade configuration which comprise a family of dynamical subsystems described by

$$\dot{x} = f_{c,p}(x,e),\tag{8a}$$

$$\dot{e} = f_{o,p}(e,d),\tag{8b}$$

where p belongs to a *finite* index set \mathcal{P} . The vector fields $f_{c,p}: \mathbb{R}^{n_c} \times \mathbb{R}^{n_o} \to \mathbb{R}^{n_c}$ and $f_{o,p}: \mathbb{R}^{n_o} \times \mathbb{R}^{n_d} \to \mathbb{R}^{n_o}$ are assumed to be continuous for each $p \in \mathcal{P}$. It is also assumed that $f_{c,p}(0,0) = 0$, and $f_{o,p}(0,0) = 0$, and the stability of the origin $\{0\} \in \mathbb{R}^{n_c+n_o}$ is the topic of interest in the sequel. The switched system generated by (8) is

$$\dot{x} = f_{c,\sigma}(x, e) \tag{9a}$$

$$\dot{e} = f_{o,\sigma}(e,d),\tag{9b}$$

where $\sigma: \mathbb{R}_{\geq 0} \to \mathcal{P}$ denotes the piecewise constant right-continuous switching signal. The function σ changes its value at switching times which are denoted by $\{t_i\}_{i\in\mathbb{N}}$. At these switching times, we allow the state values to have jumps defined by the maps

$$x^+ = g_c(x, e) \tag{10a}$$

$$e^+ = g_o(e, d),$$
 (10b)

so that $x(t_i^+) = (x(t_i))^+$, and $e(t_i^+) = (e(t_i))^+$ denote the value of the state variables just after the switching times. We say that the switching signal σ has an average dwell-time τ_a , denoted $\sigma \in \mathcal{S}_{\tau_a}$ if there exists $N_0 \geq 1$ such that for each $t > s \geq 0$, it holds that

$$N_{\sigma(t,s)} \le N_0 + \frac{t-s}{\tau_a} \tag{11}$$

where $N_{\sigma(t,s)}$ is the number of switching in the interval (s,t]. The constant N_0 is called the *chatter bound* giving the tolerance number of fast switchings.

Problem 1. Given that each subsystem in (8a) (with e as input) and (8b) (with d as input) admits an ISS Lyapunov function w.r.t. the origin, how can we

- 1. compute the lower bound on τ_a , and
- 2. construct an ISS Lyapunov function for the hybrid system (9)-(10),

such that, for each $\sigma \in \mathcal{S}_{\tau_a}$, we have

$$\left|\left(x(t,j),e(t,j)\right)\right| \leq \beta(\left|\left(x(0,0),e(0,0)\right)\right|,t,j) + \gamma\left(\|d\|_{(t,j)}\right)$$

for some $\beta \in \mathcal{KLL}$, and $\gamma \in \mathcal{K}$.

3. Stability of Cascade System

To find a solution to the problem mentioned above, we proceed in several steps which allow us to arrive at the result given in Theorem 1.

3.1. Individual Lyapunov Functions

The first step is to formally state the stability assumptions imposed on the dynamical subsystem (8a) and (8b) which are formally listed below:

(L1) For each $p \in \mathcal{P}$, there exists a continuously differentiable function $V_{o,p} : \mathbb{R}^{n_o} \to \mathbb{R}_{\geq 0}$, and there exist class \mathcal{K}_{∞} functions $\overline{\alpha}_{o,p}, \underline{\alpha}_{o,p}, \alpha_{o,p}$ and $\gamma_{o,p}$ such that

$$\underline{\alpha}_{o,p}(|e|) \le V_{o,p}(e) \le \overline{\alpha}_{o,p}(|e|)$$

$$\left\langle \frac{\partial V_{o,p}}{\partial e}, f_{o,p}(e,d) \right\rangle \le -\alpha_{o,p}(V_{o,p}(e)) + \gamma_{o,p}(|d|)$$

hold for every $(e,d) \in \mathbb{R}^{n_o} \times \mathbb{R}^{n_d}$.

(L2) For each $p \in \mathcal{P}$, there exists a continuously differentiable function $V_{c,p} : \mathbb{R}^{n_c} \to \mathbb{R}_{\geq 0}$, and there exist class \mathcal{K}_{∞} functions $\overline{\alpha}_{c,p}$, $\underline{\alpha}_{c,p}$, $\alpha_{c,p}$, and $\gamma_{c,p}$ such that

$$\underline{\alpha}_{c,p}(|x|) \le V_{c,p}(x) \le \overline{\alpha}_{c,p}(|x|)$$

$$\left\langle \frac{\partial V_{c,p}}{\partial x} f_{c,p}(x,e) \right\rangle \le -\alpha_{c,p}(V_{c,p}(x)) + \gamma_{c,p}(V_{o,p}(e)),$$

hold for every $(x, e) \in \mathbb{R}^{n_c} \times \mathbb{R}^{n_o}$.

(L3) As $s \to 0^+$, we have $\gamma_{c,p}(s) = O(\alpha_{o,p}(s))$, that is, if we let

$$\overline{\nu}_p(s) := \frac{\gamma_{c,p}(s)}{\alpha_{o,p}(s)}, \quad \text{for } s > 0, \tag{13}$$

then there exists a constant M > 0, such that

$$\lim_{s \to 0^+} \overline{\nu}_p(s) \le M.$$

In addition, we introduce the following assumption² on the jump maps introduced in (10).

(A1) For each $(x, e, d) \in \mathbb{R}^{n_c + n_o + n_d}$, the jump maps at switching times satisfy

$$|g_c(x,e)| \le \widehat{\alpha}_c(|(x,e)|)$$

$$|g_o(e,d)| \le \widehat{\alpha}_o(|e|) + \widehat{\rho}_o(|d|)$$

for some class \mathcal{K}_{∞} functions $\widehat{\alpha}_c$, $\widehat{\alpha}_o$, $\widehat{\rho}_c$, and $\widehat{\rho}_o$.

²It is also possible to consider more general jump maps of the form $x^+ = g_c(x,e,d)$ and $e^+ = g_o(x,e,d)$, provided that the inequalities in **(A1)** take the form $|g_c(x,e,d)| \leq \widehat{\alpha}_c(|(x,e)|) + \widehat{\rho}_c(|d|)$ and $|g_o(x,e,d)| \leq \widehat{\alpha}_o(|(x,e)|) + \widehat{\rho}_o(|d|)$. The results of this paper would carry just by changing the map ρ in (24).

Using the assumptions introduced above, it is possible to construct a candidate Lyapunov function $V_p(x, e)$ for each subsystem $p \in \mathcal{P}$. This construction is primarily inspired from the work of (Tanwani et al., 2015).

Remark 2. The assumption (L3) is introduced to provide a construction of the candidate Lyapunov function V_p explicitly in terms of $V_{o,p}$ and $V_{c,p}$. One can always modify the function $V_{c,p}$ to $\widehat{V}_{c,p}$ such that the resulting $\widehat{\gamma}_{c,p}$ satisfies (L3); this is a direct consequence of (Sontag and Teel, 1995, Theorem 1) as we can choose $\widehat{\gamma}_{c,p}(s) = O(\alpha_{o,p}(s)) \cdot \underline{\alpha}_{o,p}(s)$ for s sufficiently small in the neighborhood of origin.

Proposition 2. Consider the family of dynamical subsystems (8) satisfying (L1), (L2), and (L3), along with the jump dynamics (10) satisfying (A1). For each $p \in \mathcal{P}$, introduce the continuously differentiable function V_p ,

$$V_p(x,e) := \int_0^{V_{o,p}(e)} \nu_p(s) \, ds + V_{c,p}(x), \tag{14}$$

where $\nu_p: \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is a continuous and nondecreasing function with $\nu_p(s) \geq 4\overline{\nu}_p(s)$, for each s > 0. It then holds that, for some $\underline{\alpha}_p, \overline{\alpha}_p \in \mathcal{K}_{\infty}$,

$$\underline{\alpha}_p(|(x,e)|) \le V_p(x,e) \le \overline{\alpha}_p(|(x,e)|), \quad \forall (x,e) \in \mathbb{R}^{n_c + n_o}.$$
(15)

There also exist $\alpha_p, \gamma_p \in \mathcal{K}_{\infty}$ such that

$$\left\langle \nabla V_p(x,e), \begin{pmatrix} f_{c,p}(x,e) \\ f_{o,p}(e,d) \end{pmatrix} \right\rangle \le -\alpha_p(V_p(x,e)) + \gamma_p(|d|),$$
 (16)

for every $(x, e, d) \in \mathbb{R}^{n_c + n_o + n_d}$. Moreover, there exist $\chi, \rho \in \mathcal{K}_{\infty}$ such that for each $(p, q) \in \mathcal{P} \times \mathcal{P}, q \neq p$,

$$V_q(x^+, e^+) \le \chi(V_p(x, e)) + \rho(|d|),$$
 (17)

for every $(x, e, d) \in \mathbb{R}^{n_c + n_o + n_d}$.

