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Finite element analysis of electromagnetic waves in two-dimensional transformed bianisotropic media

YAN LIU. BORIS GRALAK, AND SEBASTIEN GUENNEAU²

¹School of Aerospace Science and Technology, Xidian University, Xi'an 710071, China 'Aix-Marseille Université, CNRS, Centrale Marseille, Institut Fresnel, Marseille, France vanliu@xidian.edu.cn

Abstract: We analyze the wave propagation in two-dimensional bianisotropic media with the Finite Element Method (FEM). Starting from the Maxwell-Tellegen's equations in bianisotropic media, we derive some system of coupled Partial Differential Equations (PDEs) for longitudinal electric and magnetic field components. These PDEs are implemented in FEM using a solid mechanics formulation. Perfectly Matched Layers (PMLs) are also discussed to model unbounded bianisotropic media. The PDEs and PMLs are then implemented in a finite element software, and transformation optics is further introduced to design some bianisotropic media with interesting functionalities, such as cloaks, concentrators and rotators. In addition, we propose a design of metamaterial with concentric layers made of homogeneous media with isotropic permittivity, permeability and magnetoelectric parameters that mimic the required effective anisotropic tensors of a bianisotropic cloak in the long wavelength limit (homogenization approach). Our numerical results show that transformation based electromagnetic metamaterials can be extended to bianisotropic media.

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1. Introduction

In the last decade, metamaterials have attracted much attention due to their extraordinary properties, such as negative refraction, ultra refraction, anomalous dispersion and so on [1–3]. Metamaterials are artificial materials engineered to gain desired properties that cannot be found in nature and usually consist of periodically arranged materials, which affect electromagnetic waves in an unconventional manner. More precisely, they exhibit new and unusual electromagnetic properties at the macro-scale, due to their structural features smaller than the operational wavelength of the electromagnetic wave. Metamaterials include the negative index materials (NIMs), single negative metamaterials, bi-isotropic and bianisotropic media, chiral media and so on. As an application, the physicist John Pendry proposed that NIMs enable a flat perfect lens [1], which allows deep sub-wavelength imaging. In 2006, Pendry and Leonhardt [4, 5] proposed that a geometric transformation of space could distort light trajectories around a bounded region, which is thus made invisible. This powerful mathematical technique is called transformation optics, and it presents great potential as it drives the fabrication of metamaterials with specially designed properties [6–9], such as invisibility cloaks, rotators, concentrators, etc.

In NIMs, or single negative metamaterials (e.g. composites with $\varepsilon < 0$ or $\mu < 0$), it is supposed that metamaterials have independent electric and magnetic responses described by ε and μ . However, magnetoelectric coupling does occur in many electromagnetic metamaterials (even in the static case [10]), wherein the electric and magnetic fields are induced reciprocally. Anisotropic (resp. isotropic) media with such kind of property are referred as bianisotropic (resp. bi-isotropic) [11]. Bianisotropic media can be achieved by variations of the split ring resonator (SRR) geometry [12], direct laser writing and silver shadow evaporation on a polymeric template [13], stack of dielectric materials [14, 15], etc.

In [16, 17], we have theoretically proved that the Pendry-Ramakrishna lens theorem is applica-

ble to complementary bianisotropic media. This uses the fact that as a powerful mathematical technique - transformation optics allows for a transformation of space between two different coordinate systems, while the Maxwell-Tellegen's equations for bianisotropic media are also proved to be form invariant under space transformation. Although it is easy to understand the principle used to design novel devices such as cloaks, concentrators, rotators and so on, a numerical model is required to illustrate their electromagnetic response.

In this paper, we would like to discuss in details the implementation of bianisotropic media in a finite element model, and further show numerical results for an invisibility cloak, concentrator and rotator, the former being even achieved with a simplified multilayered design. On the other hand, we know that, Partial Differential Equations (PDEs) are generally used to model many physical phenomena such as fluid dynamics and electromagnetism. In complex media, solutions to the governing equations can be difficult to derive in closed-form by traditional analytical routes - Fourier or Laplace transform methods, power series expansion and so on, hence one often has to resort to numerical approximation of the solutions. As a particular class of numerical techniques, Finite Element Method (FEM) is efficient to solve problems in heterogeneous anisotropic media [18, 19], and it has been widely applied in engineering design and analysis in mechanics starting from the late seventies, while researchers in photonics started to use it rather in the nineties. Notably, the COMSOL Multiphysics package is a finite element analysis software much used nowadays to solve various physics and engineering applications in the metamaterials' community, as it allows entering coupled systems of PDEs, such as in opto-mechanics, thermal elasticity, etc. Here, we make use of it to set up a coupled PDE system for a bianisotropic structure. To do this, we start from the Maxwell-Tellegen's equations in bianisotropic media and derive two coupled PDEs with the longitudinal electric and magnetic fields as the unknowns. This PDE system has the same structure as that for (coupled pressure and shear) in-plane elastic waves. Then we discuss open boundary conditions known in the photonics literature as perfectly matched layers (PMLs). To illustrate the usefulness of our numerical algorithm, we apply transformation optics to bianisotropic media in order to design some functional devices, i.e. invisibility cloak, concentrator and rotator. Their electromagnetic (EM) properties studied thanks to our PDEs model implemented in COMSOL Multiphysics prove that these well-known devices work equally well in bianisotropic media.

2. A coupled PDE system for bianisotropic media

We first recall the time-harmonic Maxwell's equations (assuming a time dependence exp(−*i*ω*t*) with t the time variable and ω the angular wave frequency)

$$
\nabla \times E = i\omega B, \n\nabla \times H = -i\omega D,
$$
\n(1)

where *E* and *H* are the electric and magnetic field respectively, while *D* and *B* are the electric displacement and magnetic flux density.

