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Stability, stabilizability and exact controllability of a class of linear neutral type systems

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Abstract

Linear systems of neutral type are considered using the infinite dimensional approach. The main problems are asymptotic, non-exponential stability, exact controllability and regular asymptotic stabilizability. The main tools are the moment problem approach, the Riesz basis of invariant subspaces and the Riesz basis of family of exponentials.

1 Introduction

Many applied problems from physics, mechanics, biology, and other fields can be described by delay differential equations. A large class of such systems are systems of neutral type. In this paper, we consider a general class of neutral systems with distributed delays given by the equation

$$\begin{cases} \frac{d}{dt}[z(t) - Kz_t] = Lz_t + Bu(t), & t \geq 0, \\ z_0 = \varphi, \end{cases} \quad (1)$$

where $z_t : [-1, 0] \rightarrow \mathbb{C}^n$ is the history of z defined by $z_t(s) = z(t + s)$. The difference and delay operators K and L , respectively, are defined by

$$Kf = A_{-1}f(-1) \quad \text{and} \quad Lf = \int_{-1}^0 A_2(\theta) \frac{d}{d\theta} f(\theta) d\theta + \int_{-1}^0 A_3(\theta) f(\theta) d\theta$$

for $f \in H^1([-1, 0], \mathbb{C}^n)$, where A_{-1} is a constant $n \times n$ matrix, A_2, A_3 are $n \times n$ matrices whose elements belong to $L_2(-1, 0)$, and B is a constant $n \times r$ matrix.

A more general case may be when

$$Kf = \int_{-1}^0 d\mu(\theta) f(\theta), \quad f \in C([-1, 0], \mathbb{C}^n),$$

where μ is a matrix-valued function of bounded variation and continuous at zero. But we limit ourselves to the above mentioned case when $Kf = A_{-1}f(-1)$ because that the properties of systems are mainly characterized by the structure of the matrix A_{-1} . For more general forms of the operator K it may be expected that the situation is analogous, but it is not clear how the properties studied here may be connected. Anyway, it is a domain for further investigation. Distributed delay may arise in the natural modeling or after some feedback. Our purpose is to investigate the problems of asymptotic stability, of stabilizability by linear feedback and of exact controllability.

For that problems, the neutral type systems are less studied than the retarded systems when $K = 0$. The difficulties are related to the following particular properties of neutral type systems: there may exist an infinite number of eigenvalues in the right half plane, in particular near the imaginary axis; the choice of the phase-space is crucial, in contrast to the case of retarded functional differential equations where solutions are more smooth than the initial data; some feedback may change the structure of the system, etc.

2 The operator model

In [13] and several other works, the framework is based on the description of neutral type systems in the space of continuous functions $C([-1, 0]; \mathbb{C}^n)$. However, for several control problems, the Hilbert space structure is more convenient in the study of our class of systems. In Hilbert spaces one can use the fundamental tool of Riesz basis (or orthonormal basis modulo a bounded isomorphism). We consider the operator model of neutral type systems introduced by Burns and al. in product spaces. This approach was also used in [41] for the construction of a spectral model. In [44] the authors consider the particular case of discrete delay, which served as a model in [28, 30] to characterize the stabilizability of a class of systems of neutral type.

The state space is $M_2(-1, 0; \mathbb{C}^n) = \mathbb{C}^n \times L_2(-1, 0; \mathbb{C}^n)$, briefly M_2 , and permits (1) to be rewritten as

$$\frac{d}{dt}x(t) = \mathcal{A}x(t) + \mathcal{B}u(t), \quad x(t) = \begin{pmatrix} y(t) \\ z_t(\cdot) \end{pmatrix}, \quad (2)$$

where the operators \mathcal{A} and \mathcal{B} are defined by

$$\mathcal{A} \begin{pmatrix} y(t) \\ z_t(\cdot) \end{pmatrix} = \begin{pmatrix} \int_{-1}^0 A_2(\theta) \dot{z}_t(\theta) d\theta + \int_{-1}^0 A_3(\theta) z_t(\theta) d\theta \\ dz_t(\theta)/d\theta \end{pmatrix}, \quad \mathcal{B}u = \begin{pmatrix} Bu \\ 0 \end{pmatrix} \quad (3)$$

The domain of \mathcal{A} is given by

$$\mathcal{D}(\mathcal{A}) = \{(y, z(\cdot)) : z \in H^1(-1, 0; \mathbb{C}^n), y = z(0) - A_{-1}z(-1)\} \subset M_2$$

and the operator \mathcal{A} is the infinitesimal generator of a C_0 -semigroup e^{At} . The relation between the solutions of the delay system (1) and the system (2) is $z_t(\theta) = z(t + \theta)$.

