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On the Input-Output Distinguishability of Single Output Continuous Linear Time-Invariant Systems

Koffi M. D. Motchon, Komi M. Pekpe, Jean-Philippe Cassar, and Stephan De Bièvre

Abstract—This technical note addresses the distinguishability problem for continuous linear time-invariant systems using their input and output data. The proposed approach is based on the characterization of the class of initial state vectors and control input signals that make the outputs of different continuous linear time-invariant systems indistinguishable. This class is defined here as the “indistinguishability zone”. From this characterization, a simple necessary and sufficient condition for the distinguishability of single-input, single-output systems is established. Furthermore, it is shown that multiple-input, single-output systems are never distinguishable.

Index Terms—Linear time-invariant system, distinguishability, Markov parameters, Switched systems.

I. INTRODUCTION

The property of distinguishability of two systems refers to their capacity to generate different output signals for a given input signal. This property was firstly introduced in the 1970s by Grewal *et al.* [1], [2], who have shown its importance for the identifiability of dynamical systems. Distinguishability is also important when studying the observability of switched systems as underlined recently in [3], [4], [5]. It characterizes in this context the ability to determine the active mode of the system from the input and output data.

Several notions of distinguishability may be encountered in the literature, depending on the field of application. In this technical note, we call “strict distinguishability” the notion of distinguishability proposed in [4]. Two linear continuous time-invariant (LTI) systems are strictly distinguishable if for any initial state vectors and control inputs of the systems (not both zero), their corresponding outputs are different. An adaptation of this notion of distinguishability for discrete LTI dynamical systems is proposed in [6].

Studies of strict distinguishability deal with the determination of necessary and (or) sufficient conditions that allow to test whether or not two different systems are strictly distinguishable. These conditions are often “rank conditions” (see for example [4], [6], [7]) on an appropriate matrix that depends on the matrices used to model the systems in a state-space representation. However, to the best of our knowledge, there is currently no research on the determination of the class of initial state vectors and input signals that generate identical output signals for dynamical systems which are not strictly distinguishable, a problem we will attack here.

Specifically, this paper is concerned with the construction of the set of initial state vectors and control inputs that lead to indistinguishable outputs for two different continuous LTI systems. We will introduce the notion of “indistinguishability zone” to refer to this class.

The paper is organized as follows. Section II is devoted to the formulation of the problem. It also contains the preliminary results necessary to obtain the characterization of the indistinguishability zone given in Section III. This characterization is used in Section IV to establish simple conditions for strict distinguishability of single-input, single-output (SISO) and multiple-input, single-output (MISO)

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systems. Finally, some remarks and conclusions are highlighted in Section V.

For more information on the state of the art and examples illustrating the theory, the interested reader can inspect [8].

Throughout the remainder of this paper, we adopt the following notation, for all $m_1, m_2 \in \mathbb{N}^*$:

- $\mathbb{R}_+ = [0; +\infty[$: set of positive real numbers.
- \mathbb{R}^{m_1} : Euclidean m_1 -space ; 0_{m_1} : zero vector of \mathbb{R}^{m_1} .
- I_{m_1} : identity matrix of $\mathbb{R}^{m_1 \times m_1}$.
- $\mathbb{R}^{m_1 \times m_2}$: space of matrices with m_1 rows and m_2 columns.
- $0_{m_1 \times m_2}$: zero matrix of $\mathbb{R}^{m_1 \times m_2}$.
- $\mathbb{R}^{m_1 \times (\bullet)}$: space of matrices with m_1 rows.
- $\deg(\cdot)$: degree function.
- $\mathcal{H}_\gamma = \{s \in \mathbb{C} : \Re(s) > \gamma\}$: half-plane to the right of the vertical line $\Re(s) = \gamma$ where $\Re(s)$ denotes the real part of s .
- $\mathcal{C}_{\mathcal{P}}^{ord}(\mathbb{R}_+, \mathbb{R}^{m_1})$: set of functions from \mathbb{R}_+ to \mathbb{R}^{m_1} that are continuous and piecewise differentiable on \mathbb{R}_+ and that have an exponential order at infinity (see [9]).
- $*$: convolution product.

II. STATEMENT OF THE PROBLEM AND PRELIMINARIES

A. Statement of the problem

Consider two different single-output linear time-invariant systems S_1 and S_2 of the form

$$S_i \begin{cases} \dot{x}_i(t) = A_i x_i(t) + B_i u(t), \\ y_i(t) = C_i x_i(t), \\ x_i(0) = x_i^o, \end{cases} \quad (1)$$

where $x_i(t) \in \mathbb{R}^n$ and $y_i(t) \in \mathbb{R}$ are respectively the state vector and the output vector of S_i and $u(t) \in \mathbb{R}^l$ is the input vector conjointly applied to S_1 and S_2 ; A_i , B_i and C_i are constant matrices with appropriate dimensions.

In the rest of this note, we denote by $y_i(\cdot, x_i^o, u)$ the output signal y_i of S_i corresponding to the initial state vector x_i^o and the control input u . The indistinguishability zone of S_1 and S_2 is defined as follows.

Definition 1. The *indistinguishability zone*, $\mathcal{Z}_{ind}(S_1, S_2)$ of the systems S_1 and S_2 is the subset of $\mathbb{R}^n \times \mathbb{R}^n \times \mathcal{U}$ given by

$$\mathcal{Z}_{ind}(S_1, S_2) = \{(x_1^o, x_2^o, u) : y_1(\cdot, x_1^o, u) \equiv y_2(\cdot, x_2^o, u)\}$$

where \mathcal{U} denotes the set of admissible inputs of the two systems and the relation $y_1(\cdot, x_1^o, u) \equiv y_2(\cdot, x_2^o, u)$ means that the signals y_1 and y_2 are identical on \mathbb{R}_+ .