Let us assume the constitutive relations in a bianisotropic medium described by

$$
D = \varepsilon E + i\xi H, B = -i\zeta E + \mu H,
$$
 (2)

with ε the permittivity, μ the permeability, ξ and ζ the magnetoelectric coupling parameters. All these material parameters can be treated as rank-2 tensors. We now substitute Eq. (2) into Eq. (1) and obtain the Maxwell-Tellegen's equations

$$
\nabla \times E = \omega \zeta E + i\omega \mu H ,
$$

\n
$$
\nabla \times H = -i\omega \varepsilon E + \omega \zeta H .
$$
 (3)

Now we consider an orthogonal coordinate system (x_1, x_2, x_3) , and assume that the material parameters and the electromagnetic field are invariant in one direction, say, the x_3 direction, that is, ε , μ , ξ , ζ , E , H are independent of x_3 . is, ε , μ , ξ , ζ , E , H are independent of x_3 .
Let us define $F = (F_1, F_2, F_3)^T$ H

Let us define $E = (E_1, E_2, E_3)^T$, $H = (H_1, H_2, H_3)^T$, and further assume following the minology used in the book by E. Zolla et al. [20], that ε , μ , ε and ζ are z-anisotronic tensors terminology used in the book by F. Zolla et al. [20], that ε , μ , ξ and ζ are *z*-anisotropic tensors, i.e. such that \overline{a} $0⁻¹$ \overline{r} $0⁻¹$

$$
\varepsilon = \begin{bmatrix} \varepsilon_{11} & \varepsilon_{12} & 0 \\ \varepsilon_{21} & \varepsilon_{22} & 0 \\ 0 & 0 & \varepsilon_{33} \end{bmatrix}, \quad \mu = \begin{bmatrix} \mu_{11} & \mu_{12} & 0 \\ \mu_{21} & \mu_{22} & 0 \\ 0 & 0 & \mu_{33} \end{bmatrix},
$$

$$
\xi = \begin{bmatrix} \xi_{11} & \xi_{12} & 0 \\ \xi_{21} & \xi_{22} & 0 \\ 0 & 0 & \xi_{33} \end{bmatrix}, \quad \zeta = \begin{bmatrix} \zeta_{11} & \zeta_{12} & 0 \\ \zeta_{21} & \zeta_{22} & 0 \\ 0 & 0 & \zeta_{33} \end{bmatrix}.
$$
(4)

We then plug Eq. (4) into Eq. (3), and have

$$
\partial_{x_2} E_3 = \omega \zeta_{11} E_1 + \omega \zeta_{12} E_2 + i \omega \mu_{11} H_1 + i \omega \mu_{12} H_2 , \qquad (5)
$$

$$
-\partial_{x_1} E_3 = \omega \zeta_{21} E_1 + \omega \zeta_{22} E_2 + i \omega \mu_{21} H_1 + i \omega \mu_{22} H_2 , \qquad (6)
$$

$$
\partial_{x_1} E_2 - \partial_{x_2} E_1 = \omega \zeta_{33} E_3 + i \omega \mu_{33} H_3 , \qquad (7)
$$

$$
\partial_{x_2} H_3 = -i\omega \varepsilon_{11} E_1 - i\omega \varepsilon_{12} E_2 + \omega \xi_{11} H_1 + \omega \xi_{12} H_2, \qquad (8)
$$

$$
-\partial_{x_1}H_3 = -i\omega\varepsilon_{21}E_1 - i\omega\varepsilon_{22}E_2 + \omega\xi_{21}H_1 + \omega\xi_{22}H_2, \qquad (9)
$$

$$
\partial_{x_1} H_2 - \partial_{x_2} H_1 = -i\omega \varepsilon_{33} E_3 + \omega \xi_{33} H_3. \tag{10}
$$

We rewrite Eqs. (5) and (6) in a matrix form

$$
\begin{bmatrix}\n\frac{\partial_{x_1} E_3}{\partial_{x_2} E_3}\n\end{bmatrix} = \omega \begin{bmatrix}\n-\zeta_{22} & \zeta_{21} \\
\zeta_{12} & -\zeta_{11}\n\end{bmatrix} \begin{bmatrix}\nE_2 \\
-E_1\n\end{bmatrix} + i\omega \begin{bmatrix}\n-\mu_{22} & \mu_{21} \\
\mu_{12} & -\mu_{11}\n\end{bmatrix} \begin{bmatrix}\nH_2 \\
-H_1\n\end{bmatrix}.
$$
\n(11)

Similarly, Eqs. (8) and (9) become

$$
\begin{bmatrix}\n\frac{\partial_{x_1} H_3}{\partial_{x_2} H_3}\n\end{bmatrix} = i\omega \begin{bmatrix}\n\varepsilon_{22} & -\varepsilon_{21} \\
-\varepsilon_{12} & \varepsilon_{11}\n\end{bmatrix} \begin{bmatrix}\nE_2 \\
-E_1\n\end{bmatrix} + \omega \begin{bmatrix}\n-\xi_{22} & \xi_{21} \\
\xi_{12} & -\xi_{11}\n\end{bmatrix} \begin{bmatrix}\nH_2 \\
-H_1\n\end{bmatrix}.
$$
\n(12)

If we define

$$
\varepsilon_T = \begin{bmatrix} \varepsilon_{22} & -\varepsilon_{21} \\ -\varepsilon_{12} & \varepsilon_{11} \end{bmatrix}, \quad \mu_T = \begin{bmatrix} -\mu_{22} & \mu_{21} \\ \mu_{12} & -\mu_{11} \end{bmatrix},
$$

$$
\xi_T = \begin{bmatrix} -\xi_{22} & \xi_{21} \\ \xi_{12} & -\xi_{11} \end{bmatrix}, \quad \zeta_T = \begin{bmatrix} -\zeta_{22} & \zeta_{21} \\ \zeta_{12} & -\zeta_{11} \end{bmatrix},
$$
(13)

and

$$
\underline{A} = \begin{bmatrix} \partial_{x_1} \\ \partial_{x_2} \end{bmatrix}, \quad \underline{E} = \begin{bmatrix} E_2 \\ -E_1 \end{bmatrix}, \quad \underline{H} = \begin{bmatrix} H_2 \\ -H_1 \end{bmatrix}, \tag{14}
$$

then Eqs. (11) and (12) turn out to be

$$
\underline{A}E_3 = \omega \zeta_T \underline{E} + i\omega \mu_T \underline{H}\,,\tag{15}
$$

$$
\underline{A}H_3 = i\omega\varepsilon_T \underline{E} + \omega\xi_T \underline{H} \,. \tag{16}
$$