Proof. Fix $p \in \mathcal{P}$. Introduce the class \mathcal{K}_{∞} function ℓ_p : $\mathbb{R}_{>0} \to \mathbb{R}_{>0}$ as follows:

$$\ell_p(s) = \int_0^s \nu_p(r) dr \tag{18}$$

where ν_p was introduced in (14), so that

$$V_p(x, e) = (\ell_p \circ V_{o,p})(e) + V_{c,p}(x).$$

The bound (15) is seen to hold since ℓ_p is a class \mathcal{K}_{∞} function. Using (L1), we now obtain

$$\langle \nabla(\ell_p \circ V_{o,p})(e), f_{o,p}(e,d) \rangle$$

$$\leq \nu_p(V_{o,p}(e))(-\alpha_{o,p}(V_{o,p}(e)) + \gamma_{o,p}(|d|)). \quad (19)$$

To analyze the right-hand side of (19), first consider the case where $\gamma_{o,p}(|d|) \leq \frac{1}{2}\alpha_{o,p}(V_{o,p}(e))$, so that

$$\langle \nabla(\ell_p \circ V_{o,p})(e), f_{o,p}(e,d) \rangle \le -\frac{1}{2} \nu_p(V_{o,p}(e)) \alpha_{o,p}(V_{o,p}(e));$$

else, by introducing $\theta_p(s) := \alpha_{o,p}^{-1}(2\gamma_{o,p}(s)),$

$$\frac{1}{2}\alpha_{o,p}(V_{o,p}(e)) \le \gamma_{o,p}(|d|) \Leftrightarrow V_{o,p}(e) \le \alpha_{o,p}^{-1}(2\gamma_{o,p}(|d|))$$
$$= \theta_p(|d|)$$

so that $\nu_p(V_{o,p}(e)) \le \nu_p(\theta_p(|d|))$, because ν_p is by construction nondecreasing, and

$$\langle \nabla(\ell_p \circ V_{o,p})(e), f_{o,p}(e,d) \rangle \le -\nu_p(V_{o,p}(e))\alpha_{o,p}(V_{o,p}(e)) + \nu_p(\theta_p(|d|))\gamma_{o,p}(|d|).$$

From these two cases, the inequality (19) results in

$$\langle \nabla(\ell_p \circ V_{o,p})(e), f_{o,p}(e,d) \rangle \le$$

$$-\frac{1}{2} \nu_p(V_{o,p}(e)) \alpha_{o,p}(V_{o,p}(e)) + \nu_p(\theta_p(|d|)) \gamma_{o,p}(|d|). \quad (20)$$

Using **(L2)**, (20), and the fact that $\nu_p(s) \geq 4\overline{\nu}_p(s)$, for each s > 0 with $\overline{\nu}_p$ given in (13), we can now derive (16) as follows:

$$\begin{split} \left\langle \nabla V_p(x,e), \begin{pmatrix} f_{c,p}(x,e) \\ f_{o,p}(e,d) \end{pmatrix} \right\rangle \\ &\leq -\frac{1}{2} \nu_p(V_{o,p}(e)) \alpha_{o,p}(V_{o,p}(e)) + \nu_p(\theta_p(|d|)) \gamma_{o,p}(|d|) \\ &- \alpha_{c,p}(V_{c,p}(x)) + \gamma_{c,p}(V_{o,p}(e)) \\ &\leq -\gamma_{c,p}(V_{o,p}(e))) - \alpha_{c,p}(V_{c,p}(x)) + \nu_p(\theta_p(|d|)) \gamma_{o,p}(|d|) \\ &\leq -\alpha_p(V_p(x,e)) + \gamma_p(|d|), \end{split}$$

where we used the definitions $\gamma_p(s) := \nu_p(\theta_p(s))\gamma_{o,p}(s)$,

$$\alpha_p(s) := \min \left\{ \alpha_{c,p} \left(\frac{1}{2} s \right), \gamma_{c,p} \left(\frac{1}{2} \ell_p^{-1}(s) \right) \right\}, \qquad (21)$$

and the triangle-inequality type result from (Kellett, 2014, Lemma 10) to derive the last inequality.

Next, to derive (17), we observe that

$$V_q(x^+, e^+) = (\ell_q \circ V_{o,q})(g_o(e, d)) + V_{c,q}(g_c(x, e)).$$

Using (A1), it then follows that

$$V_{q}(x^{+}, e^{+}) \leq \ell_{q} \circ \overline{\alpha}_{o,q}(|g_{o}(e, d)|) + \overline{\alpha}_{c,q}(|g_{c}(x, e)|)$$

$$\leq \ell_{q} \circ \overline{\alpha}_{o,q}(2\widehat{\alpha}_{o}(|(x, e)|)) + \ell_{q} \circ \overline{\alpha}_{o,q}(2\widehat{\rho}_{o}(|d|)$$

$$+ \overline{\alpha}_{c,q}(\widehat{\alpha}_{c}(|(x, e)|))$$

$$\leq \chi(V_{p}(x, e)) + \rho(|d|)$$
(22)

where, recalling (15), we used the definitions

$$\chi(s) := \max_{p,q \in \mathcal{P}} \left\{ \ell_q \circ \overline{\alpha}_{o,q} \left(2\widehat{\alpha}_o \circ \underline{\alpha}_p^{-1}(s) \right) + \overline{\alpha}_{c,q} (\widehat{\alpha}_c \circ \underline{\alpha}_p^{-1}(s)) \right\}$$
(23)

and

$$\rho(s) := \max_{q \in \mathcal{P}} \left\{ \ell_q \circ \overline{\alpha}_{o,q}(2\widehat{\rho}_o(s)) \right\}, \tag{24}$$

which establishes the desired bound since both functions are class \mathcal{K}_{∞} .

3.2. Stability of Overall Hybrid System

We now use the result of Proposition 2 to compute lower bounds on the dwell-time which result in cascade switched system being globally ISS. To do so, we find it convenient to express the switched system in the framework of the hybrid system adopted in (Goebel et al., 2012). This is done by introducing the augmented state variable $\xi := (x, e, p, \tau) \in \mathcal{X} := \mathbb{R}^{n_c + n_o} \times \mathcal{P} \times [0, N_0]$, where p is a discrete variable denoting a subsystem, and τ plays the role of a scaled timer. The hybrid model capturing the dynamics of the switched system driven by an external disturbance $d \in \mathbb{R}^{\overline{d}}$, and where the switching signals have an average dwell-time τ_a , is

$$(\xi, d) \in \mathcal{C} : \begin{cases} \dot{x} = f_{c,p}(x, e) \\ \dot{e} = f_{o,p}(e, d) \\ \dot{p} = 0 \\ \dot{\tau} \in [0, \frac{1}{\tau_a}] \end{cases}$$

$$(25a)$$

$$(\xi, d) \in \mathcal{D} : \begin{cases} x^{+} = g_{c}(x, e) \\ e^{+} = g_{o}(e, d) \\ p^{+} \in \mathcal{P} \setminus \{p\} \\ \tau^{+} = \tau - 1 \end{cases}$$
 (25b)

where the flow set $\mathcal{C} := \mathcal{X} \times \mathbb{R}^{\overline{d}}$, and the jump set $\mathcal{D} := \mathbb{R}^{n_c + n_o} \times \mathcal{P} \times [1, N_0] \times \mathbb{R}^{\overline{d}}$. We denote the set-valued mapping on the right-hand side of (25a) by $\mathcal{F}(\xi, d)$, and the mapping on the right-hand side of (25b) is denoted by $\mathcal{G}(\xi, d)$. We are interested in studying the ISS property of the system (25) (driven by the disturbance d) with respect to the compact set

$$\mathcal{A}_0 := \{0\}^{n_c + n_o} \times \mathcal{P} \times [0, N_0] \tag{26}$$

by finding an appropriate ISS Lyapunov function. To do so, we introduce the function $\varphi : \mathbb{R}_{>0} \to \mathbb{R}_{>0}$ defined as

$$\varphi(s) = \begin{cases} \exp\left(\int_1^s \frac{2}{\psi(r)} dr\right), & s > 0\\ 0, & s = 0 \end{cases}$$
 (27)

where $\psi : \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is a continuously differentiable class \mathcal{K}_{∞} function, with $\psi'(0) = 0$, and

$$\psi(s) \le \min\{c_0 s, \alpha_p(s) \mid p \in \mathcal{P}\}, \quad s \in [0, c_1],$$

and
$$\psi(s) \le \min\{\alpha_p(s) \mid p \in \mathcal{P}\}, \quad s \ge c_1,$$

for some $c_0, c_1 > 0$. We recall that $\alpha_p \in \mathcal{K}_{\infty}$ were introduced as the decay function for V_p in (16). The function φ is now used in the following result:

Theorem 1. Consider system (25) and suppose that (L1), (L2), (L3), (A1) hold. Let $\chi \in \mathcal{K}_{\infty}$ be as in (17). If, for some $\varepsilon > 0$, the average dwell-time τ_a satisfies

$$\tau_a > \zeta^* := \sup_{s \ge 0} \int_s^{(1+\varepsilon)\chi(s)} \frac{1}{\psi(r)} dr, \tag{28}$$

then for each $\tau_a > \zeta > \zeta^*$,

$$W(x, e, p, \tau) := \exp(2\zeta\tau)\varphi(V_p(x, e)), \tag{29}$$

is an ISS Lyapunov function for the hybrid system (25) w.r.t. the compact set A_0 , and input d.

Remark 3. To gain an insight about the constraints imposed by the stability condition (28) on the system structure, we study particular instances where $\alpha(s) := \min\{\alpha_p(s) \mid p \in \mathcal{P}\}$ exhibits linear, super-linear and sublinear growth. It can be seen that if the jump map χ does not grow too fast compared to $\alpha(s)$, then ζ^* in (28) is finite. For the sake of simplicity, let $\alpha(s) := as^k$, with a, k > 0, and choose $c_0 = a$ in the definition of ψ .