Example 2. Consider two harmonic systems S_1 and S_2 which have the following state-space representation:

$$S_i \begin{cases} \begin{bmatrix} \dot{q}_i(t) \\ \ddot{q}_i(t) \end{bmatrix} = \begin{bmatrix} 0_{\tilde{n} \times \tilde{n}} & I_{\tilde{n}} \\ -M_i^{-1} K_i & 0_{\tilde{n} \times \tilde{n}} \end{bmatrix} \begin{bmatrix} q_i(t) \\ \dot{q}_i(t) \end{bmatrix} + \begin{bmatrix} 0_{\tilde{n} \times \tilde{n}} \\ M_i^{-1} \end{bmatrix} u(t), \\ y_i(t) = q_i(t), \end{cases} \quad (2)$$

where $M_i \in \mathbb{R}^{\tilde{n} \times \tilde{n}}$ and $K_i \in \mathbb{R}^{\tilde{n} \times \tilde{n}}$ are real, symmetric positive-definite matrices. One may determine the indistinguishability zone of S_1 and S_2 as follows. Let $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$. Then $q_1 \equiv q_2 =: q$ and this implies that $\dot{q}_1 \equiv \dot{q}_2 \equiv \dot{q}$. Consequently, when $M_2 - M_1$ is invertible, we deduce from the following expression

$$\ddot{q}_i(t) = -M_i^{-1} K_i q_i(t) + M_i^{-1} u(t), \quad i = 1, 2, \quad (3)$$

which derives from (2) that $u(t)$ and $q(t)$ satisfy

$$u(t) = G q(t), \quad (4)$$

where $G = (M_2^{-1} - M_1^{-1})^{-1} (M_2^{-1} K_2 - M_1^{-1} K_1)$. By substituting $u(t) = G q(t)$ into (3) for $i = 1$, one obtains that q is the solution of the second order differential equation $\ddot{q} = -H q$ with $H := M_1^{-1} K_1 - M_1^{-1} G = M_2^{-1} K_2 - M_2^{-1} G$. Let us suppose \sqrt{H} exists and is invertible. Then one obtains

$$q(t) = \cos(t\sqrt{H}) q^o + (\sqrt{H})^{-1} \sin(t\sqrt{H}) \dot{q}^o$$

with $q^o := q(0) = q_1(0) = q_2(0)$ and $\dot{q}^o := \dot{q}(0) = \dot{q}_1(0) = \dot{q}_2(0)$. Therefore, the expression (4) of u becomes:

$$u(t) = G \cos(t\sqrt{H}) q^o + G (\sqrt{H})^{-1} \sin(t\sqrt{H}) \dot{q}^o. \quad (5)$$

We conclude that if $(x_1^o, x_2^o, u) \in \mathcal{L}_{ind}(S_1, S_2)$ then $x_1^o = x_2^o$ (i.e. $q_1(0) = q_2(0) =: q^o$ and $\dot{q}_1(0) = \dot{q}_2(0) =: \dot{q}^o$) and u is given explicitly by (5).

Conversely, suppose $x_1^o = x_2^o$ and that u is given by (5). Then, $u(t) = G q(t)$ with q solution of $\ddot{q} = -H q$. Thus, from the formula of H and (3), one obtains $\ddot{q}_1 \equiv \ddot{q} \equiv \ddot{q}_2$. As $x_1^o = x_2^o$, one has $q_1(0) = q_2(0)$ and $\dot{q}_1(0) = \dot{q}_2(0)$. Consequently, the relation $\ddot{q}_1 \equiv \ddot{q}_2$ implies that $q_1 \equiv q_2$. Hence $y_1(\cdot, x_1^o, u)$ and $y_2(\cdot, x_2^o, u)$ are identical on \mathbb{R}_+ .

Remark 3. The zero triplet of $\mathbb{R}^n \times \mathbb{R}^n \times \mathcal{U}$, $(0_n, 0_n, 0_{\mathcal{U}})$ with $0_{\mathcal{U}}$ the zero function of \mathcal{U} , always belongs to $\mathcal{L}_{ind}(S_1, S_2)$.

In the sequel of the paper, we will focus on the determination of the triplets (x_1^o, x_2^o, u) that belong to the indistinguishability zone $\mathcal{L}_{ind}(S_1, S_2)$ of S_1 and S_2 and deduce the existence conditions of non-zero triplets. Throughout the remainder of this technical note, we assume that

Assumption 4. $\mathcal{U} = \mathcal{C}_{\mathcal{F}}^{ord}(\mathbb{R}_+, \mathbb{R}^l)$.

This assumption ensures that the control inputs u considered in the paper are Laplace transformable [9], [10].

B. Preliminaries

The aim of this section is to establish some fundamental results which are necessary to characterize, in the next section, the indistinguishability zone of S_1 and S_2 . Lemma 5 is the main tool for our subsequent analysis. It gives a first characterization of the indistinguishability zone. For simplicity in the presentation, we will consider the augmented (or extended) system S associate to S_1 and S_2 defined by:

$$S \begin{cases} \dot{x}(t) = A x(t) + B u(t), \\ y(t) = C x(t), \\ x(0) = x_o, \end{cases} \quad (6)$$

where the matrices A , B and C are defined as follows:

$$A = \begin{bmatrix} A_1 & 0_n \\ 0_n & A_2 \end{bmatrix} ; \quad B = \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} ; \quad C = [C_1 \quad -C_2]$$

and

$$x_o = [(x_1^o)^T \quad (x_2^o)^T]^T = \begin{bmatrix} x_1^o \\ x_2^o \end{bmatrix}$$

is the initial state vector of S . As for the systems S_1 and S_2 , we denote by $y(\cdot, x_o, u)$ the output signal of S corresponding to the initial state vector x_o and the control input u . Thus the output $y(\cdot, x_o, u)$ of S is the difference $y_1(\cdot, x_1^o, u) - y_2(\cdot, x_2^o, u)$ of the S_1 and S_2 output signals. Consequently, $(x_1^o, x_2^o, u) \in \mathcal{L}_{ind}(S_1, S_2)$ if and only if

$$y(t, x_o, u) = 0 \quad \forall t \in \mathbb{R}_+. \quad (7)$$

One can easily show that $y(\cdot, x_o, u)$ has exponential order θ_y , for any $\theta_y \geq \lambda^* + \theta_u$ where $\lambda^* = \max_{\lambda \in \sigma(A)} \Re(\lambda)$ with $\sigma(A)$ the

spectrum of A and where θ_u denotes any exponential order of u . Hence it follows from Lerch's Theorem [11] that the condition in (7) holds if and only if there exists a real number $\theta > \lambda^* + \theta_u$ such that

$$Y(s, x_o, u) = 0 \quad \forall s \in \mathcal{H}_{\theta} \quad (8)$$

where $Y(\cdot, x_o, u)$ denotes the Laplace transform of the output $y(\cdot, x_o, u)$. It is straightforward to verify that $\forall s \in \mathcal{H}_{\lambda^* + \theta_u}$,

$$Y(s, x_o, u) = \frac{F(s, C, A, B) U(s) + F(s, C, A, x_o)}{\det(s I_{2n} - A)} \quad (9)$$

where for every $s \in \mathcal{H}_{\lambda^*}$, the operator $F(s, C, A, \cdot)$ is defined on $\mathbb{R}^{2n \times (\bullet)}$ by

$$M \in \mathbb{R}^{2n \times (\bullet)}, \quad F(s, C, A, M) = C \operatorname{adj}(s I_{2n} - A) M \quad (10)$$

with $\operatorname{adj}(s I_{2n} - A) = \det(s I_{2n} - A) (s I_{2n} - A)^{-1}$.

