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▶ To cite this version:

Bruno Macke, Bernard Ségard. Material slow and fast light in a zero-dispersion configuration. Journal of the Optical Society of America B, 2020, 37 (7), pp.2080. 10.1364/JOSAB.389480 . hal-02093839v2

HAL Id: hal-02093839 https://hal.science/hal-02093839v2

Submitted on 17 Sep 2020

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Material slow and fast light in a zero-dispersion configuration

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(Dated: September 8, 2020)

We study the propagation of light pulses in an absorbing medium when the frequency of their carrier coincides with a zero of the refractive index dispersion. Although slow light and, *a fortiori*, fast light are not expected in such conditions, we show that both can be obtained by selecting particular phase-components of the transmitted field. Analytical expressions of the resulting signals are obtained by a procedure of periodic continuation of the incident pulse and a proof of principle of the predicted phenomena is performed by means of a very simple electrical network, the transfer function of which mimics that of the medium.

I. INTRODUCTION

The one-dimensional propagation of coherent light pulses with a slowly varying envelope through a linear medium is usually analyzed by means of the group velocity [1]. At the optical frequency ω it reads $v_g(\omega) = c/[n(\omega) + \omega (dn/d\omega)]$ where c is the light velocity in vacuum, $n(\omega)$ is the refractive index and its derivative $dn/d\omega$ is the refractive index dispersion. When the group velocity and the medium transmission are both constant over the whole pulse spectrum, the envelope travels undistorted at the velocity $v_g(\omega_c) = c/[n(\omega_c) + \omega_c (dn/d\omega_c)]$ where ω_c is the pulse carrier frequency and $dn/d\omega_c$ is a short-hand notation of $dn/d\omega$ for $\omega = \omega_c$. When the above-mentioned double condition is not fulfilled, group velocity dispersion and transmission variation over the pulse spectrum result in a pulse reshaping.

The refractive index dispersion $dn/d\omega_c$ can take very large positive or negative values when the carrier frequency ω_c of the pulses is equal or close to the frequency ω_0 of a narrow and well-marked dip or peak of the medium transmission. When $dn/d\omega_c > 0$ (normal dispersion), the group velocity is then very small compared to the phase velocity $c/[n(\omega_c)]$ (slow light regime) while it becomes very large or even negative (fast last regime) when $dn/d\omega_c < 0$ (anomalous dispersion). The principle of causality implies that the two regimes can be obtained with a same medium, depending on the probe frequency ω_c [2]. We examine in the present article what occurs when ω_c is such that $dn/d\omega_c = 0$ (zero-dispersion configuration). Neither slow light nor fast light are expected in this case. We will however show that both can be observed by post-selecting particular phase components of the transmitted field, in analogy with the experiments involving post-selection of the field polarization [3–6]. We specifically consider the reference case of a medium with a narrow absorption line. Convincing demonstrations of slow light [7, 8] and fast light [9–13] have been performed with this system. The arrangement of our paper is as follows. In Section II, we give the transfer functions for the electric field and for its relevant phase components. The

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envelopes of the corresponding transmitted pulses are determined in Section III and we give in Section IV a proof of principle of the predicted phenomena by means of a very simple electrical network. We conclude in Section V by summarizing our main results.

II. TRANSFER FUNCTIONS OF THE MEDIUM

Slow or fast light effects can be directly evidenced by comparing the pulse transmitted by the medium (probe) to that which would be transmitted in vacuum (reference). The transfer functions linking the Fourier transforms of the corresponding fields to that of the incident field read, respectively, $\exp\left[-i\tilde{n}(\omega)\omega\ell/c\right]$ and $\exp\left[-i\omega\ell/c\right]$, where $\tilde{n}(\omega)$ is the complex refractive index of the medium and ℓ its thickness [14]. In the time domain, the reference field is simply delayed by the luminal transit time ℓ/c and this naturally leads to use times retarded by ℓ/c to describe the transmitted fields. In this *retarded time picture* the transfer function for the probe becomes

$$H(\omega) = \exp\{-i[\tilde{n}(\omega) - 1]\omega\ell/c\}$$
(1)

and the reference field is equal to the incident field in real time.