In the particular case when $A_2(\theta) = A_3(\theta) = 0$, we use the notation $\tilde{\mathcal{A}}$ for \mathcal{A} . The properties of $\tilde{\mathcal{A}}$ can be expressed mainly in terms of the properties of matrix A_{-1} only. Some important properties of \mathcal{A} are close to those of $\tilde{\mathcal{A}}$.

3 Spectral analysis

Let us denote by μ_1, \dots, μ_ℓ , $\mu_i \neq \mu_j$ if $i \neq j$, the eigenvalues of A_{-1} and the dimensions of their rootspaces (generalized eigenspaces) by p_1, \dots, p_ℓ , $\sum_{k=1}^\ell p_k = n$. Consider the points $\lambda_m^{(k)} \equiv \ln |\mu_m| + i(\arg \mu_m + 2\pi k)$, $m = 1, \dots, \ell$; $k \in \mathbb{Z}$ and the circles $L_m^{(k)}$ of fixed radius $r \leq r_0 \equiv \frac{1}{3} \min\{|\lambda_m^{(k)} - \lambda_i^{(j)}|, (m, k) \neq (i, j)\}$ centered at $\lambda_m^{(k)}$.

Theorem 3.1. *The spectrum of \mathcal{A} consists of the eigenvalues only which are the roots of the equation $\det \Delta(\lambda) = 0$, where*

$$\Delta_{\mathcal{A}}(\lambda) = \Delta(\lambda) \equiv -\lambda I + \lambda e^{-\lambda} A_{-1} + \lambda \int_{-1}^0 e^{\lambda s} A_2(s) ds + \int_{-1}^0 e^{\lambda s} A_3(s) ds. \quad (4)$$

The corresponding eigenvectors of \mathcal{A} are $\varphi = (C - e^{-\lambda} A_{-1} C, e^{\lambda \theta} C)$, with $C \in \text{Ker } \Delta(\lambda)$.

There exists N_1 such that for any $|k| \geq N_1$, the total multiplicity of the roots of the equation $\det \Delta(\lambda) = 0$, contained in the circle $L_m^{(k)}$, equals p_m .

The description of the location of the spectrum of \mathcal{A} we use Rouché theorem.

3.1 Basis of invariant subspaces

The most desired situation for concrete systems is to have a Riesz basis formed by eigenvectors of A or, at least, by generalized eigenvectors. In more general situations, one studies the existence of bases formed by subspaces. We remind that a sequence of nonzero subspaces $\{V_k\}_i^\infty$ of the space V is called basis (of subspaces) of the space V , if any vector $x \in V$ can be uniquely presented as $x = \sum_{k=1}^\infty x_k$, where $x_k \in V_k$, $k = 1, 2, \dots$. We say that the basis $\{V_k\}_i^\infty$ is orthogonal if V_i is orthogonal to V_j when $i \neq j$. A basis $\{V_k\}$ of subspaces is called a Riesz basis if there are an orthogonal basis of subspaces $\{W_k\}$ and a linear bounded invertible operator R , such that $RV_k = W_k$.

The best "candidates" to form the basis of subspaces are generalized eigenspaces of the generator of a semigroup, but there are simple examples (see Example 3.3 below) showing that generalized eigenspaces do not form such a basis in the general case.

One of the main ideas of our approach is to construct a Riesz basis of finite-dimensional subspaces which are invariant for the generator of the semigroup (see (2)).

In [31, 32] we obtained the following general result.

