



# Circularly polarised planar antenna excited by coplanar waveguide feedline

Hakim Aïssat, Laurent Cirio, Marjorie Grzeskowiak, Jean-Marc Laheurte, Odile Picon

## ► To cite this version:

Hakim Aïssat, Laurent Cirio, Marjorie Grzeskowiak, Jean-Marc Laheurte, Odile Picon. Circularly polarised planar antenna excited by coplanar waveguide feedline. *Electronics Letters*, 2004, 40 (7), pp.402-403. 10.1049/el:20040276 . hal-00692958

**HAL Id: hal-00692958**

**<https://hal.science/hal-00692958>**

Submitted on 24 Jun 2019

**HAL** is a multi-disciplinary open access archive for the deposit and dissemination of scientific research documents, whether they are published or not. The documents may come from teaching and research institutions in France or abroad, or from public or private research centers.

L'archive ouverte pluridisciplinaire **HAL**, est destinée au dépôt et à la diffusion de documents scientifiques de niveau recherche, publiés ou non, émanant des établissements d'enseignement et de recherche français ou étrangers, des laboratoires publics ou privés.



## Open Archive Toulouse Archive Ouverte (OATAO)

OATAO is an open access repository that collects the work of some Toulouse researchers and makes it freely available over the web where possible.

This is an author's version published in: <https://oatao.univ-toulouse.fr/24014>

**Official URL :** <https://doi.org/10.1049/el:20040276>

### To cite this version :

Aïssat, Hakim and Cirio, Laurent and Grzeskowiak, Marjorie and Laheurte, Jean-Marc and Picon, Odile Circularly polarised planar antenna excited by coplanar waveguide feedline. (2004) Electronics Letters, 40 (7). 402-403. ISSN 0013-5194

Any correspondence concerning this service should be sent to the repository administrator:

[tech-oatao@listes-diff.inp-toulouse.fr](mailto:tech-oatao@listes-diff.inp-toulouse.fr)

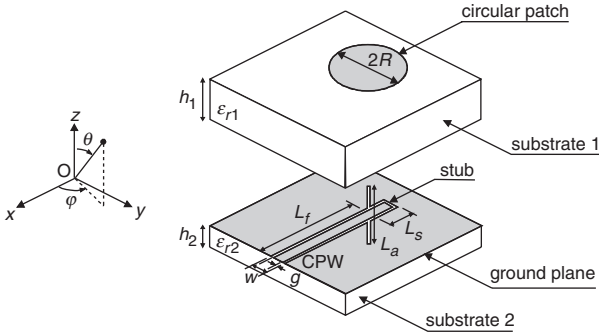
# Circularly polarised planar antenna excited by coplanar waveguide feedline

H. Aïssat, L. Cirio, M. Grzeskowiak, J.-M. Laheurte and O. Picon

A circular patch antenna fed by a coplanar waveguide (CPW) line is proposed to obtain circular polarisation (CP). The CP results from the combined excitation of the patch by an inclined slot (integrated in the CPW feedline) and the CPW feedline termination. The asymmetrical characteristics of the excitation involve the excitation of both the odd and the even modes in the CPW line.

**Introduction:** As the feedline and the ground plane are located on the same layer, CPW-fed antennas facilitate the integration of active devices without via hole connections (no backside processing). This feature makes the CPW-fed antenna a good radiating element for radio frequency identification (RFID) systems where the transponder is generally limited to the basic connection of a dedicated integrated circuit to the antenna. In various RFID applications, the relative positions of the transponder and the reader are unknown, which makes the use of CP highly advisable for a correct communication. This Letter presents a circularly polarised CPW-fed antenna dedicated to RFID applications and more specifically to dedicated short range communication (DSRC) at 5.8 GHz.

CP is often generated by geometrical perturbations on a printed element such as asymmetries [1], notches [2] or stubs [3]. These perturbations yield two orthogonal degenerate modes with linear polarisations of equal amplitude and  $90^\circ$  out of phase in a narrow bandwidth. In our work, the CP is obtained by combining two non-orthogonal modes independently excited by an inclined slot and the open termination end of the CPW line (Fig. 1). For this reason, the selected radiating element is a perfectly circular patch, which can generate a linear polarisation in any direction and also permit a symmetrical radiation pattern to be obtained. The correct phase difference and magnitude of the excited modes are obtained by tuning the various available parameters (length  $L_s$  of the line extension under the patch, length  $L_a$  of the inclined slot, thickness  $h_1$  of the substrate).