For the sake of simplicity, we consider a dilute medium with a Lorentzian absorption line of half width at half maximum γ very small compared to the resonance frequency ω_0 (narrow resonance limit). Under these conditions, the complex susceptibility $\chi(\omega)$ of the medium is such that $|\chi(\omega)| \ll 1 \ \forall \omega$ and, in SI units, the complex refractivity $\tilde{n}(\omega) - 1$ is reduced to

$$\widetilde{n}(\omega) - 1 = \sqrt{1 + \chi(\omega)} - 1 \approx \chi(\omega)/2$$
 (2)

In addition, the susceptibility is only significant when $|\omega \pm \omega_0| \ll \omega_0$. For $\omega > 0$, the classical Lorentz model [1, 15] and the semi-classical model of two-levels atoms [16, 17] both lead to a relation of the form

$$\chi(\omega) \approx \frac{\gamma'}{\omega_0 - \omega + i\gamma} \tag{3}$$



Figure 1: Modulus (on the left) and phase (on the right) of the field transfer function vs. the optical detuning $(\omega - \omega_0)$ for $\alpha \ell = \pi/2$. The vertical dash-dotted line indicates the detuning $(\omega_c - \omega_0) = \gamma$ of the pulse carrier considered in the article.

with $0 < \gamma' \ll \gamma$. Eqs. (1, 2, 3) finally yield

$$H(\omega) = \exp\left[-\frac{\alpha\ell}{1+i(\omega-\omega_0)/\gamma}\right].$$
 (4)

where $\alpha = \omega_0 \gamma'/(2\gamma c)$ naturally appears as the medium absorption coefficient on resonance for the amplitude. By denoting $\Phi(\omega)$ the phase of $H(\omega)$, the group advance (the opposite of the group delay) is given by the relation $a_g(\omega_c) = d\Phi/d\omega \mid_{\omega=\omega_c} [14]$ and Eq.4 yields

$$a_g(\omega_c) = \left(\frac{\alpha\ell}{\gamma}\right) \left[\frac{1 - \Delta^2/\gamma^2}{\left(1 + \Delta^2/\gamma^2\right)^2}\right]$$
(5)

where $\Delta = (\omega_c - \omega_0)$ is the detuning of the pulse carrier frequency from resonance. The group advance attains its maximum $a_g(\omega_0) = \alpha \ell / \gamma$ for $\Delta = 0$, is positive (fast light regime) when $|\Delta| < \gamma$ and negative (slow light regime) when $|\Delta| > \gamma$. It cancels when $\Delta = \pm \gamma$. We will consider in the following the case where $\Delta = \gamma$. Quite similar results are obtained when $\Delta = -\gamma$. Figure 1 shows the amplitude transmission $|H(\omega)|$ and phase $\Phi(\omega)$ as functions of the detuning $\omega - \omega_0$ in the reference case $\alpha \ell = \pi/2$. The vertical dash-dotted line indicates the carrier frequency $\omega_c = \omega_0 + \gamma$ considered in the following.

The transfer function $H_{\Delta}(\Omega)$ for the pulse envelope is derived from Eq.(4) by passing in a frame rotating at the frequency $\omega_c[16]$. We get

$$H_{\Delta}(\Omega) = \exp\left[-\frac{\alpha\ell}{1+i\left(\Omega+\Delta\right)/\gamma}\right] \tag{6}$$

where $\Omega = \omega - \omega_c$ with $|\Omega| \ll \omega_c$. On exact resonance $(\Delta = 0), H_0(\Omega) = H_0^*(-\Omega)$, where the asterisk indicates complex conjugate. The corresponding impulse response $h_0(t)$, inverse Fourier transform of $H_0(\Omega)$ [14], is then real [18, 19]. If the envelope x(t) of the incident pulse is real (unchirped pulse) as assumed in the following, the envelope y(t) of the transmitted pulse will be also real or, otherwise said, probe and reference fields will be

in phase. In addition the relation $H_0(\Omega) = H_0^*(-\Omega)$ implies that the amplitude transmission $|H_0(\Omega)|$ and the phase $\Phi_0(\Omega)$ of $H_0(\Omega)$ are, respectively, even and odd functions of Ω . As discussed in [19], the group advance $a_{g0} = (d\Phi_0/d\Omega)|_{\Omega=0} = \alpha \ell/\gamma$ can then be identified to the advance of the center-of-gravity (COG) of y(t) over that of x(t) while $H_0(0) = e^{-\alpha \ell}$ is the ratio of the two envelopes areas [19]. These results hold whatever the pulse distortion is.