Lemma 5 introduces a condition under which the triplet (x_1^o, x_2^o, u) belongs to the indistinguishability zone of S_1 and S_2 .

Lemma 5. Let $(x_1^o, x_2^o, u) \in \mathbb{R}^n \times \mathbb{R}^n \times \mathcal{U}$. Then the following statements are equivalent:

- (i) $(x_1^o, x_2^o, u) \in \mathcal{L}_{ind}(S_1, S_2)$.
- (ii) There exists a real number θ such that $\theta > \lambda^* + \theta_u$ and

$$F(s, C, A, B) U(s) + F(s, C, A, x_o) = 0 \quad (11)$$

for every $s \in \mathcal{H}_{\theta}$.

This result directly derives from (9) and from the equivalence between (7) and (8). To characterize the indistinguishability zone of S_1 and S_2 using Lemma 5, the polynomial forms of $F(\cdot, C, A, B)$ and $F(\cdot, C, A, x_o)$ given by (13) are needed. This relation can be established using the following expression of $\operatorname{adj}(s I_{2n} - A)$ [12], [13]:

$$\operatorname{adj}(s I_{2n} - A) = Q_{2n-1} s^{2n-1} + \dots + Q_1 s + Q_0 \quad (12)$$

where the matrices $Q_k \in \mathbb{R}^{2n \times 2n}$, $k = 0, \dots, 2n-1$ are defined by

$$\begin{cases} Q_{2n-1} = I_{2n} \\ Q_{2n-1-r} = \alpha_{2n-r} I_{2n} + Q_{2n-r} A, \quad r = 1, 2, \dots, 2n-1 \end{cases}$$

and the α_k , $k = 0, 1, \dots, 2n-1$ denote the coefficients of the characteristic polynomial of A i.e.

$$\det(s I_{2n} - A) = s^{2n} + \alpha_{2n-1} s^{2n-1} + \dots + \alpha_1 s + \alpha_0.$$

For every $s \in \mathcal{H}_{\lambda^*}$, $F(s, C, A, M)$ can be written in the polynomial form:

$$F(s, C, A, M) = \sum_{r=0}^{2n-1} D_{2n-1-r}(M) s^{2n-1-r}. \quad (13)$$

It follows from (10) and (12) that, for every $r = 0, 1, \dots, 2n-1$, the matrices $D_{2n-1-r}(M) = C Q_{2n-1-r} M$ can be expanded as follows:

$$D_{2n-1-r}(M) = \sum_{k=0}^r \alpha_{2n-r+k} C A^k M \quad (14)$$

with $\alpha_{2n} = 1$.

Lemma 6 below gives the crucial properties of the matrix $F(s, C, A, M)$ that are needed in the next sections.

Lemma 6. Let $M \in \mathbb{R}^{n \times q}$. Then for every $r_0 \in \{0, 1, \dots, 2n-1\}$, the following statements are equivalent:

- (i) $\forall r = 0, 1, \dots, r_0, \quad D_{2n-1-r}(M) = 0_{1 \times q}$.
- (ii) $\forall r = 0, 1, \dots, r_0, \quad C A^r M = 0_{1 \times q}$.

Proof. cf. Appendix A □

The matrix $CA^k B = C_1 A_1^k B_1 - C_2 A_2^k B_2$ that appears in the formula of $F(s, C, A, B)$ corresponds to the k^{th} Markov parameter of S . Relatively to the Markov parameters of S , consider the index k_0 defined by

$$k_0 = \begin{cases} \min(\mathcal{E}_{\mathbb{N}}) & \text{if } \mathcal{E}_{\mathbb{N}} \neq \emptyset, \\ 2n & \text{if not,} \end{cases} \quad (15)$$

where $\mathcal{E}_{\mathbb{N}} = \{k \in \mathbb{N} : CA^k B \neq 0_{1 \times l}\}$.

Example 7. Consider the two harmonic systems of Example 2 and suppose $\tilde{n} = 1$ (i.e. $n = 2$). One has $k_0 = 1$ if $M_1 \neq M_2$; $k_0 = 3$ if $M_1 = M_2$ and $K_1 \neq K_2$; and $k_0 = 4$ if $M_1 = M_2$ and $K_1 = K_2$.

The index k_0 provides information on the similarity between S_1 and S_2 in terms of their Markov parameters. Using the well known Cayley-Hamilton theorem, it is easy to verify that $0 \leq k_0 \leq 2n$. Therefore, the following three cases will be discussed in the next sections:

- case 1: $k_0 = 2n$.
- case 2: $k_0 = 2n - 1$.
- case 3: $0 \leq k_0 < 2n - 1$.

The first case corresponds to the situation where S_1 and S_2 have the same Markov parameters. In the second case, only the $(2n - 1)^{\text{th}}$ Markov parameter of S_1 and S_2 are different. The last case corresponds to the case where there exists at least one couple of Markov parameters of the same order $k < 2n - 1$ that are different.

III. CHARACTERIZATION OF THE INDISTINGUISHABILITY ZONE

This section aims to characterize the indistinguishability zone of S_1 and S_2 . The case where $k_0 = 2n$ and the case where $k_0 = 2n - 1$ are considered first in Theorem 9. The more complex, and more generic, situation where $k_0 < 2n - 1$ is discussed in Theorem 16. In the sequel of the paper, we adopt the following notation:

Notation 8.

- $B = [B^1 \ B^2 \ \dots \ B^l]$ where B^j denotes the j^{th} column of B .
- $\mathcal{O}^{[k]} = [C^T \ (CA)^T \ \dots \ (CA^k)^T]^T$ denotes the observability matrix at order $k \in \mathbb{N}$ of S .
- $\mathcal{O} := \mathcal{O}^{[2n-1]}$ denotes the observability matrix of S .

A. Characterization of the indistinguishability zone: the cases $k_0 = 2n$ and $k_0 = 2n - 1$.

When $k_0 = 2n$ or $k_0 = 2n - 1$, the indistinguishability zone of S_1 and S_2 can be characterized as follows:

Theorem 9.

