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Technical note for the FS-MMSE-IC receiver

T. Petitpied^T, R. Tajan^T, G. Ferré^T, P. Chevalier^{†‡}, S. Traverso[†]

^T IMS, Univ. Bordeaux, Bordeaux INP, CNRS (UMR 5218), F-33405 Talence (FRANCE)

[†] Thales, 4 avenue des Louvresses, 92230 Gennevilliers (FRANCE)

[‡] CEDRIC laboratory – Conservatoire National des Arts et Métiers 292 rue Saint-Martin, 75141 Paris Cedex 3 (FRANCE)

 $Email: for ename.name @\{^{T}ims-bordeaux.fr, \ ^{\dagger}thalesgroup.com, \ ^{\ddagger}cnam.fr\}$

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1 Definitions

For any scalars or vectors \mathbf{x} and \mathbf{y} with convenient sizes, we define:

$$\begin{cases} x_n &= \mathbf{e}_n^{\dagger} \mathbf{x} \\ \mathbf{R}_{\mathbf{x}\mathbf{y}} &= \mathbb{E}_{\mathbf{x},\mathbf{y}} [\mathbf{x}\mathbf{y}^{\dagger}] \\ \mathbf{R}_{\mathbf{x}} &= \mathbf{R}_{\mathbf{x}\mathbf{x}} \end{cases}$$

where \mathbf{e}_n is the $(n+1)^{\text{th}}$ element of the canonic base. The operator $\mathbb{E}_{\mathbf{x},\mathbf{y}}[\mathbf{x}\mathbf{y}^{\dagger}]$ represents the statistical average with respect to \mathbf{x} and \mathbf{y} .

For any matrix \mathbf{A} , we define \mathbf{A}^* the conjugate matrix of \mathbf{A} , \mathbf{A}^T its transpose matrix and $\mathbf{A}^{\dagger} = (\mathbf{A}^T)^*$.

2 Context & Problem formulation

We suppose having the following expressions:

$$\begin{cases} \mathbf{r} &= \mathbf{HUs} + \mathbf{w} \\ \mathcal{I}(\mathbf{f}, p) &= \mathbf{f}^{\dagger} \mathbf{r} + p \\ \mathcal{J}(\mathbf{f}, p) &= \mathbb{E}_{\mathcal{I}(\mathbf{f}, p)} \left| \mathcal{I}(\mathbf{f}, p) - s_n \right|^2 \end{cases}$$

where:

- \mathbf{f} : a $(2N \times 1)$ vector,
- p: a scalar,
- \mathbf{H} : a $(2N \times 2N_s)$ deterministic convolution matrix,
- **U** : a $(2N_s \times N_s)$ up-sampling matrix,
- $\mathbf{D} = \mathbf{U}^{\dagger}$: a down-sampling matrix,
- \mathbf{s} : a $(N_s \times 1)$ probabilistic vector such that $\mathbb{E}[\mathbf{s}] = \mathbf{s}^d$ and $\mathbf{R}_{s_n, s_m} = (v_n^d + |s_n^d|^2)\delta(n-m) \Longrightarrow \mathbb{E}\left[\left(\mathbf{s} \mathbf{s}^d\right)\left(\mathbf{s} \mathbf{s}^d\right)^{\dagger}\right] = \mathbf{V}^d = \operatorname{diag}(\mathbf{v}^d),$
- \mathbf{w} : a $(2N \times 1)$ probabilistic vector such that $\mathbb{E}[\mathbf{w}] = \mathbf{0}_N$ and $\mathbf{R}_{\mathbf{w}} = \sigma_w^2 \mathbf{I}_N$,

Typically, the vector \mathbf{r} represents a band-limited received signal after the convolution operation of the shaping filter and the channel \mathbf{H} , corrupted by a centered circular white noise \mathbf{w} with average power σ_w^2 . Moreover, this holds if and only if the shaping filter bandwidth is bounded by $\frac{2}{T_s}$. Hence, \mathbf{r} represents a Fractionanally-Spaced (FS) system model where $\mathcal{I}(\mathbf{f}, p)$ is the estimator of \mathbf{s} with Mean Square Error (MSE) $\mathcal{J}(\mathbf{f}, p)$. Furthermore, considering the prior information ($\mathbf{s}^d, \mathbf{V}^d$) on \mathbf{s} leads to an *Interference-Cancellation (IC)* structure.

