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# Performance Analysis of an EEMD-based Hilbert Huang Transform as a Bearing Failure Detector in Wind Turbines

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**Abstract**—Sustainability and viability of wind farms are highly dependent on the reduction of the operational and maintenance costs. The most efficient way of reducing these costs would be to continuously monitor the condition of these systems. This allows for early detection of the degeneration of the generator health, facilitating a proactive response, minimizing downtime, and maximizing productivity. This paper deals then with the assessment of a demodulation technique for bearing failure detection through wind turbines generator stator current. The proposed technique is based on a modified version of the Hilbert Huang transform. In this version, the use of the EEMD algorithm allows overcoming the well-known mixed mode.

**Keywords**—Wind turbines, bearing failure detection, amplitude demodulation, Hilbert Huang transform, EEMD.

## I. INTRODUCTION

Wind turbines failure detection is certainly one of the most important key in maintenance cost reduction. Despite the long experience accumulated by several technologies applied in electric machine, the task in wind turbines is still an art. It has become even more challenging as far as wind energy conversion system are deployed onshore or offshore where there are substantial wind resources, leading to a best electricity generating opportunities, so it yields to high maintenance costs because they are inaccessible or hardly accessible [1]. With the development of these wind farms due to increasing land or sea constraints, new challenges arise particularly with regard to maintenance. Indeed, maintenance is significantly restricted during periods of high wind speed and significant wave height. In this context, cost-effective, predictive and proactive maintenance of wind turbines assumes more importance. Wind turbine Condition Monitoring Systems (CMS) provide then an early indication of component incipient failure, allowing the operator to plan system repair prior to complete failure. So the CMS will be an important tool for lifting uptime and maximizing productivity; in other words when cost-effective availability targets must be reached. The experience feedback of wind turbine industries shows important features of failure rate values and trends [2-3]; and states that a large fraction of wind turbine downtime is due to drive train and bearing failures, particularly in the generator and gearbox. For failure diagnosis problem, it is important to know if a failure exists or not in the generator; and in addition, to identify the failed element of the

system and to find the failure causes via the processing of available measurements. To today condition monitoring techniques for wind turbines have not been resolved and have not reached their full potential, because CMS are highly linked to the detection philosophy and should be applied only when the detection methods are reliable [4-5]. A well-known method for assessing impending problems is to use current sensors installed within the wind turbine generator as transducer for failure detection [6].

Many techniques and tools are developed for condition monitoring of wind turbine electric generator in order to prolong their life span [7]. Some of the technology used for monitoring includes existing and pre-installed sensors, such for speed, torque, vibrations, temperature, flux density, etc. These sensors are managed together in different architectures and coupled with algorithms to allow an efficient monitoring of the system condition. Those methods are outcome from electric motor condition monitoring. From the theoretical and experimental point of view, the well-established methods are: electrical quantities signature analysis (current, power...), vibration monitoring, temperature monitoring and oil monitoring. In the case of wind turbines generator, some research works on fault detection were carried out using the electrical quantities of the generator, such as the diagnosis of unbalance and failure in the blades of a small wind turbine by measuring the power spectrum density at the turbine generator terminal [8]. The advantage of signature analysis of the generator electrical quantities is that those quantities are easily accessible during operation.

Analyzing the generator electrical quantities usually involves the use of signal processing techniques. For steady state operations, the Fast Fourier Transform (FFT) and other techniques based upon it are widely used in the literature [8-9]. However, in the case of variable speed wind turbines, the FFT is difficult to interpret and to extract the features of variations in time-domain, since the operation is predominately non-stationary due the behavior of the wind speed. To overcome this problem, procedures based on time-frequency representations (Spectrogram, Quadratic Wiegner-Ville, etc.) or time-scale analysis (wavelet) have been proposed in the literature of the electric machines community [10-12]. Nevertheless, these methods are formulated through integral transforms and analytic signal representations, so their accuracy depends on data length

and stationarity. Also these techniques have drawbacks such as high complexity, poor resolution and/or may suffer from artifacts (cross-terms...) and it is not easier to track the frequencies introduced by the failure.