- (i) If $k_0 = 2n$ then $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$ if and only if

$$\begin{bmatrix} (x_1^o)^T & (x_2^o)^T \end{bmatrix}^T \in \text{Ker}(\mathcal{O}). \quad (16)$$

- (ii) If $k_0 = 2n - 1$ then $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$ if and only if x_1^o and x_2^o satisfy (16) and

$$u \in \{v \in \mathcal{U} : \forall t \in \mathbb{R}_+, v(t) \in \text{Ker}(CA^{2n-1}B)\}. \quad (17)$$

Proof.

- (i) Suppose $k_0 = 2n$. Then from the definition (15) of k_0 , $CA^r B = 0_{1 \times l}$ for every $r = 0, 1, \dots, 2n - 1$. Consequently from Lemma 6, one has $D_{2n-1-r}(B) = 0_{1 \times q}$ for every $r = 0, 1, \dots, 2n - 1$. Therefore $F(\cdot, C, A, B)$ vanishes identically and from Lemma 5, one has $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2, \mathbb{R}_+)$ if and only if the polynomial matrix function $F(\cdot, C, A, x_o)$ vanishes identically. Moreover as one can deduce from Lemma 6 and (13) - (14) that $F(\cdot, C, A, x_o)$ vanishes identically if and

only if $CA^r x_o = 0$ for every $r = 0, 1, \dots, 2n - 1$. Hence (16) holds.

- (ii) Suppose $k_0 = 2n - 1$. To begin with the proof, notice that as $CA^r B = 0_{1 \times l}$ for every $r = 0, 1, \dots, 2n - 2$ then it follows from Lemma 6 that $D_{2n-1-r}(B) = 0_{1 \times l}$ for every $r = 0, 1, \dots, 2n - 2$. Consequently,

$$F(s, C, A, B) = D_0(B) = CA^{2n-1}B \quad \forall s \in \mathcal{H}_{\lambda^*}. \quad (18)$$

Now let $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$. Its follows from (18) and from Lemma 5 that there exists a real $\theta > \lambda^* + \theta_u$ such that

$$CA^{2n-1}BU(s) = -F(s, C, A, x_o) \quad \forall p \in \mathcal{H}_{\theta}. \quad (19)$$

Since u is continuous, it follows from the initial value theorem [9] that $CA^{2n-1}Bu(0) = -\lim_{s \rightarrow +\infty} sF(s, C, A, x_o)$. Since $F(\cdot, C, A, x_o)$ is a polynomial, this can be true only if $F(\cdot, C, A, x_o)$ vanishes identically. Thus, $D_{2n-1-r}(x_o) = 0$ for every $r = 0, 1, \dots, 2n - 1$. Consequently we conclude from Lemma 6 that (16) holds. Finally, (19) becomes

$$CA^{2n-1}BU(s) = 0, \quad \forall s \in \mathcal{H}_{\theta} \subset \mathcal{H}_{\lambda^* + \theta_u} \subset \mathcal{H}_{\theta_u},$$

and we conclude from Lerch's Theorem that $CA^{2n-1}Bu(t) = 0$ for every $t \geq 0$. This shows that u satisfies (17).

Conversely let $(x_1^o, x_2^o, u) \in \mathbb{R}^n \times \mathbb{R}^n \times \mathcal{U}$ satisfy (16) and (17). Therefore, $F(\cdot, C, A, x_o)$ vanishes identically and one has $F(s, C, A, B)U(s) = CA^{2n-1}BU(s) = 0$ for every $s \in \mathcal{H}_{\theta_u} \cap \mathcal{H}_{\lambda^*}$. We conclude that for every $\theta > \lambda^* + \theta_u$,

$$F(s, C, A, B)U(s) + F(s, C, A, x_o) = 0.$$

Hence $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$. □

Remark 10. The results of Theorem 9 go through unaltered for multiple output systems using the same approach. This is no longer the case for the results of the next subsection, in particular Theorem 16.

B. Characterization of the indistinguishability zone: the case $k_0 < 2n - 1$.

In what follows, $J(S)$ is the non-empty set defined by

$$J(S) = \left\{ j \in \{1, 2, \dots, l\} : CA^{k_0}B^j \neq 0 \right\}.$$

This subsection aims to characterize the indistinguishability zone of S_1 and S_2 by solving (11) in the case where $k_0 < 2n - 1$. Note that, given x_o , (11) is a linear equation for U . It can be solved by adding to a particular solution U^{par} of this equation, the general solution U^{hom} of the homogeneous equation

$$F(s, C, A, B)U^{hom}(s) = 0 \quad \forall s \in \mathcal{H}_{\theta_1^*} \quad (20)$$

where

$$\theta_1^* = \max(\lambda^*, \delta_F)$$

with $\delta_F = \max_{1 \leq j \leq l} \{\Re(s) : F(s, C, A, B^j) = 0\}$.

We turn to solve first the homogeneous equation. As it is assumed that the control inputs belong to $\mathcal{C}_{\mathcal{P}}^{ord}(\mathbb{R}_+, \mathbb{R}^l)$, U^{hom} has to be determined so that $u^{hom} \in \mathcal{C}_{\mathcal{P}}^{ord}(\mathbb{R}_+, \mathbb{R}^l)$. The form of u^{hom} is given by Theorem 12. The proof of this result uses Lemma 11 and the expression (22) of $\text{Ker}(F(s, C, A, B))$ when $s \in \mathcal{H}_{\theta_1^*}$.

As $F(s, C, A, B)$ is a non-zero row vector for every $s \in \mathcal{H}_{\delta_F}$, it is clear that

$$\dim(\text{Ker}(F(s, C, A, B))) = l - 1 \quad (21)$$

and that for any $j_0 \in J(S)$,

$$\text{Ker}(F(s, C, A, B)) = \text{span} \left\{ \begin{array}{l} \Psi_1^{j_0}(s), \dots, \Psi_{j_0-1}^{j_0}(s), \\ \Psi_{j_0+1}^{j_0}(s), \dots, \Psi_l^{j_0}(s) \end{array} \right\} \quad (22)$$

where the components $\Psi_{i k}^{j_0}(s)$, $k = 1, 2, \dots, l$ of the vectors $\Psi_i^{j_0}(s) \in \mathbb{R}^l$ ($i \neq j_0$) are defined as follows:

$$\Psi_{i k}^{j_0}(s) = \begin{cases} 1 & \text{if } k = i, \\ -\frac{F(s, C, A, \mathbf{B}^i)}{F(s, C, A, \mathbf{B}^{j_0})} & \text{if } k = j_0, \\ 0 & \text{if } k \neq j_0 \text{ and } k \neq i. \end{cases} \quad (23)$$

Notice that it follows from Lemma (11)-(ii) and the definition of δ_F that for any $j_0 \in J(S)$, $F(s, C, A, \mathbf{B}^{j_0}) \neq 0$ for every $s \in \mathcal{H}_{\delta_F}$. Consequently, the component functions $\Psi_{i k}^{j_0}$ are well defined on \mathcal{H}_{δ_F} .