Problem formulation: We search for $\tilde{s}_n^e = \mathcal{I}(\mathbf{f}_n, p_n)$, $\tilde{v}_n^e = \mathcal{J}(\mathbf{f}_n, p_n)$ such that:

$$(\mathbf{f}_n, p_n) = \operatorname*{argmin}_{(\mathbf{f}, p)} \mathcal{J}(\mathbf{f}, p)$$

Consequently, \tilde{v}_n^e represents the Minimum MSE of the considered system model (FS-MMSE-IC) and with estimator \tilde{s}_n^e of s_n .

3 Problem resolution

3.1 Expression of the MSE minimizers

$$\mathcal{J}(\mathbf{f}, p) = \mathbb{E}_{\mathcal{I}(\mathbf{f}, p)} |\mathcal{I}(\mathbf{f}, p) - s_n|^2$$

$$= \mathbf{f}^{\dagger} \mathbf{R}_{\mathbf{r}} \mathbf{f} + \mathbf{f}^{\dagger} \mathbb{E}[\mathbf{r}] p^* - \mathbf{f}^{\dagger} \mathbf{R}_{\mathbf{r}s_n} + p \mathbb{E}[\mathbf{r}^{\dagger}] \mathbf{f} + |p|^2 - p \mathbb{E}[s_n^*] - \mathbf{R}_{s_n \mathbf{r}} \mathbf{f} - \mathbb{E}[s_n] p^* + \mathbf{R}_{s_n}$$

Minimizing $\mathcal{J}(\mathbf{f}, p)$ leads us to nullify its the first partial derivatives.

$$\frac{\partial \mathcal{J}(\mathbf{f}, p)}{\partial \mathbf{f}} = 0 \iff 2\mathbf{R}_{\mathbf{r}}\mathbf{f}_n + 2\mathbb{E}[\mathbf{r}]p_n^* - 2\mathbf{R}_{\mathbf{r}s_n} = 0$$
$$\iff \mathbf{f}_n = \mathbf{R}_{\mathbf{r}}^{-1} \left(\mathbf{R}_{\mathbf{r}s_n} - \mathbb{E}[\mathbf{r}]p_n^*\right)$$

$$\frac{\partial \mathcal{J}(\mathbf{f}, p)}{\partial p^*} = 0 \iff 2\mathbb{E}[\mathbf{r}^{\dagger}]\mathbf{f}_n + 2p_n^* - 2\mathbb{E}[s_n^*] = 0$$
$$\iff p_n = \mathbb{E}[s_n] - \mathbf{f}_n^{\dagger}\mathbb{E}[\mathbf{r}]$$

We check that such an extremum corresponds to a minimum:

$$\begin{cases} \frac{\partial^2 \mathcal{J}(\mathbf{f}, p)}{\partial \mathbf{f}^2} = 2\mathbf{R_r} \ge 0\\ \frac{\partial^2 \mathcal{J}(\mathbf{f}, p)}{\partial p^{*2}} = 2 \ge 0 \end{cases}$$

3.2 Computation of the MSE minimizers

Let us first compute:

$$\begin{split} \mathbb{E}[s_n] &= s_n^d \\ \mathbb{E}[\mathbf{r}] &= \mathbf{H} \mathbf{U} \mathbf{s}^d \\ \mathbf{R}_{\mathbf{r}} &= \mathbf{H} \mathbf{U} \mathbf{R}_{\mathbf{s}} \mathbf{D} \mathbf{H}^\dagger + \sigma_w^2 \mathbf{I}_N \\ \mathbf{R}_{\mathbf{r}s_n} &= \mathbf{H} \mathbf{U} \mathbf{R}_{\mathbf{s}s_n} \end{split}$$

