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Hierarchical robust performance analysis of uncertain large scale systems

Khaled Laib, Anton Korniienko, Marc Dinh, Gérard Scorletti and Florent Morel

Abstract—In this paper, the problem of robust performance analysis of interconnected uncertain systems with hierarchical structure is investigated. The computational load associated to such problems does not allow a direct application of robustness analysis usual tools. To overcome this difficulty, we exploit the hierarchical structure of the problem and propose an algorithm to perform robustness analysis using IQC “propagation” along the hierarchical structure. This algorithm allows to establish a trade-off between computation time required to perform the analysis and the conservatism of the obtained results. Furthermore, it is possible to perform parallel computation using the proposed algorithm.

Index Terms—Uncertain large scale systems, robustness analysis, IQC analysis, LMI optimization, hierarchical approach, IQC propagation.

I. INTRODUCTION

Robustness analysis of uncertain large scale systems (LSS) to ensure a certain level of performance in the worst case scenario is a major topic in the automatic control community. These LSS (networks) are obtained by interconnecting smaller subsystems within the objective of ensuring some global tasks. Therefore, it is natural to consider the hierarchical structure of networks: subsystems define the local level while their interconnection define the global level. In this paper, we are interested in robustness analysis of uncertain large scale systems with hierarchical structure.

Robustness analysis of uncertain LSS is a problem with many challenges. Major difficulties that frequently arise in this problem are: robustness analysis and large scale aspects. Although the robustness analysis is an NP hard problem [1], many efficient methods have been developed based on relaxations as convex optimization problem under Linear Matrix Inequality (LMI) constraints [2], see *e.g.* the μ -upper bound [3] in the μ -analysis approach [4] or the Integral Quadratic Constraint (IQC) approach [5].

The second aspect is the large scale associated to networks. Even when we consider the interconnection of systems without any uncertainties, the analysis problem remains complicated and the network stability is not easy to certify. In this case, the

robustness of the LSS is discussed with respect to the network size and its interconnection topology.

The objective is to establish decentralized conditions to ensure the stability of the LSS *i.e.* conditions that subsystems have to satisfy with respect to their interconnection matrix to guarantee the overall stability of the network. These conditions are obtained in different frameworks: [6] and [7] for dissipativity approaches, [8] and [9] for S -hull convexification approaches, [10] and [11] for graph theory approaches, *etc.*

Nevertheless, when considering both aspects (robustness and large scale), the complexity and the computation time dramatically increase due to the large size optimization problem we have to solve. Therefore, the robustness analysis usual tools cannot be practically applied directly.

In order to reduce this computational load, researchers focused on exploiting the particular characteristics of the subsystems and their interconnections. The authors of [12] propose a decomposable approach to investigate the robustness of uncertain LSS. The obtained conditions involve the structured singular values of the individual systems and the eigenvalues of the interconnection matrix. However, these results are valid only when the subsystems are homogeneous. Within the framework of IQC, characterizations of local systems and/or interconnections are used to obtain robust stability conditions in the case where the interconnection matrix is normal [13], unitarily diagonalized [14] or sparse with a chordal pattern [15] and [16].

However for a given LSS, it could be difficult to model the network with a normal, unitarily diagonalized or interconnection matrix with a chordal pattern. Furthermore, even if the network presents one of these structures, the previous methods allow only to investigate the stability. Introducing extra signals to investigate the performance may change the structure. In addition, these results do not exploit an important aspect of the problem which is the hierarchical structure. In this context, the *Hierarchical approach* was initially introduced in [17], in the case of conic uncertainty *i.e.* non structured dynamical uncertainty, to split the overall analysis problem into several low dimensional ones. Each uncertain system in the network can be characterized with conic properties which are a special case of IQC. The overall analysis is then performed in a hierarchical manner by propagating conic properties along the hierarchical structure. Nevertheless, this approach was not really exploited and implemented because it has neither been explained nor formulated how to obtain these IQC characterizing uncertain systems.

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In this paper, we adapt the hierarchical approach and extend it to the case of structured uncertainties. As a first contribution of this paper, we define a set of IQC characterizing different informations on the uncertain system: gain uncertainty, phase uncertainty, mixed gain-phase uncertainty. The IQC set thus defines a basis to characterize uncertain systems. In order to reduce the conservatism of this approach, a size measure is defined and minimized for each of the basis element. The problem of minimizing each size measure is formulated as an LMI optimization problem which can be solved efficiently. A second contribution of this paper is an algorithm to investigate the robust performance of the overall uncertain large scale system using basis propagation along the hierarchical structure. This algorithm allows to establish a trade-off between computation time required for the analysis and the conservatism of the results. Preliminary results can be found in [18]–[20].

Paper outline

This paper is organized as follows: Section II presents the formulation of uncertain LSS performance analysis problem. Section III presents the usual approach to perform robustness analysis of uncertain systems using dissipativity properties. Section IV presents the proposed approach to solve the problem of robustness analysis of uncertain LSS using basis propagation. The formulation and practical computation of the basis elements are presented in Section V. The benefits of the hierarchical approach, compared to the direct analysis method, are illustrated through two application examples in Section VI. We discuss in Section VII the algorithmic complexity and the computation time required to perform robustness analysis for a sub-class of uncertain LSS with hierarchical structure. Conclusions and perspectives are presented in Section VIII.

Notations

\mathbb{R} and \mathbb{C} are the sets of real and complex numbers respectively. $\overline{\mathbb{R}}^+ = \mathbb{R}^+ \cup \{+\infty\}$ where \mathbb{R}^+ is the set of real positive numbers including 0. $\overline{\mathbb{C}}^+$ denotes the closed right half of complex plane including imaginary axis and infinity. The maximum singular value of a matrix M is denoted $\bar{\sigma}(M)$. If the index of a matrix or a signal is not relevant and can be understood from the context, then this index will be omitted or replaced with "•". The real and imaginary parts of the complex entity \bullet are denoted $\text{Re}(\bullet)$ and $\text{Im}(\bullet)$ respectively. \mathbf{RH}_∞ (respectively \mathbf{RL}_∞) denotes the set of matrices of stable (resp. non causally stable) rational transfer functions.

We consistently denote elementary uncertainties by Δ and interconnections by M which can be partitioned into $M = \begin{pmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{pmatrix}$. For several matrices M_j , $\mathbf{bdiag}_j(M_j)$ denotes the block diagonal matrix composed of M_j . We denote by $\Delta \star M$ the set $\{\Delta \star M, \forall \Delta \in \Delta\}$, referred to as an uncertain system, defined by $\Delta \star M = M_{22} + M_{21}\Delta(I - M_{11}\Delta)^{-1}M_{12}$ with \star standing for the Redheffer star product and it will be referred to as the Linear Fractional Transformation (LFT) interconnection of M and Δ .

Finally, we denote by $\mathcal{L}_S(\bullet, \Phi_{11}, \Phi_{12}, \Phi_{22})$ the matrix

$$\begin{pmatrix} \bullet \\ I \end{pmatrix}^* \begin{pmatrix} -\Phi_{22} & | & -\Phi_{12}^* \\ -\Phi_{12} & | & -\Phi_{11} \end{pmatrix} \begin{pmatrix} \bullet \\ I \end{pmatrix}$$

and we denote by $\mathcal{L}_P(\bullet, \Phi_{11}, \Phi_{12}, \Phi_{22}, X, Y, Z, \epsilon)$ the matrix

$$\begin{pmatrix} \bullet \\ I \end{pmatrix}^* \begin{pmatrix} -\Phi_{22} & 0 & | & -\Phi_{12}^* & 0 \\ 0 & X - \epsilon I & | & 0 & 0 \\ -\Phi_{12} & 0 & | & -\Phi_{11} & Y \\ 0 & Y^* & | & 0 & Z - \epsilon I \end{pmatrix} \begin{pmatrix} \bullet \\ I \end{pmatrix}.$$

II. PROBLEM FORMULATION

A. LSS Hierarchical Structure

Inspired by [17] and [18], an uncertain large scale system is defined by a tree composed of leaves interconnected through branches. These leaves and branches will be organized by levels. An example of tree is illustrated in Fig. 1 where a hierarchical structure arises with $l = 4$ levels.

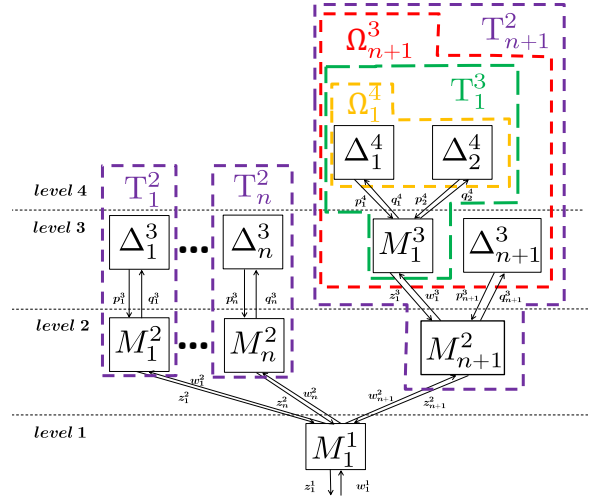


Fig. 1: Uncertain linear large scale system T_1^1 with hierarchical structure of four levels

An index i is associated to each level *i.e.* $i = 1, \dots, l$. At each level i , two types of components may be found: leaves and branches.

The leaves represent the elementary uncertain components and they are denoted $\Delta_{j_\Delta}^i \in \Delta$ with $j_\Delta \in \{1, \dots, N_\Delta^i\}$ where N_Δ^i is the number of elementary uncertain components at level i and Δ is the uncertainty set traditionally considered in robust analysis literature. It is given as a block diagonal combination of elementary uncertainties

$$\Delta = \left\{ \Delta \left| \begin{array}{l} \|\Delta\|_\infty < 1 \\ \Delta = \mathbf{bdiag}(\delta_{1r_1}^r, \dots, \delta_{nr_r}^r, \\ \delta_{1c_1}^c, \dots, \delta_{nc_c}^c, \\ \Delta_1, \dots, \Delta_{n_f}) \end{array} \right. \right\} \quad (1)$$

where

- $\delta_j^r \in \mathbb{R}$ is a real parametric uncertainty,
- $\delta_j^c \in \mathbb{C}$ is a complex uncertainty,
- Δ_j is a LTI system which represents a dynamical uncertainty with k_m^j inputs and k_f^j outputs.

The input output signals of $\Delta_{j\Delta}^i$ are $q_{j\Delta}^i$ and $p_{j\Delta}^i$ respectively. The elementary uncertain components $\Delta_{j\Delta}^i$, of level i , are the end of the tree *i.e.* leaves since they are only connected to the certain components of the level $i - 1$.

The branches are certain components denoted $M_{j_M}^i$, assumed to be LTI systems, with $j_M \in \{1, \dots, N_M^i\}$ where N_M^i is the number of certain components at the level i . In contrast with $\Delta_{j\Delta}^i$, which are only connected to certain components of the level below, $M_{j_M}^i$ are connected to both levels: below and above. The certain components $M_{j_M}^i$ are connected to certain and to uncertain components from level $i + 1$ and to certain components from level $i - 1$. The signals $w_{j_M}^i$ and $p_{j_M}^{i+1}$ (and possibly $z_{j_M}^{i+1}$) are the input signals of $M_{j_M}^i$ while $z_{j_M}^i$ and $q_{j_M}^{i+1}$ (and possibly $w_{j_M}^{i+1}$) are the output signals.

After showing the different levels of the LSS with its different certain and uncertain components (which can be seen as an horizontal decomposition), it is possible to regroup the components connected vertically. The result will be an uncertain system denoted T_j^i , where i stands for the hierarchical level and j for the index of the uncertain system at this hierarchical level. The signals w_j^i and z_j^i are the input and output signals of the uncertain system T_j^i . The number of uncertain systems at each level i is $N_T^i = N_M^i$.

Each uncertain system T_j^i is the LFT interconnection of $M_{j_M}^i$ with either just only elementary uncertain components of the next hierarchical level ($\Delta_{j\Delta}^{i+1}$ for example) or a block diagonal composition of elementary uncertain components $\Delta_{j\Delta}^{i+1}$ and uncertain systems $T_{j_M}^{i+1}$ from the next hierarchical level.

