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Phase IQC for the Hierarchical Performance Analysis of Uncertain Large Scale Systems

Khaled Laib, Anton Kornienko, Gérard Scorletti and Florent Morel

Abstract—This paper investigates the performance analysis of uncertain large scale systems. Due to their complexity, the usual robustness analysis methods based on e.g. μ or Integral Quadratic Constraints cannot be practically applied. In order to address this problem, in [1], we propose to represent a large scale system as an interconnection of sub-systems and to perform a hierarchical analysis by propagating the IQC characterization of each uncertain sub system through the interconnection. For a given computational time, the conservatism of the analysis dramatically depends on the class of IQC under consideration. In this paper, we propose a new class of IQC which characterizes the phase of uncertain system. An application to the robustness analysis of a PLL network reveal that the use of this class of IQC improves the trade-off between conservatism and computation time.

Index Terms—Uncertain large scale systems, phase uncertainty, cone sector, hierarchical approach

I. INTRODUCTION

Large scale systems (LSS) such as networks or interconnected systems have become important nowadays. With the technological development and the miniaturization of components, high complexity systems are designed in order to achieve a high level of performance, see e.g. Phase Locked Loop (PLL) networks in synchronous multi-core microprocessor systems [2], [3]. However, during the fabrication process, technological dispersions, system ageing, *etc.* could dramatically affect the performance level: so it is crucial to a priori ensure that the desired level of performance is obtained when the system is realized. Since the first step of the design is to obtain a mathematical model of the system, the differences between the realized system and the model can be expressed as an uncertain model. Ensuring a certain level of performance then reduces to a worst case performance analysis problem (robustness analysis). Robustness analysis investigates the stability and the performance of uncertain Linear Time Invariant (LTI) models. Within this framework, even if the underlying problem is NP hard, many efficient methods were developed based on relaxations as convex optimization problem under Linear Matrix Inequality (LMI) constraints [4], see e.g. the μ upper bound [5] in the μ -analysis approach [6] or the Integral Quadratic Constraint (IQC) approach [7].

Nevertheless, these methods can not be practically applied to uncertain large scale systems, since the computation time of the robustness analysis becomes dramatically important.

In order to avoid the direct μ -analysis approach and to reduce the computational load, many techniques are used depending on the nature of the interconnection topology: normal [8] or unitarily diagonalized [9] and [10] (with a normal adjacency matrix). In these works, the authors exploited the particular structure of the interconnection topology and IQC characterization of sub-systems and/or interconnections to derive scalable robust stability conditions. In [11], the authors propose scalable stability test based on Nyquist-like conditions. However, it could be applied only for SISO interconnected systems. An interesting approach is proposed in [12] and [13] where the authors exploit the sparse structure of the interconnection. Based on an IQC characterization of the interconnection, a sparse frequency depended LMI condition is obtained ensuring robust global stability. This condition can be then solved efficiently based on Cholesky factorization techniques [14] assuming a chordal patterns for interconnection topology. However, it could be difficult, for a given LSS, to model it with an interconnection which has a chordal pattern especially if, besides the stability, global system performance is under consideration.

In this paper we rather propose an alternative to [12] and [13] approach for robust performance analysis of LSS without any assumption on the interconnection topology: *Hierarchical approach*. Initially introduced by Safonov [16], it exploits the hierarchical structure of the interconnection and splits the overall analysis problem into several low dimensional problems. The coupling between these problems is ensured by appropriate IQC conditions such that it implies overall system robust analysis result including stability and performance.

According to the hierarchical approach, a large scale system is represented as a tree with leafs (an interconnection of N systems). Each system j can be described as the interconnection of sub-systems which are the leafs of system j and so on until having sub-systems that can only be described as the certain interconnection of parametric or dynamical uncertainties. Since the effect of an uncertainty and a level of (e.g. H_∞) performance can be expressed as Integral Quadratic Constraints on input-output signals, the hierarchical approach consists on the recursive propagation of the IQC of the uncertainties to the IQC which defines the performance of the large scale system *i.e.* propagate the local input-output behaviour evaluated using IQC through the network layer by layer, see [1], [15] for the details. Even if the hierarchical robustness analysis is possibly more conservative than the usual one, the benefit is to reduce the computational time. Furthermore, in order to perform the

propagation, it is necessary to compute a set of IQCs satisfied by the input and output of an interconnection whose sub-systems are defined by a set of IQCs. A set of IQCs is generated by combining elementary classes of IQC. In [1], [15], we investigate the computation of different classes of IQC. Nevertheless, these classes were not adapted for describing (uncertain) phase which is crucial e.g. in vibration control applications. In this paper, in order to improve the trade-off between the conservatism and the computation time of the hierarchical approach, we investigate the computation of an IQC corresponding to the phase information of an interconnection of sub-systems. If the phase of a Single Input Single Output (SISO) system can be easily defined, its definition for a Multiple Input Multiple Output (MIMO) system is more difficult. To address this problem, researchers had defined many concepts such as principal phases [17], phase spread [18], phase matching [19], multi-variable phase margin [20], phase envelope [21], phase sensitive structured value [22] and structured phase margins [23].

In this paper, we reveal that the definition based on the numerical range of a complex matrix Γ [18] is a nice candidate to evaluate the uncertain phase since it can be expressed as a quadratic constraint on z and w with $z = \Gamma w$. In [22], the authors considered that each uncertainty block can be phase characterized inside a cone sector and then investigate the stability according to those phase information. The uncertainty phase characterization of [22] can be seen as a special case of the phase characterization presented in this paper in the sense that for the phase rotation a matrix is used rather than a scalar in [22]. The advantage of this rotation matrix is to allow to characterize the phase uncertainty according to any point in the complex plane and not just the origin as in [22]. In this paper, the problem considered is different and more challenging than the robust stability analysis with phase information considered in [22].