- Linear decay: We first consider the case k=1, so that $\alpha(s)=as$, and we let $\psi(s)=as$, for $s\geq 0$. This gives $\zeta^*=\sup_{s\geq 0}\frac{1}{a}\log\frac{(1+\varepsilon)\chi(s)}{s}$, which is finite if $\lim_{s\to\infty}\chi(s)=O(s)$, and $\lim_{s\to 0}\chi(s)=O(s)$. If $\chi(s)=\mu s$, then $\zeta^*=\frac{1}{a}\log((1+\varepsilon)\mu)$, which resembles the bound given in (Liberzon, 2003, Chapter 3) by taking $\varepsilon>0$ arbitrarily small.
- Super-linear decay: Next consider the case where $\alpha(s) = as^k$ with k > 1. Choose $c_1 = 1$, then we can let $\psi(s) = as^k$, for every $s \geq 0$. This yields $\zeta^* = \frac{1}{k-1} \sup_{s \geq 0} \left[\frac{1}{as^{k-1}} \frac{1}{a\chi(s)^{k-1}} \right]$. With $\chi \in \mathcal{K}_{\infty}$, it is seen that ζ^* is finite and positive if $\lim_{s \to 0^+} \left[\frac{1}{as^{k-1}} \frac{1}{a\chi(s)^{k-1}} \right] \leq 0$, which holds if $\chi(s) = o(s)$ when $s \to 0$.
- Sub-linear decay: Lastly, consider the case where $\alpha(s) = as^k$ with k < 1. Choose $c_1 = 1$, then there exist $\underline{c} < 1$, $\overline{c} > 1$ and a continuously differentiable function ψ such that $\psi(s) = as$, $s \in [0,\underline{c}]$, and $\psi(s) = as^k$ for $s \geq \overline{c}$. With this choice of ψ , the lower bound ζ^* is a finite positive scalar, if $\lim_{s\to 0^+} \chi(s) = O(s)$, and $\lim_{s\to\infty} \chi(s) = o(s)$.

Remark 4. A function similar to φ defined in (27) also appears in (Praly and Wang, 1996) to transform nonlinear decay rates to linear ones in inequalities associated with Lyapunov functions, while keeping the modified Lyapunov function differentiable. In the proof of Theorem 1, this function serves the same purpose. Here, the construction of ψ is modified slightly because the differentiability of φ is only a concern at the origin. If one keeps the construction of ψ as in (Praly and Wang, 1996), with $\psi = \min\{s, \alpha(s)\}$, $\alpha(s) = \min_{p \in \mathcal{P}} \{\alpha_p(s)\}$, then $\zeta^* \geq \ln \frac{\chi(s)}{s}$, and one is restricted to consider jump maps χ which grow at most linearly away from the origin.

Proof of Theorem 1. The proof is based on showing that W satisfies the conditions listed in Proposition 1. It is seen that φ is differentiable (away from the origin) on $\mathbb{R}_{>0}$, and it is shown in (Praly and Wang, 1996, Lemma 12) that the function φ is also continuously differentiable in the

neighborhood of the origin with $\varphi'(0) = 0$. Therefore, W is also continuously differentiable.

Since φ is a class \mathcal{K}_{∞} function, one can easily verify that

$$\underline{\alpha}(|\xi|_{\mathcal{A}_0}) \le W(\xi) \le \overline{\alpha}(|\xi|_{\mathcal{A}_0})$$

for some functions $\underline{\alpha}$, $\overline{\alpha}$ of class \mathcal{K}_{∞} . From the definition of \mathcal{A}_0 and the assumption that $f_{c,p}(0,0)=f_{o,p}(0,0)=0$, $p\in\mathcal{P}$, it immediately follows that for each $\xi\in\mathcal{A}_0$ such that $(\xi,0)\in\mathcal{D}$, we have $\mathcal{G}(\xi,0)\in\mathcal{A}_0$. Also, along any continuous motion resulting from (25a), with initial condition $\xi\in\mathcal{A}_0$ satisfying $(\xi,0)\in\mathcal{C}$, the system trajectory stays within \mathcal{A}_0 . Hence, \mathcal{A}_0 is forward invariant with d=0.

Let f be an element of $\mathcal{F}(\xi, d)$. When $(\xi, d) \in \mathcal{C}$, it follows from (16) that

$$\begin{split} \langle \nabla W(\xi), f \rangle &\leq \frac{2\zeta \exp(2\zeta\tau)}{\tau_a} \varphi(V_p(x, e)) \\ &+ \exp(2\zeta\tau) \varphi(V_p(x, e)) \left[-\frac{2\alpha_p(V_p(x, e))}{\psi(V_p)} + \frac{2\gamma_p(|d|)}{\psi(V_p)} \right] \end{split}$$

and hence

$$\langle \nabla W(\xi), f \rangle \leq W(\xi) \left[\frac{2\zeta}{\tau_a} - \frac{2\alpha_p(V_p)}{\psi(V_p)} + \frac{2\gamma_p(|d|)}{\psi(V_p)} \right].$$

Since $\psi(s) \leq \alpha_p(s)$ by construction, and ζ is chosen to satisfy $\tau_a > \zeta$, we get an $a := 2(1 - \zeta/\tau_a) > 0$ such that

$$\langle \nabla W(\xi), f \rangle \le -aW(\xi) + \frac{2W(\xi)}{\psi(V_p(x, e))} \gamma_p(|d|), \quad \xi \in \mathcal{C}$$

which is of the same form 3 as (6a). It is readily checked that the asymptotic ratio condition (7) holds since

$$\lim_{|(x,e)| \to \infty} \frac{1}{\psi(V_p(x,e))} = 0.$$

The next step is to show that (6b) holds under condition (28) for the jump maps (25b). Let g denote an element of $\mathcal{G}(\xi,d)$. It is seen that for each $(\xi,d) \in \mathcal{D}$, that is, whenever $\tau \in [1, N_0]$, it follows from (17) that

$$\max_{g \in \mathcal{G}} W^{+} = \max_{g \in \mathcal{G}} \exp(2\zeta \tau^{+}) \varphi(V_{p^{+}}(x^{+}, e^{+}))$$

$$\leq \exp(2\zeta \tau - 2\zeta) \varphi\left(\chi(V_{p}(x, e)) + \rho(|d|)\right), \quad (30)$$

where χ and ρ were introduced in (23) and (24). Take an $\varepsilon > 0$ for which (28) holds, then it follows from (Kellett, 2014, Lemma 10) that:

$$\begin{split} \varphi\left(\chi(V_p(x,e)) + \rho(|d|)\right) &\leq \varphi((1+\varepsilon)\chi(V_p(x,e))) \\ &+ \varphi\left(\frac{1+\varepsilon}{\varepsilon}\rho(|d|)\right). \end{split}$$

For $(x, e) \neq 0$, the bound on the first term on the right-hand side is given by

$$\varphi((1+\varepsilon)\chi(V_p(x,e))) = \exp\left(\int_1^{(1+\varepsilon)\chi(V_p(x,e))} \frac{2dr}{\psi(r)}\right)$$

$$= \exp\left(\int_{V_p(x,e)}^{(1+\varepsilon)\chi(V_p(x,e))} \frac{2dr}{\psi(r)}\right) \cdot \exp\left(\int_1^{V_p(x,e)} \frac{2dr}{\psi(r)}\right)$$

$$\leq \exp(2\zeta^*) \varphi(V_p(x,e))$$

where ζ^* is defined as in (28). Letting

$$\widetilde{\rho}(s) := \exp(2\zeta N_0 - 2\zeta) \varphi\left(\frac{1+\varepsilon}{\varepsilon}\rho(s)\right),$$
 (31)

and noting that, for $\tau \in [0, N_0]$,

$$\widetilde{\rho}(s) \ge \exp\left(2\zeta\tau - 2\zeta\right) \, \varphi\left(\frac{1+\varepsilon}{\varepsilon}\rho(s)\right),$$

we obtain

$$\max_{g \in \mathcal{G}} W^+ \le \exp(2\zeta^* - 2\zeta) \exp(2\zeta\tau) \varphi(V_p(x, e)) + \widetilde{\rho}(|d|)$$

$$\le \exp(2\zeta^* - 2\zeta) W(\xi) + \widetilde{\rho}(|d|).$$

Having chosen ζ such that $\zeta^* - \zeta < 0$, we see that (6b) holds. The result of Proposition 1 thus ensures that W is an ISS Lyapunov function for (25) w.r.t. the set \mathcal{A}_0 , with input d.

3.3. Linear Case

We use the following linear example to illustrate the Theorem 1. Assume that the system (8) is linear

$$\dot{x} = A_p x + B_p e \tag{32a}$$

$$\dot{e} = F_n e + G_n d \tag{32b}$$

with the matrices A_p , F_p being Hurwitz. For the sake of simplicity, we assume that the states (x, e) do not go through any jump dynamics, and remain unchanged at switching instances. This way, we let the maps in (10) be

$$g_c(x, e) = x$$
, and $g_o(e, d) = e$.