For the ease of notation, and for the purpose of this paper, both cases will be combined into block diagonal augmented uncertainty that will be denoted $\Omega_{j\Delta}^{i+1} \in \Omega$ where Ω is the extension of the elementary uncertainty set (1), that is

$$\Omega = \left\{ \Omega \mid \Omega = \mathbf{bdiag} (\Omega_j) \right\} \quad (2)$$

where Ω_j is either an elementary uncertain block *i.e.* $\Omega_j \in \Delta$, or an uncertain LTI system that belongs to a bounded and connected set without a priori “known” bound. It is now possible to model the LSS as

$$\begin{aligned} \forall i \in \{1, \dots, l - 1\} \\ \forall j \in \{1, \dots, N_T^i\} \\ z_j^i = \underbrace{\Omega_{j\Delta}^{i+1} \star M_{j_M}^i}_{T_j^i} w_j^i \end{aligned} \quad (3)$$

with

$$\Omega_{j\Delta}^{i+1} = \mathbf{bdiag} \left(\mathbf{bdiag}_{m \in \mathcal{N}_\Delta(M_j^i)} (\Delta_m^{i+1}), \mathbf{bdiag}_{n \in \mathcal{N}_T(M_j^i)} (T_n^{i+1}) \right) \quad (4)$$

where $\mathcal{N}_T(M_j^i)$ is the index set of the uncertain systems $T_{j_M}^{i+1}$ connected to $M_{j_M}^i$ and $\mathcal{N}_\Delta(M_j^i)$ is the index set of the elementary uncertain components $\Delta_{j\Delta}^{i+1}$ connected to $M_{j_M}^i$ respectively.

For illustration purposes, let us consider the LSS presented in Fig. 1 where w_1 and z_1 are the input and the output signals respectively. They define the system T_1^1 for which we want to investigate the performance as it will be formally defined later. At level 2, the components can be regrouped into $n + 1$ uncertain systems T_j^2 , for $\forall j \in \{1, \dots, n + 1\}$, interconnected through M_1^1 to form T_1^1 . Each T_j^2 is the LFT interconnection of M_j^2 with Ω_j^3 . Two types of Ω_j^3 appear: either $\Omega_j^3 = \Delta_j^3$ for $j = \{1, \dots, n\}$ or $\Omega_{n+1}^3 = \mathbf{bdiag} (\Delta_{n+1}^3, T_1^3)$. This last uncertain system T_1^3 , used to construct Ω_{n+1}^3 , is in itself the interconnection of M_1^3 with Ω_1^4 which is the block diagonal combination of two elementary uncertain components: $\Omega_1^4 = \mathbf{bdiag} (\Delta_1^4, \Delta_2^4)$.

B. Problem statement

The LTI system performance is achieved if the maximum singular value $\bar{\sigma}(\bullet)$ of the system frequency response respects, along the frequencies, some user-defined frequency dependent constraints (see [21]). Since the uncertainties will impact the system frequency response, we denote γ_ω as an upper bound on the system maximum singular value $\bar{\sigma}(\bullet)$ over all the uncertainties along the frequencies. Hence, the robust performance analysis boils down to check if the computed minimal value γ_ω , for each frequency $\omega \in \overline{\mathbf{R}}^+$, is less than the user defined bound for all possible uncertainties.

Now, the robust performance analysis problem of uncertain LSS can be formulated.

Problem 2.1 (Robust Performance): Given an uncertain LSS defined by (3) and (4) with $\Delta_j^i \in \Delta$, $M_j^i \in \mathbf{RH}_\infty$. Given l , N_T^i , $\mathcal{N}_T(M_j^i)$ and $\mathcal{N}_\Delta(M_j^i)$, test if the LSS is stable $\forall \Delta_j^i \in \Delta$ for all i, j . Furthermore, for every frequency $\omega \in \overline{\mathbf{R}}^+$ solve

$$\begin{aligned} \min_{\gamma_\omega} \quad & \gamma_\omega \\ \text{subject to} \quad & \bar{\sigma}(T_1^1(j\omega)) < \gamma_\omega \quad \forall \Delta_j^i \in \Delta \end{aligned}$$

Road-map

Problem 2.1 is the main problem under consideration in this paper and it may be solved using robustness analysis usual tools which will be presented in Section III if the LSS size is not important. However, when the LSS size becomes important, the efficient resolution of Problem 2.1 is not possible using these tools due to the large size of the associated optimization problem and the dramatic increase of computation time. Algorithm 1 presented in Section IV allows to solve Problem 2.1 efficiently by exploiting the LSS hierarchical structure. Algorithm 1 allows to split the overall analysis problem into several low dimensional ones. In order to solve each of these problems, the input output signals of each uncertain subsystem in the LSS are characterized using and combining different relevant information such as gain uncertainty, phase uncertainty, *etc.*. This characterization is presented in Section V and it represents a generalization of the robustness analysis usual tools presented in Section III.

The gain uncertainty problem (Problem 5.1) is solved using Theorem 5.1, the phase uncertainty problem (Problem 5.3) is solved using Theorem 5.3 and the mixed gain/phase uncertainty problem (Problem 5.2) is solved using Theorem 5.2.

III. ROBUSTNESS ANALYSIS OF UNCERTAIN SYSTEMS

A. Uncertain systems

An uncertain system will be defined as an interconnection $T = \Omega \star M$ with $M \in \mathbf{RH}_\infty$ and $\Omega \in \mathbf{\Omega}$ where $\mathbf{\Omega}$ is bounded and connected set of LTI systems as in (2). Introducing internal signals and using frequency domain, we obtain the following system description

$$\begin{aligned} p(\mathbf{j}\omega) &= \Omega(\mathbf{j}\omega) q(\mathbf{j}\omega) \\ \begin{pmatrix} q(\mathbf{j}\omega) \\ z(\mathbf{j}\omega) \end{pmatrix} &= M(\mathbf{j}\omega) \begin{pmatrix} p(\mathbf{j}\omega) \\ w(\mathbf{j}\omega) \end{pmatrix} \end{aligned} \quad (5)$$

where $w(\mathbf{j}\omega) \in \mathbb{C}^{n_w}$ and $z(\mathbf{j}\omega) \in \mathbb{C}^{n_z}$ are the input and the output signals respectively and they will be used to define and evaluate the system performance as it will be explained later. The signals $p(\mathbf{j}\omega) \in \mathbb{C}^{n_p}$ and $q(\mathbf{j}\omega) \in \mathbb{C}^{n_q}$ are internal signals.

B. Robustness analysis

We will use dissipativity properties to characterize the performance of an uncertain system T . In the field of robust control theory, IQC are often used to characterize the system behavior in form of input-output signal relations (also known as graph relations).

Definition 3.1 (IQC [5]): The two signals $w(\mathbf{j}\omega)$ and $z(\mathbf{j}\omega)$ are said to satisfy the IQC defined by $\Phi = \Phi^*$ of \mathbf{RL}_∞ , if $\exists \epsilon > 0$ and

$$\int_{-\infty}^{+\infty} \begin{pmatrix} z(\mathbf{j}\omega) \\ w(\mathbf{j}\omega) \end{pmatrix}^* \Phi(\mathbf{j}\omega) \begin{pmatrix} z(\mathbf{j}\omega) \\ w(\mathbf{j}\omega) \end{pmatrix} d\omega \geq \epsilon \int_{-\infty}^{+\infty} \begin{pmatrix} z(\mathbf{j}\omega) \\ w(\mathbf{j}\omega) \end{pmatrix}^* \begin{pmatrix} z(\mathbf{j}\omega) \\ w(\mathbf{j}\omega) \end{pmatrix} d\omega$$

Since the problem considered in this paper is the performance analysis of uncertain LTI systems i.e. $z(\mathbf{j}\omega) = T(\mathbf{j}\omega)w(\mathbf{j}\omega)$, the integral term could be dropped which gives a Quadratic Constraint (QC). Furthermore if

$$\Phi(\mathbf{j}\omega) = \begin{pmatrix} X(\mathbf{j}\omega) & Y(\mathbf{j}\omega) \\ Y(\mathbf{j}\omega)^* & Z(\mathbf{j}\omega) \end{pmatrix}$$

with $X = X^*$, $Z = Z^*$ and Y are transfer functions of \mathbf{RL}_∞ , the QC thus defines a dissipativity property and the uncertain system $\Omega \star M$ is said to be $\{X(\mathbf{j}\omega), Y(\mathbf{j}\omega), Z(\mathbf{j}\omega)\}$ dissipative if

$$\begin{aligned} \begin{pmatrix} \Omega(\mathbf{j}\omega) \star M(\mathbf{j}\omega) \\ I \end{pmatrix}^* \begin{pmatrix} X(\mathbf{j}\omega) & Y(\mathbf{j}\omega) \\ Y(\mathbf{j}\omega)^* & Z(\mathbf{j}\omega) \end{pmatrix} \begin{pmatrix} \Omega(\mathbf{j}\omega) \star M(\mathbf{j}\omega) \\ I \end{pmatrix} \\ \geq \epsilon \begin{pmatrix} \Omega(\mathbf{j}\omega) \star M(\mathbf{j}\omega) \\ I \end{pmatrix}^* \begin{pmatrix} \Omega(\mathbf{j}\omega) \star M(\mathbf{j}\omega) \\ I \end{pmatrix} \end{aligned} \quad (6)$$

Remark 3.1: The performance constraint of the LSS T_1^1 in Problem 2.1 can be expressed in terms of dissipativity properties as following $\bar{\sigma}(T_1^1(\mathbf{j}\omega)) < \gamma_\omega$, $\forall \Delta_j^i \in \mathbf{\Delta} \iff T_1^1$ is $\{-I, 0, \gamma_\omega^2 I\}$ dissipative $\forall \Delta_j^i \in \mathbf{\Delta}$.

In order to keep the discussion as general as possible, we will use, in this paper, a general performance measure defined by a general $\{X(\mathbf{j}\omega), Y(\mathbf{j}\omega), Z(\mathbf{j}\omega)\}$ dissipativity.

The following theorem gives necessary and sufficient conditions to perform the robust performance analysis of an uncertain system $\Omega \star M$ for every $\Omega \in \mathbf{\Omega}$

Theorem 3.1: Let $\mathbf{\Omega}$ be a bounded and connected set of LTI systems. The uncertain system $\Omega \star M$ is stable and $\{X(\mathbf{j}\omega), Y(\mathbf{j}\omega), Z(\mathbf{j}\omega)\}$ dissipative for every $\Omega \in \mathbf{\Omega}$ if and only if

- 1) There exists $\Omega_0 \in \mathbf{\Omega}$ such that $\Omega_0 \star M$ is stable.
- 2) There exists a function $\Phi = \Phi^* \in \mathbf{RL}_\infty$ such that Ω is $\{\Phi_{11}(\mathbf{j}\omega), \Phi_{12}(\mathbf{j}\omega), \Phi_{22}(\mathbf{j}\omega)\}$ dissipative for every $\Omega \in \mathbf{\Omega}$
- 3) There exists $\epsilon > 0$ such that $\mathcal{L}_{\mathcal{P}}(M(\mathbf{j}\omega), \Phi_{11}(\mathbf{j}\omega), \dots, \dots, \Phi_{12}(\mathbf{j}\omega), \Phi_{22}(\mathbf{j}\omega), X(\mathbf{j}\omega), Y(\mathbf{j}\omega), Z(\mathbf{j}\omega), \epsilon) \geq 0$

Proof: The proof can be found in [22]. ■

Theorem 3.1 presents necessary and sufficient conditions for the uncertain system $\Omega \star M$ to be $\{X(\mathbf{j}\omega), Y(\mathbf{j}\omega), Z(\mathbf{j}\omega)\}$ dissipative for every $\Omega \in \mathbf{\Omega}$. This theorem requires

- the existence of $\Omega_0 \in \mathbf{\Omega}$ such that $\Omega_0 \star M$ is stable;
- the verification of conditions 2) and 3) for every $\omega \in \overline{\mathbf{R}}^+$.

Satisfying condition 1) is not difficult and it is sufficient to take any $\Omega_0 \in \mathbf{\Omega}$ and test if $\Omega_0 \star M$ is stable. If it is the case, we can continue testing conditions 2) and 3). Otherwise, we can conclude that $\Omega \star M$ is not stable for all $\Omega \in \mathbf{\Omega}$.

Testing conditions 2) and 3) along the frequencies $\omega \in \overline{\mathbf{R}}^+$ is impossible from a practical point of view due to the infinity of tests to be performed.

To overcome this difficulty and as it is usual for LTI systems, a frequency by frequency approach can be performed without loss of generality [23] and a frequency gridding \mathcal{W} is defined where \mathcal{W} is a finite set of frequencies.