In the case of SISO transfer function, the proposed definition reduces to the usual one. Furthermore, using the separation of graph theorem [24], we reveal that an IQC corresponding to the phase information of an interconnection of sub-systems can be computed using quasiconvex optimization involving LMI constraints. We then use the proposed IQC in order to reduce the conservatism of the hierarchical analysis of a PLL network.

This paper is organized as follows: section II presents the problem formulation of the uncertain phase characterization. Some preliminary background is presented followed by the proposed approach in Section III. The main results are presented in Section IV with some illustrative examples in Section V. The advantage of using the phase uncertainty to perform the hierarchical analysis on a PLL network is illustrated in Section VI.

Notations $\mathbb{RH}_\infty^{n \times m}$ denotes the set of matrices rational transfer functions with m inputs and n outputs. A^* (respectively A^T) is the complex conjugate (respectively transpose) of a the matrix A . $Re(A)$ (respectively $Re(x)$) represents the real part of a complex matrix A (respectively the complex

vector x) and $Im(A)$ (respectively $Im(x)$) represents the imaginary part. To simplify the notations, $A^\mathbb{R}$ (respectively $x^\mathbb{R}$) will be used to denote the real part and $A^\mathbb{I}$ (respectively $x^\mathbb{I}$) for the imaginary part. I_n and $0_{n \times n}$ is the identity and the zero matrices respectively, when their dimensions are not specified, it is assumed they are known from the context. The \star denotes the Redheffer star product [25]. $\mathbf{bdiag}(A, B, \dots)$ denotes the block diagonal matrix whose diagonal blocks are A, B, \dots .

II. PROBLEM FORMULATION

Let be the uncertain system $\mathcal{G} = \{M \star \Delta \mid \Delta \in \underline{\Delta}\}$, that is, an uncertain system is represented as the interconnection of a certain part $M \in \mathbb{RH}_\infty^{(n_z+n_q) \times (n_w+n_p)}$ and an uncertain part $\Delta \in \underline{\Delta}$ where $\underline{\Delta}$ denotes the set of uncertainties traditional in robust analysis literature:

$$\underline{\Delta} = \left\{ \Delta \mid \begin{array}{l} \|\Delta\|_\infty < 1 \\ \Delta = \mathbf{bdiag}(\delta_1^r I_{n_1}, \dots, \delta_{n_r}^r I_{n_{n_r}}, \\ \delta_1^c I_{c_1}, \dots, \delta_{n_c}^c I_{c_{n_c}}, \Delta_1, \dots, \Delta_{n_f}) \end{array} \right\}$$

where

- $\delta_j^r \in \mathbb{R}$ is a real n_j times repeated uncertainty
- $\delta_j^c \in \mathbb{C}$ is a complex n_c times repeated uncertainty
- $\Delta_j \in \mathbb{C}^{k_l^j \times k_m^j}$ is a full bloc of complex uncertainties.

see Fig. 1. In the sequel, for the sake of brevity, the uncertain system is denoted $M \star \Delta$. Furthermore, for a given frequency ω_0 , let us denote \mathcal{G}_{ω_0} the set $\{M(j\omega_0) \star \Delta(j\omega_0) \mid \Delta \in \underline{\Delta}\}$.

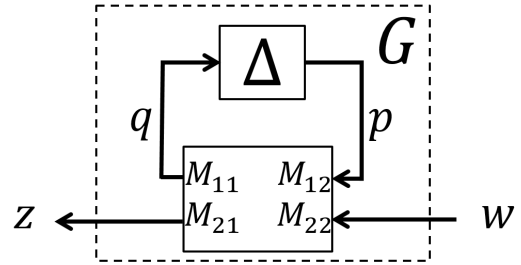


Fig. 1. Uncertain linear system

The numerical range of a complex matrix Γ , denoted $\mathcal{N}(\Gamma)$, can be used to define the phase of MIMO systems. It is defined to be a compact and convex set of \mathbb{C} given by [18]:

$$\mathcal{N}(\Gamma) = \{w^* z \mid z = \Gamma w, w \in \mathbb{C}^{n_w} \text{ and } \|w\| = 1\} \quad (1)$$

In the case of a MIMO system G , Γ represents the frequency response of G at the frequency ω_0 . In order to define the phase of an uncertain MIMO system $M \star \Delta$, the numerical range is extended to the union of the numerical ranges $\mathcal{N}(G(j\omega_0))$ for any $G \in \mathcal{G}$ which will be referred to as *union of numerical ranges*. Let us define in the complex plane the cone sector as the sector containing all these numerical ranges. It is defined by a spread angle α and the angle γ measured between the bisectrix of α and the real axis direction, see Fig. II where a sampling of the union of

numerical ranges, for a given frequency ω_0 , and the cone sector with a centre at the origin are represented. For any