This paper deals then with the assessment of a demodulation technique for bearing failure detection through wind turbines generator stator current [13]. The proposed technique is based on a modified version of the Hilbert Huang transform. In this version, the use of the EEMD algorithm allows overcoming the well-known mixed mode. The proposed technique is tested using experimental data from a 0.75kW test bench.

## II. ROLLING ELEMENT BEARING FAULT

The failure of rolling element bearings of the electric generator is the most common failure mode associated with a long downtime of wind turbines. Because of their construction, rolling element bearings generate precisely identifiable signature on vibration. The characteristic frequencies of rolling element bearings depend on the geometrical size of the various elements [14]. Those frequencies present an effective route for monitoring progressive bearing degradation. It is therefore possible to detect on the stator side the frequencies associated with the bearings using an accelerometer mounted directly on the bearing housing, which is not often easily accessible. It is also true that vibration monitoring has made out its efficiency; and it is highly suitable for rolling element bearings, however it represents an issue when requiring a good vibration baseline. To tackle this problem, an alternative procedure for bearing failures detection in electrical machines is proposed by analyzing the stator side electrical quantities, such as the current or the instantaneous power [14]. Indeed, bearing failures generate predictable frequencies in the stator current. In fact, a bearing failure is assumed to produce an air gap eccentricity. The effect of the eccentricity on the magnetic flux distribution is depicted in Fig. 1. Due to the eccentric rotor motion, an *unbalanced magnetic pull* is produced; this gives rise to torque oscillations which lead to an amplitude and/or phase modulation of stator current [15]. It is therefore sufficient to demodulate the current to achieve failure detection. In this paper, the authors will assess an alternative technique detecting bearing failures regardless the stator current frequency content.

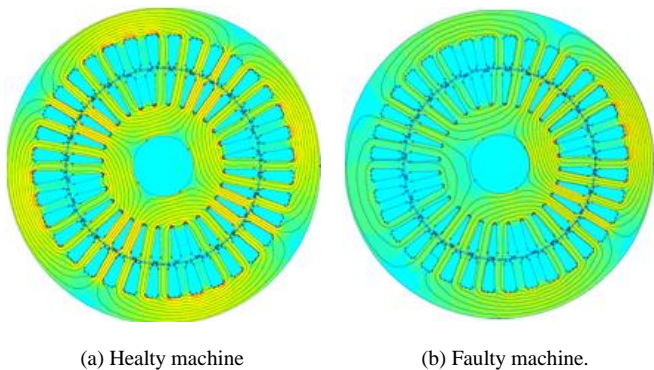


Fig. 1. Effect of the eccentricity on the magnetic field.

## III. SIGNAL PROCESSING TOOLS

This work focuses on mechanical failures that lead to stator current amplitude modulations. These include bearing failure and air gap irregularities.

For amplitude modulated signals, the gathered current  $i(t)$  is assumed to be multi-components and can be expressed as

$$i(t) = \sum_{k=1}^M a_k(t) \sin(\phi_k(t)) \quad (1)$$

where  $a_k(t) = a_k(1 + m_{ka} \sin(2\pi f_{ka} t + \phi_{ka}))$

and  $\phi_k(t) = 2\pi f_{k0} t + m_{kp} \sin(2\pi f_{ka} t + \phi_{kp})$

However, due to sampling procedure (1) is rewritten as follow.

$$i(n) = \sum_{k=1}^M a_k(n) \sin(2\pi f_k / F_s + \phi_k(n)) \quad (2)$$

Where  $n = 0 \dots N - 1$  is the sample index,  $N$  is the number of logged samples,  $\phi_k$  is the phase parameter and  $F_s$  is the sampling frequency.

For failure detection, a possible approach relies on the use of amplitude demodulation techniques to estimate the instantaneous amplitude ( $IA$ ). Then, statistical features can be extracted to detect if  $IA$  is time-varying or not.

### A. Amplitude Demodulation

For amplitude modulated signals, many techniques for amplitude demodulation were investigated. The most popular include the Hilbert transform (HT) [16] and the Teager Energy Operator [17]. Furthermore for three-phase system, it has been recently shown that the Concordia transform can be used to perform demodulation [18]. In this study, one phase current is considered. In this context, if the current is assumed to be mono-component, (2) is reduced to

$$i(n) = a(n) \sin(2\pi f / F_s + \phi(n)) \quad (3)$$

and the Hilbert transform can be chosen to estimate the instantaneous amplitude since it is usually more robust against noise than the Teager energy operator and easier to implement, because its computation is closely related to FFT which is the most built-in function in embedded targets.