Theorem 3.2. *There exists a sequence of invariant for \mathcal{A} (see (2)) finite-dimensional subspaces which constitute a Riesz basis in M_2 .*

More precisely, these subspaces are $\{V_m^{(k)}, |k| \geq N, m = 1, \dots, \ell\}$ and a $2(N+1)n$ -dimensional subspace spanned by all eigen- and rootvectors, corresponding to all eigenvalues of \mathcal{A} , which are outside of all circles $L_m^{(k)}, |k| \geq N, m = 1, \dots, \ell$.

Here $V_m^{(k)} \equiv P_m^{(k)} M_2$, where

$$P_m^{(k)} M_2 = \frac{1}{2\pi i} \int_{L_m^{(k)}} R(\mathcal{A}, \lambda) d\lambda$$

are spectral projectors; $L_m^{(k)}$ are circles defined before.

We emphasize that the operator \mathcal{A} may not possess in a Riesz basis of generalized eigenspaces. We illustrate this on the following

Example 3.3. Consider the particular case of the system (1):

$$\dot{x}(t) = A_{-1}\dot{x}(t-1) + A_0x(t), \quad A_{-1} = \begin{pmatrix} 1 & 1 \\ 0 & 1 \end{pmatrix}, \quad A_0 = \begin{pmatrix} \alpha & 0 \\ 0 & \beta \end{pmatrix}. \quad (5)$$

One can check that the characteristic equation is $\det \Delta(\lambda) = (\alpha - \lambda + \lambda e^{-\lambda})(\beta - \lambda + \lambda e^{-\lambda}) = 0$ and for $\alpha \neq \beta$ there are two sequences of eigenvectors, such that $\|v_n^1 - v_n^2\| \rightarrow 0$, as $n \rightarrow \infty$. It is clear that such family vectors do not form a Riesz basis.

4 Stability

By stability we mean here asymptotic stability. For our neutral type system, as for several infinite dimensional systems, we have essentially two notions of asymptotic stability : exponential (or uniform) stability and strong stability.

Definition 4.1. *A linear system in a Banach space \mathcal{X} is exponentially stable if the e^{At} semigroup verifies: $\exists M_\omega > 1, \exists \omega > 0, \forall x, \|e^{At}x\| \leq M_\omega e^{-\omega t} \|x\|$. The system is strongly stable if $\forall x, \|e^{At}x\| \rightarrow 0$, as $t \rightarrow \infty$.*

The problem of exponential stability was widely described in several classical works. An sufficiently exhaustive analysis may be found in [39] (see also the references therein and the bibliographic notes). In our case the exponential stability is completely determined by the spectrum of the operator \mathcal{A} . It is a well known result for some linear neutral type systems: the spectrum has to be bounded away from the imaginary axis (cf. [14, Theorem 6.1]).

Theorem 4.2. *The system (2) is exponentially stable if and only if $\sigma(\mathcal{A}) \subset \{\lambda : \operatorname{Re} \lambda \leq -\alpha < 0\}$.*

We can partially reformulate in terms of the matrix A_{-1} the condition on the spectrum $\sigma(\mathcal{A})$.

Theorem 4.3. *System (2) is exponentially stable if and only if the following conditions are verified*

- i) $\sigma(\mathcal{A}) \subset \{\lambda : \operatorname{Re} \lambda < 0\}$
- ii) $\sigma(A_{-1}) \subset \{\lambda : |\lambda| < 1\}$.

It can be interesting how the condition ii) of Theorem 4.3 may be formulated for the case of a general linear operator K in the system (1).

We would like to study more deeply the problem of asymptotic nonexponential stability. To this end, we recall some important abstract result in this domain. We have the following

Theorem 4.4. *Let $e^{At}, t \geq 0$ be a C_0 -semigroup in the Banach space X and A be the infinitesimal generator of the semigroup. Assume that $(\sigma(\mathcal{A}) \cap (i\mathbb{R}))$ is at most countable and the operator \mathcal{A}^* has no pure imaginary eigenvalues. Then e^{At} is strongly asymptotically stable (i.e. $e^{At}x \rightarrow 0, t \rightarrow +\infty$ as $x \in X$) if and only if one of the following conditions is valid:*

- i) There exists a norm $\|\cdot\|_1$, equivalent to the initial one $\|\cdot\|$, such that the semigroup e^{At} is contractive according to this norm: $\|e^{At}x\|_1 \leq \|x\|_1$, $\forall x \in X, t \geq 0$;
- ii) The semigroup e^{At} is uniformly bounded: $\exists C > 0$ such that $\|e^{At}\| \leq C$, $t \geq 0$.