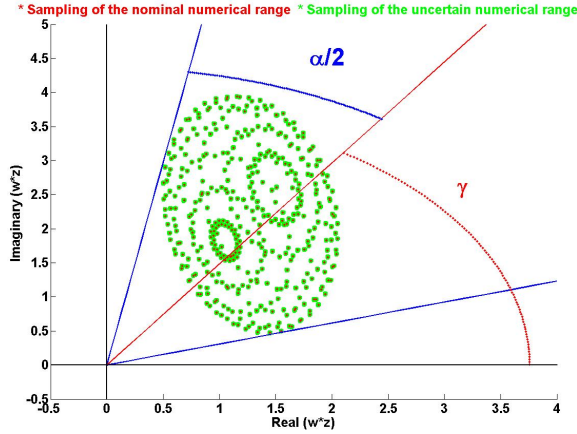


Fig. 2. Cone sector containing at a frequency ω_0 , $\mathcal{N}(G(j\omega_0))$ for any $G \in \mathcal{G}$

numerical range $\mathcal{N}(\Gamma)$ which is not contained in the right half plane, one can make $\mathcal{N}(\Gamma)$ rotated by an angle $-\gamma$ such that the resulting numerical range $\mathcal{N}(e^{-j\gamma}\Gamma)$ will be centred around the real axis in the right half plane with a new $\tilde{\gamma} = 0$. To improve the flexibility of the results, this rotation $e^{-j\gamma}$ can be generalized to be an homothetic transformation and a rotation using a scaling matrix $\Omega \in \mathbb{C}^{n_z \times n_w}$. The resulting numerical range is $\mathcal{N}(\Omega^*\Gamma)$ and it belongs to the sector centred at the origin with an angle spread α denoted $\text{sec}(0, \alpha)$. This scaling matrix is a generalization of the rotation introduced in [22]. The uncertain phase problem can then be formulated as follows.

Problem 2.1: Let \mathcal{G} be an uncertain system. For a given frequency ω_0 , find the smallest α such that:

$$\exists \Omega \in \mathbb{C}^{n_z \times n_w}, \forall \Gamma \in \mathcal{G}_{\omega_0}, \mathcal{N}(\Omega^*\Gamma) \subset \text{sec}(0, \alpha).$$

Remark 2.1: For SISO LTI systems without uncertainty, the numerical range reduces to one point $\mathcal{N}(\Gamma) = G(j\omega_0)$; in this case $\alpha = 0$ and $\Omega = e^{j \arg(G(j\omega_0))}$.

In the general case, one can define an offset characterized by $C \in \mathbb{C}^{n_z \times n_w}$ and seek the smallest sector with the corresponding notation $\text{sec}(C, \alpha)$. This problem can be solved by finding the cone sector containing all the numerical ranges of $\Gamma - C$. Hence, one can search for the smallest α such that the numerical range $\mathcal{N}(\Omega^*(\Gamma - C))$ will be on the right half plane and centred at the origin.

Problem 2.2: Let \mathcal{G} be an uncertain system. For a given frequency ω_0 and a given offset C , find the smallest α such that:

$$\exists \Omega \in \mathbb{C}^{n_z \times n_w}, \forall \Gamma \in \mathcal{G}_{\omega_0}, \mathcal{N}(\Omega^*(\Gamma - C)) \subset \text{sec}(0, \alpha).$$

III. PRELIMINARY AND PROPOSED APPROACH

A. Preliminary

Definition 3.1: The stable system G is said to be $\{X(j\omega), Y(j\omega), Z(j\omega)\}$ dissipative, with $X(j\omega) =$

$X^*(j\omega) \in \mathbb{C}^{n_z \times n_z}$, $Y(j\omega) \in \mathbb{C}^{n_z \times n_w}$ and $Z(j\omega) = Z^*(j\omega) \in \mathbb{C}^{n_w \times n_w}$ if for every $z(j\omega)$ and $w(j\omega)$ such that $z(j\omega) = G(j\omega)w(j\omega)$:

$$\begin{pmatrix} z(j\omega) \\ w(j\omega) \end{pmatrix}^* \begin{pmatrix} X(j\omega) & Y(j\omega) \\ Y(j\omega)^* & Z(j\omega) \end{pmatrix} \begin{pmatrix} z(j\omega) \\ w(j\omega) \end{pmatrix} < 0 \quad (2)$$

Remark 3.1: Dissipativity properties define a set of relations describing the input-output behaviour of a system G . They represent a set of Quadratic Constraints (QC) involving the input-output signals in the case of LTI systems. In a more general framework for non linear systems, they can be generalized to a set of Integral Quadratic Constraints (IQC) [7]. Note that in the case of LTI systems, IQC are simplified to QC.

Without loss of generality, the frequency dependence will be dropped in the sequel. A frequency gridding is defined and the different operations and ideas will be introduced for a given frequency ω_0 i.e. the $\{X(j\omega_0), Y(j\omega_0), Z(j\omega_0)\}$ dissipativity will be written as:

$$\begin{pmatrix} z \\ w \end{pmatrix}^* \begin{pmatrix} X & Y \\ Y^* & Z \end{pmatrix} \begin{pmatrix} z \\ w \end{pmatrix} < 0 \quad (3)$$

The following result is a direct application of a Theorem 4.1 of [24].

Theorem 3.1: The uncertain system $M \star \Delta$ is $\{X, Y, Z\}$ dissipative for every $\Delta \in \underline{\Delta}$ if and only if there exists an hermitian matrix $\Phi = \Phi^*$ such that:

1)

$$\begin{pmatrix} \Delta \\ I \end{pmatrix}^* \underbrace{\begin{pmatrix} \Phi_{11} & \Phi_{12} \\ \Phi_{12}^* & \Phi_{22} \end{pmatrix}}_{\Phi} \begin{pmatrix} \Delta \\ I \end{pmatrix} > 0 \quad \forall \Delta \in \underline{\Delta}$$

and

2)

$$\begin{pmatrix} M \\ I \end{pmatrix}^* \begin{pmatrix} \Phi_{22} & 0 & \Phi_{12}^* & 0 \\ 0 & X & 0 & Y \\ \Phi_{12} & 0 & \Phi_{11} & 0 \\ 0 & Y^* & 0 & Z \end{pmatrix} \begin{pmatrix} M \\ I \end{pmatrix} < 0$$