Lemma 11. Let $q \in \{1, 2, \dots, l\}$.

- (i) If $q \notin J(S)$ then $\deg[F(\cdot, C, A, \mathbf{B}^q)] < 2n - 1 - k_0$.
- (ii) If $q \in J(S)$ then $\deg[F(\cdot, C, A, \mathbf{B}^q)] = 2n - 1 - k_0$.

Proof. Since $C A^r B = 0_{1 \times l}$ for every $r = 0, 1, \dots, k_0 - 1$, it follows from Lemma 6 that $D_{2n-1-r}(B) = 0_{1 \times l}$ for every $r = 0, 1, \dots, k_0 - 1$. Consequently, $D_{2n-1-r}(\mathbf{B}^q) = 0$ for every $r = 0, 1, \dots, k_0 - 1$ and hence

$$\begin{aligned} F(s, C, A, \mathbf{B}^q) &= C A^{k_0} \mathbf{B}^q s^{2n-1-k_0} \\ &\quad + \sum_{r=k_0+1}^{2n-1} D_{2n-1-r}(\mathbf{B}^q) s^{2n-1-r}. \end{aligned}$$

The rest of the proof follows from the fact that $C A^{k_0} \mathbf{B}^q = 0$ if $q \notin J(S)$ and $C A^{k_0} \mathbf{B}^q \neq 0$ if not. \square

Theorem 12. Suppose $k_0 < 2n - 1$ and let $j_0 \in J(S)$. Then the following statements are equivalent:

- (i) $\mathbf{u}^{hom} \in \mathcal{U}$ and the Laplace transform \mathbf{U}^{hom} of \mathbf{u}^{hom} solves (20).
- (ii) There exist $l - 1$ functions $\omega_k \in \mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R})$, $k \in \{1, 2, \dots, l\} \setminus \{j_0\}$ such that

$$\mathbf{u}_k^{hom} = \omega_k, \quad \forall k \in \{1, 2, \dots, l\} \setminus \{j_0\} \quad (24)$$

and

$$\mathbf{u}_{j_0}^{hom} = - \sum_{j_0 \neq k=1}^l \omega_k * \mathcal{L}^{-1} \left[\frac{F(\cdot, C, A, \mathbf{B}^k)}{F(\cdot, C, A, \mathbf{B}^{j_0})} \right] \quad (25)$$

where \mathcal{L}^{-1} denotes the inverse Laplace transform operator and for every $k \in \{1, 2, \dots, l\}$, \mathbf{u}_k^{hom} stands for the k^{th} component of \mathbf{u}^{hom} .

Proof. (i) \implies (ii) Suppose \mathbf{U}^{hom} solves (20). Then it follows from (22) that there exists $l - 1$ functions Ω_k , $k \in \{1, 2, \dots, l\} \setminus \{j_0\}$ such that for every $s \in \mathcal{H}_{\theta_1^+}$,

$$\mathbf{U}^{hom}(s) = \sum_{j_0 \neq k=1}^l \Omega_k(s) \Psi_k^{j_0}(s). \quad (26)$$

Thus it follows from (26) and (23) that the components \mathbf{U}_k^{hom} of \mathbf{U}^{hom} are such that

$$\mathbf{U}_k^{hom}(s) = \Omega_k(s) \quad \forall k \in \{1, 2, \dots, l\} \setminus \{j_0\} \quad (27)$$

and

$$\mathbf{U}_{j_0}^{hom}(s) = - \sum_{j_0 \neq k=1}^l \Omega_k(s) \frac{F(s, C, A, \mathbf{B}^k)}{F(s, C, A, \mathbf{B}^{j_0})}. \quad (28)$$

Relation (27) implies that Ω_k is the Laplace transform of \mathbf{u}_k^{hom} and we can define $\omega_k = \mathbf{u}_k^{hom} \in \mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R})$. Furthermore, since

$\omega_k \in \mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R})$ and since $\frac{F(s, C, A, \mathbf{B}^k)}{F(s, C, A, \mathbf{B}^{j_0})}$ is proper or strictly proper (see Lemma 11),

$$\omega_k * \mathcal{L}^{-1} \left[\frac{F(\cdot, C, A, \mathbf{B}^k)}{F(\cdot, C, A, \mathbf{B}^{j_0})} \right] \in \mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R})$$

and it follows from the property of the Laplace transform that

$$\omega_k * \mathcal{L}^{-1} \left[\frac{F(\cdot, C, A, \mathbf{B}^k)}{F(\cdot, C, A, \mathbf{B}^{j_0})} \right] = \mathcal{L}^{-1} \left[\Omega_k(\cdot) \frac{F(\cdot, C, A, \mathbf{B}^k)}{F(\cdot, C, A, \mathbf{B}^{j_0})} \right]. \quad (29)$$

Hence using Lerch's Theorem and the linearity of the operator \mathcal{L}^{-1} , it is straightforward to verify that (28) and (29) imply (25).

(ii) \implies (i) Suppose there exist $l - 1$ functions $\omega_k \in \mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R})$, $k \in \{1, 2, \dots, l\} \setminus \{j_0\}$ for which \mathbf{u}^{hom} satisfies (24) and (25). Then denoting by Ω_k the Laplace transform of ω_k and applying the Laplace transform to (24) and (25), one obtains respectively (27) and (28). Consequently, we conclude from (22) and (23) that \mathbf{U}^{hom} solves (20). \square

We now turn to the task of determining a particular solution of (11). Lemmas 13 and 14 are technical results needed for the proof of Theorem 15.