In this case and given a frequency $\omega_0 \in \mathcal{W}$, we define the static uncertain system $\Omega(\mathbf{j}\omega_0) \star M(\mathbf{j}\omega_0)$ where $M(\mathbf{j}\omega_0)$ is a given matrix and $\Omega(\mathbf{j}\omega_0) \in \mathbf{\Omega}(\mathbf{j}\omega_0)$ where $\mathbf{\Omega}(\mathbf{j}\omega_0)$ is given by

$$\mathbf{\Omega}(\mathbf{j}\omega_0) = \left\{ \Omega(\mathbf{j}\omega_0) \mid \Omega(\mathbf{j}\omega_0) = \mathbf{bdiag}_j(\Omega_j(\mathbf{j}\omega_0)) \right\} \quad (7)$$

Furthermore, the static uncertain system $\Omega(\mathbf{j}\omega_0) \star M(\mathbf{j}\omega_0)$ is said to be $\{X(\mathbf{j}\omega_0), Y(\mathbf{j}\omega_0), Z(\mathbf{j}\omega_0)\}$ dissipative if the condition (6) is respected with $\omega = \omega_0$ for every $\Omega(\mathbf{j}\omega_0) \in \mathbf{\Omega}(\mathbf{j}\omega_0)$ with $X(\mathbf{j}\omega_0) = X(\mathbf{j}\omega_0)^*$ and $Z(\mathbf{j}\omega_0) = Z(\mathbf{j}\omega_0)^*$.

Assuming the existence of $\Omega_0 \in \mathbf{\Omega}$ such that $\Omega_0 \star M$ is stable and after introducing the frequency gridding set \mathcal{W} , the robust performance analysis problem boils down to a set of matrix problems which can be interpreted as the robustness analysis of the static uncertain system $\Omega(\mathbf{j}\omega_0) \star M(\mathbf{j}\omega_0)$. Each of these problems is given by

Problem 3.1: Let $\Omega(\mathbf{j}\omega_0) \star M(\mathbf{j}\omega_0)$ be a static uncertain system where $M(\mathbf{j}\omega_0)$ is a given matrix and $\Omega(\mathbf{j}\omega_0) \in \mathbf{\Omega}(\mathbf{j}\omega_0)$ where $\mathbf{\Omega}(\mathbf{j}\omega_0)$ is given by (7). Given the matrices $X(\mathbf{j}\omega_0)$, $Y(\mathbf{j}\omega_0)$ and $Z(\mathbf{j}\omega_0)$ with $X(\mathbf{j}\omega_0) = X(\mathbf{j}\omega_0)^*$ and $Z(\mathbf{j}\omega_0) = Z(\mathbf{j}\omega_0)^*$. Test if the static uncertain system $\Omega(\mathbf{j}\omega_0) \star M(\mathbf{j}\omega_0)$ is $\{X(\mathbf{j}\omega_0), Y(\mathbf{j}\omega_0), Z(\mathbf{j}\omega_0)\}$ dissipative for every $\Omega(\mathbf{j}\omega_0) \in \mathbf{\Omega}(\mathbf{j}\omega_0)$.

The following theorem gives necessary and sufficient conditions to perform the robust performance analysis of a static uncertain system $\Omega(j\omega_0) \star M(j\omega_0)$.

Theorem 3.2: Given the matrices $M(j\omega_0)$, $X(j\omega_0)$, $Y(j\omega_0)$ and $Z(j\omega_0)$ and the set $\Omega(j\omega_0)$ of Problem 3.1, the static uncertain system $\Omega(j\omega_0) \star M(j\omega_0)$ is $\{X(j\omega_0), Y(j\omega_0), Z(j\omega_0)\}$ dissipative for every $\Omega(j\omega_0) \in \Omega(j\omega_0)$ if and only if

- 1) There exists a matrix $\Phi(j\omega_0) = \Phi(j\omega_0)^*$ such that $\Omega(j\omega_0)$ is $\{\Phi_{11}(j\omega_0), \Phi_{12}(j\omega_0), \Phi_{22}(j\omega_0)\}$ dissipative for every $\Omega(j\omega_0) \in \Omega(j\omega_0)$
- 2) There exists $\epsilon > 0$ such that $\mathcal{L}_{\mathcal{P}}(M(j\omega_0), \Phi_{11}(j\omega_0), \dots, \Phi_{12}(j\omega_0), \Phi_{22}(j\omega_0), X(j\omega_0), Y(j\omega_0), Z(j\omega_0), \epsilon) \geq 0$

Proof: The proof is similar to the proof of Theorem 3.1 without considering the stability part. ■

In the sequel, for ease of notation, the frequency dependence will be dropped *i.e.* we write $\{X, Y, Z\}$ dissipativity instead of $\{X(j\omega_0), Y(j\omega_0), Z(j\omega_0)\}$ dissipativity. Furthermore, if the static uncertain system $\Omega \star M$ is $\{X, Y, Z\}$ dissipative for every $\Omega \in \Omega$, then the static uncertain system set $\Omega \star M$ is $\{X, Y, Z\}$ dissipative. Moreover, for the sake of simplicity, the static uncertain system set will be referred to as the static uncertain system.

Theorem 3.2 presents necessary and sufficient conditions for the static uncertain system $\Omega \star M$ to be $\{X, Y, Z\}$ dissipative. Testing these conditions is a convex optimization problem. Nevertheless, it is infinite dimensional since condition 1) has to be tested for all $\Omega \in \Omega$ which is impossible from a computational point of view.

In order to obtain a finite dimensional convex optimization problem, let us introduce the set

$$\Phi_{\Omega} = \left\{ \Phi_{\Omega} = \begin{pmatrix} \mathbf{bdiag}_j((\Phi_{11})_j) & \mathbf{bdiag}_j((\Phi_{12})_j) \\ \mathbf{bdiag}_j((\Phi_{12}^*)_j) & \mathbf{bdiag}_j((\Phi_{22})_j) \end{pmatrix} \right\}$$

such that condition 1) is satisfied for all $\Omega \in \Omega$. Let us introduce as well the sets associated to each $\Omega_j \in \Phi_{\Omega_j}$ as

$$\Phi_{\Omega_j} = \left\{ \Phi_{\Omega_j} = \begin{pmatrix} (\Phi_{11})_j & (\Phi_{12})_j \\ (\Phi_{12}^*)_j & (\Phi_{22})_j \end{pmatrix} \right\}$$

Furthermore, for each Ω_j , let us define $\mathcal{B}_{\Omega_j}^{diss}$ sets of conically independent elements $\Phi_j^k \in \Phi_{\Omega_j}$ with $k \in \{1, \dots, n_j^d\}$, *i.e.* each Φ_j^k cannot be expressed as the conic combination of Φ_j^l , $l \in \{1, \dots, n_j^d\} \setminus \{k\}$. Please note, since $\Phi_j^k \in \Phi_{\Omega_j}$ for all $k \in \{1, \dots, n_j^d\}$, the uncertainty Ω_j is $\{(\Phi_{11})_j^k, (\Phi_{12})_j^k, (\Phi_{22})_j^k\}$ dissipative with

$$\begin{pmatrix} (\Phi_{11})_j^k & (\Phi_{12})_j^k \\ (\Phi_{12}^*)_j^k & (\Phi_{22})_j^k \end{pmatrix} = \Phi_j^k \in \mathcal{B}_{\Omega_j}^{diss} \quad (8)$$

Then, let us define $\Phi(\mathcal{B}_{\Omega}^{diss})$ a set of block diagonal conic combinations of the elements of $\mathcal{B}_{\Omega_j}^{diss}$ as

$$\Phi(\mathcal{B}_{\Omega}^{diss}) = \left\{ \Phi^{\Omega} = \begin{pmatrix} \exists \alpha_{kj} \geq 0, k \in \{1, \dots, n_j^d\} \\ \Phi^{\Omega} = \begin{pmatrix} \Phi_{11}^{\Omega} & \Phi_{12}^{\Omega} \\ (\Phi_{12}^{\Omega})^* & \Phi_{22}^{\Omega} \end{pmatrix} \\ \Phi_{gh}^{\Omega} = \mathbf{bdiag}_j \left(\sum_{k=1}^{n_j^d} \alpha_{kj} (\Phi_{gh})_j^k \right) \\ \text{with } g, h \in \{1, 2\} \\ \text{and } \Phi_j^k \in \mathcal{B}_{\Omega_j}^{diss} \end{pmatrix} \right\} \quad (9)$$

Since Φ_{Ω} is a convex cone, we have $\Phi(\mathcal{B}_{\Omega}^{diss}) \subseteq \Phi_{\Omega}$.

A counterpart of Theorem 3.2 with sufficient conditions only is given in the following corollary.

Corollary 3.1: Given the matrices M , X , Y and Z and the set Ω of Problem 3.1 and if there exist basis sets $\mathcal{B}_{\Omega_j}^{diss}$ of conically independent elements Φ_j^i as in (8) for which Ω_j is $\{(\Phi_{11})_j^k, (\Phi_{12})_j^k, (\Phi_{22})_j^k\}$ dissipative $\forall k$, then the static uncertain system $\Omega \star M$ is $\{X, Y, Z\}$ dissipative if there exists a $\Phi^{\Omega} \in \Phi(\mathcal{B}_{\Omega}^{diss})$ as in (9) and there exists $\epsilon > 0$ such that

$$\mathcal{L}_{\mathcal{P}}(M, \Phi_{11}^{\Omega}, \Phi_{12}^{\Omega}, \Phi_{22}^{\Omega}, X, Y, Z, \epsilon) \geq 0 \quad (10)$$

Proof: This is an immediate consequence of Theorem 3.2 after the parametrization of Φ as in (9). ■

Remark 3.2: In contrast with Theorem 3.2 and if there exist the basis $\mathcal{B}_{\Omega_j}^{diss}$, Corollary 3.1 presents sufficient conditions for the static uncertain system $\Omega \star M$ to be $\{X, Y, Z\}$ dissipative. Testing these conditions is a finite dimensional convex optimization problem under LMI constraints with α_{ij} as decision variables. It can be solved efficiently [2]. The consequence of introducing this parametrization $\Phi(\mathcal{B}_{\Omega}^{diss})$ is a possible conservatism. However, it can be reduced by an appropriate choice of $\mathcal{B}_{\Omega_j}^{diss}$, depending on the class of the uncertainty sets Ω_j . The conservatism depends also on how precise each element of $\mathcal{B}_{\Omega_j}^{diss}$ characterizes Ω_j . Furthermore, the larger $\mathcal{B}_{\Omega_j}^{diss}$ sets are, the less conservative results can be obtained.

In the case where the uncertainty set Ω is an elementary uncertainty set *i.e.* $\Omega = \Delta$, the associated basis can be easily obtained from [3] and [4] and it will be denoted by $\mathcal{B}_{\Delta}^{diss}$. In this case $\mathcal{B}_{\Delta}^{diss}$ can be defined as the well-known D or DG scaling sets according to (9).

C. Application to uncertain LSS

A solution to Problem 2.1 can be obtained by applying the tools presented in this section if the LSS is expressed as (5). After gathering all the different Δ_j^i in $\tilde{\Delta}$, defining the global uncertainty set $\tilde{\Delta}$, while the different M_j^i and the interconnections are gathered in \tilde{M} using LFT algebra, the global system T_1^1 will be given by

$$T_1^1 = \tilde{\Delta} \star \tilde{M} \quad (11)$$

with

$$\tilde{\Delta} = \mathbf{bdiag}_i \left(\mathbf{bdiag}_j \left(\Delta_j^i \right) \right), \quad \Delta_j^i \in \Delta, \quad (12)$$

Please note that $\widetilde{M} \in \mathbf{RH}_\infty$. It is a direct consequence of the fact that the LSS is designed to be stable (in the nominal case).

As explained earlier, if there exists $\widetilde{\Delta}_0 \in \widetilde{\Delta}$ such that $\widetilde{\Delta}_0 \star \widetilde{M}$ is stable and after considering a finite frequency set \mathcal{W} , then Problem 2.1 can be treated as a set of robust performance analysis problems of the static uncertain system $\widetilde{\Delta} \star \widetilde{M}$. In these problems, we denote γ_{ω_0} as an upper bound on the system maximum singular value $\bar{\sigma}(\bullet)$ over all the uncertainties for a given frequency $\omega_0 \in \mathcal{W}$. After dropping the frequency dependence, each of these problems is given by

Problem 3.2: Let the static uncertain system $T_1^1 = \widetilde{\Delta} \star \widetilde{M}$ where \widetilde{M} is a given matrix and $\widetilde{\Delta} \in \widetilde{\Delta}$ a set of the uncertain matrices $\widetilde{\Delta}$. Given a frequency ω_0 , solve

$$\begin{aligned} & \min_{\gamma_{\omega_0}} \quad \gamma_{\omega_0} \\ & \text{subject to} \quad \bar{\sigma}(T_1^1) < \gamma_{\omega_0} \quad \forall \widetilde{\Delta} \in \widetilde{\Delta} \end{aligned}$$

We can try to solve Problem 3.2 using a direct application of Corollary 3.1. Since $\widetilde{\Delta}$ is the block diagonal composition of elementary uncertainties $\Delta_j^i \in \Delta$, the associated basis $\mathcal{B}_{\widetilde{\Delta}}^{diss}$ can easily be chosen from [3] and [4]

This approach will be referred to as the *Direct LSS Robust Performance Analysis* (Direct LSS RPA) method and it is summarized in the following corollary.