We can now choose quadratic Lyapunov functions to satisfy (L1) and (L2). This is done by computing symmetric positive definite matrices $P_{o,p}$, $P_{c,p} > 0$ such that, for each $p \in \mathcal{P}$,

$$F_p^{\top} P_{o,p} + P_{c,p} F_p \le -Q_{o,p}$$

 $A_p^{\top} P_{c,p} + P_{c,p} A_p \le -Q_{c,p}$

for some symmetric positive definite matrices $Q_{o,p}, Q_{c,p} > 0$. By letting

$$V_{c,p}(x) = x^{\mathsf{T}} P_{c,p} x$$
, and $V_{o,p} = e^{\mathsf{T}} P_{o,p} e^{\mathsf{T}}$

we get

$$\underline{a}_{o,p}|x|^2 \le V_{o,p}(x) \le \overline{a}_{o,p}|x|^2$$

The function $\frac{2W(\xi)}{\psi(V_p(x,e))}$ is obviously nonnegative. The continuity follows from recalling that φ is continuously differentiable with $\varphi'(0)=0$, and observing that $\frac{2W(\xi)}{\psi(V_p(x,e))}=\exp{(2\zeta\tau)}\frac{\varphi(V_p(x,e))}{\psi(V_p(x,e))}=\exp{(2\zeta\tau)}\varphi'(V_p(x,e)).$

with $\underline{a}_{o,p} = \lambda_{\min}(P_{o,p})$, and $\overline{a}_{o,p} = \lambda_{\max}(P_{o,p})$. Similarly, it holds that

$$\underline{a}_{c,p}|x|^2 \le V_{c,p}(x) \le \overline{a}_{c,p}|x|^2$$

with $\underline{a}_{c,p} = \lambda_{\min}(P_{c,p})$, and $\overline{a}_{c,p} = \lambda_{\max}(P_{c,p})$. It can be readily shown that

$$\langle \nabla V_{o,p}, F_p x + G_p e \rangle \le -a_{o,p} V_{o,p}(x) + \overline{\gamma}_{o,p} |d|^2$$

by letting

$$a_{o,p} = \frac{\lambda_{\min}(Q_{o,p})}{2\lambda_{\max}(P_{o,p})}$$
 and $\overline{\gamma}_{o,p} = 2\frac{\|P_{o,p}G_p\|^2}{\lambda_{\min}(Q_{o,p})}$

and likewise

$$\langle \nabla V_{c,p}, A_p x + B_p e \rangle \le -a_{c,p} V_{c,p}(x) + \overline{\gamma}_{c,p} V_{o,p}(e)$$

with

$$a_{c,p} = \frac{\lambda_{\min}(Q_{c,p})}{2\lambda_{\max}(P_{c,p})} \quad \text{ and } \quad \overline{\gamma}_{c,p} = 2\frac{\|P_{c,p}B_p\|^2}{\lambda_{\min}(Q_{c,p})}.$$

For each $p \in \mathcal{P}$, the function $\overline{\nu}_p(s)$ in (13) turns out to be a constant as

$$\overline{\nu}_p(s) = \frac{\overline{\gamma}_{c,p} \, s}{a_{o,p} \, s} = \frac{\overline{\gamma}_{c,p}}{a_{o,p}} =: \overline{\nu}_p.$$

Thus, we can choose the Lyapunov function in (14) to be

$$V_p(x, e) = 4 \,\overline{\nu}_p V_{o,p}(e) + V_{c,p}(x)$$

which leads to

$$\left\langle \nabla V_p(x,e), \begin{pmatrix} A_p x + B_p e \\ F_p e + G_p d \end{pmatrix} \right\rangle \leq -a_p \, V_p(x,e) + \overline{\gamma}_p |d|^2$$

in which

$$\begin{split} a_p &= \min \left\{ a_{c,p}, 0.75 \, a_{o,p} \right\}, \\ \overline{\gamma}_p &= 4 \overline{\nu}_p \overline{\gamma}_{o,p}. \end{split}$$

For the given jump maps at switching times, the maps in (17) can be chosen such that $\rho \equiv 0$, and

$$\chi(s) = \overline{\chi}s, \quad \overline{\chi} = \max_{p,q \in \mathcal{P}} \left\{ \frac{\overline{\nu}_q}{\overline{\nu}_p} \frac{\lambda_{\max}(P_{o,q})}{\lambda_{\min}(P_{o,p})}, \frac{\lambda_{\max}(P_{c,q})}{\lambda_{\min}(P_{c,p})} \right\}. \tag{33}$$

To construct the Lyapunov function W in (29), we let

$$a := \min_{p \in \mathcal{P}} \{a_p\}$$

so that $\psi(s) = as$ satisfies the desired conditions with $c_0 = a$ and $c_1 > 0$ arbitrary. We now choose W such that $W(x, e, p, \tau) = 0$ if (x, e) = 0, and for $(x, e) \neq 0$,

$$\begin{split} W(x,e,p,\tau) &= \exp(2\zeta\tau) \exp\left(\int_1^{V_p(x,e)} \frac{2\,dr}{a\,r}\right) \\ &= \exp(2\zeta\tau) V_p^{2/a}(x,e). \end{split}$$

This construction leads to the following result:

Corollary 1. The switched linear system (9) with subsystems described by (32) is input-to-state stable with respect to the origin and disturbance d if

$$\tau_a > \frac{1}{a} \ln(\overline{\chi}),$$

where $a = \min_{p \in \mathcal{P}} \{a_p\}$, and $\overline{\chi}$ is given in (33).

4. Output Feedback Stabilization with Sampling

In this section, we will use the theoretical results from the previous section to study the problem of feedback stabilization with dynamic output feedback for switched nonlinear systems with time-sampled measurements. Our starting point is a nominal setup where an output feedback controller is already designed for each subsystem. We assume this controller to be robust with respect to measurement errors, and this property is formalized using ISS notion. Next, we implement these controllers when the output measurements sent by the plant, and the control inputs sent by the controller are time-sampled and subjected to zero-order hold between two updates. Our goal is to design the sampling algorithms for control signals and measurements such that the resulting closed-loop system is asymptotically stable. Inspired by the work of (Tanwani et al., 2015), we introduce the dynamic filters and eventbased update rules to design the sampling algorithms.

4.1. Problem Setup

The switched nonlinear plant which we want to stabilize, is described by

$$\dot{x} = f_{c,\sigma}(x, u) \tag{34a}$$

$$y = h_{\sigma}(x), \tag{34b}$$

and the corresponding controller is described by

$$\dot{z} = f_{o,\sigma}(z, u, y) \tag{35a}$$

$$u = k_{\sigma}(z). \tag{35b}$$

Here, for a finite index set \mathcal{P} , $\sigma: \mathbb{R}_{\geq 0} \to \mathcal{P}$ is the switching signal, and we emphasize that the controller is driven by the same switching signal as the plant. For each $p \in \mathcal{P}$, the mappings $f_{c,p}: \mathbb{R}^n \times \mathbb{R}^{n_u} \to \mathbb{R}^n$, $h_p: \mathbb{R}^n \to \mathbb{R}^{n_y}$, $f_{o,p}: \mathbb{R}^n \times \mathbb{R}^{n_u} \times \mathbb{R}^{n_y} \to \mathbb{R}^n$, and $k_p: \mathbb{R}^n \to \mathbb{R}^{n_u}$ are assumed to be continuous on their respective domains. The underlying working principle behind the controller (35) is that the variable z in (35a) acts as a full-state estimate for the variable x governed by (34a). The dynamics of the estimation error, denoted e:=z-x, are assumed to be ISS with respect to measurement errors, as described in (L4) below. Also, the static feedback controller k_p in (35b), $p \in \mathcal{P}$, has the property that the corresponding subsystem $\dot{x} = f_{c,p}(x,k_p(x))$ is ISS with respect to errors in measurement of x, as stated in (L5). With these properties, we can design sampling algorithms for output y and input u,

where the difference between the last updated value of the output (resp. input) and its current value acts as an error in measurement.

In our setup of sampling, we use a zero-order-hold between sampling times so that the inputs and outputs stay constant between two successive updates. Thus, we introduce two additional states in our model, with piecewise constant trajectories, which keep track of the sampled values. These are $x_d: \mathbb{R}_{\geq 0} \to \mathbb{R}^n$ and $z_d: \mathbb{R}_{\geq 0} \to \mathbb{R}^n$, and are to be seen as the time-sampled versions of x and z respectively. We set $y_d = h_{\sigma}(x_d)$ to denote the sampled values of the output that are sent to the controller, and $u_d = k_{\sigma}(z_d)$ to denote the sampled control values, which are sent to the plant. Whenever a new sampled value of the output (resp. control input) is to be sent, we update $x_d^+ = x$ so that $y_d^+ = h_{\sigma}(x)$ (resp. $z_d^+ = z$ so that $u_d = k_{\sigma}(z)$). We emphasize that the plant and controller use the same value of the switching signal σ at all times.

The overall schematic of the closed-loop system is given in Figure 2. We denote the measurement error due to sampling in the output by $d_y := y_d - y$, and $d_z := z_d - z$ denotes the error which appears in the plant dynamics due to sampling of the input. By constructing a Lyapunov function for the augmented system using the cascade principle, we next show that the state of system (34)-(35) converges to the origin for appropriately designed sampling algorithms.

The assumptions imposed on the nominal system (34)-(35) are now listed below:

(A2) For each $p \in \mathcal{P}$, there exists a class \mathcal{K} function $\alpha_{h,p}$ such that the function $h_p : \mathbb{R}^n \to \mathbb{R}^{n_y}$ satisfies:

$$|y| = |h_p(x)| \le \alpha_{h,p}(|x|), \quad \forall p \in \mathcal{P}, \forall x \in \mathbb{R}^n.$$

(L4) There exist continuously differentiable functions $V_{o,p}: \mathbb{R}^n \to \mathbb{R}_{\geq 0}, \ p \in \mathcal{P}, \ \text{class } \mathcal{K} \ \text{function } \alpha_{o,p}, \ \text{and} \ \text{class } \mathcal{K}_{\infty} \ \text{functions } \overline{\alpha}_{o,p}, \ \underline{\alpha}_{o,p} \ \text{such that, for every} \ (x, z, u, y, d_y) \in \mathbb{R}^{2n + n_u + 2n_y},$

$$\underline{\alpha}_{o,p}(|e|) \leq V_{o,p}(e) \leq \overline{\alpha}_{o,p}(|e|) \tag{36a}$$

$$\left\langle \frac{\partial V_{o,p}}{\partial e}(e), f_{c,p}(x,u) - f_{o,p}(z,u,y+d_y) \right\rangle \leq$$

$$-\alpha_{o,p}(V_{o,p}(e)) + \gamma_{o,p}(|d_y|), \tag{36b}$$

where e = z - x.