Furthermore, from Eq. (15), we have

$$
\underline{E} = \omega^{-1} \zeta_T^{-1} \underline{A} E_3 - i \zeta_T^{-1} \mu_T \underline{H} \,. \tag{17}
$$

Substituting it into Eq. (16), we obtain

$$
\underline{A}H_3 = i\omega\varepsilon_T \left(\omega^{-1}\zeta_T^{-1}\underline{A}E_3 - i\zeta_T^{-1}\mu_T\underline{H}\right) + \omega\xi_T\underline{H}
$$

= $i\varepsilon_T \zeta_T^{-1}\underline{A}E_3 + \omega\varepsilon_T \zeta_T^{-1}\mu_T\underline{H} + \omega\xi_T\underline{H}$. (18)

This can be recast as

$$
\underline{H} = B^{-1} \left(\underline{A} H_3 - i \varepsilon_T \zeta_T^{-1} \underline{A} E_3 \right) , \qquad (19)
$$

with

$$
B = \omega \xi_T + \omega \varepsilon_T \zeta_T^{-1} \mu_T \,. \tag{20}
$$

Let us now plug Eq. (19) into Eq. (10)

$$
\begin{aligned}\n\left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \underline{H} &= \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left(B^{-1} \underline{A} H_3 - i B^{-1} \varepsilon_T \zeta_T^{-1} \underline{A} E_3\right) \\
&= \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[\begin{array}{cc}\n-\frac{i}{\omega} \left(\mu_T + \zeta_T \varepsilon_T^{-1} \xi_T\right)^{-1}\right] \underline{A} E_3 \\
&+ \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[\begin{array}{cc}\n\frac{1}{\omega} \left(\xi_T + \varepsilon_T \zeta_T^{-1} \mu_T\right)^{-1}\right] \underline{A} H_3 \\
&= -i \omega \varepsilon_{33} E_3 + \omega \xi_{33} H_3\n\end{array}\right]\n\end{aligned}\n\tag{21}
$$

In the same way, from Eq. (16), we have

$$
\underline{H} = \omega^{-1} \xi_T^{-1} \underline{A} H_3 - i \xi_T^{-1} \varepsilon_T \underline{E} \,, \tag{22}
$$

which is then applied to Eq. (15)

$$
\underline{A}E_3 = \omega \zeta_T \underline{E} + i\omega \mu_T \left(\omega^{-1} \xi_T^{-1} \underline{A} H_3 - i \xi_T^{-1} \varepsilon_T \underline{E} \right)
$$

= $\omega \zeta_T \underline{E} + i\mu_T \xi_T^{-1} \underline{A} H_3 + \omega \mu_T \xi_T^{-1} \varepsilon_T \underline{E}$, (23)

from which we have

$$
\underline{E} = C^{-1} \underline{A} E_3 - i C^{-1} \mu_T \xi_T^{-1} \underline{A} H_3, \qquad (24)
$$

with

$$
C = \omega \zeta_T + \omega \mu_T \xi_T^{-1} \varepsilon_T \,. \tag{25}
$$

We plug Eq. (24) into Eq. (7)

$$
\begin{aligned}\n\left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] & \underline{E} = \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left(\frac{C^{-1} \underline{A} E_3 - i C^{-1} \mu_T \xi_T^{-1} \underline{A} H_3\right) \\
&= \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[\frac{1}{\omega} \left(\zeta_T + \mu_T \xi_T^{-1} \epsilon_T\right)^{-1}\right] \underline{A} E_3 \\
&- \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[\frac{i}{\omega} \left(\epsilon_T + \xi_T \mu_T^{-1} \zeta_T\right)^{-1}\right] \underline{A} H_3 \\
&= \omega \zeta_{33} E_3 + i \omega \mu_{33} H_3.\n\end{aligned} \tag{26}
$$

Finally, we rewrite Eqs. (21) and (26) by multiplying by ω on both sides of the equations, and we obtain

$$
\begin{aligned}\n\left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[-i\left(\mu_T + \zeta_T \ \varepsilon_T^{-1} \ \xi_T\right)^{-1}\right] \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right]^{\mathrm{T}} E_3 \\
+ \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[\left(\xi_T + \varepsilon_T \ \zeta_T^{-1} \ \mu_T\right)^{-1}\right] \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right]^{\mathrm{T}} H_3 = -i\omega^2 \varepsilon_{33} E_3 + \omega^2 \xi_{33} H_3, \\
\left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[\left(\zeta_T + \mu_T \ \xi_T^{-1} \ \varepsilon_T\right)^{-1}\right] \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right]^{\mathrm{T}} E_3 \\
+ \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right] \left[-i\left(\varepsilon_T + \xi_T \ \mu_T^{-1} \ \zeta_T\right)^{-1}\right] \left[\begin{array}{cc}\n\partial_{x_1} & \partial_{x_2}\n\end{array}\right]^{\mathrm{T}} H_3 = \omega^2 \zeta_{33} E_3 + i\omega^2 \mu_{33} H_3.\n\end{aligned} \tag{27}
$$

which are two coupled equations with the longitudinal fields E_3 and H_3 as unknowns.

3. Implementation of the PDE system in Comsol Multiphysics

The standard form of PDEs in COMSOL Multiphysics is described by the following equation reminiscent of a governing equation in elasticity theory

$$
\nabla \cdot (-c : \nabla u - \alpha u + \gamma) + au + \beta \cdot \nabla u = f. \tag{28}
$$

where *c* is a rank-4 tensor, α and β are vectors, γ is a rank 2-tensor, *a* is a scalar and *f* is a forcing term.

Comparing Eqs. (27) and (28), we can see that there is a neat isomorphism, i.e. we establish the correspondences between the Maxwell-Tellegen's equations for electromagnetic waves and the Navier equations for solid mechanical waves, which generalises the correspondence between in-plane elasticity and a reduced form of the Maxwell's equations in the paper of G. W. Milton and A. B. Movchan [21]. These correspondences can be straightforwardly implemented with a finite element package, as well as a finite difference time domain (FDTD) package written for linear elastodynamics, e.g. the freeware SIMSONIC [22], in order to solve electromagnetic wave problems in complex bianisotropic media. Although we use here the FEM based software - COMSOL to deal with the analysis of bianisotropic media in the following sections, we would like to point out that the correspondences should work equally well in time and frequency domains.