Corollary 3.2 (Direct LSS RPA): Given a frequency ω_0 , Problem 3.2 can be solved with the following optimization problem

$$\min_{\substack{\gamma_{\omega_0}^2 \\ \widetilde{\Phi} \in \Phi(\mathcal{B}_{\widetilde{\Delta}}^{diss})}} \gamma_{\omega_0}^2$$

subject to

$$\exists \epsilon > 0 \quad \text{s.t.} \quad \mathcal{L}_{\mathcal{P}}(\widetilde{M}, \widetilde{\Phi}_{11}, \widetilde{\Phi}_{12}, \widetilde{\Phi}_{22}, -I, 0, \gamma_{\omega_0}^2 I, \epsilon) \geq 0. \quad (13)$$

Proof: Corollary 3.2 is a direct application of Corollary 3.1 with $X = -I$, $Y = 0$ and $Z = \gamma_{\omega_0}^2 I$ (see Remark 3.1) and the minimization of $\gamma_{\omega_0}^2$ to obtain the lowest upper bound on $\bar{\sigma}(T_1^1)$ at a given frequency ω_0 . ■

Remark 3.3: If there are no performance signals (w_1^1 and z_1^1), it is possible to investigate the robust stability of a network using this direct approach. The robust stability condition is given by $|\det(I - \widetilde{\Delta}(s)\widetilde{M}_{11}(s))| > 0$ for every $\widetilde{\Delta} \in \widetilde{\Delta}$ and for every $s \in \overline{\mathbb{C}}^+$, see [24]. To test this condition, we have to find $\widetilde{\Delta}_0 \in \widetilde{\Delta}$ such that $\widetilde{\Delta}_0 \star \widetilde{M}_{11}$ is stable. Thereafter, we have to solve a set of matrix problems concerning the static interconnection $\widetilde{\Delta}(j\omega_0) \star \widetilde{M}_{11}(j\omega_0)$ at each frequency $\omega_0 \in \mathcal{W}$. Each of these problems can be solved using Corollary 3.2 by replacing condition (13) with $\mathcal{L}_{\mathcal{S}}(\widetilde{M}_{11}, \widetilde{\Phi}_{11}, \widetilde{\Phi}_{12}, \widetilde{\Phi}_{22}) \geq 0$. This approach will be referred to as the *Direct LSS Robust Stability Analysis* (LSS RSA).

As it will be shown later with numerical examples, the direct methods cannot be practically applied when the size of the LSS becomes too important. This is due to the important number of decision variables in the optimization problem in Corollary 3.2 which leads to a dramatic increase of the computation time.

In the next section, we propose a method that allows to take advantage of the hierarchical structure to practically investigate the robust performance of uncertain LSS within a reasonable computation time.

IV. HIERARCHICAL APPROACH

In the previous section, we have shown that the performance can be characterized using dissipativity properties. Nevertheless, Corollary 3.1 reveals that to ensure robust performance, defined by the $\{X, Y, Z\}$ dissipativity property, it is sufficient (and sometimes necessary) to exhibit a dissipativity property, defined by $\Phi^\Omega \in \Phi(\mathcal{B}_{\Omega}^{diss})$, satisfied by all $\Omega \in \Omega$ i.e. Ω is $\{\Phi_{11}^\Omega, \Phi_{12}^\Omega, \Phi_{22}^\Omega\}$ dissipative such that

$$\mathcal{L}_{\mathcal{P}}(M, \Phi_{11}^\Omega, \Phi_{12}^\Omega, \Phi_{22}^\Omega, X, Y, Z, \epsilon) \geq 0$$

holds true. Therefore, the dissipativity properties are suitable to characterize both uncertainty and performance.

Corollary 3.1 allows to characterize the performance with $\{X, Y, Z\}$ dissipativity property knowing that all the uncertainties Ω are $\{\Phi_{11}^\Omega, \Phi_{12}^\Omega, \Phi_{22}^\Omega\}$ dissipative where Φ^Ω is constructed from the basis of the uncertainty Ω_j as in (8) and (9). Thereafter, if we have a method, which will use necessarily Corollary 3.1, that allows to find a basis \mathcal{B}_T^{diss} for the static system $T = \Omega \star M$ from the basis $\mathcal{B}_{\Omega}^{diss}$ of the uncertainty, then Problem 3.2, which corresponds to Problem 2.1 at a given frequency ω_0 (as explained in the previous section), can be solved efficiently in a hierarchical manner. This method will be known as “basis propagation” and it consists in propagating the basis from one hierarchical level to another starting from level l where the uncertainty basis is known since $\Omega_j^l = \Delta_j^i$. This basis propagation method will be performed from basis of level i to basis of level $i - 1$ until reaching level 1 where the objective is to minimize γ_{ω_0} .

The proposed Hierarchical Robust Performance Analysis (HRPA) approach is summarized in the following algorithm