(L5) There exist continuously differentiable functions $V_{c,p}: \mathbb{R}^n \to \mathbb{R}_{\geq 0}, \ p \in \mathcal{P}, \ \text{class } \mathcal{K} \ \text{functions } \alpha_{c,p}, \gamma_{c,p}, \ \text{and class } \mathcal{K}_{\infty} \ \text{functions } \overline{\alpha}_{c,p}, \ \underline{\alpha}_{c,p} \ \text{such that}$

$$\underline{\alpha}_{c,p}(|x|) \leq V_{c,p}(x) \leq \overline{\alpha}_{c,p}(|x|)$$

$$\left\langle \frac{\partial V_{c,p}}{\partial x}, f_{c,p}(x, k_p(x+e+d_z)) \right\rangle \leq -\alpha_{c,p}(V_{c,p}(x))$$

$$+ \gamma_{c,p}(V_{o,p}(e)) + \gamma_{c,p}(|d_z|),$$
(37a)

hold for every $(x, z, u, y, d_y) \in \mathbb{R}^{2n+n_u+2n_y}$.

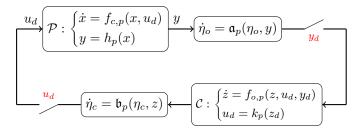


Figure 2: Nonlinear switched systems with sampled-data control.

For the class of plants and controllers satisfying the aforementioned hypotheses, we are now interested in designing the sampling algorithms, and characterizing the class of switching signals which result in an overall asymptotically stable system.

4.2. Sampling Algorithms

As mentioned in the introduction, we are interested in analyzing the stability of the closed-loop system under event-based sampling rules. To do so, the auxiliary signals x_d , z_d are thus modeled as

$$\begin{cases} \dot{x}_d = 0, & \begin{cases} x_d^+ = x, & \text{if event}_1 = \text{true} \\ \dot{z}_d = 0, & z_d^+ = z, & \text{if event}_2 = \text{true} \end{cases}$$

and by setting $y_d = h_{\sigma}(x_d)$ and $u_d = k_{\sigma}(z_d)$, the dynamics of the system with time-sampled inputs and outputs are given by

$$\dot{x} = f_{c,\sigma}(x, u_d) = f_{c,\sigma}(x, k_{\sigma}(z_d))
\dot{z} = f_{o,\sigma}(z, u_d, y_d) = f_{o,\sigma}(z, k_{\sigma}(z_d), h_{\sigma}(x_d)).$$

To define the events at which the outputs and inputs are updated, we introduce the following dynamic filters:

$$\dot{\eta}_o := -\beta_{o,p}(\eta_o) + \rho_{o,p}(|y|) + \gamma_{o,p}(|h_p(x) - h_p(x_d)|)$$
(38a)

$$\dot{\eta}_c := -\beta_{c,p}(\eta_c) + \rho_{c,p}\left(\frac{|z|}{2}\right) + \gamma_{c,p}(|z - z_d|)$$
 (38b)

where, for each $p \in \mathcal{P}$, $\beta_{o,p}$, $\beta_{c,p}$, $\rho_{o,p}$, $\rho_{o,p}$, $\gamma_{o,p}$, $\gamma_{c,p}$ are class \mathcal{K}_{∞} functions, and the initial conditions for η_o and η_c are chosen to be some positive numbers. We say that

$$\begin{cases} \texttt{event}_1 = \texttt{true} & \text{if } |y - y_d| \ge \mu_{o,\sigma}(\eta_o) \\ \texttt{event}_2 = \texttt{true} & \text{if } |z - z_d| \ge \mu_{c,\sigma}(\eta_o), \end{cases}$$
(39)

for some class \mathcal{K}_{∞} functions $\mu_{o,p}, \mu_{c,p}$. Note that event₁ and event₂ may occur at different times, or simultaneously, and that each one corresponds to a different update rule.

4.3. Hybrid Model

Just as we did in Section 3.2, it is convenient to write the entire system with controlled plant dynamics, controller, and sampling algorithms using the framework of hybrid systems. To do so, we first introduce

$$\xi := (x, z, x_d, z_d, \eta_o, \eta_c, p, \tau) \in \mathbb{R}^{4n+2} \times \mathcal{P} \times [0, N_0]$$

to describe the state the closed-loop system. The variables u_d and y_d are obtained by setting $u_d = k_p(z_d)$ and $y_d = h_p(x_d)$.

The flow set is now described as

$$C = \{ \xi : |y - h_p(x_d)| \le \mu_{o,p}(\eta_o) \} \cap \{ \xi : |z - z_d| \le \mu_{c,p}(\eta_c) \}$$
$$\cap \{ \xi : \eta_o \ge 0 \land \eta_c \ge 0 \} \cap \{ \xi : \tau \in [0, N_0] \},$$

so that the state variables evolve according to a differential equation/inclusion when $\xi \in \mathcal{C}$. By construction, jumps in at least one of the state variables occur either due to switching, or when the condition for event_i , i=1,2, holds true. The jump set \mathcal{D} therefore corresponds to the switching event, or the sampling event, and they may not occur at the same time. The jump set is defined as

$$\mathcal{D} = \mathcal{D}_{sw} \cup \mathcal{D}_u \cup \mathcal{D}_y \tag{40}$$

where

$$\mathcal{D}_{sw} := \{ \xi : \tau \in [1, N_0] \}$$

$$\mathcal{D}_u := \{ \xi : |z - z_d| \ge \mu_{c,p}(\eta_c) \}$$

$$\mathcal{D}_y := \{ \xi : |y - h_p(x_d)| \ge \mu_{o,p}(\eta_o) \}$$
(41)

so that the variables may get updated instantaneously when $\xi \in \mathcal{D}$. The evolution equations for the augmented variable ξ can now be described as follows:

$$\xi \in \mathcal{C} : \begin{cases} \dot{x} = f_{c,p}(x, k_p(z_d)) \\ \dot{z} = f_{o,p}(z, k_p(z_d), h_p(x_d)) \\ \dot{x}_d = 0 \\ \dot{z}_d = 0 \\ \dot{\eta}_o = -\beta_{o,p}(\eta_o) + \rho_{o,p}(|y|) + \gamma_{o,p}(|y - y_d|) \\ \dot{\eta}_c = -\beta_{c,p}(\eta_c) + \rho_{c,p}(\frac{|z|}{2}) + \gamma_{c,p}(|z - z_d|) \\ \dot{p} = 0 \\ \dot{\tau} \in [0, \frac{1}{\tau_a}] \end{cases}$$
(42a)

where, in the description of η_o -dynamics, we recall that $y = h_p(x)$ and $y_d(x) = h_p(x_d)$. The jump maps for this system are:

$$\xi \in \mathcal{D}: \left\{ \xi^{+} \in \mathcal{G}(\xi) = \mathcal{G}_{sw}(\xi) \cup \mathcal{G}_{u}(\xi) \cup \mathcal{G}_{y}(\xi) \right. \tag{42b}$$

$$\mathcal{G}_{sw}(\xi) = \begin{bmatrix} x \\ z \\ x \\ \eta_o \\ \eta_c \\ \mathcal{P} \setminus \{p\} \\ \tau - 1 \end{bmatrix}, \mathcal{G}_u(\xi) = \begin{bmatrix} x \\ z \\ x_d \\ z \\ \eta_o \\ \eta_c \\ p \\ \tau \end{bmatrix}, \mathcal{G}_y(\xi) = \begin{bmatrix} x \\ z \\ x \\ z_d \\ \eta_o \\ \eta_c \\ p \\ \tau \end{bmatrix}. (43)$$

It is noted that this system satisfies the basic assumptions required for the existence of solutions (Goebel et al., 2009, Assumption 6.5). We are interested in asymptotic stability of the compact target set \mathcal{A} defined as

$$\mathcal{A} := \left\{0\right\}^{4n+2} \times \mathcal{P} \times [0, N_0] \tag{44}$$

for the hybrid system (42). Our design problem can thus be formulated as follows:

Problem statement: For each $p \in \mathcal{P}$, find the design functions $\beta_{o,p}$, $\beta_{c,p}$, $\rho_{o,p}$, $\rho_{c,p}$, $\gamma_{o,p}$, $\gamma_{c,p}$, $\mu_{o,p}$, $\mu_{c,p}$ appearing in (38), (39), and the lower bound on the average dwell-time τ_a , such that the set \mathcal{A} defined in (44) is globally asymptotically stable for the hybrid system (42).