We rewrite the formulae in Eq. (27) as

$$
\begin{aligned}\n& [\partial_{x_1} \quad \partial_{x_2}] \left[\left(\mu_T + \zeta_T \ \varepsilon_T^{-1} \ \xi_T \right)^{-1} \right] \left[\ \partial_{x_1} \ \partial_{x_2} \right]^{T} E_3 \\
&+ [\partial_{x_1} \ \partial_{x_2}] \left[i \left(\xi_T + \varepsilon_T \ \zeta_T^{-1} \ \mu_T \right)^{-1} \right] \left[\ \partial_{x_1} \ \partial_{x_2} \right]^{T} H_3 = \omega^2 \varepsilon_{33} E_3 + i \omega^2 \xi_{33} H_3, \\
& [\partial_{x_1} \ \partial_{x_2}] \left[i \left(\zeta_T + \mu_T \ \xi_T^{-1} \ \varepsilon_T \right)^{-1} \right] \left[\ \partial_{x_1} \ \partial_{x_2} \right]^{T} E_3 \\
&+ [\partial_{x_1} \ \partial_{x_2}] \left[\left(\varepsilon_T + \xi_T \ \mu_T^{-1} \ \zeta_T \right)^{-1} \right] \left[\ \partial_{x_1} \ \partial_{x_2} \right]^{T} H_3 = i \omega^2 \zeta_{33} E_3 - \omega^2 \mu_{33} H_3.\n\end{aligned} \tag{29}
$$

Comparing with Eq. (28), if we consider the coupled longitudinal electric and magnetic fields, the unknown u in Eq. (28) is

$$
u = \left[\begin{array}{c} E_3 \\ H_3 \end{array} \right],\tag{30}
$$

and the coefficients are

$$
c = \begin{bmatrix} c_{11} & c_{12} \\ c_{21} & c_{22} \end{bmatrix}, \quad a = \begin{bmatrix} \omega^2 \varepsilon_{33} & i \omega^2 \zeta_{33} \\ i \omega^2 \zeta_{33} & -\omega^2 \mu_{33} \end{bmatrix},
$$
(31)

with entries

$$
c_{11} = (\mu_T + \zeta_T \ \varepsilon_T^{-1} \ \xi_T)^{-1}, \quad c_{12} = i \left(\xi_T + \varepsilon_T \ \zeta_T^{-1} \ \mu_T \right)^{-1},
$$

\n
$$
c_{21} = i \left(\zeta_T + \mu_T \ \xi_T^{-1} \ \varepsilon_T \right)^{-1}, \quad c_{22} = \left(\varepsilon_T + \xi_T \ \mu_T^{-1} \ \zeta_T \right)^{-1}.
$$

\n(32)

Simply stated, when the tensors of ε , μ , ξ , ζ in Eq. (13) are represented by diagonal matrices in a Cartesian coordinate system

$$
v = diag(v_{11}, v_{22}, v_{33}), \quad v = \varepsilon, \mu, \xi, \zeta.
$$
 (33)

then the entries in Eq. (32) become

$$
c_{11} = \begin{bmatrix} \frac{-\varepsilon_{22}}{\mu_{22}\varepsilon_{22} - \xi_{22}\zeta_{22}} & 0 & 0\\ 0 & \frac{-\varepsilon_{11}}{\mu_{11}\varepsilon_{11} - \xi_{11}\zeta_{11}} \end{bmatrix}, c_{12} = \begin{bmatrix} \frac{i\zeta_{22}}{\mu_{22}\varepsilon_{22} - \xi_{22}\zeta_{22}} & 0 & i\zeta_{11}\\ 0 & \frac{i\zeta_{11}}{\mu_{11}\varepsilon_{11} - \xi_{11}\zeta_{11}} \end{bmatrix}
$$

$$
c_{21} = \begin{bmatrix} \frac{i\xi_{22}}{\mu_{22}\varepsilon_{22} - \xi_{22}\zeta_{22}} & 0 & 0\\ 0 & \frac{i\xi_{11}}{\mu_{11}\varepsilon_{11} - \xi_{11}\zeta_{11}} \end{bmatrix}, c_{22} = \begin{bmatrix} \frac{\mu_{22}}{\mu_{22}\varepsilon_{22} - \xi_{22}\zeta_{22}} & 0 & 0\\ 0 & \frac{\mu_{11}}{\mu_{11}\varepsilon_{11} - \xi_{11}\zeta_{11}} \end{bmatrix}
$$
(34)

Note that each component of the coefficients c_{ij} ($i, j = 1, 2$) is a fraction with a denominator $(\varepsilon_{kk} \mu_{kk} - \xi_{kk} \zeta_{kk})$ ($k = 1, 2$). In order to avoid any infinite entries of *c* while solving the PDEs, it should satisfy

$$
\varepsilon_{kk}\mu_{kk} - \xi_{kk}\zeta_{kk} \neq 0, \ (k = 1, 2).
$$
 (35)

,

Remark: This condition is essential and generally is met since the magnetoelectric coupling parameters for bianisotropic media are usually small.

4. Problem setup with open boundary conditions

In the last section, we have discussed the general PDEs in bianisotropic media, which allows us to introduce the PDEs model in COMSOL Multiphysics; however, in order to mimic the scattering and propagation properties of bianisotropic media in an open space, we need to consider the boundary conditions. We usually introduce a transformation of an infinite domain into a finite one, which should enforce that the wavelength is contracted to infinitely small values as it approaches the outer boundary of the transformed domain within a perfectly matched layer (PML). At this outer boundary, Dirichlet data could be set, thereby enforcing a vanishing field [23, 24].

The PML is shown to be equivalent to an analytic continuation of Maxwell's equations to complex variables's spatial domain [25]. The form of the electromagnetic field inside a PML region can be obtained from the solutions in real space from a mapping to a complex space through this analytic continuation. Physically speaking, an equivalent artificial medium is achieved from this map in the PML region that is described by complex, anisotropic and inhomogeneous parameters, even if the original ones are real, isotropic and homogeneous. The analytic continuation means to alter the eigenfunctions of Maxwell's equations in such a way that the propagating modes are mapped continuously to exponentially decaying modes, which allows for reflectionless absorption of electromagnetic waves. In other words, an absorbing medium dissipating the outgoing wave can be achieved through proper complex map: an artificial medium situated in a region can be added to a finite medium in order to mimic an open space within which the field is damped to a negligible value [26]. Importantly, the impedance of the equivalent medium is the same as that of the initial medium since all the parameters undergo the same transform, what ensures non-reflecting features on the interface between the medium and the PMLs.