In the zero-dispersion configuration $(\Delta = \gamma)$, $H_{\gamma}(\Omega) \neq H_{\gamma}^{*}(-\Omega)$ and the impulse response $h_{\gamma}(t)$ is complex. The envelope y(t) of the transmitted pulse is also complex. Its real part $y_{I}(t)$ and imaginary part $y_{Q}(t)$ are then the envelopes of the components of the probe field, respectively, in phase (I) and in quadrature (Q) with the reference field. The impulse responses $h_{I,Q}(t)$ linking the envelopes $y_{I,Q}(t)$ of these two components to that of the reference pulse read:

$$h_I(t) = Re[h_{\gamma}(t)] = \frac{1}{2} \left[h_{\gamma}(t) + h_{\gamma}^*(t) \right]$$
 (7)

$$h_Q(t) = Im[h_\gamma(t)] = \frac{1}{2i} \left[h_\gamma(t) - h_\gamma^*(t) \right]$$
 (8)

Coming back in the frequency domain, we get the corresponding transfer functions

$$H_{I}(\Omega) = \frac{1}{2} \left[H_{\gamma}(\Omega) + H_{\gamma}^{*}(-\Omega) \right]$$
(9)

$$H_Q(\Omega) = \frac{1}{2i} \left[H_\gamma(\Omega) - H_\gamma^*(-\Omega) \right]$$
(10)

These transfer functions are closely related to those encountered in Ref.[6] where the post-selection was made on the field polarization. As $H_0(\Omega)$ in the resonant case, $H_{I,Q}(\Omega) = H_{I,Q}^*(-\Omega)$ and the modulus $|H_{I,Q}(\Omega)|$ and phase $\Phi_{I,Q}(\Omega)$ are, respectively, even and odd functions of Ω . As made in the resonant case, we derive from Eqs.(6,9, 10) the advance $a_{gI,gQ}$ of the COG of $y_{I,Q}(t)$ over that of x(t) and the ratio $|H_{I,Q}(0)|$ of its algebraic area over that of x(t). We get

$$\gamma a_{gI} = \frac{\alpha \ell}{2} \tan\left(\frac{\alpha \ell}{2}\right) \tag{11}$$

$$H_I(0) = \cos\left(\frac{\alpha\ell}{2}\right) \exp\left(-\frac{\alpha\ell}{2}\right) \tag{12}$$

$$\gamma a_{gQ} = -\frac{\alpha \ell}{2} \cot\left(\frac{\alpha \ell}{2}\right) \tag{13}$$

$$H_Q(0) = \sin\left(\frac{\alpha\ell}{2}\right) \exp\left(-\frac{\alpha\ell}{2}\right). \tag{14}$$

The most significant results are obtained when $H_I(\Omega)$ and $H_Q(\Omega)$ are both minimum-phase-shift functions. This requires that all the zeroes of their continuation in the complex plane have a positive imaginary part [14]. It is easily shown that this condition is met when $\alpha \ell < \pi$. It then results from Eqs.(11, 13) that the COG of $y_I(t)$ is advanced (fast light regime) while that of $y_Q(t)$ is delayed (slow light regime). We additionally remark that the transmissions $|H_I(\Omega)|$ and $|H_Q(\Omega)|$ are, respectively, minimal and maximal for $\Omega = 0$. Figure 2 illustrates the



Figure 2: Same as Fig.1 for the envelope transfer functions vs. the frequency Ω . The blue (red) line refers to the in-phase component I (the quadrature component Q). Note that, for the considered optical thickness, the group advances a_{gI} and a_{gQ} (slope of the corresponding phase at $\Omega = 0$) have the same modulus but opposite sign.

previous points in the reference case $\alpha \ell = \pi/2$ already considered in Fig.1. We are then in the remarkable situation where $a_{gQ} = -a_{gI}$ and $H_I(0) = H_Q(0)$. This means that the COG of $y_I(t)$ and $y_Q(t)$ are shifted by the same amount in absolute value and that their *algebraic* area are equal.

A special attention will be paid in the following on fast light which suffers from the most severe restrictions. Eq.(11) might lead to expect that advances as large as wanted could be obtained for $\alpha \ell \rightarrow \pi$. It is worth remarking that this equation only gives the advance of the pulse COG and that the corresponding transmission $H_I(0)$ cancels. Anyway, obtaining large absolute advances is not an aim *per se* and the challenge in fast light experiments is to attain ratios of the advances over the pulse duration as large as possible with moderate distortion. The practical limitations to these fractional advances are examined in the following section.