Remark 3.2: Theorem 3.1 presents necessary and sufficient conditions for the uncertain system $M \star \Delta$ to be $\{X, Y, Z\}$ dissipative. Testing these conditions can be expressed as the optimization problem of finding Φ such that conditions 1) and 2) are satisfied. However, condition 1) makes the optimization problem infinite dimensional since it has to be tested for all $\Delta \in \underline{\Delta}$, which is difficult from a computational point of view. The complexity can be reduced by introducing an affine set Φ_{Δ} such that for any $\Phi \in \Phi_{\Delta}$, condition 1) is satisfied. The set Φ_{Δ} depends on the nature of $\underline{\Delta}$. In this case, Theorem 3.1 gives only sufficient conditions for all $\Delta \in \underline{\Delta}$ and the conditions of Theorem 3.1 define a finite dimensional LMI optimization problem that consists in finding one $\Phi \in \Phi_{\Delta}$ such that the second condition of Theorem 3.1 is satisfied. From a computational point of view, this problem can be efficiently solved. The consequence of this parametrization of Φ is a possible conservatism in the obtained results. However, this last can be reduced by an

appropriate choice of Φ_{Δ} depending on the class of the uncertainties $\underline{\Delta}$.

Remark 3.3: If conditions 1) and 2) of Theorem 3.1 are satisfied for all ω and since $\underline{\Delta}$ is a connected set that contains 0 then the stability is guaranteed, see Theorem 4.1 of [24]. Furthermore, when $\underline{\Delta}$ is not a ball of center 0, conditions 1) and 2) still imply the results of Theorem 3.1 if $\underline{\Delta}$ is a connected set with Δ_0 such that $M \star \Delta_0$ is stable.

B. Proposed approach

Find a cone sector $\sec(0, \alpha)$ of Problem 2.1 (that contains $\mathcal{N}(\Gamma), \forall \Gamma \in \mathcal{G}_{j\omega_0}$) can be formulated as find b such that:

$$\begin{aligned} \operatorname{Re}(w^* \Omega^* z) - \beta \operatorname{Im}(w^* \Omega^* z) &> 0 \quad \beta = \pm b \\ \forall z, w \quad \text{such that } \exists \Delta \in \underline{\Delta} \text{ and } z &= M \star \Delta w \end{aligned} \quad (4)$$

$b > 0$ defines a slope of two lines forming the cone sector and is related to α by $\tan\left(\frac{\alpha}{2}\right) = \frac{1}{b}$. Please note that the matrix Ω is used to rotate all numerical ranges in the right half plane such that $\gamma = 0$, see Fig. 3. Inequalities (4) represent a couple of Quadratic Constraints

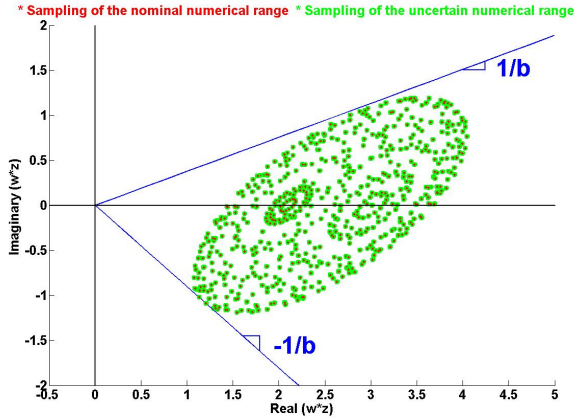


Fig. 3. Illustration of $\mathcal{N}(\Gamma)$ and the cone sector $\sec(0, \alpha)$

(QC) characterizing the input output behaviour of the system G with its input and output signals w and z . The cone sector can hence be formulated as:

$$\begin{pmatrix} z \\ w \end{pmatrix}^* \begin{pmatrix} 0 & \Omega(-I + j\beta I) \\ (\Omega(-I + j\beta I))^* & 0 \end{pmatrix} \begin{pmatrix} z \\ w \end{pmatrix} < 0 \quad (5)$$

For any other given $C \neq 0$ i.e. as defined in Problem 2.2, the cone sector $\sec(C, \alpha)$ is given by:

$$\begin{pmatrix} z \\ w \end{pmatrix}^* \begin{pmatrix} X & Y \\ Y^* & Z \end{pmatrix} \begin{pmatrix} z \\ w \end{pmatrix} < 0 \quad (6)$$

Where:

$$X = 0 \quad Y = \Omega(-I + j\beta I) \quad Z = -(Y^* C + C^* Y)$$

With $\beta = \pm b$, the inequalities of (6) define two dissipativity properties $\{X_1, Y_1, Z_1\}$ for $\beta = +b$ and $\{X_2, Y_2, Z_2\}$ for $\beta = -b$.

IV. MAIN RESULTS

Theorem 4.1: For a given C , given α and for the uncertain system $M \star \Delta$. If there exist: $\Omega, (\Phi_{111}, \Phi_{121}, \Phi_{221}) \in \Phi_{\Delta}$ and $(\Phi_{112}, \Phi_{122}, \Phi_{222}) \in \Phi_{\Delta}$ such that:

$$\begin{pmatrix} M \\ I \end{pmatrix}^* \begin{pmatrix} \Phi_{22i} & 0 & \Phi_{12i}^* & 0 \\ 0 & X_i & 0 & Y_i \\ \Phi_{12i} & 0 & \Phi_{11i} & 0 \\ 0 & Y_i^* & 0 & Z_i \end{pmatrix} \begin{pmatrix} M \\ I \end{pmatrix} < 0 \quad i = 1, 2$$

Where:

$$X_1 = 0, \quad X_2 = 0$$

$$Y_1 = \Omega(-I + j \cot\left(\frac{\alpha}{2}\right) I), \quad Z_1 = -(Y_1^* C + C^* Y_1)$$

$$Y_2 = \Omega(-I - j \cot\left(\frac{\alpha}{2}\right) I), \quad Z_2 = -(Y_2^* C + C^* Y_2)$$