F. L. Teixeira et al. [24, 25] derive the Maxwellian PML's for arbitrary bianisotropic and dispersive media. In the Cartesian coordinate system, if an analytic continuation is defined by the transformation function

$$
\tilde{u} = \int_0^u s_u(u') du' \,. \tag{36}
$$

with $s_u(u')$ the complex stretching variables [27] and *u* stands for *x*, *y*, *z*. In the complex space, the nabla operator in Maxwell's equations changes as

$$
\nabla \to \tilde{\nabla} = \hat{x}\frac{\partial}{\partial \tilde{x}} + \hat{y}\frac{\partial}{\partial \tilde{y}} + \hat{z}\frac{\partial}{\partial \tilde{z}} = \hat{x}\frac{1}{s_x}\frac{\partial}{\partial x} + \hat{y}\frac{1}{s_y}\frac{\partial}{\partial y} + \hat{z}\frac{1}{s_z}\frac{\partial}{\partial z},\tag{37}
$$

it reads as

$$
\tilde{\nabla} = \bar{\vec{S}} \cdot \nabla , \qquad (38)
$$

with

$$
\bar{S} = \hat{x}\hat{x}\left(\frac{1}{s_x}\right) + \hat{y}\hat{y}\left(\frac{1}{s_y}\right) + \hat{z}\hat{z}\left(\frac{1}{s_z}\right). \tag{39}
$$

since $s_u(u)$ and $\partial/\partial u'$ commute for $u \neq u'$, \overline{S} is a diagonal tensor, and det $\overline{S} = (s_x s_y s_z)^{-1}$.

Based on this, we then can expand the Maxwell's equations in the new complex space, and derive that the PMLs with bianisotropic constitutive parameters are

$$
v_{\text{PML}} = \left(\det \bar{\bar{S}}\right)^{-1} \left[\bar{\bar{S}} \cdot v \cdot \bar{\bar{S}}\right], \quad v = \varepsilon, \mu, \xi, \zeta. \tag{40}
$$

Take a full wave simulation for two dimensional structures in COMSOL Multiphysics as an illustration, the computational domain is terminated by PMLs in Cartesian coordinates. The PMLs are designed to decrease the distribution of the waves along the *x*, *y* and *xy* directions, respectively. Specifically, assuming the parameter of the matrix $v = \text{diag}(v_{xx}, v_{yy}, v_{zz})$ ($v =$ ε , μ , ξ , ζ), then we choose \overline{S}

$$
PML_x: s_x = a - b \times i, s_y = 1, s_z = 1,\nPML_y: s_x = 1, s_y = a - b \times i, s_z = 1,\nPML_{xy}: s_x = a - b \times i, s_y = a - b \times i, s_z = 1.
$$
\n(41)

with *a*, *b* $\in \mathbb{Z}^+$. Then the transformed parameters in the PMLs are

$$
v' = \text{diag}\left(v_{xx}L_{xx}, v_{yy}L_{yy}, v_{zz}L_{zz}\right), \quad v = \varepsilon, \mu, \xi, \zeta. \tag{42}
$$

where

$$
L_{xx} = \frac{s_y s_z}{s_x}, \quad L_{yy} = \frac{s_x s_z}{s_y}, \quad L_{zz} = \frac{s_x s_y}{s_z} \,. \tag{43}
$$

can be derived by introducing the definition of s_x , s_y and s_z in Eq. (41) along different directions. Similar concept of PMLs can be extended to the anisotropic/bianisotropic matrix and other coordinate system [25, 28].

5. FEM implementation

To check both our PDEs model and PMLs for bianisotropic media, some conceptual devices designed from transformation optics are numerically studied in the following subsections, these structures are all achieved by bianisotropic media.

5.1. Invisibility cloak

Electromagnetic (EM) metamaterials such as invisibility cloaks can be designed through the blowup of a point [4, 5], using transformation optics which is a versatile mathematical tool enabling a deeper analytical insight into the scattering properties of EM fields in metamaterials. In this subsection, a cloak consisting of a bianisotropic medium is investigated, and the geometric transformation proposed by J. Pendry [4] is proved to be applicable to design such kind of cloak.

We start with a free space with permittivity $\varepsilon = \varepsilon_0$, permeability $\mu = \mu_0$ and magnetoelectric coupling parameters $\xi = \xi_0$, $\zeta = \zeta_0$ in the Cartesian coordinate system, and first map it onto polar coordinates (r, θ, z) as defined by

$$
x = r \cos \theta, \ y = r \sin \theta, \ z = z.
$$
 (44)

Then we introduce a geometric transformation defined by Eq. (45) which maps the field within a disk of radius $r = R_2$ onto an annulus $R_1 < r' \le R_2$, i.e. the original point $r = 0$ in the

original disk is blowup to a disk with $r' = R_1$ in coordinates (r', θ', z') . Based on this coordinate
transformation [4, 9], the rays are distorted and quided around the cloak transformation [4, 9], the rays are distorted and guided around the cloak.

The mapping function is defined as

$$
\begin{cases}\nr' = R_1 + r(R_2 - R_1)/R_2, & 0 \le r \le R_2 \\
\theta' = \theta, & 0 < \theta \le 2\pi \\
z' = z.\n\end{cases} \tag{45}
$$

which provides the Jacobian matrix

$$
\mathbf{J}_{rr'} = \frac{\partial(r, \theta, z)}{\partial(r', \theta', z')} = \text{diag}\left(\frac{R_2}{R_2 - R_1}, 1, 1\right), \quad 0 \le r \le R_2.
$$
 (46)

Finally, we go back to the Cartesian coordinates (x', y', z') , which are radially contracted, d the compound Jacobian matrix is and the compound Jacobian matrix is

$$
\mathbf{J}_{xx'} = \mathbf{J}_{xr}\mathbf{J}_{rr'}\mathbf{J}_{r'x'}\,. \tag{47}
$$

More specifically, it is

$$
\mathbf{J}_{xx'} = \mathbf{J}_{xr}\mathbf{J}_{rr'}\mathbf{J}_{r'x'} = \mathbf{R}(\theta) \operatorname{diag}\left(\frac{R_2}{R_2 - R_1}, \frac{r}{r'}, 1\right) \mathbf{R}(-\theta'). \tag{48}
$$

where $\mathbf{R}(\theta)$ is the rotation matrix in the *xy*-plane about an angle θ .