III. ENVELOPES OF THE INCIDENT AND TRANSMITTED PULSES

We consider an incident pulse of carrier frequency $\omega_c = \omega_0 + \gamma$ and of envelope

$$x(t) = \cos^2\left(\frac{\pi t}{2\tau}\right) \Pi\left(\frac{t}{2\tau}\right) \tag{15}$$

where $\Pi(u)$ designates the rectangle function equal to 1 for -1/2 < u < 1/2 and 0 elsewhere. This pulse is very close to the Gaussian pulse usually considered in the literature. It has a full width at half maximum τ (taken as time unit in the following) and a strictly finite overall duration 2τ . We exploit this point by continuing the envelope x(t) at every time by the periodic signal

$$\tilde{x}(t) = \cos^2\left(\frac{\pi t}{2\tau}\right) = \frac{1 + \cos\left(\pi t/\tau\right)}{2} \tag{16}$$

 $\tilde{x}(t)$ contains only three frequencies, namely 0 and $\pm \Omega_1$ with $\Omega_1 = \pi/\tau$. As shows Fig.3, the signals $\tilde{y}_{I,Q}(t)$ obtained by substituting $\tilde{x}(t)$ to x(t) reproduce very well the main features of the exact envelopes $y_{I,Q}(t)$ obtained by fast Fourier transform (FFT). We get:

$$\tilde{y}_{I,Q}(t) = \frac{1}{2} \left\{ H_{I,Q}(0) + |H_{I,Q}(\Omega_1)| \cos\left[\Omega_1 t + \Phi_{I,Q}(\Omega_1)\right] \right\}$$
(17)

 $y_{I,Q}(t)$ has a maximum of amplitude $A_{I,Q} = [H_{I,Q}(0) + |H_{I,Q}(\Omega_1)|]/2$ in advance over that of the reference pulse by $a_{I,Q} = \Phi_{I,Q}(\Omega_1)/\Omega_1$. The corresponding fractional advances read

$$a_{I,Q}/\tau = \Phi_{I,Q}(\Omega_1)/(\Omega_1\tau) = \Phi_{I,Q}(\Omega_1)/\pi.$$
 (18)

The advances $a_{I,Q}$ of the maximum have the same sign that the corresponding group advances $a_{gI,gQ}$ ($a_I > 0$, $a_Q < 0$). Since $H_I(0) < |H_I(\Omega_1)|$, the amplitude A_I of the advanced signal is larger than its asymptotic value $H_I(0)$ when $\Omega_1/\gamma \to 0$ ($\gamma \tau \to \infty$), the opposite occurring for the amplitude A_Q of the delayed signal [$A_Q < H_Q(0)$]. Equation (17) also enables us to determine the full duration at half maximum τ_I and τ_Q of both signals. They read

$$\tau_{I,Q} = \left(\frac{2\tau}{\pi}\right) \arccos\left[\frac{|H_{I,Q}(\Omega_1)| - H_{I,Q}(0)}{2|H_{I,Q}(\Omega_1)|}\right]$$
(19)

with $\tau_I < \tau$ (pulse narrowing) and $\tau_Q > \tau$ (pulse broadening). All these results are consistent with those expected for systems having a dip or a peak of transmission.

In agreement with relativistic causality, both phase components start at the same time $-\tau$ that the incident pulse in our retarded time picture (Fig.3). As soon as it has a significant amplitude, the component $y_Q(t)$ is simply broadened and the (negative) advance a_Q of its maximum is generally close to the group advance a_{gQ} . The behaviour of the in-phase component $y_I(t)$ is richer. The pulse distortion of $y_I(t)$ is manifested in a narrowing (as already mentioned) and, moreover, in the appearance of a secondary lobe [19]. The latter is also well reproduced by the periodic model (Fig.3). Quite generally, its maximum occurs at $t = [\pi - \Phi_I(\Omega_1)]/\Omega_1$ and its relative amplitude compared to that of the main lobe reads

$$D = \frac{|H_I(\Omega_1)| - H_I(0)}{|H_I(\Omega_1)| + H_I(0)}$$
(20)