**Fig. 1** Geometry of circularly polarised microstrip antenna

$\epsilon_{r1}=2.2$  (DUROID 5880),  $h_1=3.175$  mm,  $R=9$  mm,  $\epsilon_{r2}=4.5$  (TMM4),  $h_2=1.524$  mm,  $w=3.5$  mm,  $g=0.5$  mm,  $L_s=5.5$  mm,  $L_a=20$  mm,  $L_f=30$  mm

**CP basics:** It is well known that the superposition of two orthogonal fields  $\vec{E}_x$  and  $\vec{E}_y$  given by:

$$\vec{E}_x = E_0 e^{j\pi/2} \vec{u}_x \quad (1)$$

$$\vec{E}_y = E_0 \vec{u}_y \quad (2)$$

in their phasor form results in a right-hand CP plane wave defined by:

$$\vec{E} = E_0 (\vec{u}_y + j\vec{u}_x) e^{-jkz} \quad (3)$$

Let us now consider two non-orthogonal linear field components  $\vec{E}_x$  and  $\vec{E}_\xi$  with an angle  $\xi = (\vec{E}_x, \vec{E}_\xi)$ :

$$\vec{E}_x = E_0 e^{j\phi} \vec{u}_x \quad (4)$$

$$\vec{E}_\xi = E_0 \vec{u}_\xi \quad (5)$$

To obtain a CP plane wave, it can be easily shown that the electrical phase  $\phi$  must comply with:

$$\tan \phi = -\tan \xi \quad (6)$$

In this case, the CP plane wave is given by:

$$\vec{E} = E_0 \sin \xi (\vec{u}_y + j\vec{u}_x) e^{-jkz} \quad (7)$$

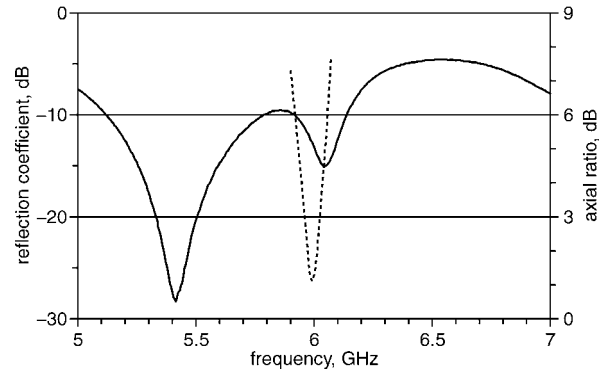
Note that, when  $\xi$  deviates from  $\pi/2$ , the amplitude of the radiated field is necessarily smaller than in the optimal  $\xi = \pi/2$  case.

**Antenna design:** The circular radiating patch is printed on the upper side of the first substrate; the ground plane, the CPW feedline and the  $45^\circ$ -inclined coupling slot are on the upper side of the second substrate. The relative permittivities and substrate thicknesses are  $\epsilon_{r1}=2.2$ ,  $\epsilon_{r2}=4.5$ ,  $h_1=3.175$  mm and  $h_2=1.524$  mm.  $\epsilon_{r1}$  has been chosen small to favour the patch radiation and the coupling between the patch and the slot.  $\epsilon_{r2}$  has been chosen high to reduce the size of the CPW feedline compared to the radiating element. The line is dimensioned using [4] for a  $50 \Omega$  characteristic impedance ( $w=3.5$  mm and  $g=0.5$  mm). The circular patch (radius  $R=9$  mm) is dimensioned using the cavity method [5] to resonate at 5.8 GHz.

The coupling slot, whose dimensions are given on Fig. 1, is centred below the patch with a  $-45^\circ$  inclination angle with respect to the  $x$ -axis. The electromagnetic energy is coupled from the CPW feedline to the circular patch through the coupling slot.