Then, the cone sector $\sec(0, \alpha)$ contains all the numerical ranges of the uncertain system $M \star \Delta - C$ scaled by Ω :

$$\mathcal{N}(\Omega^*(M \star \Delta - C)) \subset \sec(0, \alpha) \quad \forall \Delta \in \underline{\Delta}$$

Proof: Theorem 4.1 is a direct application of Theorem 3.1, with $\{X, Y, Z\}$ dissipativity (6) for $b = \cot\left(\frac{\alpha}{2}\right)$ and by choosing some particular structure defined by Φ_{Δ} , see Remark 3.2. ■

Theorem 4.1 defines an optimization problem. The decision variables are: $\Omega, \Phi_{111}, \Phi_{121}, \Phi_{221}, \Phi_{112}, \Phi_{122}$, and Φ_{222} . It proposes only sufficient conditions for the existence of a cone sector $\sec(0, \alpha)$ containing all the numerical ranges. Testing them is a finite dimensional feasibility problem involving LMI which can be efficiently solved [26]. If a cone sector exists, it is possible to search for the smallest one. To this purpose, we introduce a size measure for the cone sector. The objective is to minimize α . Since $\tan\left(\frac{\alpha}{2}\right) = \frac{1}{b}$ and $b > 0$:

$$\begin{aligned} \min_{\alpha} \quad \alpha &\Leftrightarrow \min_{\alpha} \quad \tan(\alpha/2) \\ &\Leftrightarrow \min_b \quad 1/b \end{aligned}$$

Then, $1/b$ represents a size measure for the cone. Since the conditions of Theorem 4.1 are only sufficient, minimizing $1/b$ such that they are satisfied implies the minimization of an upper bound $\tilde{\alpha}$ on the spread angle α . Nevertheless, how this upper bound is close to real spread depends on the choice of Φ_{Δ} , see Remark 3.2.

After defining the size measure of the cone sector $\sec(C, \alpha)$, Problem 2.2 can be addressed such that an upper bound on the angle α can be computed efficiently.

Theorem 4.2: An upper bound on the angle α defined in Problem 2.2 can be obtained by finding $\Omega, \hat{D}_1, \hat{G}_1, \hat{D}_2, \hat{G}_2, \hat{D}_2$, and \hat{G}_2 that minimize λ in the following Generalized Eigenvalues Problem (GEVP):

$$\begin{aligned} \min_{\lambda, \Omega} \quad & \lambda \\ & \hat{D}_1, \hat{G}_1, \hat{D}_1, \hat{G}_1 \\ & \hat{D}_2, \hat{G}_2, \hat{D}_2, \hat{G}_2 \end{aligned}$$

1)

$$\lambda \begin{pmatrix} \tilde{D}_1 & 0 \\ 0 & \tilde{D}_2 \end{pmatrix} + \begin{pmatrix} \tilde{D}_1 & 0 \\ 0 & -\tilde{D}_2 \end{pmatrix} > 0$$

$$\lambda \begin{pmatrix} M & 0 \\ I & 0 \\ 0 & M \\ 0 & I \end{pmatrix}^* \begin{pmatrix} B_1 & 0 \\ 0 & B_2 \end{pmatrix} \begin{pmatrix} M & 0 \\ I & 0 \\ 0 & M \\ 0 & I \end{pmatrix} + \dots$$

$$\dots + \begin{pmatrix} M & 0 \\ I & 0 \\ 0 & M \\ 0 & I \end{pmatrix}^* \begin{pmatrix} A_1 & 0 \\ 0 & A_2 \end{pmatrix} \begin{pmatrix} M & 0 \\ I & 0 \\ 0 & M \\ 0 & I \end{pmatrix} > 0$$

2)

$$\begin{pmatrix} \tilde{D}_1 & 0 \\ 0 & \tilde{D}_2 \end{pmatrix} > 0$$

$$\begin{pmatrix} M & 0 \\ I & 0 \\ 0 & M \\ 0 & I \end{pmatrix}^* \begin{pmatrix} B_1 & 0 \\ 0 & B_2 \end{pmatrix} \begin{pmatrix} M & 0 \\ I & 0 \\ 0 & M \\ 0 & I \end{pmatrix} > 0$$

with $i = \{1, 2\}$, B_i and A_i are given by:

$$B_i = \begin{pmatrix} -\tilde{D}_i & 0 & -\tilde{G}_i^* & 0 \\ 0 & 0 & 0 & \Omega \\ -\tilde{G}_i & 0 & \tilde{D}_i & 0 \\ 0 & \Omega^* & 0 & -(\Omega^* C + C^* \Omega) \end{pmatrix}$$

$$A_i = (-1)^{i-1} \begin{pmatrix} -\tilde{D}_i & 0 & -\tilde{G}_i^* & 0 \\ 0 & 0 & 0 & -j\Omega \\ -\tilde{G}_i & 0 & \tilde{D}_i & 0 \\ 0 & (-j\Omega)^* & 0 & -j(\Omega^* C - C^* \Omega) \end{pmatrix}$$