When it is small compared to unity, D constitutes a good indicator of the pulse distortion. The corresponding duration τ_I and peak-amplitude A_I of $y_I(t)$ given by the



Figure 3: Normalized envelopes $y_I(t)$ (solid blue line) and $y_Q(t)$ (solid red line) vs. retarded time expressed in units of the incident pulse duration $\tau = \pi/\Omega_1$. The dashed lines (same colours) are the solutions obtained by periodically continuing the incident pulse. The envelope of the pulse transmitted in vacuum (reference pulse) is given for comparison (dotted black line). In real time all the envelopes should be shifted to the right by the luminal transit time ℓ/c . Parameters: $\alpha \ell$ $(\Omega_1/\gamma) = \pi/2$ (1.17) for a) and $3\pi/4$ (0.345) for b).

periodic model read:

$$\tau_I = \frac{2\tau}{\pi} \arccos\left[\frac{D}{1+D}\right] \tag{21}$$

$$A_I = \frac{H_I(0)}{1-D} \tag{22}$$

The envelopes shown Fig.3a and Fig.3b have been obtained, respectively, for $\alpha \ell = \pi/2$ (reference case) and for $\alpha \ell = 3\pi/4$. In both cases, the pulse duration τ has been chosen such that D = 20%. As illustrated Fig.4, the distortion of the corresponding intensity profiles (currently considered in optics) is quite moderate and the fractional advances are $\sqrt{2}$ times larger than those of the envelopes. We note in particular that the advanced intensity profiles compare favourably with those observed in the reference experiments involving a gain-doublet medium [20].

For a given distortion, relativistic causality imposes severe limitations to the fractional advance a_I/τ of $y_I(t)$. As for every fast light system, the larger is the dynamics of the system transmission, the larger is the fractional advance [21]. In the present case, the transmission dynamics [maximum over minimum of $|H_I(\Omega)|$] is reduced to $1/H_I(0)$. Eq.(22) thus involves that the amplitude A_I



Figure 4: Same as Fig.3 for the intensity profiles $y_I^2(t)$ and $y_Q^2(t)$.



Figure 5: Fractional advance a_I/τ (solid blue line, left scale) and amplitude A_I (dashed blue line, right scale) of the maximum of $y_I(t)$ vs. the optical thickness $\alpha \ell$ for D = 20%.

and the fractional delay a_I/τ evolve in opposite directions as functions of the optical thickness $\alpha \ell$. Figure 5 shows the results obtained for D = 20% in a broad domain of variation of $\alpha \ell$. In the reference case $\alpha \ell = \pi/2$ (conditions of Fig.3a), we get $a_I/\tau \approx 17\%$ and $A_I \approx 0.40$ while $a_I/\tau \approx 22\%$ and $A_I \approx 0.15$ when $\alpha \ell = 3\pi/4$ (conditions of Fig.3b). In the latter case, sufficient amplitude is conciliated with a fractional advance which is not far below its asymptotic value ($\approx 27\%$, see below). For the sake of completeness, we give Fig.6 the results obtained in the conditions of Fig.5 for the delayed envelope $y_Q(t)$.



Figure 6: Same as Fig.5 for the fractional delay $-a_Q/\tau$ (solid red line, left scale) and amplitude A_Q (dashed red line, right scale) of the maximum of $y_Q(t)$.

The variations of its amplitude A_Q are moderate, while the fractional delay $-a_Q/\tau$ continuously decreases to 0 when $\alpha \ell \to \pi$. We additionally mention that, correlatively, the pulse duration τ_Q (the ratio a_Q/a_{gQ}) regularly decreases (increases) to τ (to 1).

When $\alpha \ell \to \pi$, the pulse duration required to obtain a fixed distortion *D* becomes very long. Denoting $\varepsilon = (\pi - \alpha \ell)/\pi$, we get at the leading order in ε

$$\frac{\Omega_1}{\gamma} = \frac{\pi}{\gamma\tau} = \beta\epsilon \tag{23}$$

where $\beta = 2\sqrt{D}/(1-D)$. Putting this value in the transfer functions $H_{I,Q}(\Omega_1)$, we get, always at the leading order in ε , the following asymptotic expressions of the pulse advances and amplitudes