Algorithm 1 Hierarchical Robust Performance Analysis

- ```

% First Step: level $l-1$
1: for $j \leftarrow 1$ to N_T^{l-1} do
2: Find a basis $\mathcal{B}_{T_j^{l-1}}^{diss}$ for each static uncertain system
 T_j^{l-1} from the given basis $\mathcal{B}_{\Delta_j^i}^{diss}$ of elementary uncertainties Δ_j^i .
3: end for
% Intermediate Steps: level $l-2$ to level 2
4: for $i \leftarrow l-2$ to 2 do
5: for $j \leftarrow 1$ to N_T^i do
6: Find a basis $\mathcal{B}_{T_j^i}^{diss}$ for each static uncertain
 system T_j^i using the given basis $\mathcal{B}_{\Delta_m^{i+1}}^{diss}$ of the elementary
 uncertainties Δ_m^{i+1} and the basis $\mathcal{B}_{T_n^{i+1}}^{diss}$ of the static uncertain
 systems T_n^{i+1} obtained at level $i+1$.
7: end for
8: end for

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-

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**Algorithm 1 (Continued)**


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% Last Step: level 1

- 9: Find the smallest  $\gamma_{\omega_0}$  such that the static system  $T_1^1$  is  $\{-I, 0, \gamma_{\omega_0}^2 I\}$  dissipative by applying Corollary 3.1 using the given basis  $\mathcal{B}_{\Delta_m^2}^{diss}$  of the elementary uncertainties  $\Delta_m^2$  and the basis  $\mathcal{B}_{T_n^2}^{diss}$  of the static uncertain systems  $T_n^2$  obtained at level 2.
- 

*Remark 4.1:* The robust stability problem of Remark 3.3 can be solved efficiently using basis propagation. As explained in Remark 3.3, the robust stability problem boils down to a set of matrix problems at each frequency  $\omega_0 \in \mathcal{W}$ . Each of these problems can be solved using Algorithm 1 by replacing condition (10) of Corollary 3.1 in the last step of Algorithm 1 with  $\mathcal{L}_S \left( (M_{11})_1^1, \Phi_{11}^\Omega, \Phi_{12}^\Omega, \Phi_{22}^\Omega \right) \geq 0$ . This will define the Hierarchical Robust Stability Analysis Algorithm. Therefore, Algorithm 1 can be used for both certification: stability and performance after considering the corresponding static system robustness analysis problems.

*Remark 4.2:* Algorithm 1 allows to investigate the performance and the stability of uncertain LSS in an efficient manner after considering the corresponding static systems problems. The global LMI given in Corollary 3.2 will be replaced with several hierarchical LMI of Corollary 3.1 linked with appropriate condition in the next hierarchical levels. Furthermore, given a hierarchical level  $i$ , the different uncertain systems  $T_{\bullet}^i$  do not interfere with each other since their interconnections appear in level  $i-1$  and levels below. Therefore, the analysis performed at each level  $i$  (the performance analysis of the different  $T_{\bullet}^i$ ) can be performed separately and in parallel. The consequence of this parallel analysis will be a very important decrease of the computation time as it will be shown in subsection VI-B.

*Remark 4.3:* In the case of parametric uncertainties and when the size of the uncertain LSS is not too important such that Corollary 3.2 can be practically applied, the basis  $\mathcal{B}_{\hat{\Delta}}^{diss}$  can be chosen in the form of the well-known DG scaling from [3]. In order to reduce the conservatism resulting from choosing this parametrization, the basis  $\mathcal{B}_{\hat{\Delta}}^{diss}$  can be chosen in the form of DGL scaling from [25] instead of DG scaling. In this case and since the basis is larger, the number of decision variables is more important and Corollary 3.2 may not be practically applied even if the size of the LSS is not too important. Nevertheless, Algorithm 1 allows to overcome this issue. Since the analysis in each level is performed on small size systems  $T_j^i$ , it is possible to choose the basis  $\mathcal{B}_{\Delta_m^{i+1}}^{diss}$  in the form of DGL scaling. The consequence will be less conservative results in the overall analysis compared to those when using the hierarchical approach with DG scaling.

In the next section, we present three types of elements for the basis  $\mathcal{B}_{T_{\bullet}}^{diss}$  and we formulate the problems of computing each one of them as a convex optimization problem allowing to minimize its size measure.

## V. PRACTICAL FORMULATION AND COMPUTATION OF BASIS ELEMENTS

In the previous section, we revealed how it is possible to solve Problem 2.1 (subsequently Problem 3.2) with Algorithm 1 using basis propagation from level  $i$  to level  $i-1$  with the assumption that we are able to find the basis elements of each static uncertain system given the basis of its uncertainty. However, the conservatism of the overall result highly depends on the choice of the propagated basis. For this reason, it is important to define and compute the “best” basis for a given uncertain system. In the sequel, we consider a static uncertain system  $T$  such that  $z = Tw$  as in (5) with a given uncertainty basis  $\mathcal{B}_{\Omega}^{diss}$ .

Corollary 3.1 gives sufficient conditions to obtain  $\{X, Y, Z\}$  dissipativity property for the static uncertain system  $T$  by solving the following feasibility optimization problem

$$\mathcal{L}_{\mathcal{P}}(M, \Phi_{11}^\Omega, \Phi_{12}^\Omega, \Phi_{22}^\Omega, X^k, Y^k, Z^k, \epsilon) \geq 0$$

with  $\Phi^\Omega \in \Phi(\mathcal{B}_{\Omega}^{diss})$  is of the form (9) with *a priori* known  $\mathcal{B}_{\Omega_j}^{diss}$ . This feasibility optimization problem is convex and can be solved efficiently. The obtained  $\{X, Y, Z\}$  dissipativity property will define an element of the basis  $\mathcal{B}_T^{diss}$ . However, in order to ensure that this characterization is the best, it is important to define a size measure for each element *i.e.* characterize  $T$  with the optimal  $\{X, Y, Z\}$  in the sense that this size is minimized. In addition, one can characterize the static uncertain system  $T$  not with just one dissipativity property but with  $N$  different dissipativity properties which will be used to construct the basis  $\mathcal{B}_T^{diss}$

$$\begin{pmatrix} X^k & Y^k \\ (Y^k)^* & Z^k \end{pmatrix} \in \mathcal{B}_T^{diss}, \forall k = 1, \dots, N^d$$

In order to construct the “best” basis  $\mathcal{B}_T^{diss}$ , all its elements should be conically independent and capture information, of different nature, characterizing the static uncertain system  $T$  such as gain or phase information.

For the sake of simplicity, the term “static” will be dropped and the static uncertain system will be referred to as the uncertain system.

In this section, for every frequency  $\omega_0$  and in order to have a geometric interpretation of each element of  $\mathcal{B}_T^{diss}$ , we characterize the uncertain system  $T$  in the signals space using the system input and output signals  $w$  and  $z$ . Within this context, a system  $T$  is said to be  $\{X, Y, Z\}$  dissipative if

$$\begin{pmatrix} z \\ w \end{pmatrix}^* \begin{pmatrix} X & Y \\ Y^* & Z \end{pmatrix} \begin{pmatrix} z \\ w \end{pmatrix} > 0 \quad (14)$$

Therefore, the search of each element of  $\mathcal{B}_T^{diss}$  consists in the following: find  $\{X, Y, Z\}$  dissipativity property such that (14) is ensured for all non null input signals  $w$  with its resulting output  $z = \Omega \star Mw$  for all  $\Omega \in \Omega$ .

### A. Ellipsoid

If  $X < 0$ , constraint (14) rewrites in the signals space as

$$(z - z_c)^* (-X)(z - z_c) < w^* (Z - Y^* X^{-1} Y) w$$



with  $z_c = -X^{-1}Yw$  and  $(Z - Y^*X^{-1}Y)$  is hermitian positive definite matrix since  $X < 0$  and constraint (14) holds.

Then, for all non null input signal  $w$ , the corresponding output signal  $z$  belongs to the ellipsoid  $\mathcal{E}_w$  ( $n_z$  dimensional space) centered at  $z_c$  and characterized with  $\mathcal{P}_w$  such that

$$\mathcal{E}_{\mathcal{P}_w} = \{z \in \mathbb{C}^{n_z} \mid (z - z_c)^* \mathcal{P}_w (z - z_c) < 1\} \quad (15)$$

where the matrix  $\mathcal{P}_w$  determines how far the ellipsoid extends in every direction.  $\mathcal{P}_w$  is given by

$$\mathcal{P}_w = \frac{-X}{w^* (Z - Y^*X^{-1}Y) w}$$

We are now interested in finding  $X$ ,  $Y$  and  $Z$  corresponding to the smallest ellipsoid  $\mathcal{E}_{\mathcal{P}_w}$  for all inputs such that  $\|w\| = 1$  and the uncertain system  $\Omega \star M$  is  $\{X, Y, Z\}$  dissipative. For this purpose, a size measure  $v$  for the ellipsoid  $\mathcal{E}_{\mathcal{P}_w}$  can be defined as its volume  $v = \text{vol}(\mathcal{E}_{\mathcal{P}_w})$ , which is given by

$$v = \beta \sqrt{\det(\mathcal{P}_w^{-1})}$$

where  $\beta$  is a positive scalar which depends on  $n_z$ .

*Problem 5.1:* Let  $\Omega \star M$  be an uncertain system. Find  $X$ ,  $Y$  and  $Z$  which

$$\begin{array}{llll} \text{minimize} & \text{maximize} & \text{maximize} & v^2 \\ \text{over} & \text{over } \Omega \in \Omega & \text{over } w & \end{array}$$

$$\text{s.t.} \begin{cases} \|w\| = 1 \\ (\Omega \star M) w \in \mathcal{E}_{\mathcal{P}_w} \end{cases}$$

*Theorem 5.1:* An upper bound  $\widetilde{v}_{opt}$  on  $v_{opt}$  optimal value of Problem 5.1 can be obtained by finding  $X$ ,  $Y$ ,  $Z$ ,  $\Phi_{11}^\Omega$ ,  $\Phi_{12}^\Omega$  and  $\Phi_{22}^\Omega$  with  $\Phi^\Omega = \begin{pmatrix} \Phi_{11}^\Omega & \Phi_{12}^\Omega \\ (\Phi_{12}^\Omega)^* & \Phi_{22}^\Omega \end{pmatrix} \in \Phi(\mathcal{B}_\Omega^{diss})$  that minimize

$$\log(\det(-X^{-1}))$$

subject to

$$\begin{array}{l} 1) \exists \epsilon > 0 \text{ s.t. } \mathcal{L}_{\mathcal{P}}(M, \Phi_{11}^\Omega, \Phi_{12}^\Omega, \Phi_{22}^\Omega, X, Y, Z, \epsilon) \geq 0 \\ 2) \begin{pmatrix} I & 0 \\ 0 & 0 \end{pmatrix} \geq \begin{pmatrix} Z & Y^* \\ Y & X \end{pmatrix}. \end{array}$$

The upper bound  $\widetilde{v}_{opt}$  is given by  $\widetilde{v}_{opt} = \beta \sqrt{\det(-\widetilde{X}^{-1})}$  where  $\widetilde{X} = \text{argmin} \log(\det(-X^{-1}))$  such that condition 1) and condition 2) of Theorem 5.1 hold.

This optimization problem is a determinant maximization under linear matrix inequality constraints [26] and is convex.

*Proof:* Problem 5.1 rewrites

$$\begin{array}{llll} \text{minimize} & \text{maximize} & \text{maximize} & v^2 \\ \text{over} & \text{over } \Omega \in \Omega & \text{over } \|w\| = 1 & \\ \text{subject to} & \Omega \star M \text{ is } \{X, Y, Z\} \text{ dissipative} & & \end{array}$$

Let us introduce the logarithm function on  $v^2$ , which is strictly increasing according to its argument

$$\log(v^2) = 2 \log(\beta) + \log(\det(-X^{-1})) + \log(w^* (Z - Y^*X^{-1}Y) w)$$

<sup>1</sup>The normalization of the input signal  $w$  is absorbed in  $\mathcal{P}_w$ .

Maximizing  $\log(v^2)$  over the inputs  $\|w\| = 1$  and with  $\bar{\sigma}$  as the maximal singular value of  $(Z - Y^*X^{-1}Y)$ , it is possible to write

$$\max_{\|w\|=1} \log(v^2) = 2 \log(\beta) + \log(\det(-X^{-1})) + \log(\bar{\sigma})$$

Since  $(Z - Y^*X^{-1}Y)$  is hermitian definite positive, then its maximal singular value is equal to its largest eigenvalue  $\lambda_{max}$ . We thus have  $\lambda_{max} I \geq (Z - Y^*X^{-1}Y)$ . Furthermore, as a dissipativity property is defined up to a strictly positive multiplicative coefficient and as  $\{X, Y, Z\}$  dissipativity defines the same ellipsoid as  $\{\tau X, \tau Y, \tau Z\}$  dissipativity for any  $\tau > 0$ . Then, one can search for  $X$ ,  $Y$  and  $Z$  such that  $\lambda_{max} = 1$  without loss of generality. Since  $\beta$  is a constant, the optimization problem is equivalent to

$$\begin{array}{ll} \text{minimize} & \text{maximize} \quad \log(\det(-X^{-1})) \\ \text{over} & \text{over } \Omega \in \Omega \\ & \begin{array}{l} X, Y, Z \\ X < 0 \end{array} \\ \text{subject to} & \begin{array}{l} \Omega \star M \text{ is } \{X, Y, Z\} \text{ dissipative} \\ I \geq (Z - Y^*X^{-1}Y) \end{array} \end{array}$$

The last optimization problem is solved by applying Corollary 3.1 which gives condition 1) of Theorem 5.1 while condition 2) is obtained by applying Schur's lemma [27] on  $I \geq (Z - Y^*X^{-1}Y)$ . Please note that since Corollary 3.1 presents sufficient conditions, we are only able to compute an upper bound  $\widetilde{v}_{opt}$  on the optimal volume  $v_{opt}$ . ■

*Remark 5.1:* The interest of finding a bounded and connected set as an ellipsoid, for which belongs the output  $z$ , is to characterize the gain of the uncertain system  $T$ . At each frequency  $\omega_0$ , the frequency response of the uncertain system is embedded in this ellipsoid *i.e.* it is possible to compute boundaries for the system gain at this frequency  $\omega_0$ .