The Theorem 4.4 was obtained initially in [38] for a bounded operator \mathcal{A} . The main idea were later used in [20] for the case of unbounded operator \mathcal{A} , see also [2] for another approach. The proof in [20] follow the scheme of the first result [38]. The development of this theory concerns a large class of differential equations in Banach space (see [39] and references therein). A more general result on the asymptotic behavior of the semigroup with respect to an arbitrary asymptotic was recently obtained in [35].

Our main result on the asymptotic stability of the neutral type system (1)–(2) is the following one.

Theorem 4.5. *Assume $\sigma(\mathcal{A}) \subset \{\lambda : \operatorname{Re}\lambda < 0\}$ and $\sigma_1 \equiv \sigma(A_{-1}) \cap \{\lambda : |\lambda| = 1\} \neq \emptyset$. Then the following three mutually exclusive possibilities exist:*

- i) *the part of the spectrum σ_1 consists of simple eigenvalues only, i.e. to each eigenvalue corresponds a one-dimensional eigenspace and there are no rootvectors. In this case system (2) is asymptotically stable.*
- ii) *the matrix A_{-1} has a Jordan block, corresponding to $\mu \in \sigma_1$. In this case system (2) is unstable.*
- iii) *there are no Jordan blocks, corresponding to eigenvalues in σ_1 , but there exists $\mu \in \sigma_1$ whose eigenspace is at least two-dimensional. In this case system (2) can be stable as well as unstable. Moreover, there exist two systems with the same spectrum, such that one of them is stable while the other one is unstable.*

The last case may be illustrated by a non trivial example (see also [32] for an example given partially in the M_2 -space framework).

Example 4.6. (Rabah-Sklyar-Barkhaev [29]) Consider the system

$$\dot{z}(t) - A_{-1}\dot{z}(t-1) = A_0z(t)$$

with

$$A_{-1} = \begin{pmatrix} -1 & 0 \\ 0 & -1 \end{pmatrix}, \quad A_0 = \begin{pmatrix} -1 & \gamma \\ 0 & -1 \end{pmatrix}, \quad \gamma = 0 \quad \text{or} \quad 1.$$

We have: $\sigma(\mathcal{A}) = \{\lambda : \lambda e^\lambda + \lambda + e^\lambda = 0\}$ in \mathbb{C}^- , this can be proved by Pontriaguin Theorem [27]. The multiplicity of eigenvalues is clearly 2, and they do not depend of γ . The system is stable for $\gamma = 0$ and unstable for $\gamma \neq 0$.

5 Stabilizability

We say that the system (2) is stabilizable if there exists a linear feedback control $u(t) = F(z_t(\cdot)) = F(z(t+\cdot))$ such that the system (2) becomes asymptotically stable.

It is obvious that for linear systems in finite dimensional spaces the linearity of the feedback implies that the control is bounded in every neighbourhood of the origin. For infinite dimensional spaces the situation is much more complicated. The boundedness of the feedback law $u = F(z_t(\cdot))$ depends on the topology of the state space.

When the asymptotic stabilizability is achieved by a feedback law which does not change the state space and is bounded with respect to the topology of the state space, then we call it *regular* asymptotic stabilizability. Under our assumption on the state space, namely $H^1([-1, 0], \mathbb{C}^n)$, the natural linear feedback is

$$Fz(t + \cdot) = \int_{-1}^0 F_2(\theta) \dot{z}(t + \theta) dt + \int_{-1}^0 F_3(\theta) z(t + \theta) dt, \quad (6)$$

where $F_2(\cdot), F_3(\cdot) \in L_2(-1, 0; \mathbb{C}^n)$.