$$\frac{a_I}{\tau} = \frac{1}{\pi} \arctan(\beta) \tag{24}$$

$$\frac{a_{gI}}{\tau} = \frac{\beta}{\pi} \tag{25}$$

$$A_I = \frac{\pi e^{-\pi/2}}{4} \left(1 + \sqrt{1 + \beta^2} \right) \varepsilon \tag{26}$$

$$\frac{a_Q}{\tau} = \frac{a_{gQ}}{\tau} = -\frac{\beta\pi\epsilon^2}{4} \tag{27}$$

$$A_Q \approx e^{-\pi/2} \tag{28}$$

Eqs.(24, 25) shows that, strictly speaking, $a_I \rightarrow a_{gI}$ only when $D \ll 1$ but the fractional advance a_I/τ is then very small. As for each fast light system, significant fractional advances are paid by some distortion. The previous results are quite consistent with those shown Figs.(5, 6) for D = 20%, that is $\beta = \sqrt{5}/2$. For $\alpha \ell \rightarrow \pi$, we get in particular $a_I/\tau \rightarrow 27\%$, $A_I \rightarrow 0$, $a_Q/\tau \rightarrow 0$, and $A_Q \rightarrow e^{-\pi/2} \approx 0.21$.



Figure 7: Electrical network used in our experiments. r designates the resistance r_L of the inductor plus an eventual additional resistance r_a . Indicative values of the parameters: $L = 2.17 \, mH$, $C = 153 \, pF$, quality factor of the inductor Q = O(100) at $\omega = 1/\sqrt{LC}$, and $r_a = 0$ or $33 \,\Omega$, $R = 1008 \,\Omega$ or $365 \,\Omega$.

IV. EXPERIMENTS WITH AN ELECTRICAL NETWORK

In the optical experiments, the information on the phase shift induced by the medium can be obtained by means of a frequency change translating both reference and probe fields in the radiofrequency domain [22, 23]. The experiments are greatly simplified by working directly in this domain. As back as 1961, Rupprecht [24] evidenced significant advances of the *envelope* of the pulse transmitted by a radiofrequency network with negative group delay (NGD). More recent demonstrations of advanced pulse-envelope can be found in [25–29]. However, as far as we know, the idea of phase post-selection to evidence NGD effects is absent in all these experiments.

Figure 7 shows the very simple four port network employed in our experiments. As the absorbing medium, it is purely passive. In the narrow resonance limit, the transfer function relating the Fourier transform of the output signal $V_{out}(t)$ to that of the input signal $V_{in}(t)$ reads

$$H(\omega) = \frac{\eta + i(\omega - \omega_0)/\gamma}{\frac{1}{\eta} + i(\omega - \omega_0)/\gamma}$$
(29)

where $\eta = \sqrt{r/(r+R)}$ (0 < η < 1), $\omega_0 = 1/\sqrt{LC}$ and $\gamma = \sqrt{r(R+r)}/(2L)$ with $\gamma \ll \omega_0$. The general relation $a_g(\omega_c) = d\Phi/d\omega \mid_{\omega=\omega_c}$ giving the group advance yields

$$a_g(\omega_c) = \frac{[(1/\eta) - \eta] (1 - \Delta^2/\gamma^2)}{\gamma [1 + \Delta^2/(\eta\gamma)^2] [1 + (\eta\Delta)^2/\gamma^2]}$$
(30)

where $\Delta = \omega_c - \omega_0$ is the detuning of the pulse carrier frequency from resonance. As for the absorbing medium, the group advance is positive when $|\Delta| < \gamma$, is negative when $|\Delta| > \gamma$, cancels when $|\Delta| = \gamma$ and is maximum on resonance where it takes the simple form $a_g(\omega_0) = [(1/\eta) - \eta] / \gamma$. The experimental transmission and phase obtained for $\eta = 0.226$, $\omega_0/2\pi = 274.9 \, kHz$ and $\gamma/2\pi = 8.53 \, kHz$ are shown Fig.8. They are in excellent agreement with those derived from Eq.(29). Note,



Figure 8: Transmission in dB (on the left) and phase (on the right) of the signal transfer function vs $\omega/2\pi$ for $r_a = 0$ and $R = 1008 \,\Omega$ (solid line) and corresponding theoretical results for $\eta = 0.226$, $\omega_0/2\pi = 274.9 \, kHz$ and $\gamma/2\pi = 8.53 \, kHz$ (dashed line). The vertical dash-dotted line indicates the zero-dispersion frequency $\omega_0 + \gamma$.

however, that the values of η , ω_0 and γ somewhat differ from those given below Eq.(29) which are obtained by considering ideal components without including the selfresonant behaviour of the capacitor and inductor [27].