## B. Band

In the signal space and if  $X = 0$ , constraint (14) rewrites as  $\xi^* z - \eta > 0$  with  $\xi = 2Yw \|2Yw\|^{-1}$  and  $\eta = -w^* Z w \|2Yw\|^{-1}$ . This last inequality expresses that for a given non null input signal  $w$ , the output signal  $z$  belongs to a half plane which is characterized by the hyperplane:  $\{z \mid \xi^* z = \eta\}$  where  $\xi$  is a vector normal to the hyperplane and  $\eta$  is twice the distance of the hyperplane to the origin (the dot product of any point of the hyperplane with  $\xi$ ). In addition to a half plane, it is possible to define a band  $\mathcal{B}_{(\eta_1, \eta_2)_w}$  as the intersection of two parallel half planes with the same normal direction  $\xi$  but opposite sign, *i.e.* two parallel half planes defined by two dissipativity properties  $\{0, Y, Z_1\}$  and  $\{0, -Y, Z_2\}$ , that is

$$\mathcal{B}_{(\eta_1, \eta_2)_w} = \left\{ z \in \mathbb{C}^{n_z} \mid \begin{array}{l} \xi^* z - \eta_1 > 0 \\ -\xi^* z - \eta_2 > 0 \end{array} \right\}$$

$$\text{with } \xi = \frac{2Yw}{\|2Yw\|}, \eta_1 = -\frac{w^* Z_1 w}{\|2Yw\|} \text{ and } \eta_2 = -\frac{w^* Z_2 w}{\|2Yw\|}.$$

Our objective now is to find the band  $\mathcal{B}_{(\eta_1, \eta_2)_w}$  with the smallest size measure, for all inputs such that  $\|w\| = 1$  and provided that the uncertain system  $\Omega \star M$  is  $\{0, Y, Z_1\}$

and  $\{0, -Y, Z_2\}$  dissipative. We can define a weighted size measure  $d$  as the weighted width of the band

$$d = \|2Yw\| |\eta_2 - \eta_1|$$

**Problem 5.2:** Let  $\Omega \star M$  be an uncertain system. Find  $Y$ ,  $Z_1$  and  $Z_2$  which

$$\begin{array}{llll} \text{minimize} & \text{maximize} & \text{maximize} & d \\ \text{over } Y, Z_1, Z_2 & \text{over } \Omega \in \Omega & \text{over } w & \end{array}$$

$$\text{subject to } \begin{cases} \|w\| = 1 \\ (\Omega \star M) w \in \mathcal{B}_{(\eta_1, \eta_2)} w \end{cases}$$

**Theorem 5.2:** An upper bound  $\widetilde{d}_{opt}$  on  $d_{opt}$  the optimal value of Problem 5.2 can be obtained by finding  $Y$ ,  $Z_1$ ,  $Z_2$ ,  $(\Phi_{11}^\Omega)^i$ ,  $(\Phi_{12}^\Omega)^i$  and  $(\Phi_{22}^\Omega)^i$  with  $(\Phi^\Omega)^i = \begin{pmatrix} (\Phi_{11}^\Omega)^i & (\Phi_{12}^\Omega)^i \\ (\Phi_{12}^\Omega)^{i*} & (\Phi_{22}^\Omega)^i \end{pmatrix} \in \Phi(\mathcal{B}_\Omega^{diss})$  where  $i = \{1, 2\}$  that minimize

$$d^2$$

subject to

$$\begin{array}{l} 1) \exists \epsilon_1 > 0 \text{ s.t. } \mathcal{L}_{\mathcal{P}} \left( M, (\Phi_{11}^\Omega)^1, (\Phi_{12}^\Omega)^1, (\Phi_{22}^\Omega)^1, 0, Y, Z_1, \epsilon_1 \right) \geq 0; \\ 2) \exists \epsilon_2 > 0 \text{ s.t. } \mathcal{L}_{\mathcal{P}} \left( M, (\Phi_{11}^\Omega)^2, (\Phi_{12}^\Omega)^2, (\Phi_{22}^\Omega)^2, 0, -Y, Z_2, \epsilon_2 \right) \geq 0; \\ 3) \begin{pmatrix} d^2 I & (Z_1 - Z_2)^* \\ (Z_1 - Z_2) & I \end{pmatrix} \geq 0. \end{array}$$

The upper bound  $\widetilde{d}_{opt}$  is given by  $\widetilde{d}_{opt} = \sqrt{\text{argmin } d^2}$  such that conditions 1 and 2 of Theorem 5.2 hold.

This optimization problem is the minimization of a linear cost under LMI constraints [27].

*Proof:* Problem 5.2 rewrites

$$\begin{array}{llll} \text{minimize} & \text{maximize} & \text{maximize} & |w^* (Z_1 - Z_2) w| \\ \text{over } Z_1, Z_2 & \text{over } \Omega \in \Omega & \text{over } \|w\| = 1 & \end{array}$$

$$\text{subject to } \begin{array}{l} \Omega \star M \text{ is } \{0, Y, Z_1\} \text{ dissipative} \\ \Omega \star M \text{ is } \{0, -Y, Z_2\} \text{ dissipative} \end{array}$$

The maximum of  $|w^* (Z_1 - Z_2) w|$  over  $\|w\| = 1$  is equal to  $d$  constrained by

$$(Z_1 - Z_2)^* (Z_1 - Z_2) \leq d^2 I.$$

The optimization is thus equivalent to

$$\begin{array}{llll} \text{minimize} & \text{maximize} & & d^2 \\ \text{over } Z_1, Z_2 & \text{over } \Omega \in \Omega & & \end{array}$$

$$\text{subject to } \begin{array}{l} \Omega \star M \text{ is } \{0, Y, Z_1\} \text{ dissipative} \\ \Omega \star M \text{ is } \{0, -Y, Z_2\} \text{ dissipative} \\ (Z_1 - Z_2)^* (Z_1 - Z_2) \leq d^2 I \end{array}$$

Finally, condition 1) and condition 2) of Theorem 5.2 are obtained by applying Corollary 3.1 while condition 3) is obtained by applying Schur's lemma [27] on  $d^2 I - (Z_1 - Z_2)^* (Z_1 - Z_2) \geq 0$ . Again, since Corollary 3.1 presents sufficient conditions, we are only able to compute an upper bound  $\widetilde{d}_{opt}$  on the optimal width  $d_{opt}$ . ■

### C. Cone Sector

The phase uncertainty presents another important characterization of the uncertain system behavior. In contrast with the system gain, and beside for single input single output systems, there is no unique definition of multiple input multiple

output systems phase. Furthermore, taking into account uncertainties in the system makes the phase characterization more complicated. It is possible to compare the direction variation between the input and the output signals to measure the phase of a system by measuring input direction variation introduced by the system. Furthermore, within this context, the phase uncertainty is characterized using the notion of numerical range as shown in [20] and [28]. The numerical range of a complex matrix  $\Gamma$  is defined to be a compact and convex set of  $\mathbb{C}$  and it is given by [29]

$$\mathcal{N}(\Gamma) = \{w^* z \mid z = \Gamma w, w \in \mathbb{C}^{n_w} \text{ and } \|w\| = 1\} \quad (16)$$

In order to define the phase of an uncertain system  $\Omega \star M$ , the numerical range is extended to the *union of numerical ranges*  $\mathcal{N}(\Omega \star M)$  for any  $\Omega \in \Omega$ . Let us define in the complex plane the cone sector centered at the origin and containing all these numerical ranges. It is defined by a spread angle  $\alpha$  such that  $0 < \alpha < \pi$  and the angle  $\beta$  measured between the bisectrix of  $\alpha$  and the real axis direction. Please refer to [20] for more details. The angle  $\beta$  can be set to zero by introducing a scaling matrix  $\Psi \in \mathbb{C}^{n_z \times n_w}$ . More generally, it is possible to introduce an offset  $z_c$  characterized with a matrix  $C \in \mathbb{C}^{n_z \times n_w}$  such that  $z_c = Cw$  and search for the cone sector  $\mathcal{C}_{(0, \alpha)}$  containing  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  with

$$\mathcal{C}_{(0, \alpha)} = \left\{ w^* z_s \begin{array}{l} \begin{pmatrix} z_s \\ w \end{pmatrix}^* \begin{pmatrix} 0 & Y_1 \\ Y_1^* & Z_1 \end{pmatrix} \begin{pmatrix} z_s \\ w \end{pmatrix} > 0 \\ \begin{pmatrix} z_s \\ w \end{pmatrix}^* \begin{pmatrix} 0 & Y_2 \\ Y_2^* & Z_2 \end{pmatrix} \begin{pmatrix} z_s \\ w \end{pmatrix} > 0 \\ w \in \mathbb{C}^{n_w} \text{ and } \|w\| = 1 \end{array} \right\} \quad (17)$$

where

- $z_s$  belongs to the signal set for which the dot product with the signal  $w$  gives the cone sector  $\mathcal{C}_{(0, \alpha)}$ ;
- $Y_1 = \Psi \left( I + \mathbf{j} \cot \left( \frac{\alpha}{2} \right) I \right)$  and  $Z_1 = -(Y_1^* C + C^* Y_1)$ ;
- $Z_2 = \Psi \left( I - \mathbf{j} \cot \left( \frac{\alpha}{2} \right) I \right)$  and  $Z_2 = -(Y_2^* C + C^* Y_2)$ .

with  $0 < \alpha < \pi$ . We are now interested in finding  $Y_i$  and  $Z_i$  corresponding to the smallest  $\mathcal{C}_{(0, \alpha)}$  for all inputs such that  $\|w\| = 1$ . This inclusion ensures that the uncertain system  $\Omega \star M$  is  $\{0, Y_1, Z_1\}$  and  $\{0, Y_2, Z_2\}$  dissipative. This problem can be solved by finding  $\Psi$  such that the union of the numerical ranges  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  is in the right half plane and then search for  $\alpha$ . Furthermore, to obtain the smallest cone, a size measure  $a$  for  $\mathcal{C}_{(0, \alpha)}$  is needed. It can be defined as the tangent of the angle  $\alpha$

$$a = \tan \left( \frac{\alpha}{2} \right)$$

**Problem 5.3:** Let  $\Omega \star M$  be an uncertain system and given a complex matrix  $C$ . Find  $\Psi$  and  $\alpha$  such that the union of the numerical ranges  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  is in the right half plane and which

$$\begin{array}{llll} \text{minimize} & \text{maximize} & \text{maximize} & a \\ \text{over } \Psi, a & \text{over } \Omega \in \Omega & \text{over } w & \end{array}$$

$$\text{s.t. } \begin{cases} \|w\| = 1 \\ \mathcal{N}(\Psi^*(\Omega \star M - C)) \subset \mathcal{C}_{(0, \alpha)} \end{cases}$$

*Theorem 5.3:* The union of numerical ranges  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  of Problem 5.3 is located in the right half plane if  $\exists \Psi$  and  $\exists \widehat{\Phi} \in \Phi(\mathcal{B}_\Omega^{diss})$  and  $\exists \epsilon > 0$  such that

$$\begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} \geq 0 \quad (18)$$

where

$$B = \begin{pmatrix} -\widehat{\Phi}_{22}^\Omega & 0 & -(\widehat{\Phi}_{12}^\Omega)^* & 0 \\ 0 & -\epsilon I & 0 & \Psi \\ -\widehat{\Phi}_{12}^\Omega & 0 & -\widehat{\Phi}_{11}^\Omega & 0 \\ 0 & \Psi^* & 0 & -(\Psi^*C + C^*\Psi) - \epsilon I \end{pmatrix}.$$

Furthermore, an upper bound  $\widetilde{a}_{opt}$  on  $a_{opt}$ , optimal value of Problem 5.3, can be obtained by finding  $a$ ,  $\widehat{\Phi}^\Omega$  and  $(\widetilde{\Phi}^\Omega)^i$  with  $\widehat{\Phi}^\Omega + a^{-1}(\widetilde{\Phi}^\Omega)^1 \in \Phi(\mathcal{B}_\Omega^{diss})$  and  $\widehat{\Phi}^\Omega - a^{-1}(\widetilde{\Phi}^\Omega)^2 \in \Phi(\mathcal{B}_\Omega^{diss})$  that minimize

$$a$$

subject to

- 1)  $a \begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} + \begin{pmatrix} M \\ I \end{pmatrix}^* A_1 \begin{pmatrix} M \\ I \end{pmatrix} \geq 0$
- 2)  $a \begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} + \begin{pmatrix} M \\ I \end{pmatrix}^* A_2 \begin{pmatrix} M \\ I \end{pmatrix} \geq 0$
- 3) Condition (18)

where  $A_i = (-1)^{i-1} \times \dots$

$$\dots \times \begin{pmatrix} -(\widetilde{\Phi}_{22}^\Omega)^i & 0 & -((\widetilde{\Phi}_{12}^\Omega)^i)^* & 0 \\ 0 & 0 & 0 & \mathbf{j}\Psi \\ -(\widetilde{\Phi}_{12}^\Omega)^i & 0 & -(\widetilde{\Phi}_{11}^\Omega)^i & 0 \\ 0 & (\mathbf{j}\Psi)^* & 0 & \mathbf{j}(\Psi^*C - C^*\Psi) \end{pmatrix}.$$

The upper bound  $\widetilde{a}_{opt}$  is given by  $\widetilde{a}_{opt} = \operatorname{argmin} a$  such that conditions 1), 2) and 3) of Theorem 5.3 hold.

Minimizing  $a$  in Theorem 5.3 such that conditions 1), 2) and 3) hold is a generalized eigenvalues problem which has been proved that it is a quasiconvex optimization problem [27] and it can be solved efficiently [30].

*Proof:* The union of the numerical ranges  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  is in the right half plane if and only if

$$\operatorname{Re}(\mathcal{N}(\Psi^*(\Omega \star M - C))) \geq 0.$$

By (16), which defines the numerical range, the previous condition rewrites as

$$\begin{pmatrix} z \\ w \end{pmatrix}^* \begin{pmatrix} 0 & \Psi \\ \Psi^* & -(\Psi^*C + C^*\Psi) \end{pmatrix} \begin{pmatrix} z \\ w \end{pmatrix} \geq 0 \quad (19)$$

where  $z = \Omega \star M$ . This constraint expresses that the uncertain system  $\Omega \star M$  is  $\{0, \Psi, -(\Psi^*C + C^*\Psi)\}$  dissipative for all  $\Omega \in \Omega$ . By applying Corollary 3.1, (19) is implied by (18) with  $\widehat{\Phi} \in \Phi(\mathcal{B}_\Omega^{diss})$ . Then, the minimization of  $a$  is well-posed when the union of numerical ranges  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  is in the right half plane, that

is  $0 < \frac{\alpha}{2} < \frac{\pi}{2}$ . With  $a = \tan\left(\frac{\alpha}{2}\right)$ ,  $Y_i$  and  $Z_i$  introduced in (17) become

$$\begin{aligned} Y_i &= \Psi + (-1)^{i-1} \mathbf{j} a^{-1} \Psi \\ Z_i &= -(\Psi^*C + C^*\Psi) + (-1)^i \mathbf{j} a^{-1} (\Psi^*C - C^*\Psi) \end{aligned} \quad (20)$$

The problem of minimizing  $a$  rewrites

$$\begin{array}{lll} \text{minimize} & \text{maximize} & \text{maximize} \quad a \\ \text{over } \Psi, a & \text{over } \Omega \in \Omega & \text{over } \|w\| = 1 \\ \text{subject to} & \Omega \star M \text{ is } \{0, Y_1, Z_1\} \text{ dissipative} & \\ & \Omega \star M \text{ is } \{0, Y_2, Z_2\} \text{ dissipative} & \end{array}$$

where  $Y_i$  and  $Z_i$  are given in (20). Nevertheless, since the conditions of Corollary 3.1 are only sufficient, we are able to compute only an upper bound  $\widetilde{a}$  on  $a_{opt}$ , solution of Problem 5.3. Applying Corollary 3.1 gives

$$\begin{aligned} \mathcal{L}_{\mathcal{P}} \left( M, (\Phi_{11}^\Omega)^1, (\Phi_{12}^\Omega)^1, (\Phi_{22}^\Omega)^1, 0, Y_1, Z_1, \epsilon \right) &\geq 0 \\ \mathcal{L}_{\mathcal{P}} \left( M, (\Phi_{11}^\Omega)^2, (\Phi_{12}^\Omega)^2, (\Phi_{22}^\Omega)^2, 0, Y_2, Z_2, \epsilon \right) &\geq 0 \end{aligned} \quad (21)$$