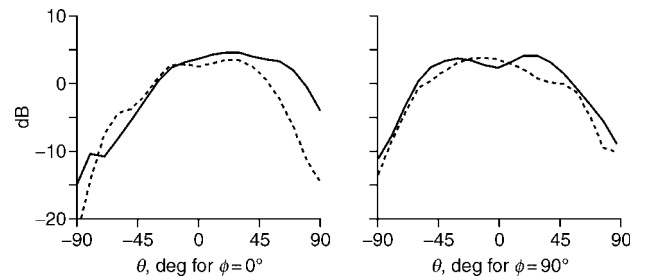
The use of a coupling slot integrated in the CPW feedline allows the appearance of two near-degenerate modes and right-hand circular polarisation (RHCP) radiation. Left-hand circular polarisation (LHCP) radiation can be obtained when a  $+45^\circ$  rotation angle is applied to the inclined cross-coupling slot.

**Simulations and experimental results:** The dimensions of the RHCP microstrip antenna have been optimised with the 2.5D commercial software Ensemble [6] for good matching at 5.8 GHz. The measured return loss ( $\Gamma$ ) is shown in Fig. 2. We observed that the operating frequency is shifted to approximately 6 GHz with  $\Gamma=-15$  dB. The measured impedance bandwidth ( $\Gamma < -10$  dB) is about 3.6%. The first resonance at 5.42 GHz corresponds to one of the two linear modes required to generate CP.



**Fig. 2** Measured reflection coefficient and axial ratio against frequency

— reflection coefficient  
--- axial ratio



**Fig. 3** Measured radiation patterns at 5.99 GHz

—  $E_\theta$   
---  $E_\phi$

The measured axial ratio (AR) is shown in Fig. 2. The RHCP bandwidth, determined for  $AR < 3$  dB, is 63 MHz ( $\sim 1\%$ ). The minimum AR is 1.1 dB at 5.99 GHz for a 5.9 dB gain. The measured

radiation pattern at 5.99 GHz is shown in Fig. 3 and a 3 dB AR aperture angle of  $\Delta\theta = 45^\circ$  has been measured in both planes around the z-axis.

**Conclusions:** A new compact CPW feed design for circularly polarised microstrip antennas has been proposed. It is based on the proper tuning of two coupling discontinuities (inclined slot and CPW end) in the feeding line. The  $\Gamma$  and AR bandwidth match most RFID requirements and the AR beamwidth is larger than in CP antennas based on geometrical perturbations. The large CP beamwidth makes the antenna a good candidate for identification applications.

**Acknowledgments:** The authors wish to thank L. Brochier from the LEAT (Nice, France) for his help in the antenna realisation and N. Berkoukchi from ESIGETEL (Fontainebleau, France) for his interest in this study.

H. Aïssat, L. Cirio, M. Grzeskowiak, J.-M. Laheurte and O. Picon (ESYCOM, Université de Marne-la-Vallée, 5 Bd Descartes, 77454 Marne-la-Vallée Cedex2, France)

E-mail: haissat@univ-mlv.fr

H. Aïssat: Also with ESIGETEL, 1 rue du Port de Valvins, 77215 Avon, Fontainebleau Cedex, France

## References

- 1 Boti, M., Dussopt, L., and Laheurte, J.-M.: 'Circularly polarised antenna with switchable polarisation sense', *Electron. Lett.*, 2000, **36**, (18), pp. 1518–1519
- 2 Huang, C.Y., and Wong, K.L.: 'Coplanar waveguide-fed circularly polarized microstrip antenna', *IEEE Trans. Microw. Theory Tech.*, 2000, **48**, pp. 328–329
- 3 Huang, C.Y.: 'A circularly polarized microstrip antenna using a coplanar-waveguide feed with an inset tuning stub', *Microw. Opt. Technol. Lett.*, 2001, **28**, pp. 311–312
- 4 Balanis, C.A.: 'Advanced engineering electromagnetics' (John Wiley & Sons, Inc., New York, 1989), pp. 492–499
- 5 Bedair, S.S., and Wolff, I.: 'Fast, accurate and simple approximate analytic formulas for calculating the parameters of supported coplanar waveguides for (M)MIC's', *IEEE Trans. Microw. Theory Tech.*, 1992, **40**, pp. 41–48
- 6 Ansoft Ensemble v8: 'Software based on the method of moments'