4.4. Stability Analysis

To state the main result of this section on asymptotic stability of the set \mathcal{A} for system (42), we introduce the design criteria that must be satisfied by the functions introduced in the sampling algorithms (38), (39). Recalling the function ν_p introduced in (14), and the the hypotheses (**L4**), (**L5**), the following conditions are imposed on the functions $\beta_{o,p}$, $\beta_{c,p}$, $\mu_{o,p}$, $\mu_{c,p}$, $\rho_{o,p}$ and $\rho_{c,p}$, for each $p \in \mathcal{P}$:

- **(D1)** $\beta_{o,p}$ and $\beta_{c,p}$ are differentiable functions of class \mathcal{K} .
- (D2) Let θ_p be a function of class \mathcal{K}_{∞} defined as:

$$\theta_p(s) := \alpha_{o,p}^{-1}(2\gamma_{o,p}(s)).$$

Choose the functions $\mu_{o,p}$ and $\mu_{c,p}$ such that, for some $\lambda \in (0,1)$,

$$(\gamma_{o,p} \circ \mu_{o,p})(s)[1 + (\nu_p \circ \theta_p \circ \mu_{o,p})(s)] \le (1 - \lambda)\beta_{o,p}(s)$$
$$2(\gamma_{c,p} \circ \mu_{c,p})(s) \le (1 - \lambda)\beta_{c,p}(s).$$

(D3) The functions $\rho_{o,p}$ and $\rho_{c,p}$ in (38) are positive definite and are chosen such that for each $s \geq 0$:

$$\rho_{o,p} \circ \alpha_{h,p} \circ \underline{\alpha}_{c,p}^{-1}(s) \le 0.5(1-\lambda)\alpha_{c,p}(s)$$

$$\rho_{c,p}(s) \le (1-\lambda)\min\left\{\gamma_{c,p}(s), 0.5\,\alpha_{c,p}(\underline{\alpha}_{c,p}(s))\right\}.$$

It can be guaranteed that there always exists a solution to the inequalities in (D1), (D2), (D3) using the properties of \mathcal{K}_{∞} functions and the results given in (Geiselhart and Wirth, 2014, Corollary 3.2) and (Kellett, 2014).

To state the main result, we recall the definition of ℓ_p from (18) and choose $\tilde{\alpha}_p \in \mathcal{K}_{\infty}$ such that $\tilde{\alpha}_p(s) =$

$$\min \left\{ \beta_{o,p} \left(\frac{1}{4} s \right), \beta_{c,p} \left(\frac{1}{4} s \right), \alpha_{c,p} \left(\frac{1}{4} s \right), \gamma_{c,p} \left(\frac{1}{4} \ell_p^{-1}(s) \right) \right\}.$$

The function ψ is chosen to be a differentiable \mathcal{K}_{∞} function, with $\psi'(0) = 0$, and

$$\psi(s) \le \min\{c_0 s, \widetilde{\alpha}_p(s) \mid p \in \mathcal{P}\}, \quad s \in [0, c_1], \quad (45a)$$

and
$$\psi(s) \le \min\{\widetilde{\alpha}_p(s) \mid p \in \mathcal{P}\}, \quad s \ge c_1, \quad (45b)$$

for some $c_0, c_1 > 0$. Finally, let

$$\chi(s) := \max_{p,q \in \mathcal{P}} \left\{ \ell_q \circ \overline{\alpha}_{o,q} \circ \underline{\alpha}_p^{-1}(s) + \overline{\alpha}_{c,q} \circ \underline{\alpha}_p^{-1}(s) \right\}. \tag{46}$$

Theorem 2. Consider the system (42) and assume that (A2), (L3), (L4), (L5) hold. Suppose that the sampling algorithms (38) and (39) are designed such that (D1), (D2), (D3) are satisfied for some $\lambda \in (0,1)$. If the average dwell-time τ_a satisfies:

$$\lambda \tau_a > \zeta^* := \sup_{s>0} \int_s^{\chi(s)} \frac{dr}{\psi(r)}$$
 (47)

for ψ and χ given in (45) and (46), then the set A given in (44) is globally asymptotically stable for the system (42).

The fundamental idea behind the proof is to first construct a weak Lyapunov function W for system (42) with respect to set \mathcal{A} in (44). Using additional arguments based on cascade hybrid systems, it is then shown that the solutions along which the derivative of W is possibly zero, also converge to the set \mathcal{A} .

Proof. We start with the function

$$W(\xi) := \exp(\zeta \tau) \varphi(V_p(x, e) + \eta_o + \eta_c) \tag{48}$$

where $\zeta \in (\zeta^*, \lambda \tau_a)$ with ζ^* given in (47), φ is defined in (27), and the function V_p is defined as in (14). It is noted that W does not involve the variables $(x_d, z_d) \in \mathbb{R}^{2n}$, so we only have the bounds

$$\underline{\alpha}(|\xi|_{\mathcal{A}_c}) \le W(\xi) \le \overline{\alpha}(|\xi|_{\mathcal{A}_c}) \tag{49}$$

where $\mathcal{A}_c := \{0\}^{2n} \times \mathbb{R}^{2n} \times \mathbb{R}^2_{\geq 0} \times \mathcal{P} \times [0, N_0]$, and $\underline{\alpha}$, $\overline{\alpha}$ are some class \mathcal{K}_{∞} functions. When $\xi \in \mathcal{C}$, we have the derivative of W:

$$\dot{W} = W \left[2\zeta \dot{\tau} + \frac{2}{\psi(V_p + \eta_o + \eta_c)} \left(\dot{V}_p + \dot{\eta}_o + \dot{\eta}_c \right) \right]. \quad (50)$$

To show that \dot{W} is bounded by a negative semidefinite function, we compute bounds on \dot{V}_p , $\dot{\eta}_o$ and $\dot{\eta}_c$ in the flow set \mathcal{C} .

Using (L3), (L4), (L5) and the inequality derived in (20), we obtain

$$\dot{V}_{p} \leq -\frac{1}{2}\nu_{p}(V_{o,p}(e))\alpha_{o,p}(V_{o,p}(e))
+ \nu_{p}(\theta_{p}(|y-y_{d}|))\gamma_{o,p}(|y-y_{d}|)
- \alpha_{c,p}(V_{c,p}(x)) + \gamma_{c,p}(V_{o,p}(e)) + \gamma_{c,p}(|z-z_{d}|).$$
(51)

It follows from the definition of ν_p , sampling condition (39), and (51) that

$$\dot{V}_{p} \leq -\gamma_{c,p}(V_{o,p}(e)) + \nu_{p}(\theta_{p}(\mu_{o,p}(\eta_{o})))\gamma_{o,p}(\mu_{o,p}(\eta_{o}))
- \alpha_{c,p}(V_{c,p}(x)) + \gamma_{c,p}(\mu_{c,p}(\eta_{c})).$$
(52)

The derivative of η_o is seen to satisfy

$$\dot{\eta}_{o} \leq -\beta_{o,p}(\eta_{o}) + \rho_{o,p} \circ \alpha_{h,p}(|x|) + \gamma_{o,p}(\mu_{o,p}(\eta_{o}))
\leq -\beta_{o,p}(\eta_{o}) + \rho_{o,p} \circ \alpha_{h,p} \circ \underline{\alpha}_{c,p}^{-1}(V_{c,p}(x)) + \gamma_{o,p}(\mu_{o,p}(\eta_{o})).$$
(53)

The derivative of η_c can be bounded as follows:

$$\dot{\eta}_{c} \leq -\beta_{c,p}(\eta_{c}) + \rho_{c,p} \left(\frac{|x| + |e|}{2} \right) + \gamma_{c,p}(\mu_{c,p}(\eta_{c}))
\leq -\beta_{c,p}(\eta_{c}) + \rho_{c,p} \circ \underline{\alpha}_{c,p}^{-1}(V_{c,p}(x))
+ \rho_{c,p} \circ \underline{\alpha}_{o,p}^{-1}(V_{o,p}(e)) + \gamma_{c,p}(\mu_{c,p}(\eta_{c})).$$
(54)

Now combining (52), (53), (54), and using the inequalities given in (D1), (D2), and (D3), we get

$$\dot{V}_p + \dot{\eta}_o + \dot{\eta}_c \leq -\lambda [\beta_{o,p}(\eta_o) + \beta_{c,p}(\eta_c) + \alpha_{c,p}(V_{c,p}(x))
+ \gamma_{c,p}^{-1} \circ \ell_p^{-1}(\ell_p(V_{o,p}(e)))]
\leq -\lambda \tilde{\alpha}_p(V_p + \eta_o + \eta_c).$$

Substituting this expression in (50), and using the definition of ψ in (45), we obtain

$$\dot{W} \leq W \left[2\zeta \dot{\tau} - \frac{2\lambda \widetilde{\alpha}_p (V_p + \eta_o + \eta_c)}{\psi (V_p + \eta_o + \eta_c)} \right]$$

$$\leq W (2\zeta \dot{\tau} - 2\lambda)$$

$$\leq W \left(\frac{2\zeta}{\tau_a} - 2\lambda \right)$$

$$= W \left(\frac{2(\zeta - \lambda \tau_a)}{\tau_a} \right),$$

which is the desired inequality for \dot{W} over the flow set since we chose $\zeta < \lambda \tau_a$.