Lemma 13. Let $j \in J(S)$ and let $\Phi_j(\cdot, A, B, C, x_o)$ be the rational function defined by

$$\Phi_j(s, A, B, C, x_o) = \frac{F(s, C, A, x_o)}{F(s, C, A, \mathbf{B}^j)}. \quad (30)$$

If $k_0 < 2n - 1$ then $\Phi_j(\cdot, A, B, C, x_o)$ is strictly proper if and only if

$$x_o \in \text{Ker}(\mathcal{O}^{[k_0]}). \quad (31)$$

Proof. Suppose $k_0 < 2n - 1$ and let $j \in J(S)$. Since it follows from Lemma 11 that $\deg[F(\cdot, C, A, \mathbf{B}^j)] = 2n - 1 - k_0$, $\Phi_j(\cdot, A, B, C, x_o)$ is strictly proper if and only if

$$\deg[F(\cdot, C, A, x_o)] \leq 2n - 2 - k_0. \quad (32)$$

Moreover, since (32) is equivalent to $D_{2n-1-r}(x_o) = 0$ for every $r = 0, 1, \dots, k_0$, one can conclude with Lemma 6 that $\Phi_j(\cdot, A, B, C, x_o)$ is strictly proper if and only if $C A^r x_o = 0$, $\forall r = 0, 1, \dots, k_0$. This concludes the proof. \square

Lemma 14. If $k_0 < 2n - 1$ and (x_1^o, x_2^o, u) satisfies (11) then the vector $x_o = [(x_1^o)^T \quad (x_2^o)^T]^T$ satisfies (31).

Proof. Suppose $k_0 < 2n - 1$. Let (x_1^o, x_2^o, u) satisfy (11) and let $j \in J(S)$. Then $F(\cdot, C, A, \mathbf{B}^j)$ does not vanish identically and it follows from (11) that

$$U_j(s) = - \frac{F(s, C, A, x_o)}{F(s, C, A, \mathbf{B}^j)} - \sum_{j \neq k=1}^l U_k(s) \frac{F(s, C, A, \mathbf{B}^k)}{F(s, C, A, \mathbf{B}^j)} \quad (33)$$

By a similar reasoning as in the proof of Theorem 12, one can easily show from Lemma 11, and the definition of the Laplace transform of the convolution of functions that

$$\mathcal{L}^{-1} \left[U_k(\cdot) \frac{F(\cdot, C, A, \mathbf{B}^k)}{F(\cdot, C, A, \mathbf{B}^j)} \right] = u_k * \mathcal{L}^{-1} \left[\frac{F(\cdot, C, A, \mathbf{B}^k)}{F(\cdot, C, A, \mathbf{B}^j)} \right]$$

belongs to $\mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R})$. Consequently, since u_j is continuous, (33) is true only if $\frac{F(\cdot, C, A, x_o)}{F(\cdot, C, A, \mathbf{B}^j)} = \Phi_j(\cdot, A, B, C, x_o)$ is strictly proper. Hence from Lemma 13, x_o satisfies (31). \square

Theorem 15. Assume that $k_0 < 2n - 1$. Let x_1^o and x_2^o be two vectors of \mathbb{R}^n such that $x_o = [(x_1^o)^T \quad (x_2^o)^T]^T$ satisfies (31) and

let \mathbf{u}^{par} be the control input with components \mathbf{u}_j^{par} , $j = 1, 2, \dots, l$ defined by:

$$\mathbf{u}_j^{par} = \begin{cases} 0 & \text{if } j \notin J(S), \\ -\frac{\mathcal{L}^{-1}[\Phi_j(\cdot, A, B, C, x_o)]}{\#(J(S))} & \text{if not,} \end{cases} \quad (34)$$

where $\#(J(S))$ denotes the cardinality of $J(S)$. Then $\mathbf{u}^{par} \in \mathcal{U}$ and moreover, the Laplace transform \mathbf{U}^{par} of \mathbf{u}^{par} and (x_1^o, x_2^o) satisfy (11) for $s \in \mathcal{H}_{\theta_2^*}$ where $\theta_2^* = \max(\theta_{\mathbf{u}^{par}}, \lambda^*, \delta_F)$.

Proof. Since $\Phi_j(\cdot, A, B, C, x_o)$ is strictly proper (see Lemma 13) then each component \mathbf{u}_j^{par} of \mathbf{u}^{par} is a smooth function with an exponential order at infinity. Consequently, the control input \mathbf{u}^{par} satisfies Assumption 4. It follows from (30) and (34) that for every $j \in J(S)$ and for every $s \in \mathcal{H}_{\theta_2^*}$,

$$F(s, C, A, \mathbf{B}^j) \mathbf{U}_j^{par}(s) + \frac{1}{\#(J(S))} F(s, C, A, x_o) = 0. \quad (35)$$

Moreover, since it follows from (10) and (34) that

$$F(s, C, A, B) \mathbf{U}^{par}(s) = \sum_{j \in J(S)} F(s, C, A, \mathbf{B}^j) \mathbf{U}_j^{par}(s),$$

(35) implies that for every $s \in \mathcal{H}_{\theta_2^*}$

$$F(s, C, A, B) \mathbf{U}^{par}(s) + F(s, C, A, x_o) = 0.$$

This concludes the proof. \square

The indistinguishability zone of S_1 and S_2 can be characterized as follows.

Theorem 16. *Suppose $k_0 < 2n - 1$. Then $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$ if and only if $x_o = [(x_1^o)^T \ (x_2^o)^T]^T$ satisfies (31) and $u = \mathbf{u}^{hom} + \mathbf{u}^{par}$ where \mathbf{u}^{hom} and \mathbf{u}^{par} are defined in Lemma 12 and Lemma 15, respectively.*

Proof. The proof is an immediate consequence of Lemma 5, Lemma 14, Theorem 12 and Theorem 15. \square

Notice that for SISO systems S_1 and S_2 ($l = 1$), it follows from (21) that $\dim(\text{Ker}(F(s, C, A, B))) = 0$, $\forall s \in \mathcal{H}_{\delta_F}$. Consequently $\mathbf{u}^{hom} = 0_{\mathcal{U}}$. Moreover, as $B = \mathbf{B}^1$ and $J(S) = \{1\}$, it follows from Theorem 16 that the control input u that generates identical outputs y_1 and y_2 is such that

$$u = \mathbf{u}^{par} = \mathbf{u}_1^{par} = -\mathcal{L}^{-1}[\Phi_1(\cdot, A, B, C, x_o)].$$

This is the situation of Example 2, when $\tilde{n} = 1$, $M_1 \neq M_2$ and $K_1 \neq K_2$.