Proof: For the cone sector defined by the dissipativity inequality (6) and using Theorem 4.1, Problem 2.2 can be solved with finding a minimum upper bound $\tilde{\alpha}$ on α such that $\text{sec}(C, \alpha) \subset \text{sec}(C, \tilde{\alpha})$. Hence, Problem 2.2 becomes: Find $\Phi_i = \{\Phi_{11i}, \Phi_{12i}, \Phi_{22i}\} \in \Phi_{\Delta}$, X_i , Y_i , and Z_i , $i = \{1, 2\}$, that maximize b such that:

$$\begin{pmatrix} M \\ I \end{pmatrix}^* \begin{pmatrix} -\Phi_{22i} & 0 & -\Phi_{12i}^* & 0 \\ 0 & -X_i & 0 & -Y_i \\ -\Phi_{12i} & 0 & -\Phi_{11i} & 0 \\ 0 & -Y_i^* & 0 & -Z_i \end{pmatrix} \begin{pmatrix} M \\ I \end{pmatrix} > 0 \quad i = 1, 2 \quad (7)$$

where X_i , Y_i and Z_i defined as in (6) for $\beta = +b$ and $\beta = -b$. Let us define Φ_1 and Φ_2 as:

$$\Phi_1 = \hat{\Phi}_1 + b\tilde{\Phi}_1 \quad \Phi_2 = \hat{\Phi}_2 - b\tilde{\Phi}_2 \quad (8)$$

where $\hat{\Phi}_i$ and $\tilde{\Phi}_i$ are chosen in the form of DG scaling presented in [5]:

$$\hat{\Phi}_i = \begin{pmatrix} -\hat{D}_i & \hat{G}_i \\ \hat{G}_i^* & \hat{D}_i \end{pmatrix} \quad \tilde{\Phi}_i = \begin{pmatrix} -\tilde{D}_i & \tilde{G}_i \\ \tilde{G}_i^* & \tilde{D}_i \end{pmatrix}$$

with: $\hat{D}_i > 0$. In order to make sure that $\Phi_i \in \Phi_{\Delta}$, one needs to guarantee that:

$$\hat{D}_1 + b\tilde{D}_1 > 0 \quad \hat{D}_2 - b\tilde{D}_2 > 0 \quad (9)$$

Then, inequalities (7) become:

$$\begin{pmatrix} M \\ I \end{pmatrix}^* \left[\begin{pmatrix} -\hat{D}_i & 0 & -\hat{G}_i^* & 0 \\ 0 & 0 & 0 & \Omega \\ -\hat{G}_i & 0 & \hat{D}_i & 0 \\ 0 & \Omega^* & 0 & -(\Omega^* C + C^* \Omega) \end{pmatrix} \right.$$

$$\left. + (-1)^{i-1} b \begin{pmatrix} -\tilde{D}_i & 0 & -\tilde{G}_i^* & 0 \\ 0 & 0 & 0 & -j\Omega \\ -\tilde{G}_i & 0 & \tilde{D}_i & 0 \\ 0 & (-j\Omega)^* & 0 & -j(\Omega^* C - C^* \Omega) \end{pmatrix} \right] \begin{pmatrix} M \\ I \end{pmatrix} > 0 \quad (10)$$

Evaluating for $i = \{1, 2\}$ and since $b > 0$, inequalities (9) and (10) become:

$$\frac{1}{b} \hat{D}_1 + \tilde{D}_1 > 0$$

$$\frac{1}{b} \hat{D}_2 - \tilde{D}_2 > 0$$

$$\frac{1}{b} \begin{pmatrix} M \\ I \end{pmatrix}^* B_1 \begin{pmatrix} M \\ I \end{pmatrix} + \begin{pmatrix} M \\ I \end{pmatrix}^* A_1 \begin{pmatrix} M \\ I \end{pmatrix} > 0$$

$$\frac{1}{b} \begin{pmatrix} M \\ I \end{pmatrix}^* B_2 \begin{pmatrix} M \\ I \end{pmatrix} + \begin{pmatrix} M \\ I \end{pmatrix}^* A_2 \begin{pmatrix} M \\ I \end{pmatrix} > 0$$

The last inequalities can be combined together as it is shown in the first condition of Theorem 4.2. Hence, minimizing $1/b$ such that the last inequalities holds is a Generalized Eigenvalues Problem since condition 2) of Theorem 4.2 holds. ■

Remark 4.1: Theorem 4.2 is a Generalized Eigenvalues Problem, it has been proved that it is a quasiconvex optimization problem [26]. It can be solved efficiently using the projective method described in [27]. Nevertheless, since the conditions are sufficient, Theorem 4.2 allows to compute a sector $\text{sec}(C, \tilde{\alpha})$ which contains the smallest cone sector $\text{sec}(C, \alpha)$. The conservatism can be reduced by an appropriate choice of Φ_{Δ} depending on the nature of uncertainties: D scaling [6], DG scaling [5] or DGL scaling [28]. Here the DG scaling is chosen to characterize Φ_i for illustration purposes. Nevertheless, one can choose other types of scaling depending on the nature of uncertainties. In this case, the structure of parametrization matrices Φ_i will be different but the idea remains the same *i.e.* decomposing Φ_i into two parts: $\hat{\Phi}_i$ and $\tilde{\Phi}_i$.

Remark 4.2: With respect to Theorem 4.1, Theorem 4.2 introduces an extra condition (condition 2) in order to obtain a Generalized Eigenvalues problem due to the advantage of this optimization problem. This condition has two nice justifications: first, it guarantees that the interconnection $M \star \Delta, \forall \Delta \in \underline{\Delta}$ is stable. Second, it guarantee that $\mathcal{N}(\Omega^*(\Gamma - C))$ is located in the right half plane. Then, one needs to minimize the upper bound $\tilde{\alpha}$:

$$\tilde{\alpha} \in \left[-\frac{\pi}{2}, +\frac{\pi}{2} \right] \quad \text{s.t.} \quad \mathcal{N}(\Omega^*(\Gamma - C)) \subset \text{sec}(0, \tilde{\alpha})$$