with  $(\Phi^\Omega)^1 \in \Phi(\mathcal{B}_\Omega^{diss})$ ,  $(\Phi^\Omega)^2 \in \Phi(\mathcal{B}_\Omega^{diss})$ .  $Y_i$  and  $Z_i$  are given in (20). However, since  $a^{-1}$  is multiplied by  $\Psi$ , the optimization problem is bilinear in  $a^{-1}$  and  $\Psi$ . In general, optimization problems involving bilinear matrix inequalities are NP hard, except if e.g. it belongs to the particular sub-class of generalized eigenvalue problems which are quasiconvex optimization problems and they can be solved efficiently [27]. In the sequel, we reveal that the minimization of  $a$  can be rewritten as a generalized eigenvalue problem. To this purpose, it is necessary to transform the inequalities (21) in order to obtain bilinear terms which are  $a$  multiplied by positive semidefinite matrices. To this purpose, we enforce  $(\Phi^\Omega)^i = \widehat{\Phi}^\Omega$  with  $i \in \{1, 2\}$ . Nevertheless, fixing this structure to  $(\Phi^\Omega)^i$  will increase the conservatism. The latter can be reduced with a more appropriate choice of  $(\Phi^\Omega)^i$  such as  $(\Phi^\Omega)^1 = \widehat{\Phi}^\Omega + a^{-1}(\widetilde{\Phi}^\Omega)^1$  and  $(\Phi^\Omega)^2 = \widehat{\Phi}^\Omega - a^{-1}(\widetilde{\Phi}^\Omega)^2$ . This choice is interesting in the sense that  $\widehat{\Phi}^\Omega$  will be used to make  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  in the right half plane while  $(\widetilde{\Phi}^\Omega)^i$  will ensure extra degree of freedom to minimize  $\widetilde{a}$ . Developing (21) and factorizing  $a^{-1}$  give

$$\begin{aligned} \begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} + a^{-1} \begin{pmatrix} M \\ I \end{pmatrix}^* A_1 \begin{pmatrix} M \\ I \end{pmatrix} &\geq 0; \\ \begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} + a^{-1} \begin{pmatrix} M \\ I \end{pmatrix}^* A_2 \begin{pmatrix} M \\ I \end{pmatrix} &\geq 0. \end{aligned}$$

Since  $0 < \alpha < \pi$ , then  $a = \tan\left(\frac{\alpha}{2}\right) > 0$  and we obtain

$$\begin{aligned} a \begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} + \begin{pmatrix} M \\ I \end{pmatrix}^* A_1 \begin{pmatrix} M \\ I \end{pmatrix} &\geq 0; \\ a \begin{pmatrix} M \\ I \end{pmatrix}^* B \begin{pmatrix} M \\ I \end{pmatrix} + \begin{pmatrix} M \\ I \end{pmatrix}^* A_2 \begin{pmatrix} M \\ I \end{pmatrix} &\geq 0. \end{aligned}$$

Minimizing  $a$  with the latter conditions is a generalized eigenvalue problem since condition (18) holds. Please note that since Corollary 3.1 presents sufficient conditions, we are only able to compute an upper bound  $\widetilde{a}_{opt}$  on the optimal  $a_{opt}$ . ■

*Remark 5.2:* The interest of finding a cone sector is to be able to characterize the phase variation by embedding the numerical range of the uncertain system inside the cone sector. This phase uncertainty information can be very important and critical in some applications such as the active control of vibrations. Please note that for SISO system, the offset  $z_c = Cw$ , with  $\|w\| = 1$ , boils down to one point  $z_c = C$  and it is possible to interpret the cone sector  $\mathcal{C}_{(0,\alpha)}$  containing  $\mathcal{N}(\Psi^*(\Omega \star M - C))$  as the cone sector  $\mathcal{C}_{(z_c, \alpha, \arg(\Psi))}$  containing  $\mathcal{N}(\Omega \star M)$ . Furthermore, if  $C = 0$  and  $\Omega = 0$ , the numerical range boils down into one point and the cone sector is nothing else than the phase of the nominal SISO system, according to the origin, at a given frequency  $\omega_0$ .

In the next section we present two applications examples to investigate the robustness of uncertain LSS using the hierarchical approach.

## VI. APPLICATIONS OF THE HIERARCHICAL APPROACH

### A. Performance analysis of PLL network

We consider in this first example the performance analysis of the active clock distribution network of [31]. It is composed of  $N = 16$  mutually synchronized Phase Locked Loop (PLL) delivering clock signals to the chip. In order to synchronize all the network, the PLLs exchange information through an interconnection structure. This example is suitable for illustration of the proposed hierarchical analysis approach as the performance is naturally evaluated in the frequency domain.

1) *PLL network description:* The description of the  $N$  PLLs frequency responses are given by

$$T_j(j\omega_0) = \frac{k_j(j\omega_0 + a_j)}{-\omega^2 + k_j j\omega_0 + k_j a_j} \quad \forall j \in \{1, \dots, N\}$$

where  $k_j$  and  $a_j$  are the PLL parameters and  $\omega_0$  is the current frequency defined by griding. Due to the manufacturing process, technological dispersions are inevitable and the PLL parameters  $k_j$  and  $a_j$  are uncertain:  $k_j \in [0.76, 6.84] \times 10^4$  and  $a_j \in [91.1, 273.3]$ . Furthermore, all the PLLs are homogeneous *i.e.* have the same description and uncertainty set  $\Delta$ . Therefore, and after normalization, it is possible to present the PLLs as the interconnection of a certain part and uncertain one

$$T_j(j\omega_0) = \Delta_j \star M_{PLL} \quad \Delta_j \in \Delta \quad \forall j \in \{1, \dots, N\}$$

where  $\Delta$  is given by

$$\Delta = \left\{ \Delta = \begin{pmatrix} \delta_k & 0 \\ 0 & \delta_a \end{pmatrix} \quad \delta_k \in \mathbb{R}, \quad \delta_a \in \mathbb{R} \quad \|\Delta\|_\infty < 1 \right\}$$

The information exchange between PLLs is modeled by an interconnection matrix  $M_{int}$  defined in the equation (13) of [18].

Therefore, the PLL network has a three level hierarchical structure and it is a sub-class of the LSS of (3) and (4). The different parameters of this hierarchical structure are presented in TABLE I.

|                             |                                                                                                                        |
|-----------------------------|------------------------------------------------------------------------------------------------------------------------|
| $l$                         | 3                                                                                                                      |
| $N_T^i$                     | $N_T^1 = 1, N_T^2 = N$                                                                                                 |
| $\mathcal{N}_\Delta(M_j^i)$ | $\mathcal{N}_\Delta(M_j^2) = \{j\} \quad \forall j \in \{1, \dots, N_T^2\}$<br>$\mathcal{N}_\Delta(M_1^1) = \emptyset$ |
| $\mathcal{N}_T(M_j^i)$      | $\mathcal{N}_T(M_j^2) = \emptyset$<br>$\mathcal{N}_T(M_1^1) = \{1, \dots, N\}$                                         |
| $M_j^i$                     | $M_1^1 = M_{int}, M_j^2 = M_{PLL} \quad \forall j \in \{1, \dots, N_T^2\}$                                             |

TABLE I: Three level hierarchical structure parameters

The network performance is characterized by its global input and output signals  $w_g$  and  $z_g$  using the global frequency response magnitude bound (more details in [18] and [19]).

2) *Hierarchical Approach:* Algorithm 1 is applied in two steps.

- Local step: compute a basis  $\mathcal{B}_{T_j}^{diss}$  for each PLL by applying Corollary 3.1 using the given basis  $\mathcal{B}_{\Delta_j}^{diss}$  of the elementary uncertainties  $\Delta_j$ . In order to construct the basis  $\mathcal{B}_{T_j}^{diss}$ , we consider several elements
  - ellipsoid: we can use Theorem 5.1 to find a disc<sup>2</sup> center and an upper bound on the optimal radius;
  - band: we can use Theorem 5.2 to find an upper bound on the band width with free orientation;
  - cone sector: given a scalar  $C$ , we can use Theorem 5.3 to find an upper bound of the cone angle.

Please note that since all the PLLs are homogeneous, the basis  $\mathcal{B}_{T_j}^{diss}$  obtained for one PLL is valid for all the 16 PLL. Therefore, computation time of the local step is equal to computation time required to find the basis elements for one PLL.

- Global step : compute the smallest  $\gamma_{\omega_0}$  such that the network, denoted  $T_1^1$ , is  $\{-I, 0, \gamma_{\omega_0}^2 I\}$  dissipative by applying Corollary 3.1 using the basis  $\mathcal{B}_{T_j}^{diss}$  obtained in local step. In this step, one can combine the elements of  $\mathcal{B}_{T_j}^{diss}$  to characterize each PLL by: ellipsoid alone, ellipsoid and band, ellipsoid and cone, *etc.* Then, propagate these characterizations to investigate the performance of the network in the global step.

3) *Results:*

- Local step : In this step, we are interested in finding three elements of the basis  $\mathcal{B}_{T_j}^{diss}$ : ellipsoid, band and cone sector. The frequency response of a PLL is presented in Fig. 2 in the complex plane for a given frequency  $\omega_0 = 480$  rad/s. The red dot is the nominal frequency response and the green dots represent a sampling of the uncertain frequency responses obtained for a sampling of  $a_j$  and  $k_j$  for illustration purposes only. The three elements of  $\mathcal{B}_{T_j}^{diss}$  are interpreted in geometric terms and presented in Fig. 2. The obtained disc is pre-

<sup>2</sup>In single input single output case, the ellipsoid boils down to a disc.

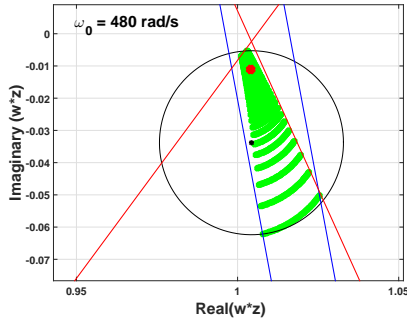


Fig. 2: Three elements of  $\mathcal{B}_{T_j}^{disc}$  characterizing each PLL

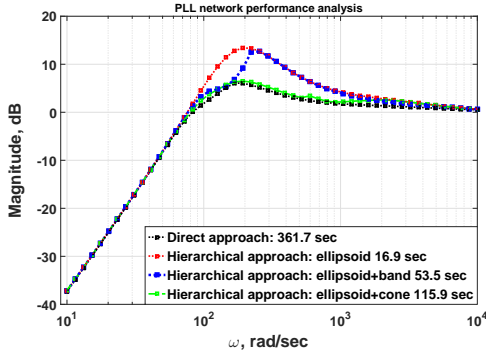


Fig. 3: The PLL network performance analysis

sented in black with its center. We can see that the radius is minimized. The blue lines represent the obtained band with a free direction and the band width is also minimized. The red lines represent the obtained cone sector with its center  $C$  fixed as  $C = c_{disc} + \mathbf{j} \times 1.1 \times r_{disc}$  where  $c_{disc}$  and  $r_{disc}$  are the disc center and radius respectively and the cone angle is minimized. Please note that since the uncertainty  $\Delta$  is parametric and in order to reduce the conservatism, the basis  $\mathcal{B}_{\Delta}^{disc}$  is chosen in the form of DGL scaling from [25], see Remark 4.3.

- Global step: The performance analysis results of the PLL network are presented in Fig. 3 and summarized in TABLE II. Please note that the direct approach corresponds to the approach of Corollary 3.2.

4) *Discussion*: All the analysis reveal that the PLL network is able to track a ramp as the slope of frequency response magnitude at low frequencies is 40 dB/dec (see [31] for more details) and the performance specifications are satisfied.

| Approach                     | Maximum peak | Computation time |
|------------------------------|--------------|------------------|
| Direct                       | 6.01 dB      | 361.7s           |
| Hierarchical: ellipsoid      | 13.44 dB     | 16.9s            |
| Hierarchical: ellipsoid+band | 12.97 dB     | 53.5s            |
| Hierarchical: ellipsoid+cone | 6.45 dB      | 115.9s           |

TABLE II: Comparison between the different approaches

The direct approach presents the less conservative results with a maximum peak of 6.01dB compared to the different hierarchical approaches. However, computation time is significant: 361.7s. TABLE II illustrates the trade-off between conservatism and computation time that can be set by the user with the hierarchical approach. Please note that TABLE II gives the overall computation time for each approach *i.e.* computation time required for both steps. When using the disc alone, the results are conservative but they are obtained faster. The results are less conservative when using the disc with the band, but they are obtained in more time. However the maximum peak is also important. When using the disc with a cone sector, the results are much less conservative. Actually, they are close to the results of the direct approach: the difference in the maximal peak value with direct approach is +0.39 dB, that corresponds to 4.73 % of ratio. In addition, the results are obtained in 32.04 % of the time needed for direct approach.

### B. Stability Analysis of a Chain of Uncertain Systems

In this second example, we consider the stability analysis of a chain of uncertain systems taken from [16]. Our objective is to compare the computation time required in our hierarchical method, the direct analysis approach and the method proposed in [16].

1) *Uncertain system chain description*: Consider a chain of  $N$  uncertain system  $T_j$ . The uncertainties are assumed to be scalar reals:  $\delta_1, \dots, \delta_N$  *i.e.* each system has one parametric uncertainty. The inputs and the outputs of each system  $T_j$  are denoted  $w_j$  and  $z_j$ .