IV. CONDITION FOR STRICT DISTINGUISHABILITY OF SINGLE OUTPUT SYSTEMS

The aim of this section is to derive, from the characterization of the indistinguishability zone, simple conditions allowing one to verify whether or not a couple of SISO or MISO dynamical LTI systems are strictly distinguishable. It was recently shown in [4], [7] that strict distinguishability is equivalent to an appropriate rank condition on a matrix valued function on the complex plane. Theorem 20 gives a necessary and sufficient condition for strict distinguishability for SISO systems. This condition requires checking the observability of S and the values of its Markov parameters. It is therefore much simpler than the condition of [7] which requires checking the rank condition of a matrix valued function. Theorem 22 confirms the result of Theorem 4.5 of [4] which states that two MISO systems S_1 and S_2 cannot be strictly distinguishable.

Definition 17 recalls the notion of strict distinguishability.

Definition 17. *The systems S_1 and S_2 are strictly distinguishable on a time interval $[0; \tau]$ if for any non-zero triplet $(x_1^o, x_2^o, u) \in \mathbb{R}^n \times \mathbb{R}^n \times L^1([0; \tau], \mathbb{R}^l)$ the outputs $y_1(\cdot, x_1^o, u)$ and $y_2(\cdot, x_2^o, u)$ are not identical on $[0; \tau]$.*

Lemmas 18 and 19 are technical results needed for the proof of Theorems 20 and 22. In Lemma 18 it is shown that when $\mathcal{U} = L^1([0; \tau], \mathbb{R}^l) \supset \mathcal{C}_{\mathcal{D}}^{ord}(\mathbb{R}_+, \mathbb{R}^l)$, the statements in Theorem 9-(ii) still hold.

Lemma 18. *Suppose $k_0 = 2n - 1$ and let $(x_1^o, x_2^o, u) \in \mathbb{R}^n \times \mathbb{R}^n \times L^1([0; \tau], \mathbb{R}^l)$. Then the following statements are equivalent:*

- (i) $y_1(\cdot, x_1^o, u)$ and $y_2(\cdot, x_2^o, u)$ are identical on $[0; \tau]$.
- (ii) $x_o = [(x_1^o)^T \ (x_2^o)^T]^T$ satisfies condition (16) and $C A^{2n-1} B u(t) = 0$ for every $t \in [0; \tau]$.

Proof. cf. Appendix B. \square

Lemma 19. *If $k_0 \neq 2n - 1$ then $\mathcal{Z}_{ind}(S_1, S_2) \neq \{(0_n, 0_n, 0_{\mathcal{U}})\}$.*

Proof.

- Suppose $k_0 < 2n - 1$ and let $x_o = [(x_1^o)^T \ (x_2^o)^T]^T \in \text{Ker}(\mathcal{O}^{[k_0]})$ be any non zero vector. Notice that the existence of a such vector follows from the fact that $\mathcal{O}^{[k_0]} \in \mathbb{R}^{(k_0+1) \times 2n}$ is not of full column rank. For this vector, consider the control input $u = \mathbf{u}^{par}$ defined by (34). Therefore, (x_1^o, x_2^o, u) is a non-zero triplet and it follows from Theorem 15 that $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$.
- Suppose $k_0 = 2n$. Let $u \in \mathcal{U}$ be any non-zero control input and let $[(x_1^o)^T \ (x_2^o)^T]^T \in \text{Ker}(\mathcal{O})$. Then (x_1^o, x_2^o, u) is non-zero and from Theorem 9, one has $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$. \square

Theorem 20. *Let S_1 and S_2 be SISO systems. Then S_1 and S_2 are strictly distinguishable on $[0; \tau]$ if and only if $k_0 = 2n - 1$ and $\text{Rank}(\mathcal{O}) = 2n$.*

Proof. First, suppose $k_0 \neq 2n - 1$ and let $(x_1^o, x_2^o, u) \in \mathcal{Z}_{ind}(S_1, S_2)$ such that $(x_1^o, x_2^o, u) \neq (0_n, 0_n, 0_{\mathcal{U}})$. The existence of a such triplet follows from Lemma 19. Then u is Lebesgue integrable on $[0; \tau]$ (because u is continuous on \mathbb{R}_+) and the outputs $y_1(\cdot, x_1^o, u)$ and $y_2(\cdot, x_2^o, u)$ are identical on $[0; \tau]$. Consequently, when $k_0 \neq 2n - 1$, S_1 and S_2 are not strictly distinguishable on $[0; \tau]$. Therefore to conclude the proof, we will show that if $k_0 = 2n - 1$ then S_1 and S_2 are strictly distinguishable if and only if $\text{Rank}(\mathcal{O}) = 2n$.

Now suppose $k_0 = 2n - 1$. It follows from definition 17 that S_1 and S_2 are strictly distinguishable if and only if

$$\left. \begin{array}{l} (x_1^o, x_2^o, u) \in \mathcal{V}, \\ y_1(t, x_1^o, u) = y_2(t, x_2^o, u), \\ \forall t \in [0; \tau], \end{array} \right\} \implies \left\{ \begin{array}{l} x_1^o = x_2^o = 0_n, \\ u(t) = 0, \\ \forall t \in [0; \tau], \end{array} \right.$$

where $\mathcal{V} = \mathbb{R}^n \times \mathbb{R}^n \times L^1([0; \tau], \mathbb{R})$. Moreover, since it follows from Lemma 18 that $y_1(\cdot, x_1^o, u)$ and $y_2(\cdot, x_2^o, u)$ are identical on $[0; \tau]$ if and only if for every $t \in [0; \tau]$, $C A^{2n-1} B u(t) = 0$ and $[(x_1^o)^T \ (x_2^o)^T]^T \in \text{Ker}(\mathcal{O})$, then S_1 and S_2 are strictly distinguishable on $[0; \tau]$ if and only if

$$\left\{ \begin{array}{l} [(x_1^o)^T \ (x_2^o)^T]^T \in \text{Ker}(\mathcal{O}), \\ C A^{2n-1} B u(t) = 0, \forall t \in [0; \tau], \end{array} \right\} \implies \left\{ \begin{array}{l} x_1^o = x_2^o = 0_n, \\ u(t) = 0, \forall t \in [0; \tau]. \end{array} \right.$$

Consequently, as $C A^{2n-1} B \neq 0$ (see definition (15) of k_0), we conclude that S_1 and S_2 are strictly distinguishable if and only if $\text{Ker}(\mathcal{O}) = \{0_{2n}\}$. This concludes the proof. \square

So, the only theoretical situation where a couple of SISO systems can be strictly distinguishable is the case where their augmented system S is observable and among its first $2n-1$ Markov parameters, only the $(2n-1)^{th}$ one is non-zero.