However, the current is not really mono component, so Hilbert transform is no longer valid. Because It is well known that the stator current is a combination of various dominant harmonic components; such as fundamental harmonic, teeth harmonic, saturation harmonic, unknown harmonics including noise; and harmonics introduced by the failure. Under such assumption, innovative techniques are investigated for tracking the failure component by separation methods [19-20].

In this paper the authors explore a separation method in order to isolate the failure effect and track the variation of the dominant component introduced by this failure. One of the emerging methods for signal separation is the Hilbert Huang transform (HHT). The HHT method has focused considerable attention and has been recently indexed to fault diagnosis of rotating machinery [21].

The HHT method proceeded on two steps. The first one consists in decomposing the signal using the Empirical Mode Decomposition (EMD) method. The EMD has been described as an adaptive time-frequency data analysis method for nonlinear and non-stationary signals [20]. Unlike standard approaches that decompose a signal (data) into series of pre-defined basis functions (harmonic, wavelet), the EMD is derived from data. The EMD produces a representation of a discrete signal in terms of elementary modes based on the local characteristic time-scale of the signal and hence leading to physical meaning. The multi-components signal is expressed as a sum of a series of intrinsic mode functions (*imfs*) and can be expressed by

$$i(n) = \sum_{m=1}^M imf_m(n) + R_M(n) \quad (4)$$

The decomposition details can be found in [20]. Nevertheless, one major drawback of the EMD is the mode mixing. This phenomenon means that the detail related to one scale can appear in two different intrinsic modes as clearly shown by Fig. 2. The mixed mode makes the individual *imf* devoid of a physical meaning. To overcome this drawback, the Ensemble EMD was introduced by [22]. The EEMD is described as a new noise-added method, which automatically mitigate the EMD mode mixing. This *friendly-noise* decomposition is base on the EMD and is described by Fig. 3.

The second step relies on applying the Hilbert transform to each *imf* obtained through the EEMD algorithm. Since the *imfs* are discrete, it is necessary to use the Discrete Hilbert Transform (DHT) [23].

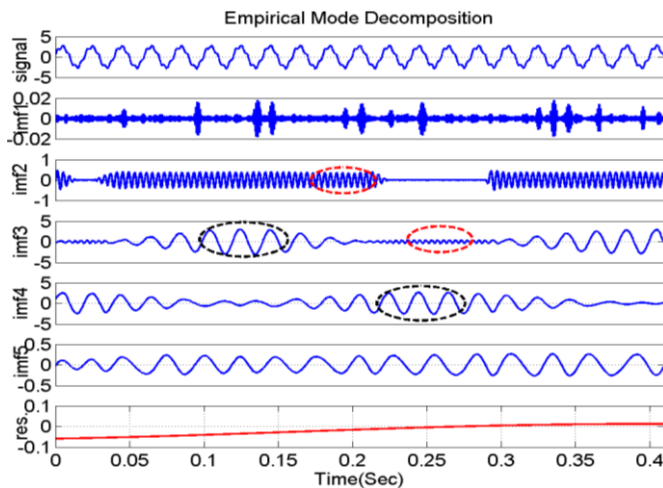


Fig. 2. Empirical mode decomposition and the mixed mode phenomena.