Then, the MSE minimizers can be computed:

$$\begin{split} \mathbf{f}_{n}^{\dagger} &= \left(\mathbf{R}_{s_{n}\mathbf{r}} - p_{n}\mathbb{E}[\mathbf{r}^{\dagger}]\right)\mathbf{R}_{\mathbf{r}}^{-1\dagger} \\ &= \left(\mathbf{R}_{s_{n}\mathbf{r}} - \mathbb{E}[s_{n}]\mathbb{E}[\mathbf{r}^{\dagger}] + \mathbf{f}_{n}^{\dagger} \left|\mathbb{E}[\mathbf{r}]\right|^{2}\right)\mathbf{R}_{\mathbf{r}}^{-1\dagger} \\ &= \left(\mathbf{R}_{s_{n}\mathbf{r}} - \mathbb{E}[s_{n}]\mathbb{E}[\mathbf{r}^{\dagger}]\right)\mathbf{R}_{\mathbf{r}}^{-1\dagger}(\mathbf{I}_{N} - |\mathbb{E}[\mathbf{r}]|^{2}\mathbf{R}_{\mathbf{r}}^{-1\dagger})^{-1} \\ &= \left(\mathbf{R}_{s_{n}\mathbf{r}} - \mathbb{E}[s_{n}]\mathbb{E}[\mathbf{r}^{\dagger}]\right)\left(\mathbf{R}_{\mathbf{r}}^{-1\dagger} - |\mathbb{E}[\mathbf{r}]|^{2}\right)^{-1} \\ &= \left(\mathbf{R}_{s_{n}\mathbf{s}} - \mathbb{E}[s_{n}]\mathbb{E}[\mathbf{s}^{\dagger}]\right)\mathbf{D}\mathbf{H}^{\dagger}\left(\mathbf{H}\mathbf{U}(\mathbf{R}_{\mathbf{s}} - \mathbb{E}|\mathbf{s}|^{2})\mathbf{D}\mathbf{H}^{\dagger} + \sigma_{w}^{2}\mathbf{I}_{2N}\right)^{-1} \\ &= v_{n}^{d}\mathbf{e}_{n}^{\dagger}\mathbf{D}\mathbf{H}^{\dagger}\left(\mathbf{H}\mathbf{U}\mathbf{V}^{d}\mathbf{D}\mathbf{H}^{\dagger} + \sigma_{w}^{2}\mathbf{I}_{2N}\right)^{-1} \end{split}$$

Using twice the Woodbury identity on any $(K_1 \times K_2)$ matrix A and $(K_2 \times K_1)$ matrix B:

$$\begin{aligned} \mathbf{A}(\mathbf{B}\mathbf{A} + \mathbf{I}_{K_2})^{-1} &= \mathbf{A}(\mathbf{B}\mathbf{I}_{K_1}\mathbf{A} + \mathbf{I}_{K_2})^{-1} \\ &= \mathbf{A}(\mathbf{I}_{K_2} - \mathbf{B}(\mathbf{I}_{K_1} + \mathbf{A}\mathbf{B})^{-1}\mathbf{A}) \\ &= (\mathbf{I}_{K_1} - \mathbf{A}\mathbf{B}(\mathbf{I}_{K_1} + \mathbf{A}\mathbf{B})^{-1})\mathbf{A} \\ &= (\mathbf{A}\mathbf{B}\mathbf{I}_{K_1} + \mathbf{I}_{K_1})^{-1}\mathbf{A} \\ &= (\mathbf{A}\mathbf{B} + \mathbf{I}_{K_1})^{-1}\mathbf{A} \end{aligned}$$

We identify $\begin{cases} \mathbf{A} = \sigma_w^{-2} \mathbf{D} \mathbf{H}^{\dagger} \\ \mathbf{B} = \mathbf{H} \mathbf{U} \mathbf{V}^d \end{cases}$ leading to:

$$\begin{split} \mathbf{f}_{n}^{\dagger} &= v_{n}^{d} \mathbf{e}_{n}^{\dagger} \mathbf{D} \mathbf{H}^{\dagger} \left(\mathbf{H} \mathbf{U} \mathbf{V}^{d} \mathbf{D} \mathbf{H}^{\dagger} + \sigma_{w}^{2} \mathbf{I}_{2N} \right)^{-1} \\ &= v_{n}^{d} \mathbf{e}_{n}^{\dagger} \left(\mathbf{D} \mathbf{H}^{\dagger} \mathbf{H} \mathbf{U} \mathbf{V}^{d} + \sigma_{w}^{2} \mathbf{I}_{N} \right)^{-1} \mathbf{D} \mathbf{H}^{\dagger} \\ &= v_{n}^{d} \mathbf{e}_{n}^{\dagger} \mathbf{\Sigma}^{-1} \mathbf{D} \mathbf{H}^{\dagger} \end{split}$$