Example 21. Consider the two harmonic systems of Example 7. If $M_1 \neq M_2$ or $K_1 = K_2$ one has $k_0 \neq 3 = 2n-1$. However, when $M_1 = M_2$ and $K_1 \neq K_2$ one has $k_0 = 3$ and $\text{Rank}(\mathcal{O}) = 4 = 2n$.

Theorem 22. Let S_1 and S_2 be MISO systems. Then S_1 and S_2 are not strictly distinguishable on $[0; \tau]$.

Proof. If $k_0 \neq 2n-1$, then using the same reasoning as that of the first part of the proof of Theorem 22, it is straightforward to verify that S_1 and S_2 are not strictly distinguishable on $[0; \tau]$.

Now suppose $k_0 = 2n-1$ and let $v^* \in \text{Ker}(CA^{2n-1}B)$ such that $v^* \neq 0_l$. The existence of a such vector v^* follows from the fact that the matrix $CA^{2n-1}B \in \mathbb{R}^{1 \times l}$ is not of full column rank. Let u be the input defined by $u(t) = v^*$, $\forall t \geq 0$ and let $x_1^o \in \mathbb{R}^n$ and $x_2^o \in \mathbb{R}^n$ such that $[(x_1^o)^T (x_2^o)^T]^T \in \text{Ker}(\mathcal{O})$. Then u is Lebesgue integrable on $[0; \tau]$ and $(x_1^o, x_2^o, u) \neq (0_n, 0_n, 0_{\mathcal{U}})$. Moreover, it follows from Theorem 9 that the outputs $y_1(\cdot, x_1^o, u)$ and $y_2(\cdot, x_2^o, u)$ are identical on $[0; \tau]$. Therefore, S_1 and S_2 are not strictly distinguishable on $[0; \tau]$. \square

V. CONCLUSION

In this paper, we have proposed a novel approach to the distinguishability of two LTI systems, based on the notion of “indistinguishability zone” of the systems. Its characterization permits us to identify exactly the set of Laplace transformable control inputs that can generate the same output signals for two distinct single output systems. Note that this result allows one in principle to check whether given inputs may lead to indistinguishable outputs for the two systems. We have presented two applications of this new approach. First, we have given a simple necessary and sufficient condition for testing the strict distinguishability of SISO systems. Second, we provided a new and simple proof of the fact that MISO systems are never strictly distinguishable, first proven in [2].

APPENDIX A PROOF OF LEMMA 6

Notice that for every $r_0 \in \{0, 1, \dots, 2n-1\}$, (ii) \implies (i) because for every $r = 0, 1, \dots, r_0$, $D_{2n-1-r}(M)$ depends linearly on $CA^k M$, $k = 0, 1, \dots, r$. Now we shall prove by recurrence on $r_0 \in \{0, 1, \dots, 2n-1\}$ that (i) \implies (ii). Let \mathcal{P}_{r_0} be the proposition (i) \implies (ii) i.e.

$$\mathcal{P}_{r_0}: \left. \begin{array}{l} D_{2n-1-r}(M) = 0_{1 \times q}, \\ r = 0, 1, \dots, r_0 \end{array} \right\} \implies \left\{ \begin{array}{l} CA^r M = 0_{1 \times q}, \\ r = 0, 1, \dots, r_0. \end{array} \right.$$

Since it follows from (14) that $D_{2n-1}(M) = CM$,

$$D_{2n-1}(M) = 0_{1 \times q} \implies CM = 0_{1 \times q}.$$

Hence \mathcal{P}_0 is true. Now, suppose $r_0 \in \{1, 2, \dots, 2n\}$ and that \mathcal{P}_{r_0-1} is true. Assume that

$$D_{2n-1-r}(M) = 0_{1 \times q} \quad \forall r = 0, 1, \dots, r_0. \quad (36)$$

Then $\forall r = 0, 1, \dots, r_0-1$, $D_{2n-1-r}(M) = 0_{1 \times q}$ and since \mathcal{P}_{r_0-1} is true,

$$CA^r M = 0_{1 \times q} \quad \forall r = 0, 1, \dots, r_0-1. \quad (37)$$

So to conclude the proof, we shall prove that $CA^{r_0} M = 0_{1 \times q}$. Since $D_{2n-1-r_0}(M) = 0_{1 \times q}$ (see (36)), it follows from (14) that

$$\sum_{k=0}^{r_0-1} \alpha_{2n-r_0+k} CA^k M + CA^{r_0} M = 0_{1 \times q}. \quad (38)$$

Hence (37) and (38) imply that $CA^{r_0} M = 0_{1 \times q}$. This concludes the proof.

APPENDIX B PROOF OF LEMMA 18

(i) \implies (ii) As $k_0 = 2n-1$, $CA^r B = 0$ for every $r = 0, 1, \dots, 2n-1$. Therefore, by successive derivations of the output equation $y(t, x_1^o, u) = Cx(t)$ and successive substitution of $\dot{x}(t) = Ax(t) + Bu(t)$ into $y^{(r)}(t, x_o, u)$, one obtains

$$y^{(r)}(t, x_o, u) = CA^r x(t) + M_r u(t), \quad r = 0, 1, \dots, 2n,$$

with $M_0 = M_1 = \dots = M_{2n-1} = 0$ and $M_{2n} = CA^{2n-1}B$. Consequently, (i) implies that for every $t \in [0; \tau]$,

$$CA^r x(t) = 0, \quad r = 0, 1, \dots, 2n-1, \quad (39)$$

and

$$CA^{2n} x(t) + CA^{2n-1} B u(t) = 0. \quad (40)$$

Thus, by rewriting (39) for $t = 0$, one obtains the condition (16). It follows from the theorem of Cayley-Hamilton that (39) implies $CA^{2n} x(t) = 0$ for every $t \in [0; \tau]$. Consequently, (40) becomes $CA^{2n-1} B u(t) = 0$ for every $t \in [0; \tau]$.

(ii) \implies (i) Let $t \in [0; \tau]$. Condition (16) implies that $Ce^{tA} x_o = 0$. Moreover, as $CA^{2n-1} B u(t) = 0$ and as $CA^r B = 0$ for every $r = 0, 1, \dots, 2n-2$, we conclude from the Cayley-Hamilton theorem that $Ce^{tA} B u(t) = 0$. Consequently, for every $t \in [0; \tau]$,

$$y(t, x_1^o, u) = Ce^{tA} x_o + C \int_0^t e^{(t-\tau)A} B u(\tau) d\tau = 0.$$

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