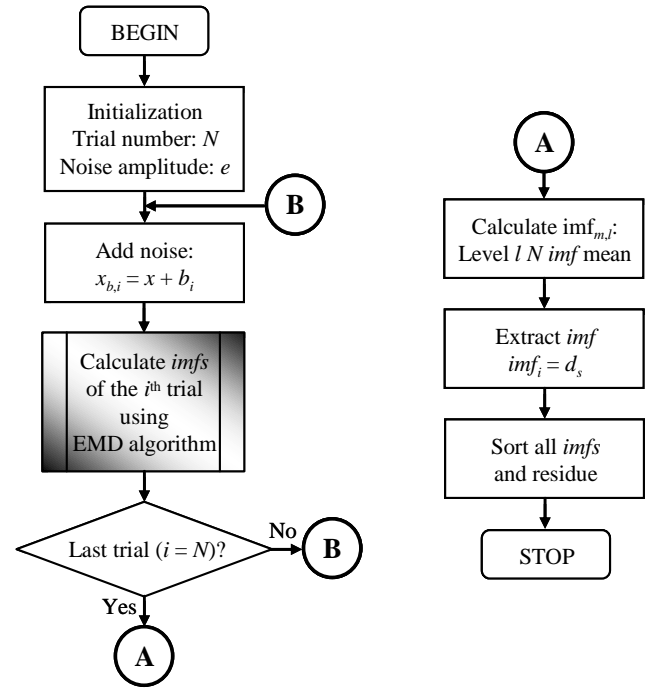


Fig. 3. EEMD algorithm flowchart.

$$H[imf_x(n)] = F^{-1} \{ F \{ imf_x(n) \} u(n) \} \quad (7)$$

where  $F\{\cdot\}$  and  $F^{-1}\{\cdot\}$  correspond to the FFT and Inverse FFT, respectively, and  $u(n)$  is defined as

$$u(n) = \begin{cases} 1, & n = 0, \frac{N}{2} \\ 2, & n = 1, 2, \dots, \frac{N}{2} - 1 \\ 0, & n = \frac{N}{2} + 1, \dots, N - 1 \end{cases} \quad (8)$$

Using (3), the instantaneous amplitude *IA*, denoted  $\hat{a}(n)$ , is given by

$$\hat{a}(n) = \sqrt{imf_x^2(n) + (H[imf_x(n)])^2} \quad (9)$$

### B. Failure Detector

Several fault detectors based on amplitude demodulation have been proposed in the literature, and most of them use complicated classifier [20]. Furthermore these methods assume that a training database is available, which can be difficult to obtain for wind turbines. In this section, we propose a low complexity detector which does not require any training set. The detector is based on the variance of the dominant *imf*. After applying EEMD and locating the most energized *imf* due to the failure occurrence, its *IA* is computed through (9). A statistical criterion is then applied to assess a failure indicator.

#### IV. TEST FACILITY DESCRIPTION

Figure 4 describes the experimental setup that is operated in the motor configuration for experimental easiness. It is composed of two parts: a mechanical part that has a tachogenerator, a three-phase induction motor and an alternator. The tachogenerator is a DC machine that generates 90 V at 3000 rpm. It is used to measure the speed. It produces linear voltage between 2500 and 3000 rpm. The alternator is a three-phase synchronous machine with a regulator and a rectifier circuit that stabilize the output voltage at 12 VDC. The advantage of using a car alternator instead of DC generator is obtaining constant output voltage at various speeds. The induction motor could be identically loaded at different speeds. Figure 5 illustrates the experimental test philosophy, while bearings with artificial deterioration are shown in Fig. 6. The induction generator and the bearings data and parameters are given in the Appendix.

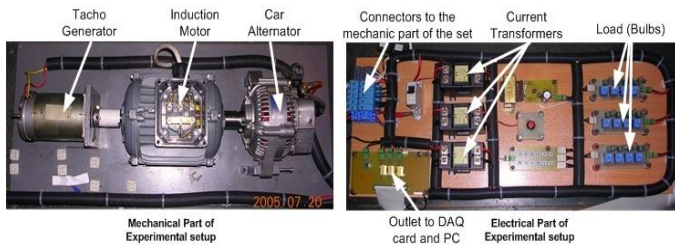


Fig. 4. Experimental set up.

#### V. FAILURE DETECTOR ASSESSMENT AND RESULTS

In this section, the results of the proposed approach are presented with experimental signals. The EEMD algorithm was adjusted for  $e = 0.3$  and  $N = 100$ . This decomposition was applied to logged stator current for several loads during operation with healthy and faulty bearing.

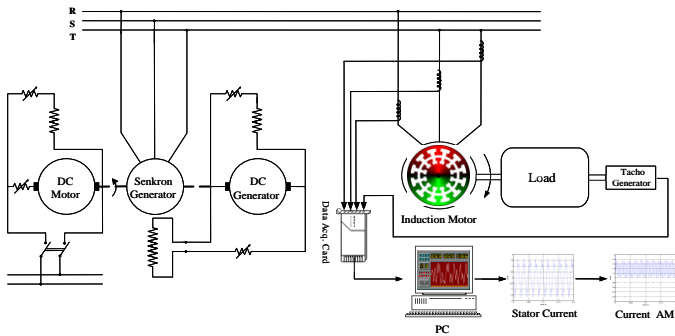


Fig. 5. Experimental test philosophy.