where $\Sigma = \mathbf{G}\mathbf{V}^d + \sigma_w^2 \mathbf{I}_N$ is the MMSE equalization matrix and $\mathbf{G} = \mathbf{D}\mathbf{H}^{\dagger}\mathbf{H}\mathbf{U}$ is also called *global filter* and gathers the up-sampling, channel and matched filtering and down-sampling operations.

$$p_n = \mathbb{E}[s_n] - \mathbf{f}_n^{\dagger} \mathbb{E}[\mathbf{r}]$$
$$= s_n^d - v_n^d \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \mathbf{G} \mathbf{s}^d$$

3.3 Computation of the FS-MMSE-IC and its estimator

The FS-MMSE-IC estimator is given by:

$$\tilde{s}_{n}^{e} = \mathbf{f}_{n}^{\dagger} \mathbf{r} + p_{n}$$

$$= s_{n}^{d} + v_{n}^{d} \mathbf{e}_{n}^{\dagger} \mathbf{\Sigma}^{-1} \left(\mathbf{D} \mathbf{H}^{\dagger} \mathbf{r} - \mathbf{G} \mathbf{s}^{d} \right)$$

$$= s_{n}^{d} + v_{n}^{d} \mathbf{e}_{n}^{\dagger} \mathbf{\Sigma}^{-1} \left(\mathbf{y} - \mathbf{G} \mathbf{s}^{d} \right)$$

where $\mathbf{y} = \mathbf{D}\mathbf{H}^{\dagger}\mathbf{r}$ represents the received signal after matched filtering and down-sampling.

The FS-MMSE-IC \tilde{v}_n^e of the estimator \tilde{s}_n^e can be computed by first deriving:

$$\begin{split} \mathbf{R}_{\tilde{s}_{n}^{e}} &= |s_{n}^{d}|^{2} + 2(s_{n}^{d})^{*}v_{n}^{d}\mathbf{e}_{n}^{\dagger}\operatorname{Re}\left(\mathbf{\Sigma}^{-1}\mathbb{E}_{\mathbf{y}}[\left(\mathbf{y} - \mathbf{G}\mathbf{s}^{d}\right)]\right) + (v_{n}^{d})^{2}\mathbf{e}_{n}^{\dagger}\mathbf{\Sigma}^{-1}\mathbb{E}_{\mathbf{y}}[\left(\mathbf{y} - \mathbf{G}\mathbf{s}^{d}\right)\left(\mathbf{y} - \mathbf{G}\mathbf{s}^{d}\right)^{\dagger}]\mathbf{\Sigma}^{-1\dagger}\mathbf{e}_{n} \\ &= |s_{n}^{d}|^{2} + (v_{n}^{d})^{2}\mathbf{e}_{n}^{\dagger}\mathbf{\Sigma}^{-1}\mathbb{E}_{\mathbf{s},\mathbf{w}}\left|\mathbf{G}(\mathbf{s} - \mathbf{s}^{d}) + \mathbf{D}\mathbf{H}^{\dagger}\mathbf{w}\right|^{2}\mathbf{\Sigma}^{-1\dagger}\mathbf{e}_{n} \\ &= |s_{n}^{d}|^{2} + (v_{n}^{d})^{2}\mathbf{e}_{n}^{\dagger}\mathbf{\Sigma}^{-1}\mathbf{D}\mathbf{H}^{\dagger}\mathbb{E}_{\mathbf{s},\mathbf{w}}\left|\mathbf{H}\mathbf{U}(\mathbf{s} - \mathbf{s}^{d}) + \mathbf{w}\right|^{2}\mathbf{\Sigma}^{-1\dagger}\mathbf{e}_{n} \\ &= |s_{n}^{d}|^{2} + (v_{n}^{d})^{2}\mathbf{e}_{n}^{\dagger}\mathbf{\Sigma}^{-1}\mathbf{D}\mathbf{H}^{\dagger}\left(\mathbf{H}\mathbf{U}\mathbf{V}^{d}\mathbf{D}\mathbf{H}^{\dagger} + \sigma_{w}^{2}\mathbf{I}_{2N}\right)\mathbf{H}\mathbf{U}\mathbf{\Sigma}^{-1\dagger}\mathbf{e}_{n} \\ &= |s_{n}^{d}|^{2} + (v_{n}^{d})^{2}\mathbf{e}_{n}^{\dagger}\mathbf{\Sigma}^{-1}\left(\mathbf{D}\mathbf{H}^{\dagger}\mathbf{H}\mathbf{U}\mathbf{V}^{d} + \sigma_{w}^{2}\mathbf{I}_{N}\right)\mathbf{D}\mathbf{H}^{\dagger}\mathbf{H}\mathbf{U}\mathbf{\Sigma}^{-1\dagger}\mathbf{e}_{n} \\ &= |s_{n}^{d}|^{2} + (v_{n}^{d})^{2}\mathbf{e}_{n}^{\dagger}\mathbf{\Sigma}^{-1}\mathbf{G}\mathbf{e}_{n} \end{split}$$