Several authors (see for example [49, 25, 26, 47] and references therein) use feedback laws which for our system may take the form

$$\sum_{i=1}^k F_i \dot{z}(t - h_i) + \int_{-1}^0 F_2(\theta) \dot{z}(t + \theta) dt + \int_{-1}^0 F_3(\theta) z(t + \theta) dt. \quad (7)$$

This feedback law is not bounded in $H^1([-1, 0], \mathbb{C}^n)$ and then stabilizability is not regular. As a counterpart, they obtain exponentially stable closed loop systems. If the original system is not formally stable (see [48]), i.e. the pure neutral part (when $A_2 = A_3 = 0$) is not stable, the non regular feedback (6) is necessary to stabilize. From the operator point of view, the regular feedback law (6) means a perturbation of the infinitesimal generator \mathcal{A} by the operator \mathcal{BF} which is relatively \mathcal{A} -bounded (cf. [17]) and verifies $\mathcal{D}(\mathcal{A}) = \mathcal{D}(\mathcal{A} + \mathcal{BF})$. Such a perturbation does not mean, in general, that $\mathcal{A} + \mathcal{BF}$ is the infinitesimal generator of a C_0 -semigroup. However, in our case, this fact is verified directly [32, 46] since after the feedback we get also a neutral type system like (1) with $\mathcal{D}(\mathcal{A}) = \mathcal{D}(\mathcal{A} + \mathcal{BF})$ (see below for more details).

From a physical point of view, \mathcal{A} -boundedness of the stabilizing feedback \mathcal{F} means that the energy added by the feedback remains uniformly bounded in every neighbourhood of 0 (see also another point of view in [48]). Hence the problem of *regular* asymptotic stabilizability for the systems (1),(2) is to find a linear relatively \mathcal{A} -bounded feedback $u = \mathcal{F}x$ such that the operator $\mathcal{A} + \mathcal{BF}$ generates a C_0 -semigroup $e^{(\mathcal{A} + \mathcal{BF})t}$ with $\mathcal{D}(\mathcal{A} + \mathcal{BF}) = \mathcal{D}(\mathcal{A})$ and for which $\|e^{(\mathcal{A} + \mathcal{BF})t}x\| \rightarrow 0$, as $t \rightarrow \infty$ for all $x \in \mathcal{D}(\mathcal{A})$. The main contribution of this paper is that under some controllability conditions on the unstable poles of the system, we can assign arbitrarily the eigenvalues of the closed loop system into circles centered at the unstable eigenvalues of the operator \mathcal{A} with radii r_k such that $\sum r_k^2 < \infty$. This is, in some sense, a generalization of the classical pole assignment problem in finite dimensional space. Precisely we have the following

Theorem 5.1. *Consider the system (1) under the following assumptions:*

- 1) *All the eigenvalues of the matrix A_{-1} satisfy $|\mu| \leq 1$.*
- 2) *All the eigenvalues $\mu_j \in \sigma_1 \stackrel{\text{def}}{=} \sigma(A_{-1}) \cap \{z : |z| = 1\}$ are simple (we denote their index $j \in I$).*

Then the system (1) is regularly asymptotic stabilizable if

- 3) $\text{rank}(\Delta_{\mathcal{A}}(\lambda) \quad B) = n$ for all $\text{Re}\lambda \geq 0$, where

$$\Delta_{\mathcal{A}}(\lambda) = -\lambda I + \lambda e^{-\lambda} A_{-1} + \lambda \int_{-1}^0 e^{\lambda s} A_2(s) ds + \int_{-1}^0 e^{\lambda s} A_3(s) ds,$$

- 4) $\text{rank}(\mu I - A_{-1} \quad B) = n$ for all $|\mu| = 1$.