In the time-resolved experiments, we use a waveform generator (Agilent 33500B) delivering both the sinewave signal of frequency ω_c and the modulation signal. It is used in the burst mode (single-shot experiment). The signals $V_{in}(t)$ and $V_{out}(t)$ are sent on two channels of a numerical oscilloscope (Keysight InfiniiVision DSOX4024A) and both are acquired on 16000 points with a 10 bit vertical resolution. Figure 9 gives an example of signals obtained in the resonant case ($\omega_c = \omega_0$) with the parameters of Fig.8. As expected, the maximum of $V_{out}(t)$ is significantly in advance over that of $V_{in}(t)$ but the two signals are in phase. On the other hand, Fig.10, obtained in the zero-dispersion configuration ($\omega_c = \omega_0 + \gamma$), confirms that the advance is then negligible but that the two signals are not in phase.

The transfer functions $H_{\gamma}(\Omega)$, $H_{I}(\Omega)$ and $H_{Q}(\Omega)$ for the envelopes in the zero-dispersion configuration ($\Delta = \gamma$) are derived from $H(\omega)$ as those of the absorbing medium. For the electrical network $H_{\gamma}(\Omega)$ reads

$$H_{\gamma}(\Omega) = \frac{\eta + i + i\Omega/\gamma}{1/\eta + i + i\Omega/\gamma}.$$
(31)

As in the absorbing medium case, the transfer functions $H_I(\Omega)$ and $H_Q(\Omega)$ are deduced from $H_{\gamma}(\Omega)$ by Eqs.(9,10). They yield in particular

$$H_I(0) = \frac{2\eta^2}{1+\eta^2}$$
(32)

$$\gamma a_{gI} = \frac{\left(1 - \eta^2\right)^2}{2\eta \left(1 + \eta^2\right)} \tag{33}$$

$$H_Q(0) = \frac{\eta(1-\eta^2)}{1+\eta^2}$$
(34)



Figure 9: Normalized output signal $V_{out}(t)$ (blue line) vs. time expressed in units of the incident pulse duration τ . This signal is obtained in the resonant case ($\omega_c = \omega_0$). The input signal $V_{in}(t)$ (black line) is given for reference. Parameters as in Fig.8 and $\tau = 295\mu s$ ($\gamma \tau = 15.8$). The advance of the maximum of $V_{out}(t)$ over that of $V_{in}(t)$ is $a = 0.221 \tau$ (corresponding group advance $a_g = 0.266 \tau$) while its relative amplitude is A = 0.052. Inset: comparison of the two signals in the vicinity of t = 0 showing that they are in phase.



Figure 10: Same as Fig.9 in the zero-dispersion configuration ($\omega_c = \omega_0 + \gamma$). The output signal has a phase differing from that of the input signal and the advance *a* of its maximum is negligible. The pulse duration $\tau = 110 \,\mu s$ has been chosen to facilitate the comparison with the results obtained in the absorbing medium for $D = 20 \,\%$. It leads to A = 0.224.

$$\gamma a_{gQ} = -\frac{2\eta}{(1+\eta^2)} \tag{35}$$

For a given distortion $D \ll 1$, the upper limit of the fractional advance a_I/τ is again approached when $H_I(0) \rightarrow$ 0, that is when the pulse amplitude $A_I \rightarrow 0$. By calculations quite similar to those leading to Eqs.(23-28), we retrieve the upper limit $a_I/\tau = (\arctan \beta)/\pi$ (27% for D = 20%) obtained in the absorbing medium case.

The post-selection of the in-phase and quadrature com-



Figure 11: Two examples of envelopes $y_I(t)$ (solid blue line) and $y_Q(t)$ (solid red line) experimentally obtained in the zerodispersion configuration. The envelope x(t) of the input signal (dotted black line) is given for reference. (a) is obtained in the conditions of Fig.10, viz. $\tau = 110 \,\mu s$, $\eta = 0.226$ and $\gamma/(2\pi) =$ $8.53 \,kHz$. We get in this case $a_I = 0.24 \,\tau$, $A_I = 0.12$ for the in-phase component and $a_Q = -0.074 \,\tau$, $A_Q = 0.20$ for the quadrature component. (b) is obtained for $\tau = 76 \,\mu s$ by using a circuit with resistances $R = 365 \,\Omega$ and $r_a = 33 \,\Omega$ leading to $\eta = 0.45$ and $\gamma/(2\pi) = 7.45 \,kHz$. We have in this case $a_I = 0.20 \,\tau$, $a_Q = -0.24 \,\tau$ and $A_I/A_Q = 1.37$.