The transformation matrix [17] is correspondingly

$$
\mathbf{T}_{xx'}^{-1} = \left[\mathbf{J}_{xx'}^{\mathrm{T}} \mathbf{J}_{xx'} / \det(\mathbf{J}_{xx'})\right]^{-1} \,. \tag{49}
$$

We denote

$$
\mathbf{T}_{xx'}^{-1} = \begin{bmatrix} T_{11} & T_{12} & 0 \\ T_{21} & T_{22} & 0 \\ 0 & 0 & T_{33} \end{bmatrix},
$$
 (50)

then the components are

$$
T_{11} = 1 - \frac{R_1}{r'} \cos^2 \theta' + \frac{R_1}{r' - R_1} \sin^2 \theta', \qquad T_{12} = T_{21} = \frac{R_1 (R_1 - 2r')}{r'(r' - R_1)} \sin \theta' \cos \theta',
$$

\n
$$
T_{22} = 1 - \frac{R_1}{r'} \sin^2 \theta' + \frac{R_1}{r' - R_1} \cos^2 \theta', \qquad T_{33} = \frac{R_2^2}{(R_2 - R_1)^2} \frac{r' - R_1}{r'}.
$$
\n(51)

Hence, the parameters in the annulus $R_1 < r' \le R_2$ are

$$
v = v_0 \mathbf{T}_{xx}^{-1}, \quad v = \varepsilon, \, \mu, \, \xi, \, \zeta \,. \tag{52}
$$

while the parameters of the outer region $r' > R_2$ are unchanged, and that of the disk $r' \leq R_1$ can
have any value without affecting the electromagnetic scattering have any value without affecting the electromagnetic scattering.

Figure 1(a) shows our designed bianisotropic cloak, the cloaking region is the innermost circle of radius $R_1 = 20$ mm wherein a L-shaped obstacle locates, the gray annulus represents the distorted space with radius $R_1 < r' \leq R_2$, where $R_2 = 40$ mm. Firstly, the EM distribution of a point source (s-polarized with the electric field along z) with frequency $f = 8.7$ GHz of a point source (*s*-polarized with the electric field along *z*) with frequency $f = 8.7$ GHz radiating in a matrix with parameters $v = v_0$ **I**₃, $(v = \varepsilon, \mu, \xi, \zeta)$ is shown in Fig. 1(b). $v_0 = \varepsilon_0$, μ_0 are the permittivity, permeability of the vacuum, respectively; while $|\xi_0 \zeta_0| \neq 1/c_0$ with c_0 the velocity of light in vacuum should be promised to ensure convergence of the numerical algorithm as indicated in Eq. (35). If there is a L-shaped obstacle with $\varepsilon = (1 + 5 \times i) \varepsilon_0 I_3$, $\mu = \mu_0 I_3$, $\xi = \zeta = 0.99/c_0I_3$ in the matrix, it leads to an interacting with the source, the plot of Re(E_z) is indicated in panel (c), where the white regions are for values outside the color scale. However, when the L-shaped obstacle is surrounded by the designed cloak, then the obstacle seems to be invisible for the outer observer, as shown in panel (d).

Fig. 1. (a) Schematic diagram of bianisotropic cloak with a L-shaped obstacle inside the cloaking region, and a point source (*s*-polarized with the electric field along *z* with $f = 8.7$ GHz) locates outside the shell. Plots of $\text{Re}(E_z)$ with the same arbitrary units for the three following panels: (b) A reference point source in a pure matrix with parameters $\varepsilon = \varepsilon_0 I_3$, $\mu = \mu_0 I_3$ and $\xi = \zeta = 0.99/c_0 I_3$; (c) Same point source in the matrix with a presence of L-shaped obstacle, the parameters of obstacle are $\varepsilon = (1 + 5 \times i) \varepsilon_0 \mathbf{I}_3$, $\mu = \mu_0 \mathbf{I}_3$ and $\xi = \zeta = 0.99/c_0$ **I**₃; (d) Same point source radiates in the presence of an L-shaped obstacle surrounded by an invisibility cloak.

5.2. Concentrator

A cylindrical concentrator is designed to focus the incident waves with wave vectors perpendicular to the cylinder axis, enhancing the electromagnetic energy density in a given area [6].

We start with a map from Cartesian coordinates to cylindrical ones, for a cylindrical lens with a radius $r = R_3$, a radial transformation is introduced to distort the space as

$$
\begin{cases}\nr' = \begin{cases}\n\frac{R_1}{R_2}r, & 0 \le r \le R_2 \\
\frac{R_3 - R_1}{R_3 - R_2}r - \frac{R_2 - R_1}{R_3 - R_2}R_3, & R_2 < r \le R_3 \\
\theta' = \theta, & 0 \le \theta < 2\pi\n\end{cases}\n\end{cases} \tag{53}
$$
\n
$$
\theta' = \theta, \quad 0 \le \theta < 2\pi
$$

A space is compressed into a cylindrical region with radius R_1 (the innermost circle) with the expense of an expansion of space between R_1 and R_3 , where an intermediate circle is located at R_2 . Similarly, we can derive the associated Jacobian matrix $J_{rr'}$, then the compound Jacobian matrix in Eq. (47) and the transformation matrix in Eq. (49). We have

$$
\mathbf{T}_{xx'}^{-1} = \begin{cases} \mathbf{R}(\theta') \operatorname{diag}\left(1, 1, \frac{R_2^2}{R_1^2}\right) \mathbf{R}(-\theta'), & 0 \le r \le R_2\\ \mathbf{R}(\theta') \operatorname{diag}\left(\frac{R_3 - R_1}{R_3 - R_2} \frac{r}{r'}, \frac{R_3 - R_2}{R_3 - R_1} \frac{r}{r'}, \frac{R_3 - R_2}{R_3 - R_1} \frac{r}{r'}\right) \mathbf{R}(-\theta'), & R_2 < r \le R_3 \end{cases} \tag{54}
$$