On the other side, introducing this condition enforces more constraints on $\Phi_1 \in \Phi_{\Delta}$ and $\Phi_2 \in \Phi_{\Delta}$ if they are chosen as in (8) with $\tilde{\Phi}_i = 0$: they must satisfy that all the numerical ranges $\mathcal{N}(\Omega^*(\Gamma - C))$ are in the right half plane and respect their dissipative properties at the same time. Hence, an extra conservatism may appear (besides that of Remark 4.1). To overcome this issue, Φ_1 and Φ_2 are chosen of the form (8). Here, \hat{D}_i and \hat{G}_i ensure that the interconnection $M \star \Delta$ is stable and $\mathcal{N}(\Omega^*(\Gamma - C))$ is in the right half plane. While \tilde{D}_i and \tilde{G}_i introduce an additional degree of freedom to ensure that $\mathcal{N}(\Omega^*(\Gamma - C)) \subset \text{sec}(0, \tilde{\alpha})$. Hence, with this choice, the conservatism is reduced.

V. ILLUSTRATION EXAMPLES

In this section, we will show some examples that illustrate our results graphically.

A. SISO Uncertain Systems

The following example is inspired from [1]. Let us consider the following SISO example with structured scalar uncertainties:

$$G^{SISO}(j\omega_0) = \frac{-\omega_0^2 + 2\xi_n\omega_n(j\omega_0) + \omega_n^2}{-\omega_0^2 + 2\xi_d\omega_d(j\omega_0) + \omega_d^2}$$

with $\omega_0 = 1 \text{ rad/s}$, $\xi_n = 0.7$ and $\xi_d = 0.01$. The uncertainties are parametric and they are defined by: $\omega_n \in [0.1, 0.5]$ and $\omega_d \in [9, 10]$. The results are illustrated in Fig.4.

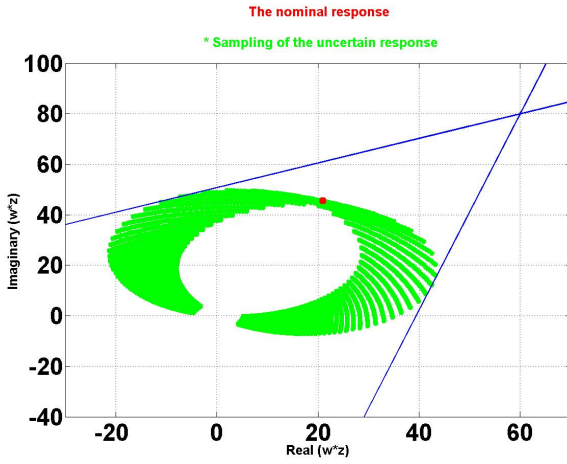


Fig. 4. Visualisation of cone sector for SISO example

The sampling of the uncertain response presented in green dots is obtained for a gridding of ω_n and ω_d equal to 0.01 and 0.001 respectively. The nominal response is equal to $21 + 45i$ and it is presented with a red dot. The cone offset C was fixed arbitrary at $C = 60 + 80i$. The obtained results shows very low conservatism and define a cone angle $\tilde{\alpha} = 49.52^\circ$.

B. MIMO Uncertain Systems

For MIMO case, a simple example can be given by:

$$G^{MIMO}(j\omega_0) = o_{dir} \times G^{SISO}(j\omega_0) \times i_{dir} - C$$

with same frequency $\omega_0 = 1 \text{ rad/s}$, $o_{dir} = (1 \ -1.5)^T$, $i_{dir} = (1 \ 2)$ and $C = 50(1 + 5i)I_2$. Fig. 5 shows a sampling of the union of numerical ranges while Fig. 6. shows a sampling of the union of numerical ranges scaled by Ω^* with the obtained cone sector.

As it can be seen in Fig 5 and Fig 6, the numerical range $\Gamma(\mathcal{N}-C)$ which was located in the left half plane was rotated and scaled in size by Ω^* . Once it is in the right half plane, a cone sector $sec(0, \tilde{\alpha})$ is found: $\tilde{\alpha} = 25^\circ$. See Fig. 6 for illustration purposes, please note that in the imaginary and the real axe scales are not equal.

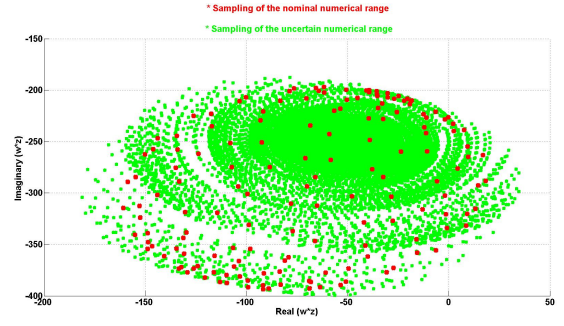


Fig. 5. Visualisation of the nominal and the uncertain numerical ranges

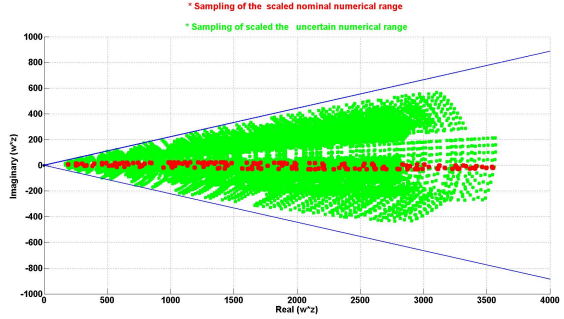


Fig. 6. Visualisation of the scaled nominal and the uncertain numerical ranges

VI. LARGE SCALE SYSTEM HIERARCHICAL ANALYSIS

Let us consider the example of the performance analysis of the active clock distribution network of [2]. It is composed of $N = 16$ mutually synchronized Phase Locked Loop (PLL) delivering the clock signals to the chip. In order to synchronize all the network, the PLLs exchange information through an interconnection structure. This example is suitable for illustration of the proposed hierarchical analysis approach as the performance is naturally evaluated in the frequency domain [15].