$$\begin{split} \mathbf{R}_{\tilde{s}_n^e, s_n} &= \mathbb{E}_{\mathbf{y}} [\left(s_n^d + v_n^d \mathbf{e}_n^\dagger \boldsymbol{\Sigma}^{-1} (\mathbf{y} - \mathbf{G} \mathbf{s}^d) \right) s_n^*] \\ &= |s_n^d|^2 + v_n^d \mathbf{e}_n^\dagger \boldsymbol{\Sigma}^{-1} \mathbf{G} \mathbb{E}_{\mathbf{s}} [\left(\mathbf{s} - \mathbf{s}^d \right) s_n^*] \\ &= |s_n^d|^2 + (v_n^d)^2 \mathbf{e}_n^\dagger \boldsymbol{\Sigma}^{-1} \mathbf{G} \mathbf{e}_n \\ &= \mathbf{R}_{\tilde{s}_n^e} \end{split}$$

$$\mathbf{R}_{s_n} = |s_n^d|^2 + v_n^d$$

This leads to the FS-MMSE-IC \tilde{v}_n^e expression:

$$\begin{split} \tilde{v}_n^e &= \mathbb{E}_{\tilde{s}_n^e} \left| \tilde{s}_n^e - s_n \right|^2 \\ &= \mathbf{R}_{\tilde{s}_n^e} - \mathbf{R}_{\tilde{s}_n^e, s_n} - \mathbf{R}_{s_n, \tilde{s}_n^e} + \mathbf{R}_{s_n} \\ &= \mathbf{R}_{s_n} - \mathbf{R}_{s_n, \tilde{s}_n^e} \\ &= |s_n^d|^2 + v_n^d - |s_n^d|^2 - (v_n^d)^2 \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \mathbf{G} \mathbf{e}_n \\ &= v_n^d \left(1 - v_n^d c_n \right) \end{split}$$

where $c_n = \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \mathbf{G} \mathbf{e}_n$.

3.4 Unbiased FS-MMSE-IC and its estimator

We can note that the estimator \tilde{s}_n^e of s_n has a residual bias :

$$\mathbb{E}\left[s_n^e|\mathbb{E}[s_n] = \mu\right] = \mathbb{E}_{\mathbf{s}_{\sim\{n\}},\mathbf{w}}\left[s_n^d + v_n^d \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \left(\mathbf{y} - \mathbf{G}\mathbf{s}^d\right) | \mathbb{E}[s_n] = \mu\right]$$

$$= s_n^d + v_n^d \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \mathbf{G} \mathbb{E}_{\mathbf{s}_{\sim\{n\}}} \left[\mathbf{s} - \mathbf{s}^d | \mathbb{E}[s_n] = \mu\right]$$

$$= s_n^d + (\mu - s_n^d) v_n^d c_n$$

$$= v_n^d c_n \mu + s_n^d (1 - v_n^d c_n)$$

where the vector $\mathbf{s}_{\sim\{n\}}$ represents all the components of \mathbf{s} but s_n .