6 Exact Controllability

The problem of controllability for delay systems was considered by several authors in different framework. One approach is based on the analysis of time delay system in a module framework (space over ring, see [22]). In this case the controllability problem is considered in a formal way using different interpretations of the Kalman rank condition. Another approach is based on the analysis of time delay systems in vector spaces with finite or infinite dimension. A powerful tool is to consider a delay system as a system in a Banach functional space, this approach was developed widely in [13]. Because the state space for delay systems is a functional space, the most important notion is the function space controllability. A first important contribution in the characterization of null functional controllability was given by Olbrot [23] by using some finite dimensional tools as (A, B) -invariant subspaces for an extended system. For retarded systems one can refer to [21] (and references therein) for the analysis of function space controllability in abstract Banach spaces. The case of neutral type systems with discrete delay was also considered in such a framework (see O'Connor and Tarn [25] and references therein). A general analysis of the time delay systems in infinite dimensional spaces is given in the book [6] where several methods and references are given.

The problem considered in this paper is close to that studied in [25]. In this work the exact controllability problem was considered for neutral type systems with discrete delay using a semigroup approach in Sobolev spaces $W_2^{(1)}$ and a boundary control problem.

We consider the problem of controllability for distributed delay system of neutral type in the space $M_2(-h, 0; \mathbb{C}^n) = \mathbb{C}^n \times L_2(-h, 0; \mathbb{C}^n)$ which is natural for control problems.

The semigroup theory developed here is based on the Hilbert space model introduced in [8]. One of our result is a generalization of the result in [25]. The main non trivial precision is the time of controllability. We generalize the results given [16] for the case of a single input and one localized delay (see also [4, 34]). The approach developed here is different from that of [25]. Our main results are based on the characterization of controllability as a moment problem and using some recent results on the solvability of this problem (see [3] for the main tools used here). Using a precise Riesz basis in the space $M_2(-h, 0; \mathbb{C}^n)$ we can give a characterization of null-controllability and of the minimal time of controllability.

The reachability set at time T is defined by

$$\mathcal{R}_T = \left\{ \int_0^T e^{At} \mathcal{B}u(t) dt : u(\cdot) \in L_2(0, T; \mathbb{C}^n) \right\}$$

It is easy to show that $\mathcal{R}_{T_1} \subset \mathcal{R}_{T_2}$ as $T_1 < T_2$. An important result is that $\mathcal{R}_T \subset \mathcal{D}(\mathcal{A}) \subset M_2$. This non-trivial fact permits to formulate the null-controllability problem in the following setting:

- i) To find maximal possible set \mathcal{R}_T (depending on T);
- ii) To find minimal T for which the set \mathcal{R}_T becomes maximal possible, i.e. $\mathcal{R}_T = \mathcal{D}(\mathcal{A})$.

Definition 6.1. The system (2) is said null-controllable at the time T if $\mathcal{R}_T = \mathcal{D}(\mathcal{A})$

The main tool is to consider the null-controllability problem as a problem of moments.

6.1 The moment problem

In order to formulate the moment problem we need a Riesz basis in the Hilbert space M_2 . We recall that a Riesz basis is a basis which may be transformed to an orthogonal basis with respect to another equivalent scalar product. Each Riesz basis possesses a biorthogonal basis. Let $\{\varphi\}$ be a Riesz basis in M_2 and $\{\psi\}$ the corresponding biorthogonal basis. Then for each $x \in M_2$ we have $x = \sum_{\varphi \in \{\varphi\}} \langle x, \psi \rangle \varphi$. In a separable Hilbert space there always exists a Riesz basis.

A state $x = \begin{pmatrix} y \\ z(\cdot) \end{pmatrix} \in M_2$ is reachable at time T by a control $u(\cdot) \in L_2(0, T; \mathbb{C}^r)$ iff the steering condition

$$x = \begin{pmatrix} y \\ z(\cdot) \end{pmatrix} = \int_0^T e^{At} \mathcal{B}u(t) dt. \quad (8)$$

holds. This steering condition may be expanded using the basis $\{\varphi\}$. A state x is reachable iff

$$\sum_{\varphi \in \{\varphi\}} \langle x, \psi \rangle \varphi = \sum_{\varphi \in \{\varphi\}} \int_0^T \langle e^{At} \mathcal{B}u(t), \psi \rangle dt \varphi,$$

for some $u(\cdot) \in L_2(-h, 0; \mathbb{R}^r)$. Then the steering condition (8) can be substituted by the following system of equalities

$$\langle x, \psi \rangle = \int_0^T \langle e^{At} \mathcal{B}u(t), \psi \rangle dt, \quad \psi \in \{\psi\}. \quad (9)$$