When $\xi \in \mathcal{D}$, we calculate the maximum of $W(\xi^+)$, over the set $\mathcal{G}(\xi) \ni g = \xi^+$, as follows:

$$\max_{g \in \mathcal{G}(\xi)} W(g) = \max_{g \in \mathcal{G}(\xi)} \exp(2\zeta \tau^{+}) \varphi \left(V_{p^{+}}(x^{+}, e^{+}) + \eta_{o}^{+} + \eta_{c}^{+} \right)$$
$$= \max_{p^{+} \in \mathcal{P}} \exp(2\zeta \tau - 2\zeta) \varphi \left(V_{p^{+}}(x, e) + \eta_{o} + \eta_{c} \right).$$

To get a bound on the right-hand side in terms of the value of ξ just prior to the jump, we recall that the function χ introduced in (46) satisfies

$$V_q \le \chi(V_p), \quad \forall p, q \in \mathcal{P}.$$

Moreover, from the definition of φ , it follows that

$$\varphi\left(V_{p^+} + \eta_o + \eta_c\right) \le \exp\left(\int_1^{\chi(V_p) + \eta_o + \eta_c} \frac{2\,dr}{\psi(r)}\right).$$

We then observe that

$$\int_{1}^{\chi(V_{p})+\eta_{o}+\eta_{c}} \frac{2 dr}{\psi(r)} = \int_{V_{p}+\eta_{o}+\eta_{c}}^{\chi(V_{p})+\eta_{o}+\eta_{c}} \frac{2 dr}{\psi(r)} + \int_{1}^{V_{p}+\eta_{o}+\eta_{c}} \frac{2 dr}{\psi(r)}.$$
(55)

Since $\eta_o + \eta_c \ge 0$ and $\frac{1}{\psi(r)}$ is decreasing, we have:

$$\int_{V_p+\eta_o+\eta_c}^{\chi(V_p)+\eta_o+\eta_c} \frac{2\,dr}{\psi(r)} \leq \int_{V_p}^{\chi(V_p)} \frac{2\,dr}{\psi(r)} \leq 2\zeta^*,$$

so that

$$\int_{1}^{\chi(V_p) + \eta_o + \eta_c} \frac{2 \, dr}{\psi(r)} \le 2\zeta^* + \int_{1}^{V_p + \eta_o + \eta_c} \frac{2 \, dr}{\psi(r)}.$$

Thus, the value of W after each jump is bounded as

$$\max_{g \in \mathcal{G}(\xi)} W(g) \le \exp(-2(\zeta - \zeta^*)) \exp(2\zeta \tau) \varphi(V_p + \eta_o + \eta_c)$$

$$\le \exp(-2(\zeta - \zeta^*)) W(\xi) \qquad \forall \, \xi \in \mathcal{D}. \tag{56}$$

Because of the bounds in (49), it thus follows that ξ converges asymptotically to the set \mathcal{A}_c . To conclude further that (x_d, z_d) also converge to $\{0\}^{2n}$, one can invoke the arguments based on the LaSalle's invariance principle (Goebel et al., 2012, Corollary 8.9(ii)), and cascaded hybrid systems (Goebel et al., 2009, Corollary 19). Following the same recipe as in (Tanwani et al., 2015, Proof of Theorem 1), we next show that the set \mathcal{A} of the closed-loop system is a globally asymptotically stable (GAS) for system (42):

Step 1 – Pre-GAS of $\{0\}$ for truncated systems: For a fixed initial condition, there exist compact set $\mathcal{M}_1 \subset \mathbb{R}^{2n}$, $\mathcal{M}_2 \subset \mathbb{R}^2 \times \mathcal{P} \times \mathbb{R}$ such that $(x, z) \in \mathcal{M}_1$ and $(\eta_o, \eta_c, p, \tau) \in$ \mathcal{M}_2 . Recalling that z_d and x_d remain constant during flows, and are reset to z and x, which belong to a compact set, there exists a compact set \mathcal{M}_d such that $(x_d, z_d) \in$ \mathcal{M}_d . Consider the truncation of system (42) to the set $\mathcal{M} := \mathbb{R}^{2n} \times \mathcal{M}_d \times \mathbb{R}^2_{>0} \times \mathcal{P} \times [0, N_0]$, which has the flow set $\mathcal{C}_{\mathcal{M}} := \mathcal{C} \cap \mathcal{M}$, the jump set $\mathcal{D}_{\mathcal{M}} := \mathcal{D}_{\mathcal{M}} \cap \mathcal{M}$. For this truncated system, it follows from the invariance principle (Goebel et al., 2012, Corollary 8.9(ii)) that the set $A_1 :=$ $\{0\}^{2n} \times \mathcal{M}_d \times \{0\}^2 \times \mathcal{P} \times [0, N_0]$ is pre-GAS. We next invoke the stability result for cascaded hybrid systems (Goebel et al., 2009, Corollary 19) to claim that the set A in (44) is pre-GAS for the truncated system. Indeed, for every system trajectory contained in A_1 , we have $\eta_o = \eta_c = 0$, and from the definition of the sets C and D, we must then have $x_d = 0$ and $z_d = 0$.

Step 2 – Bounded solutions and Pre-GAS of $\{0\}$ for (42): As shown in the first step, for each initial condition, there exist compact sets \mathcal{M}_1 , \mathcal{M}_2 and \mathcal{M}_d such that ξ is contained in the compact set $\mathcal{M}_1 \times \mathcal{M}_d \times \mathcal{M}_2$ for all times. Boundedness of the solutions now allows us to conclude that \mathcal{A} is pre-GAS for the original system (42). To see this, assume that there exists a solution for which $(x, z, x_d, z_d, \eta_o, \eta_c)$ does not converge to $\{0\}$. Since all solutions are bounded, there exists a compact set \mathcal{M}_d such that this bounded solution eventually coincides with the solution of the system truncated to $\mathbb{R}^{2n} \times \mathcal{M}_d \times \mathbb{R}^2_{\geq 0} \times \mathcal{P} \times [0, N_0]$. But, every solution of the truncated system must converge to \mathcal{A} . Hence, for (42), a bounded solution not converging to \mathcal{A} cannot exist, proving that \mathcal{A} is pre-GAS.

Step 3 – $\{0\}$ is GAS for (42): To move from preasymptotic stability to asymptotic stability of the compact set \mathcal{A} , we next show that every solution of (42) is forward complete. This is seen due to the fact that for

each $\xi \in \mathcal{C} \setminus \mathcal{D}$, the solutions would always continue to flow. Moreover, after each jump the states are reset to the set $\mathcal{C} \cup \mathcal{D}$, making it possible to extend the time domain for the solutions either by jump or flow. Hence, each solution of the system is forward complete, proving that the set \mathcal{A} is GAS.

Remark 5. For implementation purposes, it is important to show that, for event-based sampling, there is a uniform lower bound on the minimal inter-sampling time between two consecutive sampling instants. For the algorithms employed in this article, such a lower bound has been obtained for nonswitched dynamical systems in (Tanwani et al., 2015, Theorem 2) under certain additional assumptions on the functions appearing in the dynamic filters (38). For switched systems, when working under the slow switching assumption like dwell-time or average dwell-time, it suffices to have such a lower bound for an individual dynamical subsystem since this guarantees there will be no accumulations of jump events if the switching signal has no accumulation of switches.

5. Example and Simulation Result

As an illustration of Theorem 2, we consider an academic example of a switched system with two modes. The first subsystem is described by linear dynamics as follows:

$$p = 1: \begin{cases} \dot{x} = A_1 x + B_1 u \\ y = C_1 x \end{cases}$$
 (57)

The feedback controller related to this subsystem is:

$$p = 1: \begin{cases} \dot{z} = A_1 z + B_1 u_d + L_1 (y - C_1 z) \\ u = -K_1 z, \end{cases}$$
 (58)

where we choose $A_1 = \begin{bmatrix} 0.5 & -1 \\ 0 & 0.5 \end{bmatrix}$, $B_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$, $C_1 = \begin{bmatrix} 1 & 0 \end{bmatrix}$, $L_1 = \begin{bmatrix} 3.5 \\ -3 \end{bmatrix}$ and $K_1 = \begin{bmatrix} -1.5 & 2.5 \end{bmatrix}$.

The second subsystem has nonlinear dynamics described by:

$$p = 2: \begin{cases} \dot{x}_1 = x_2 + 0.25|x_1| \\ \dot{x}_2 = \text{sat}(x_1) + u_d \\ y = x_1. \end{cases}$$
 (59)

The notation sat denotes the saturation function $\operatorname{sat}(x_1) = \min\{1, \max\{-1, x_1\}\}$. The corresponding feedback controller is:

$$p = 2: \begin{cases} \dot{z}_1 = z_2 + 0.25|y| + l_1(y - z_1) \\ \dot{z}_2 = \text{sat}(y) + u_d + l_2(y - z_1) \\ u = \text{sat}(z_1) + K_2 z, \end{cases}$$
(60)

where we choose $K_2 = \begin{bmatrix} -2 & -2 \end{bmatrix}$ and $L_2 = \begin{bmatrix} l_1 \\ l_2 \end{bmatrix} = \begin{bmatrix} 2 \\ 2 \end{bmatrix}$.

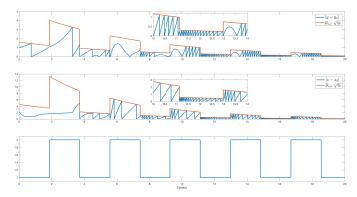


Figure 3: Simulation results: In the top plot, whenever $|y(t)-y_d(t)|$ reaches the sampling threshold $\overline{\mu}_{o,\sigma}\sqrt{\eta_o(t)}$, y_d is updated. The middle plot shows $\overline{\mu}_{c,\sigma}\sqrt{\eta_c(t)}$ and $|z(t)-z_d(t)|$. The bottom plot shows the switching signal used in the simulations.