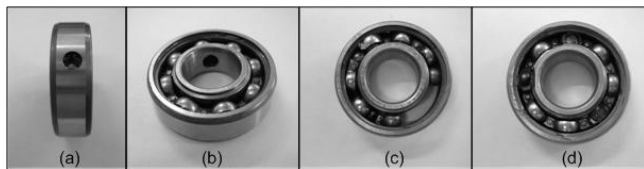
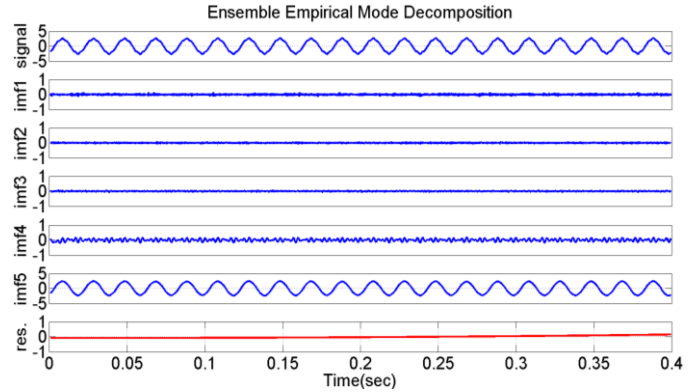
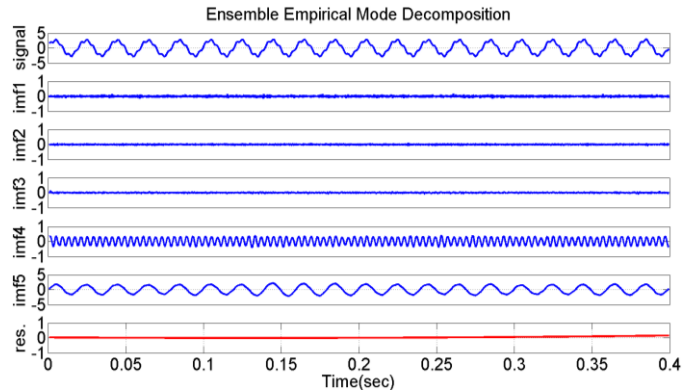


Fig. 6. Artificially deteriorated bearings: (a) outer race deterioration, (b) inner race deterioration, (c) cage deterioration, (d) ball deterioration [25].

Figure 7 depicts the first five *imfs* and the residue for a 40% loaded healthy and faulty machine. It can be seen that the 4<sup>th</sup> *imf* is more energized when the bearings are defected; whatever the bearing failure; except when it is defected by outer race, as shown in Fig. 8. In this case, the 4<sup>th</sup> *imf* seems to be no more different from the healthy one. This is mainly due a bad emulation of the failure. This *imf* can therefore be investigated for bearing failure detection. Owing to this ascertainment, the mean of the instantaneous amplitude of this *imf* for several loads and cases is therefore computed.



(a) Healthy bearings.



(b) Faulty bearings (cage damage).

Fig. 7. Stator current EEMD.

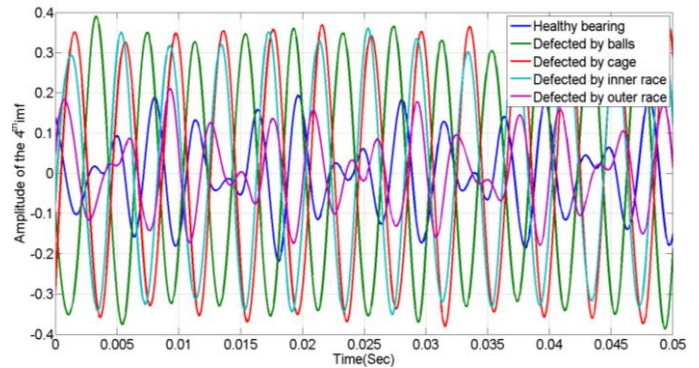


Fig. 8. 4<sup>th</sup> *imf* for healthy and faulty bearings.