ponents of the output signal $V_{out}(t)$ is experimentally performed as follows. The data collected by the numerical oscilloscope are treated by computer. In a first step, we generate a continuous sinewave, the frequency and *phase* of which coincide with those of the input signal $V_{in}(t)$. This continuous sinewave is next multiplied by the output signal $V_{out}(t)$ to deliver the envelope $y_I(t)$ of the in-phase component (I), the harmonic at $2\omega_c$ and the high frequency noise being eliminated by a finite impulse response (F.I.R.) filter. The used low pass filter (IGOR software Blackman 367) insures a rejection better than 70 dB for $\omega/(2\pi) > 200 \, kHz$. The envelope $y_Q(t)$ of the quadrature component is similarly derived by using a continuous sinewave in quadrature with that used to obtain $y_I(t)$.

Figure 11 shows the envelopes $y_I(t)$ and $y_Q(t)$ experimentally observed in the zero-dispersion configuration ($\Delta = \gamma$) in two representative cases. As it was made to obtain Fig.10 also as Figs.(3-6) for the absorbing medium, the durations τ of the incident pulse are chosen such that D = 20% for the I-component. The envelopes shown Fig.11a are observed in the conditions of Fig.10. They are quite comparable to those obtained

theoretically with an absorbing medium of optical thickness $\alpha \ell = 3\pi/4$ (see Fig.3b). In particular, the advance a_I is significantly larger than the delay $-a_Q$. On another hand, the amplitudes of the two components are such that $A_Q/A_I \approx 1.7$ and this explains why no secondary lobe is visible Fig.10 in the overall envelope y(t) = $\sqrt{y_I^2(t) + y_Q^2(t)}$. As a second example, Fig.11b shows the envelopes experimentally observed when $\eta = 0.45$ and $\gamma/(2\pi) = 7.45 \, kHz$. In this case $|a_I/a_Q|$ and A_Q/A_I are both close to unity and the envelopes are now comparable to those obtained in the case $\alpha \ell = \pi/2$ taken as reference for the absorbing medium (see Fig.3a). In case a) as in case b), the observed envelopes are in very good agreement with the envelopes derived by FFT by using the transfer function $H_{\gamma}(\Omega)$ given Eq.(31). In addition, the advances, amplitudes and pulse durations are exactly determined by the periodic model with, in particular, $\tau_I = 0.89 \tau$ and $A_I = 1.25 H_I(0)$ as predicted by Eqs.(21,22).

V. CONCLUSION

The dilute medium with a narrow absorption line is a reference system for the observation of fast and slow light. Fast light is obtained when the carrier frequency of the incident pulse coincides or is close to resonance while slow light is observed when this frequency lies in the line wings. There are thus two intermediate carrier frequencies for which the group velocity equals that of the light in vacuum. Paradoxically enough, we have shown that, in such a case, fast and slow light can be simultaneously observed. This is achieved by post-selecting particular phase components of the transmitted field. Fast light is obtained by selecting the component in phase with that of a pulse travelling the same distance in vacuum while slow light is observed on the quadrature component. The general properties of fast and slow light are retrieved with this arrangement. A particular attention is paid to fast light to which the relativistic causality imposes the most severe constraints. As usual, evidencing significant fast light effects with moderate distortion requires large transmission dynamics of the medium and long incident pulses. Finally the theoretical results obtained in optics with an absorbing medium are experimentally reproduced by using a passive electrical network running in the radiofrequency range. We expect that our work will stimulate direct demonstrations in optics or microwave. In this purpose, we emphasize that the phase post-selection procedure introduced in the present article can be applied to different frequency configurations and systems.

Funding: Contrat de Plan Etat-Région (CPER), Photonics and Society Project (P4s) ; Agence Nationale de Recherche (ANR), LABEX CEMPI Project (ANR-11-LABX-0007).

Disclosures: The authors declare no conflicts of in-

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The final results obviously do not depend on the used convention.

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