Removing the additive term $s_n^d(1-v_n^dc_n)$ and dividing the multiplicative term $v_n^dc_n$ leads to:

$$s_n^e = (v_n^d c_n)^{-1} \left(\tilde{s}_n^e - s_n^d (1 - v_n^d c_n) \right)$$
$$= s_n^d + c_n^{-1} \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \left(\mathbf{y} - \mathbf{G} \mathbf{s}^d \right)$$

We compute the new statistical matrices:

$$\mathbf{R}_{s_n^e} = |s_n^d|^2 + c_n^{-2} \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \mathbb{E}_{\mathbf{y}} [(\mathbf{y} - \mathbf{G}\mathbf{s}^d) (\mathbf{y} - \mathbf{G}\mathbf{s}^d)^{\dagger}] \mathbf{\Sigma}^{-1\dagger} \mathbf{e}_n$$
$$= |s_n^d|^2 + c_n^{-1}$$

$$\begin{split} \mathbf{R}_{s_n^e,s_n} &= \mathbb{E}_{\mathbf{y}}[\left(s_n^d + c_n^{-1}\mathbf{e}_n^{\dagger}\boldsymbol{\Sigma}^{-1}(\mathbf{y} - \mathbf{G}\mathbf{s}^d)\right)s_n^*] \\ &= |s_n^d|^2 + c_n^{-1}\mathbf{e}_n^{\dagger}\boldsymbol{\Sigma}^{-1}\mathbf{G}\mathbb{E}_{\mathbf{s}}[\left(\mathbf{s} - \mathbf{s}^d\right)s_n^*] \\ &= |s_n^d|^2 + v_n^d \end{split}$$

$$\mathbf{R}_{s_n} = |s_n^d|^2 + v_n^d$$

And finally we find we unbiased FS-MMSE-IC:

$$\begin{aligned} v_n^e &= \mathbb{E}_{s_n^e} \left| s_n^e - s_n \right|^2 \\ &= \mathbf{R}_{s_n^e} - \mathbf{R}_{s_n^e, s_n} - \mathbf{R}_{s_n, s_n^e} + \mathbf{R}_{s_n} \\ &= \left| s_n^d \right|^2 + c_n^{-1} - \left| s_n^d \right|^2 - v_n^d - \left| s_n^d \right|^2 - v_n^d + \left| s_n^d \right|^2 + v_n^d \\ &= c_n^{-1} - v_n^d \end{aligned}$$

4 Conclusion

The biased FS-MMSE-IC receiver is given by:

$$\begin{cases} \tilde{s}_n^e &= s_n^d + v_n^d \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \left(\mathbf{y} - \mathbf{G} \mathbf{s}^d \right) \\ \tilde{v}_n^e &= v_n^d \left(1 - v_n^d c_n \right) \end{cases}$$

where $c_n = \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \mathbf{G} \mathbf{e}_n$ is a scalar, $\mathbf{y} = \mathbf{D} \mathbf{H}^{\dagger} \mathbf{r}$ is the receive signal after matched filtering and down-sampling, $\mathbf{G} = \mathbf{D} \mathbf{H}^{\dagger} \mathbf{H} \mathbf{U}$ is the global effect including up-sampling, channel and matched filtering and down-sampling, $\mathbf{\Sigma} = \mathbf{G} \mathbf{V}^d + \sigma_w^2 \mathbf{I}_N$ is the MMSE equalization matrix.

Moreover, the unbiased FS-MMSE-IC receiver is given by:

$$\begin{cases} s_n^e &= s_n^d + c_n^{-1} \mathbf{e}_n^{\dagger} \mathbf{\Sigma}^{-1} \left(\mathbf{y} - \mathbf{G} \mathbf{s}^d \right) \\ v_n^e &= c_n^{-1} - v_n^d \end{cases}$$

Consequently, the FS-MMSE-IC receiver leads to a FS matched filter followed by a symbol time equalizer.