After Hilbert transform,  $n_e$  samples have been removed at the beginning and at the end of the instantaneous amplitude of the 4<sup>th</sup> *imf*, to avoid the edge effects problem of the Hilbert transform. Figure 9 displays the instantaneous amplitude *IA* of the 4<sup>th</sup> *imf* for a healthy and faulty bearings. Readable information on failure detection performance using the *IA* mean are illustrated by the bar graph of Fig. 10. Compared to the healthy case, the *IA* mean is higher in the faulty case. In particular, this criterion is multiply by 3 for a bearing failure.

In this context, a bearing failure can be detected by setting the hypothesis test threshold to an adjusted value during normal conditions and operations.

## VI. CONCLUSION

This paper dealt with the assessment of a demodulation technique for bearing failure detection through the generator stator current in wind turbines context. The proposed technique is based on a modified version of the Hilbert Huang transform. In this version, the use of the EEMD algorithm allows overcoming the well-known mixed mode. In this context, the current was first decomposed into intrinsic mode functions through the EEMD.

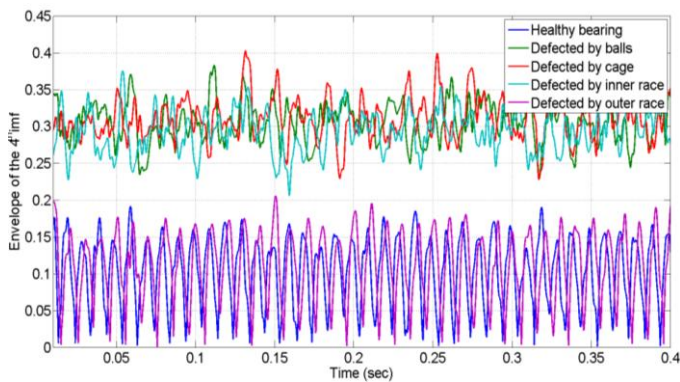


Fig. 9. Instantaneous amplitude of the 4<sup>th</sup> *imf* for healthy and faulty bearings.

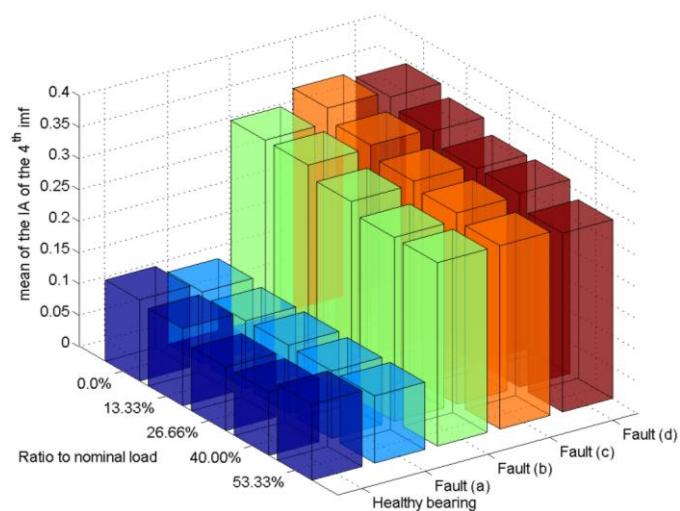


Fig 10. 4<sup>th</sup> *imf* instantaneous amplitude mean for healthy and faulty bearings.

It was then found that the 4<sup>th</sup> one is the most energized when bearing faults occur. The instantaneous amplitude of the 4<sup>th</sup> *imf* mode is then analyzed using a statistic criterion based on the mean value. The achieved results clearly show that it can be used as a reliable indicator for bearing failures regardless training data.

## APPENDIX

### INDUCTION GENERATOR AND BEARINGS PARAMETERS

Induction Generator	
Power rate	0.75KW
Voltage rate	220/3380V
Current rate	1.95/3.4A
Rate speed	2780 rpm
Frequency	50Hz
Bearings Parameters	
Type	6204.2ZR
Outside diameter	47mm
Inside diameter	20mm
Pitch diameter $D_p$	31.85mm
Number of balls $N$	8
Diameter of balls $D_B$	12mm

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