Let $\{b_1, \dots, b_r\}$ be an arbitrary basis in $\text{Im}B$, the image of the matrix B and $\mathbf{b}_i = \begin{pmatrix} b_i \\ 0 \end{pmatrix} \in M_2$, $i = 1, \dots, r$. Then the right hand side of (9) takes the form

$$\int_0^T \langle e^{At} \mathcal{B}u(t), \psi \rangle dt = \sum_{i=1}^r \int_0^T \langle e^{At} \mathbf{b}_i, \psi \rangle u_i(t) dt. \quad (10)$$

Effectiveness of the proposed approach becomes obvious if we assume that the operator \mathcal{A} possess a Riesz basis of eigenvector. This situation is characteristic, for example, for control systems of hyperbolic type when \mathcal{A} is skew-adjoint

($\mathcal{A}^* = -\mathcal{A}$) and has a compact resolvent (see, for example, [1], [16], [17]). Let in this case $\{\varphi_k\}$, $k \in \mathbb{N}$, be a orthonormal eigenbasis with $\mathcal{A}\varphi_k = i\lambda_k\varphi_k$, $\lambda_k \in \mathbb{R}$. Assuming for simplicity $r = 1$, $b_1 = b = \sum_k \alpha_k\varphi_k$, $\alpha_k \neq 0$, we have from (4), (5)

$$\frac{x_k}{\alpha_k} = \int_0^T e^{-i\lambda_k t} u(t) dt, \quad k \in \mathbb{N}, \quad (11)$$

where $x = \sum_k x_k\varphi_k$. Equalities (6) are a non-Fourier trigonometric moment problem whose solvability is closely connected with the property for the family of exponentials $e^{-i\lambda_k t}$, $k \in \mathbb{N}$, to form a Riesz basis on the interval $[0, T]$ ([1]). In particular, if $e^{-i\lambda_k t}$ forms a Riesz basis of $L_2[0, T_0]$ then one has

$$\mathcal{R}_T = \left\{ x : \sum_k \left(\frac{x_k}{\alpha_k} \right)^2 < \infty \right\} \quad \text{for all } T \geq T_0. \quad (12)$$

Obviously formula (12) gives the complete answer to the both items of the controllability problem. Returning now to neutral type systems we observe that the operator \mathcal{A} given in is not skew-adjoint and, moreover, does not possess a basis even of generalized eigenvectors. So the choice of a proper Riesz basis to transform the steering condition in a moment problem is an essentially more complicated problem.

6.2 The choice of basis

In order to design the needed basis for our case we use spectral the properties of the operator \mathcal{A} obtained in [32]. Let μ_1, \dots, μ_ℓ , $\mu_i \neq \mu_j$ be eigenvalues of A_{-1} and let the integers p_m be defined as : $\dim(A_{-1} - \mu_m I)^n = p_m$, $m = 1, \dots, \ell$. Denote by

$$\lambda_m^{(k)} = \frac{1}{h} (\ln |\mu_m| + i(\arg \mu_m + 2\pi k)), \quad m = 1, \dots, \ell; \quad k \in \mathbb{Z},$$

and let $L_m^{(k)}$ be the circles of the fixed radius $r \leq r_0 = \frac{1}{3} \min |\lambda_m^{(k)} - \lambda_i^{(j)}|$ centered at $\lambda_m^{(k)}$.

Let $\{V_m^{(k)}\}_{\substack{k \in \mathbb{Z} \\ m=1, \dots, \ell}}$ be a family of \mathcal{A} -invariant subspaces given by

$$V_m^{(k)} = P_m^{(k)} M_2, \quad P_m^{(k)} = \frac{1}{2\pi i} \int_{L_m^{(k)}} R(\mathcal{A}, \lambda) d\lambda.$$

The following theorem plays an essential role in our approach

Theorem 6.2. [31, 32] *There exists N_0 large enough such that for any $N \geq N_0$*