Let us expand this formula in the distorted cylindrical coordinates (r', θ', z') , and denote $\mathbf{T}_{xx'}^{-1}$ as in Eq. (50), finally we have

$$
0 \le r' \le R_1: \t T_{11} = T_{22} = 1, \t T_{12} = T_{21} = 0, \t T_{33} = \frac{R_2^2}{R_1^2}.
$$

\n
$$
R_1 < r' \le R_3: \t T_{11} = A \cos^2 \theta' + B \sin^2 \theta',
$$

\n
$$
T_{12} = T_{21} = (A - B) \sin \theta' \cos \theta',
$$

\n
$$
T_{22} = A \sin^2 \theta' + B \cos^2 \theta', \t T_{33} = C.
$$
\n
$$
(55)
$$

with

$$
A = 1 + \frac{R_2 - R_1}{R_3 - R_2} \frac{R_3}{r'}, \quad B = \frac{r'}{r' + \frac{R_2 - R_1}{R_3 - R_2} R_3}, \quad C = \left(\frac{R_3 - R_2}{R_3 - R_1}\right)^2 \left(1 + \frac{R_2 - R_1}{R_3 - R_2} \frac{R_3}{r'}\right).
$$
\n(56)

Apply matrix T_{xx}^{-1} , to Eq. (52), we can obtain the formulae of these parameters in the annulus as well as the innermost circle, while the outside space of the concentrator is bianisotropic matrix with $v = v_0$ **I**₃.

Fig. 2. (a) Schematic diagram of a bianisotropic concentrator; (b) Plot of $Re(E_z)$ under an *s*-polarized incidence (amplitude normalized to unity) with frequency $f = 8.7$ GHz radiating from the above, the parameters in the region $r' \le R_1$, and the annulus $R_1 < r' \le R_3$ are given by Eq. (52) with transformation matrix T^{-1} in Eq. (55) given by Eq. (52) with transformation matrix \mathbf{T}_{xx}^{-1} , in Eq. (55).

Figure 2(a) is the schematic diagram of our concentrator, the radii of the inner and outer circles are $R_1 = 20$ mm, $R_3 = 40$ mm, while the intermediate circle has a radius $R_2 = 30$ mm. The

parameters of each region are defined in Eq. (52) along with transfer matrix T_{xx}^{-1} , in Eq. (55), here the magnetoelectric coupling parameters $\xi_0 = \zeta_0 = 0.99/c_0$ are chosen as the same values as those of the invisibility cloak. An *s*-polarized (the electric field is perpendicular to the *xy* plane) plane wave with frequency 8.7 GHz is incident from above, the plot of $\text{Re}(E_z)$ is depicted in panel (b): the waves are compressed in the inner circle as expected.

5.3. Rotator

We would like to design a rotator which allows the rotation of the incident wave, the distribution of the field inside a disk $r \leq R_1$ appears as it comes from a different angle comparing with that of outside $r > R_2$, for example, a $\pi/2$ angle rotation is achieved here in the inner disk.

Be different from the geometric transformation introduced in the concentrator, an angular transformation is introduced in this case instead of the radial transformation. The mapping is performed in the region $r \leq R_2$: the domain in $r \leq R_1$ is transformed by rotating a fixed angle from the region $r' = a$ in virtual space, while the rotation angle is continually changed from the fixed angle to zero in region $R_1 < r \leq R_2$. In other words, the transformation algorithm can be expressed as

$$
r \le R_1: \qquad r' = r, \quad \theta' = \theta + \theta_0, \quad z' = z.
$$

\n
$$
R_1 < r \le R_2: \quad r' = r, \quad \theta' = \theta + \theta_0 \frac{f(R_2) - f(r)}{f(R_2) - f(R_1)}, \quad z' = z.
$$
\n
$$
r > R_2: \qquad r' = r, \quad \theta' = \theta, \quad z' = z.
$$
\n
$$
(57)
$$

where θ_0 is the rotation angle for the inner disk, and it is reduced to zero as the radius approaches to $r = R_2$. $f(r)$ is an arbitrary continuous function of *r*.

Based on this transformation, we can derive the associated Jacobian matrix $J_{rr'}$, furthermore, the compound Jacobian matrix in Eq. (47) as well as the transformation matrix in Eq. (49) can be obtained. In annulus $R_1 < r' (= r) \le R_2$, the notations in Eq. (50) become

$$
R_1 < r' \le R_2: \quad T_{11} = 1 + r'^2 D^2 \sin^2 \theta' + 2r' D \sin \theta' \cos \theta',
$$
\n
$$
T_{12} = T_{21} = r' D (\sin^2 \theta' - \cos^2 \theta') - r'^2 D^2 \sin \theta' \cos \theta',
$$
\n
$$
T_{22} = 1 + r'^2 D^2 \cos^2 \theta' - 2r' D \sin \theta' \cos \theta', \quad T_{33} = 1.
$$
\n
$$
r' \le R_1 \text{ and } r' > R_2: \quad \mathbf{T}_{xx'}^{-1} = \mathbf{I}_3.
$$
\n
$$
(58)
$$

with $D = \theta_0 f'(r) / [f(R_2) - f(R_1)]$. Again, substituting the matrix T_{xx}^{-1} , into Eq. (52), the expressions for those parameters can be derived in different regions. expressions for those parameters can be derived in different regions.

The FEM analysis for the rotator is shown in Fig. 3. The radii of the inner and outer circles are $R_1 = 20$ mm, $R_2 = 40$ mm, respectively. The parameters of each region are defined in Eq. (52) along with transfer matrix T_{xx}^{-1} , in Eq. (58), the magnetoelectric coupling parameters $\xi_0 = \zeta_0 = 0.99/c_0$ are chosen as the same values as those of the invisibility cloak. An *s*-polarized incident wave with frequency $f = 8.7$ GHz is assumed to radiate from above. A rotation of $\theta_0 = \pi/2$ of the fields in the inner circle can be observed in panel (b) comparing with the incidence. Note that, θ_0 can be an arbitrary values in $[-\pi, \pi]$.