For  $j = \{2, \dots, N-1\}$ , each system  $T_j$  has two inputs and two outputs *i.e.*  $w_j, z_j \in \mathbb{C}^2$  while  $w_j, z_j \in \mathbb{C}$  for  $j = \{1, N\}$ ; hence it is possible, after normalization, to represent each uncertain system as

$$T_j = \Delta_j \star M_j \quad \Delta_j \in \mathbf{\Delta}$$

with  $\mathbf{\Delta}$  is given by

$$\mathbf{\Delta} = \{\Delta = \delta, \delta \in \mathbb{R} \quad \|\Delta\|_{\infty} < 1\}.$$

Since each uncertain system  $T_j$  is only connected to  $T_{j-1}$  and to  $T_{j+1}$ , see Fig.4, the interconnections in this chain are defined by:  $w_j^2 = z_{j+1}^1$  and  $w_j^1 = z_{j-1}^2$  for  $j = \{2, \dots, N-1\}$  and by  $w_1 = z_2^1$  and  $w_N = z_{N-1}^2$  for  $j = \{1, N\}$ . The global system is given by

$$T_{global} = \mathbf{bdiag}_j(T_j) \star M_{global}$$

where  $M_{global}$  is the global interconnection matrix. This matrix is sparse and it corresponds to a chordal graph, please refer to [16] for more details.

The systems are generated randomly as explained in [16] where the authors considered three conditions that different systems should satisfy: each system has to be nominally and robustly stable and the chain of  $N$  systems has to be nominally stable.

A possible application corresponding to the uncertain system chain of Fig. 4 is when an electric current is flowing through a line with a length that is not negligible compared to the



Please note that in order to compare our results with those of [16], the basis  $\mathcal{B}_{\Delta}^{diss}$  is the D-scaling from [4].

The average CPU time required for the direct approach increases dramatically when increasing the size of the network and it can reach 37.40s for a chain of 200 systems. For the hierarchical approach, the analysis is performed in a parallel manner at each hierarchical level. The required CPU time, at level  $i$ , is the maximum among all the CPU time required for the analysis of each subsystem  $T_j^i$  and the global CPU time is the sum of the computation time of each level. The average CPU time required for  $N = 200$  is 0.2552s which gives a ratio of 154 between the time required in direct and hierarchical approaches. For comparison, this ratio is 10 in [16].

Please note that for small size chains (less than approximately 36), the computation time of the hierarchical approach is more important compared to the direct approach. When the chain size is increasing, the computation time of the direct approach becomes more and more important compared to the hierarchical approach. This behavior can be explained by analyzing how the network size  $N$  affects the computation time in both approach. The objective of the next section is to discuss how the computation time grows according to  $N$ .

## VII. COMPUTATION TIME FOR A SUB-CLASS OF UNCERTAIN LSS

The hierarchical structure introduced in the second example of the previous section can be generalized by regrouping systems  $n_s$  by  $n_s$  instead of two by two, to obtain  $n_s$  by  $n_s$  fully split hierarchical structure. Without loss of generality, let us suppose that  $N$  the number of systems in the chain is given by  $N = n_s^\kappa$  where  $\kappa$  is a positive integer.

The chain of uncertain systems with  $n_s$  by  $n_s$  fully split hierarchical structure is a sub-class of the LSS of (3) and (4). The different parameters of this hierarchical structure are given as in TABLE III with the following differences

- $l = \kappa + 2$ ;
- $N_T^i = n_s^{i-1}$ ;
- $\mathcal{N}_T(M_j^i) = \{n_s(j-1) + 1, \dots, n_s j\}$ ;
- $M_j^i$ , for  $i \neq l-1$ , are adapted according to  $n_s$ .

Furthermore, the stability problem considered in the previous section can be extended to a performance problem *i.e.* investigate if the chain is  $\{X_1, Y_1, Z_1\}$  dissipative for all possible uncertainties by introducing the global performance signal  $w_g$  and  $z_g$  and adapting  $M_1^1$ .

For illustration purpose of the computation time, and without lost of generality, we suppose that all the uncertainties  $\Delta_j^l$  are homogeneous. Furthermore, again and without lost of generality, we perform basis propagation in hierarchical approaches, from level  $l-1$  to level 2, with only one element of  $\mathcal{B}_{T_j^i}^{diss}$  such as the ellipsoid.

The objective is to characterize the computation time required to perform robustness analysis, at a given frequency  $\omega_0$ , using the different approaches: direct, hierarchical and parallel hierarchical.

## Computation complexity

Computation time of an algorithm characterizes how fast or slow this algorithm performs. This computation time depends on the algorithm complexity which is defined as a numerical function of the number of variables  $n$ .

Given an optimization problem under LMI constraints, the algorithmic complexity is  $\mathcal{O}(n^3)$  when using the interior point method [30]. The computation time of an algorithm depends on implementation details such as: processor speed, instruction set, *etc.* For this reason, in the sequel, the computation time is referred to as the algorithmic complexity.

### A. Direct approach

For the direct analysis approach and given a dissipativity property  $\{X_1, Y_1, Z_1\}$ , the number of decision variables of the robust performance analysis is  $\eta_{\{X_1, Y_1, Z_1\}} + \sum_{k=1}^N \eta_{\Delta_k}$  where  $\eta_{\{X_1, Y_1, Z_1\}}$  is the number of decision variables corresponding to  $\{X_1, Y_1, Z_1\}$  and  $\eta_{\Delta_k}$  is the number of decision variables corresponding to each elementary uncertainty  $\Delta_k$ . In the case where all the uncertainties are homogeneous, all the  $\eta_{\Delta_k}$  become  $\eta_{\Delta}$  and the required computation time is given by

$$T_{DA} = \mathcal{O}\left(\left(\eta_{\{X_1, Y_1, Z_1\}} + N\eta_{\Delta}\right)^3\right)$$

which is cubic according to  $N$ .

### B. Hierarchical approach

Given  $l = \kappa + 2$  and  $N_T^i = n_s^{i-1}$ . The total number of analysis to be performed for all levels is the sum of  $\kappa + 1$  first terms of a geometric progression, that is

$$\underbrace{1}_{\text{level 1}} + \underbrace{n_s + n_s^2 + \dots + n_s^{\kappa-1}}_{\text{intermediate levels}} + \underbrace{n_s^\kappa}_{\text{level } l-1} = \frac{Nn_s - 1}{n_s - 1}$$

The algorithmic complexity of the hierarchical approach depends on the number of performed tests. The latter changes according to the considered hierarchical structure *i.e.* the number of levels and the number of analysis at each level

- each of the  $N$  analysis problem, at level  $l-1$ , has  $(\eta_{\Delta} + \eta_{\{X, Y, Z\}})$  decision variables with  $\eta_{\{X, Y, Z\}}$  is the number of decision variables corresponding to one  $\{X, Y, Z\}$  dissipativity property to be propagated. The computation time is  $N \mathcal{O}\left(\left(\eta_{\Delta} + \eta_{\{X, Y, Z\}}\right)^3\right)$ .
- each of the  $N \frac{1}{n_s - 1} - \frac{n_s}{n_s - 1}$  intermediate analysis problems, from level  $l-2$  to level 2, has  $(n_s + \eta_{\{X, Y, Z\}})$  decision variables<sup>4</sup>. The required computation time is  $\left(N \frac{1}{n_s - 1} - \frac{n_s}{n_s - 1}\right) \mathcal{O}\left(\left(n_s + \eta_{\{X, Y, Z\}}\right)^3\right)$ .
- at level 1, the objective is to investigate if  $T_1^1(j\omega_0)$  is  $\{X_1, Y_1, Z_1\}$  dissipative. The analysis problem

<sup>4</sup>Here  $n_s$  corresponds to the number of  $\alpha_{kj}$  of (9), with  $k = 1$  and  $j \in \{1, \dots, n_s\}$ , corresponding to the  $n_s$  dissipativity properties propagated from level  $i+1$ .

has  $(n_s + \eta_{\{X_1, Y_1, Z_1\}})$  decision variables and the computation time is  $\mathcal{O}\left((n_s + \eta_{\{X_1, Y_1, Z_1\}})^3\right)$ .

Then the global computation time, at a given frequency  $\omega_0$ , of the hierarchical approach  $T_{HA}$  is

$$T_{HA} = N \mathcal{O}\left((\eta_\Delta + \eta_{\{X, Y, Z\}})^3\right) + \mathcal{O}\left((n_s + \eta_{\{X_1, Y_1, Z_1\}})^3\right) + \left(N \frac{1}{n_s - 1} - \frac{n_s}{n_s - 1}\right) \mathcal{O}\left((n_s + \eta_{\{X, Y, Z\}})^3\right)$$

which is affine with respect to  $N$ .

### C. Parallel hierarchical approach

In this approach, the computation time at each level  $i$  is the maximum among computation time required for the analysis problems at this level  $i$ .

- at level  $l - 1$ , the required computation time is  $\mathcal{O}\left((\eta_\Delta + \eta_{\{X, Y, Z\}})^3\right)$
- from level  $l - 2$  to level 2, since the number of intermediate levels is  $l - 3$  and since  $(l - 2) = \kappa$ , that is  $\log_{n_s}(N)$ , the required computation time is  $(\log_{n_s}(N) - 1) \mathcal{O}\left((n_s + \eta_{\{X, Y, Z\}})^3\right)$
- at level 1, the computation time is  $\mathcal{O}\left((n_s + \eta_{\{X_1, Y_1, Z_1\}})^3\right)$

The global computation time required in this parallel hierarchical approach  $T_{PHA}$  is

$$T_{PHA} = \mathcal{O}\left((\eta_\Delta + \eta_{\{X, Y, Z\}})^3\right) + \mathcal{O}\left((n_s + \eta_{\{X_1, Y_1, Z_1\}})^3\right) + \log_{n_s}(N) \mathcal{O}\left((n_s + \eta_{\{X, Y, Z\}})^3\right) - \mathcal{O}\left((n_s + \eta_{\{X, Y, Z\}})^3\right)$$

which is logarithmic according to  $N$ .

We are now able to explain the results of Fig.5. Since the objective is to test the stability of the chain,  $\eta_{\{X_1, Y_1, Z_1\}} = 0$  and the computation time  $T_{DA}$  becomes

$$T_{DA} = \mathcal{O}\left((N\eta_\Delta)^3\right)$$

For the parallel hierarchical approach with  $n_s = 2$ , the computation time  $T_{PHA}$  becomes

$$T_{PHA} = \mathcal{O}\left((2)^3\right) \log_2(N) \mathcal{O}\left((2 + \eta_{\{X, Y, Z\}})^3\right) + \mathcal{O}\left((\eta_\Delta + \eta_{\{X, Y, Z\}})^3\right)$$

The algorithmic complexity of the direct approach is cubic according to  $N$ . For  $N < 36$ , the algorithmic complexity is not really costly from computation time point of view and the analysis can be performed relatively fast. In the parallel hierarchical approach, even if finding one  $\{X, Y, Z\}$  for each uncertain systems is not costly (since we are dealing with small size systems), the overall algorithmic complexity is important compared to direct approach because of the multiplication term by  $\log_2(N)$ . However, when  $N$  is increasing, the evolution of the logarithmic function is less important than the cubic evolution which makes  $T_{DA}$  really significant compared to  $T_{PHA}$ . Fig.5 confirms this trend. Please note that the scales in this figure are logarithmic.

*Remark 7.1:* It is possible to perform the same analysis for the case when  $N \neq n_s^\kappa$ . The number of levels and systems at each level will change and the expressions of  $T_{HA}$  and  $T_{PHA}$  will be slightly different. However, the affine and logarithmic evolution of  $T_{HA}$  and  $T_{PHA}$  remain valid.

## VIII. CONCLUSION AND FUTURE WORK

In this paper, robustness analysis of uncertain large scale systems with hierarchical structure is considered. In order to reduce the computational load, a hierarchical robust performance analysis algorithm is proposed. This algorithm performs several hierarchical analysis using basis propagation from one hierarchical level to another. We have also presented how to formulate and compute several basis elements. The efficiency of this algorithm is illustrated through two examples: PLL network and chain of uncertain systems. The proposed algorithm allows to establish a trade-off between conservatism and computation time. Furthermore, we discuss the computation time for a sub-class of uncertain LSS. In contrast with cubic evolution of computation time with respect to  $N$  in the direct approach, the computation time grows in an affine manner in the hierarchical approach and logarithmic manner in the parallel hierarchical approach. Therefore, the hierarchical approaches are more suitable and adapted to perform the robustness analysis when  $N$  becomes very significant. Nevertheless, in this paper, we did not discuss how it is possible to obtain hierarchical structures neither the advantage of one hierarchical structure with respect to another. Therefore, major future work directions are

- establish a systematic approach to obtain the optimal hierarchical structure with respect to computation time and conservatism;
- combine the hierarchical approach with other approaches as the one presented in [15] and [16] *i.e.* introduce specialized solvers into the hierarchical approach.

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