For both subsystems, we introduce the same form of Lyapunov function: $V_{c,p}(x) = x^T P_{c,p} x$ and $V_{o,p}(e) = e^T P_{o,p} e$. Since the controller is driven by the sampled output y_d , we have for each $p \in \{1,2\}$:

$$\dot{V}_{o,p} \le -a_{o,p}V_{o,p}(e) + \overline{\gamma}_{o,p} |y - y_d|^2$$

where

$$a_{o,p} = \frac{\lambda_{\min}(Q_{o,p})}{2\lambda_{\max}(P_{o,p})} \quad \text{ and } \quad \overline{\gamma}_{o,p} = \frac{2\|P_{o,p}L_p\|^2}{\lambda_{\min}(Q_{o,p})}.$$

Similarly, for each $p \in \{1, 2\}$:

$$\dot{V}_{c,p} \leq -a_{c,p}V_{c,p}(x) + \overline{\gamma}_{c,p}V_{o,p}(e) + \overline{\gamma}_{c,p}\left|z - z_d\right|^2,$$

where

$$a_{c,p} = \frac{\lambda_{\min}(Q_{c,p})}{2\lambda_{\max}(P_{c,p})},$$

and

$$\overline{\gamma}_{c,p} = \frac{4\|P_{c,p}B_pK_p\|^2}{\lambda_{\min}(Q_{c,p})} \max\left\{1, \frac{1}{\lambda_{\min}(P_{o,p})}\right\}.$$

For the dynamic filters, we choose

$$\begin{split} \dot{\eta}_o &= -a_{o,p}\eta_o + \overline{\rho}_{o,p}|y|^2 + \overline{\gamma}_{o,p}|y - y_d|^2 \\ \dot{\eta}_c &= -a_{c,p}\eta_c + \overline{\rho}_{c,p}\frac{|z|^2}{4} + \overline{\gamma}_{c,p}|z - z_d|^2, \end{split}$$

where we let

$$\overline{\rho}_{o,p} := \frac{(1 - 2\varepsilon)\lambda_{\min}(P_{o,p})}{\|C_p\|^2},$$

$$\overline{\rho}_{c,p} := \min\left\{(1 - \varepsilon)\overline{\gamma}_{c,p}, \varepsilon a_{c,p}\lambda_{\min}(P_{c,p})\right\}$$

for some small $\varepsilon \in (0,0.5)$. The jump set which describes the conditions when the sampled values get updated or when there is a switching event occurs, is defined as follows:

$$\mathcal{D} = \left\{ \xi : |y - y_d| \ge \overline{\mu}_{o,p} \sqrt{\eta_o} \right\} \cup \left\{ \xi : |z - z_d| \ge \overline{\mu}_{c,p} \sqrt{\eta_c} \right\} \cup \left\{ \tau \in [1, N_0] \right\}$$

where
$$\overline{\mu}_{o,p}:=\frac{(1-arepsilon)lpha_{o,p}}{(1+\overline{
u}_p)\overline{\gamma}_{o,p}}$$
 and $\overline{\mu}_{c,p}:=\frac{(1-arepsilon)lpha_{c,p}}{2\overline{\gamma}_{c,p}}$, with

$$\overline{\nu}_p = \frac{4\overline{\gamma}_{c,p}}{\lambda_{\min}(P_{o,p})}.$$

The function χ can thus be defined as $\chi(s) = \overline{\chi}s$, where

$$\overline{\chi} = \max_{p,q \in \{1,2\}} \left\{ \frac{\overline{\nu}_p \lambda_{\max}(P_{o,p}) + \lambda_{\max}(P_{c,p})}{\overline{\nu}_q \lambda_{\min}(P_{o,q}) + \lambda_{\min}(P_{c,q})} \right\}.$$

It follows from Theorem 2 that, if the average dwell-time τ_a satisfies that:

 $au_a > \frac{\ln \overline{\chi}}{\varepsilon},$

then we have the asymptotic stability of the origin for (x,z) system. The simulation results reported in Figure 3 indeed show the convergence of (y,z,η_o,η_c) to the origin.

6. Conclusion

In this article, the construction of ISS Lyapunov functions is considered for switched nonlinear systems in cascade configuration. The stability analysis for the resulting hybrid systems is carried out under an average dwell-time condition on the switching signal, and an asymptotic ratio condition for establishing ISS. The results pave the path for studying the stabilization of switched systems with dynamic output feedback. A Lyapunov function similar to the one used for non-sampled ISS system is constructed to design the sampling algorithms and for the analysis of the closed-loop hybrid systems with sampled measurements. The results are illustrated with the help of examples and simulations. One of the limitations of the dynamic output feedback problem considered in this paper is that the controller requires exact knowledge of the switching signal. It is of interest to develop theoretical tools when there is mismatch in the switching signal between the plant and the controller. One can also consider additional measurement errors, for example, due to quantization of output and input in space, as done in (Tanwani et al., 2016). One could also potentially study the affect of random uncertainties, on top of event-based samples, as has been recently proposed in (Tanwani and Tell, 2017).

Acknowledgements

The authors would like to thank Dr. Guosong Yang for his useful comments on an earlier version of this paper.

References

References

Cai, C., Teel, A., 2009. Characterizations of input-to-state stability for hybrid systems. Systems & Control Letters 58, 47–53.

Cai, C., Teel, A., 2013. Robust input-to-state stability for hybrid systems. SIAM Journal on Control and Optimization 51, 1651– 1678

- Dashkovskiy, S., Mironchenko, A., 2013. Input-to-state stability of nonlinear impulsive systems. SIAM J. Control & Optim. 51, 1962 – 1987.
- Dashkovskiy, S., Rüffer, B., Wirth, F., 2010. Small gain theorems for large scale systems and construction of iss lyapunov functions. SIAM Journal on Control and Optimization 48, 4089–4118.
- Geiselhart, R., Wirth, F., 2014. Solving iterative functional equations for a class of piecewise linear \mathcal{K}_{∞} -functions. Journal of Mathematical Analysis and Applications 411, 652 664.
- Goebel, R., Sanfelice, R., Teel, A., 2009. Hybrid dynamical systems. IEEE Control Systems Magazine 29, 28–93.
- Goebel, R., Sanfelice, R.G., Teel, A.R., 2012. Hybrid Dynamical Systems: Modeling, Stability, and Robustness. Princeton University Press.
- Heemels, W., Johansson, K., Tabuada, P., 2012. An introduction to event-triggered and self-triggered control, in: Proc. 51st IEEE Conf. Decision & Control, Maui, HI. pp. 3270–3285.
- Hespanha, J.P., Liberzon, D., Teel, A., 2008. Lyapunov conditions for input-to-state stability of impulsive systems. Automatica 44, 2735 – 2744.
- Hespanha, J.P., Morse, A., 1999. Stability of switched systems with average dwell-time, in: Proceedings of 38th IEEE Conf. on Decision and Control, pp. 2655–2660.
- Ito, H., 2006. State-dependent scaling problems and stability of interconnected iISS and ISS systems. IEEE Transactions on Automatic Control 51, 1626–1643.
- Jiang, Z.P., Teel, A.R., Praly, L., 1994. Small-gain theorem for iss systems and applications. Mathematics of Control, Signals, and Systems 7, 95–120.
- Kellett, C., 2014. A compendium of comparison function results. Mathematics of Control, Signals, and Systems 26, 339–374. doi:10.1007/s00498-014-0128-8.
- Khalil, H.K., 2002. Nonlinear Systems. 3rd ed., Prentice-Hall, Inc. Liberzon, D., 2003. Switching in Systems and Control. Birkhäuser, Boston.
- Liberzon, D., Nesic, D., Teel, A., 2014. Lyapunov-based small-gain theorems for hybrid systems. IEEE Transactions on Automatic Control 59, 1395 – 1410.
- Liberzon, D., Shim, H., 2015. An asymptotic ratio characterization of input-to-state stability. IEEE Transactions on Automatic Control 60, 3401–3404.
- Persis, C.D., Santis, R.D., Morse, A., 2003. Switched nonlinear systems with state-dependent dwell-time switching logic. Systems & Control Letters 50, 291 302.
- Praly, L., Wang, Y., 1996. Stabilization in spite of matched unmodeled dynamics and an equivalent definition of input-to-state stability. Mathematics of Control, Signals, and Systems 9, 1–33.
- Sanfelice, R.G., 2014. Input-output-to-state stability tools for hybrid systems and their interconnections. IEEE Transactions on Automatic Control 59, 1360–1366.
- Sontag, E., Teel, A., 1995. Changing supply functions in input/state stable systems. IEEE Transactions on Automatic Control 40, 1476-1478
- Sontag, E.D., 1989. Smooth stabilization implies coprime factorization. IEEE Transactions on Automatic Control 34, 435 – 443.
- Tanwani, A., Chatterjee, D., Liberzon, D., 2018. Stabilization of deterministic control systems under random sampling: Overview and recent developments, in: Başar, T. (Ed.), Uncertainty in Complex Networked Systems. SpringerNature, Switzerland. Systems & Control: Foundations & Applications, pp. 209–246.
- Tanwani, A., Prieur, C., Fiacchini, M., 2016. Observer-based feed-back stabilization of linear systems with event-triggered sampling and dynamic quantization. Systems & Control Letters 94, 46 56.
- Tanwani, A., Teel, A., Prieur, C., 2015. On using norm estimators for event-triggered control with dynamic output feedback, in: Proc. 54th IEEE Conf. on Decision and Control, pp. 5500–5505.
- Tanwani, A., Tell, A., 2017. Stabilization with event-driven controllers over a digital communication channel with random transmissions, in: Proc. 56th IEEE Conf. on Decision and Control, pp. 6063–6068.

- Vu, L., Chatterjee, D., Liberzon, D., 2007. Input-to-state stability of switched systems and switching adaptive control. Automatica 43, 639 – 646.
- Yang, G., Liberzon, D., 2015. A lyapunov-based small-gain theorem for interconnected switched systems. Systems & Control Letters 78, 47–54.
- Yang, G., Liberzon, D., Mironchenko, A., 2016. Analysis of different Lyapunov function constructions for interconnected hybrid system, in: Proc. 55th IEEE Conference on Decision and Control, pp. 475–480.
- Zhang, G., Tanwani, A., 2019. ISS Lyapunov functions for cascade switched systems and sampled-data control. Automatica 105, 216– 227.
- Zhang, G.X., Tanwani, A., 2018. Stabilization of switched nonlinear systems with dynamic output feedback and event-based sampling, in: Proc. of American Control Conference, pp. 2198 2203.