6. A multilayered bianisotropic cloak through homogenization

Let us now consider a layered medium [29] periodic along x_1 and invariant along x_2 and x_3 . A periodic cell consists of a heterogeneous bi-isotropic medium described by piecewise constant scalar valued functions $\varepsilon(x_1)$, $\mu(x_1)$, $\xi(x_1)$, $\zeta(x_1)$. Using two-scale expansion techniques applied

Fig. 3. (a) Schematic diagram of the bianisotropic rotator; (b) Plot of $\text{Re}(E_z)$ under an *s*polarized incidence (amplitude normalized to unity) with frequency $f = 8.7$ GHz radiating from the above, the parameters in the region $r' \le R_1$, annulus $R_1 < r' \le R_2$ are given by Eq. (52) with transformation matrix T^{-1} in Eq. (58) Eq. (52) with transformation matrix \mathbf{T}_{xx}^{-1} , in Eq. (58).

to the Maxwell-Tellegen's equations like in [30] one gets the homogenized equations

$$
\nabla \times E_{\text{eff}} = \omega \zeta_{\text{eff}} E_{\text{eff}} + i\omega \mu_{\text{eff}} H_{\text{eff}} ,
$$

\n
$$
\nabla \times H_{\text{eff}} = -i\omega \varepsilon_{\text{eff}} E_{\text{eff}} + \omega \xi_{\text{eff}} H_{\text{eff}} .
$$
\n(59)

where ε_{eff} , μ_{eff} , ξ_{eff} and ζ_{eff} are anisotropic homogenized tensors such that

$$
v_{\text{eff}} = \text{diag}\left(\langle v^{-1} \rangle^{-1}, \langle v \rangle, \langle v \rangle \right), \quad v = \varepsilon, \mu, \xi, \zeta. \tag{60}
$$

where $\langle f(x_1) \rangle = \int_0^d f(x_1) dx_1$, with *d* the periodic cell size. Note also that these formulae can be deduced from the high-order homogenization approach in [15], by keeping only leading can be deduced from the high-order homogenization approach in [15], by keeping only leading order terms.

layer	1 (inner)	$\mathcal{D}_{\mathcal{L}}$	3	4	5	6
thickness [mm]		0.375	0.75	0.75	0.75	0.75
$\varepsilon_r^{-1} = \mu_r^{-1} = \xi_r^{-1} = \zeta_r^{-1}$	2.0	1680.7	0.2503	80.7492	0.2514	29.3898
layer		8	9	10	11	12
thickness [mm]	0.75	0.875	0.625	0.75	0.75	0.75
$\varepsilon_r^{-1} = \mu_r^{-1} = \xi_r^{-1} = \zeta_r^{-1}$	0.2530	16.3674	0.2548	10.9864	0.2568	8.1798
layer	13	14	15	16	17	18
thickness [mm]	0.75	0.75	0.75	0.75	0.75	0.75
$\varepsilon_r^{-1} = \mu_r^{-1} = \xi_r^{-1} = \zeta_r^{-1}$	0.2589	6.5	0.2611	5.3990	0.2632	4.6292
layer	19	20	21	22		
thickness [mm]	0.75	0.75	0.75	0.375		
$\varepsilon_r^{-1} = \mu_r^{-1} = \xi_r^{-1} = \zeta_r^{-1}$	0.2653	4.0645	0.2674	1.0		

Table 1. Parameters of the layered cloak. ($\varepsilon = \varepsilon_r \varepsilon_0$, $\mu = \mu_r \mu_0$, $\xi = \xi_r \xi_0$, $\zeta = \zeta_r \zeta_0$.)

One can use these formulae to design a concentric multilayered bianisotropic cloak of inner radius $R_1 = 14$ mm and outer radius $R_2 = 30$ mm, consisting of 22 homogeneous layers (see Table 1 for the values of the inverse relative permittivity, permeability, magnetoelectric coupling parameters) in a bianisotropic background with permittivity $\varepsilon = \varepsilon_0$, permeability $\mu = \mu_0$ and magnetoelectric coupling parameters $\xi = \xi_0 = 0.99/c_0$, $\zeta = \zeta_0 = 0.99/c_0$. Notice that almost all layers have an equal thickness of 0.75 mm and one layer in two has almost the same optical properties.

We show the results of numerical simulations in Fig. 4. One can see that the backscattering by the infinite conducting obstacle (of radius 14 mm) with a concentric multilayered bianisotropic cloak is much reduced in Fig. 4(b) comparing with Fig. 4(a), and the shadow zone behind the obstacle is also much reduced with the layered cloak. The interval of cloaking frequencies from ⁵.7 GHz to 7.7 GHz is fairly broadband.

Fig. 4. (a) Scattering of an infinite conducting obstacle (of radius 14 mm) by a point source at a frequency 5.7 GHz; (b) Scattering like in (a) when the obstacle is surrounded by a layered cloak; (c) Scattering like in (a) for a point source at a frequency 7.7 GHz; (d) Scattering like in (b) for a point source at a frequency 7.7 GHz. Color scale corresponds to $Re(E_z)$ (the same arbitrary units have been used in the four panels).

7. Conclusion

In this paper, we have discussed the PDEs model for the analysis of the propagation properties in bianisotropic media with an invariance along one direction: two coupled PDEs with scalar solutions are derived for this electromagnetic system. The spatially varying coefficients within these two PDEs can be scalar or matrix valued. Regarding the numerical model, specially designed perfectly matched layers (PMLs) are introduced in order to account for the unbounded domain. To do this, we consider a transformation which maps an infinite domain onto a finite one, and we further add the damping. This provides a reflection-less bianisotropic layer for all incident angles, and this can be considered as an extension of the PMLs introduced by J. Berenger nearly twenty years ago [23]. We implement these PMLs in scattering problems in bianisotropic media including an invisibility cloak, a field concentrator and a field rotator designed by introducing the proper geometric transformation. Importantly, these functional structures are proved to work well when they are achieved by bianisotropic media. We finally, show that one can achieve near-cloaking with a layered cloak with piecewise constant permittivity, permeability and magnetoelectric coupling parameters. The solutions of singular systems considered in this paper represent important benchmarks for the validation of non-trivial numerical calculations for bianisotropic media.

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