A. PLL network description

In this application, all the PLLs are homogeneous *i.e.* have the same description and uncertainty set $\underline{\Delta}$. Due to the manufacturing process, technological dispersions are inevitable. They can be presented as parametric uncertainties belonging to the same set $\underline{\Delta}$. Then, the description of the N PLLs is:

$$T_i(j\omega_0) = \frac{k_i(j\omega_0 + a_i)}{-\omega^2 + k_i j\omega_0 + k_i a_i} \quad \forall i \in \{1, \dots, N\}$$

Where $k_i \in [0.76, 6.84] \times 10^4$, $a_i \in [91.1, 273.3]$ and ω_0 is the current frequency defined by gridding. Furthermore, $T_i(j\omega_0)$ can be written as the interconnection of certain and uncertain part:

$$T_i(j\omega_0) = \Delta_i \star M_{PLL} \quad \Delta_i \in \underline{\Delta}$$

With Δ_i is given by:

$$\underline{\Delta} = \left\{ \Delta_i = \begin{pmatrix} \delta_{k_i} & 0 \\ 0 & \delta_{a_i} \end{pmatrix} \mid \|\Delta_i\|_\infty \leq 1 \right\}$$

The exchange of information between PLLs is modelled by an interconnection matrix defined in [1] (see equation (13)). The performance of this network can be characterized by its global input and output w_g and z_g using the global frequency response magnitude bound. (See more details in [1] and [15]).

B. Hierarchical Approach

For the PLL network, the hierarchical approach consists of two steps:

1) *Local step*: Characterize the input-output behaviour of each PLL using dissipativity properties. In addition to the cone sector, one can use other quadratic constraints that can be interpreted by simple geometric forms: disc and band, see [1] and [15]) for more details. Since all the PLLs are homogeneous, the dissipativity properties obtained for one PLL are valid for all the 16 PLL.

2) *Global step*: Evaluate the performance of the network by finding a minimal upper bound on its frequency response magnitude using the dissipativity properties describing the PLLs and obtained in the local step. See [16], [1] and [15] for more details. In this step, one can combine the different dissipativity properties of each PLL: disc, disc+band, disc+cone, *etc.* and propagate this input-output characterization of each sub-system to investigate the performance of the network.

C. The PLL network performance analysis

The performance analysis results of the PLL network presented in Fig. 7 and summarized in TABLE I. The direct

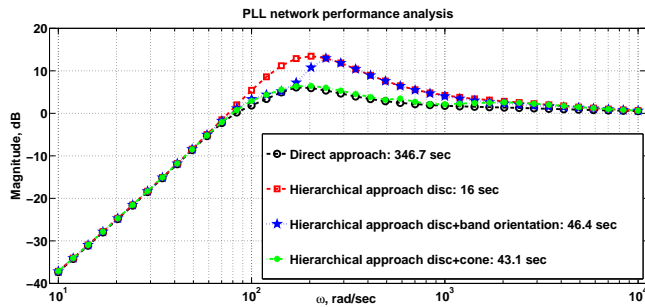


Fig. 7. The PLL network performance analysis

μ -analysis approach presents the less conservative results with a maximum peak of 6.01dB comparing to the different hierarchical approaches. However, computation time is significant: 346.7s.

Approach	Maximum peak	Computation time
μ -analysis	6.01dB	347.6s
Hierarchical: disc	13.44dB	16s
Hierarchical: disc+band	12.97dB	46.4s
Hierarchical: disc+cone	6.45dB	43.1s

TABLE I

COMPARISON BETWEEN THE DIFFERENT APPROACHES

To overcome the time issue, the authors of [1] and [15] introduced the hierarchical approach for the PLL network

performance analysis with two sub-system characterizations: disc and band QCs. The results of the disc [1] and of the disc+band [15] characterizations are presented in Fig. 7. It is clear that the performance analysis is much faster but also more conservative. This paper introduces a new QC: cone sector which considers the phase information. The cone sector is used in combination with the disc QC obtained in [1]. Namely, in the PLL application example, the cone sector offset is fixed as $C = c_{disc} + ir_{disc}$ where c_{disc} and r_{disc} are the disc centre and radius respectively obtained from the disc QC [1]. It is clear that this combination is more suitable for the PLL network where the obtained result is almost the same as the result of direct μ -analysis approach but in much less time, see Fig. 7.

In a general way, the conservatism of the hierarchical approach can be reduced by combining multiple dissipativity properties from the local step: disc+bands+cones. However, to obtain precise result it is better to combine the dissipativity properties of different nature. Of course, the computation time will increase with the number of dissipativity properties used, however since they are used in local step, and thus potentially for a system of small dimensions (local sub-system), this increase is limited within a reasonable time. The contribution of this paper, is an introduction of a new QCs that take into account the information on uncertain sub-system phase. One can thus better define a trade-off between conservatism and efficiency.

VII. CONCLUSION

In this paper, the performance analysis of uncertain large scale systems is considered. In order to reduce the complexity and the computation time, the hierarchical approach is used to investigate the performance of a PLL network. A new IQC is proposed for the local step which is the phase uncertainty characterization. SISO and MIMO examples are used to illustrate the efficiency of this characterization. This new IQC confirmed its advantage in obtaining less conservative results in the hierarchical approach comparing with the already existing IQCs in literature. For the future work, it would be interesting to investigate the appropriate decomposition of the network into sub-networks in order to have the less conservative results in a reasonable time *i.e.* add an additional indicator defining conservatism/computation time trade-off.

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