- i) $\dim V_m^{(k)} = p_m$, $k \geq N$,*
- ii) the family $\{V_m^{(k)}\}_{\substack{|k| \geq N \\ m=1, \dots, \ell}} \cup \widehat{V}_N$ forms a Riesz basis (of subspaces) in M_2 , where \widehat{V}_N is a finite-dimensional subspace ($\dim \widehat{V}_N = 2(N+1)n$) spanned by all generalized eigenvectors corresponding to all eigenvalues of \mathcal{A} located outside of all circles $L_m^{(k)}$, $|k| \geq N$, $m = 1, \dots, \ell$.*

Using this theorem we construct a Riesz basis $\{\varphi\}$ of the form

$$\{\varphi_{m,j}^k, |k| > N; m = 1, \dots, \ell; j = 1, \dots, p_m\} \cup \{\hat{\varphi}_j^N, j = 1, \dots, 2(N+1)n\}$$

where for any $m = 1, \dots, l$, and $k : |k| > N$ the collection $\{\varphi_{m,j}^k\}_{j=1, \dots, p_m}$ is in a special way chosen basis of $V_m^{(k)}$ and $\{\hat{\varphi}_j^N\}_{j=1, \dots, 2(N+1)n}$ is a basis of \hat{V}_N . In this basis equalities (4) with regard to (5) turns into a moment problem with respect to a special collection of quasipolynomials. Analyzing the mentioned moment problem by means of the methods given in [1] we obtain our main results concerning the null-controllability problem.

7 The main results

The characterization of the null-controllability is given by the following Theorem.

Theorem 7.1. *The system (2) is null-controllable by controls from $L_2(0, T)$ for some $T > 0$ iff the following two conditions hold:*

i) $\text{rank} [\Delta_{\mathcal{A}}(\lambda) \quad B] = n, \quad \forall \lambda \in \mathbb{C}$; where

$$\Delta_{\mathcal{A}}(\lambda) = -\lambda I + \lambda e^{-\lambda h} A_{-1} + \lambda \int_{-h}^0 e^{\lambda s} A_2(s) ds + \int_{-h}^0 e^{\lambda s} A_3(s) ds.$$

ii) $\text{rank} [B \quad A_{-1}B \quad \dots \quad A_{-1}^{n-1}B] = n$.

The main results on the time of controllability are as follows.

Theorem 7.2. *Let the conditions i) and ii) of Theorem 7.1 hold. Then*

i) The system (2) is null-controllable at the time T as $T > nh$;

ii) If the system (2) is of single control ($r = 1$), then the estimation of the time of controllability in i) is exact, i.e. the system is not controllable at time $T = nh$.

For the multivariable case, the time depends on some controllability indices. suppose that $\dim B = r$. Let $\{b_1, \dots, b_r\}$ be an arbitrary basis noted β . Let us introduce a set integers. We denote by $B_i = (b_{i+1}, \dots, b_r)$, $i = 0, 1, \dots, r-1$, which gives in particular $B_0 = B$ and $B_{r-1} = b_r$ and we put formally $B_r = 0$. Let us consider the integers

$$n_i^\beta = \text{rank} [B_{i-1} \quad A_{-1}B_{i-1} \quad \dots \quad A_{-1}^{n-1}B_{i-1}], \quad i = 1, \dots, r,$$

corresponding to the basis β . We need in fact the integers

$$m_i^\beta = n_{i-1}^\beta - n_i^\beta,$$

Let us denote by

$$m_{\min} = \max_{\beta} m_1^\beta \quad m_{\max} = \min_{\beta} \max_i m_i^\beta,$$

for all possible choice of a basis β .

The main result for the multivariable case is the following Theorem.

Theorem 7.3. *Let the conditions i) and ii) of the Theorem 7.1 hold, then*

i) The system (2) is null-controllable at the time $T > m_{\max}h$;

ii) The system (2) is not null-controllable at the time $T < m_{\min}h$.

The proofs are based on the construction of a special Riesz basis of \mathcal{A} -invariant subspaces in the space M_2 according to [31, 32] and on the analysis of the properties of some quasi-exponential functions to be a Riesz basis in $L_2(0, T)$ depending of the time T [3].

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