

## Magnetic Field Created by Tile Permanent Magnets

Romain Ravaud, Guy Lemarquand, Valérie Lemarquand

### ▶ To cite this version:

Romain Ravaud, Guy Lemarquand, Valérie Lemarquand. Magnetic Field Created by Tile Permanent Magnets. IEEE Transactions on Magnetics, 2009, 45 (7), pp.2920-2926. 10.1109/TMAG.2009.2014752 . hal-00401733

## HAL Id: hal-00401733 https://hal.science/hal-00401733

Submitted on 6 Jul 2009

**HAL** is a multi-disciplinary open access archive for the deposit and dissemination of scientific research documents, whether they are published or not. The documents may come from teaching and research institutions in France or abroad, or from public or private research centers. L'archive ouverte pluridisciplinaire **HAL**, est destinée au dépôt et à la diffusion de documents scientifiques de niveau recherche, publiés ou non, émanant des établissements d'enseignement et de recherche français ou étrangers, des laboratoires publics ou privés.

# Magnetic Field Created by Tile Permanent Magnets

R. Ravaud, G. Lemarquand, V. Lemarquand

#### Abstract

This paper presents the analytical calculation of the three components of the magnetic field created by 2 tile permanent magnets whose magnetization is either radial or axial. The calculations are based on the з coulombian model of permanent magnets. The magnetic field is directly calculated, without the magnetic potential. Both axial and radial magnetization of the tiles are considered. The expressions obtained give the 5 magnetic field in all the space. Such analytical expressions are very useful for the design and optimization of many industrial applications. **Index Terms** 8 Analytical model, magnetic field, permanent magnet, axial magnetization, radial magnetization 9 I. INTRODUCTION 10 FTER the parallelepiped, the most common shape for a permanent magnet in electrical engineering 11 is certainly the tile, which can also be described as a ring sector. Such tiles can be either axially 12 or radially magnetized. Therefore, the calculation of the magnetic field created by tile magnets is of great 13 utility and numerous approaches exist. 14 Numerical approaches do not allow one to perform numerous parametric studies quickly and have 15 generally a high computational cost. Consequently, authors are looking for alternative solutions. Their 16 approaches are often semianalytical ones [1], and they represent important steps toward the ideal analytical 17

Manuscript Received April 10, 2008.

1

The authors are with the Laboratoire d'Acoustique de l'Universite du Maine UMR CNRS 6613, Avenue Olivier Messiaen, 72085 Le Mans Cedex 9, France

ones. In fact, different points of view can be adopted, which all look for the same thing: the calculation of 18 the magnetic field and the magnetic forces. Kim et al. [2][3] calculate the magnetic force from the vector 19 potential, with complete elliptic integrals and a mesh-matrix method. Kwon et al.[4] work with spherical 20 coordinates and multipole expansions to calculate the far field of permanent magnet motors. Selvaggi et 21 al.[5] propose an approach with cylindrical coordinates and the use of Green's function and Fourier series 22 expansion to characterize the field from a set of permanent magnets. Babic and Akyel [6]-[8] calculate the 23 force between coils thanks to Heuman's Lambda function. Toroidal harmonics yield interesting solutions 24 to formulate the magnetic field created by permanent magnet cylinders [9]. Conway uses Bessel's function 25 to calculate the inductance of coils using the vector potential [10] and proposes also a direct calculation 26 of the magnetic field -without the vector potential- to reach the same goal [11]. 27

While these studies show variety in the starting point for the problem they want to solve, many others 28 are more specific and describe methods to calculate the magnetic field created by toroidal magnets. Perigo 29 et al.[12] present analytical-integral expressions for the axial and radial magnetic flux density components 30 of axially magnetized toroidal magnets, either on- or off-axis. The application considered is an electron 31 beam focusing system. Zhilichev [13] works with cylindrical coordinates and uses separation of variables 32 to evaluate the magnetic field from the scalar potential for tubular linear permanent magnet machines. He 33 gives 2D and 3D approximations. Furlani et al. propose solutions based on the vector potential for radially 34 polarized multipole cylindrical magnets [14] as well as for axially polarized magnets for axial field motors 35 [15][16]. Rakotoarison et al. [17] give a semianalytical method to calculate the field created by radially 36 magnetized tile magnets from the scalar potential. They use a coulombian model of the magnets and take 37 the volume charge density into account. 38

All the studies dedicated to the calculation of the magnetic field created by toroidal or tile magnets are very useful: they constitute tools for the design and the optimization of devices which use such magnets and enable to meet specific requirements. For example, tile magnets with rotating magnetizations are used to make discrete Halbach cylinders [18] which have various applications, such as electrodynamic wheels [19] for maglev devices, the creation of homogeneous fields or tailored gradient fields [20], the magnetization of brushless machines to obtain sinusoidal field variation in the airgap [21]. Tile magnets are used with peculiar profiles to reduce the cogging torque in axial flux machines [22] as well as in radial flux ones [23], but also to control the torque in permanent magnet couplings [24][25] and gears. They are constitutive parts of displacement sensors [26] and also ironless loudspeakers [27][28] demonstrating their widespread use.

This paper presents analytical expressions for the three components of the magnetic field created by tile permanent magnets. The magnetic field is directly calculated, without a previous calculation of the scalar or vector potential[29][30][31]. The expressions given are for a ring sector, but remain valid when the sector is extended to the whole ring. The axial magnetization of the tiles is considered in the first section below and the radial magnetization in the second.

54

#### II. AXIALLY MAGNETIZED TILES

55 A. Notation and Geometry

The geometry which is considered is a tile permanent magnet and the related parameters are shown 56 in Fig.1. The tile inner radius is  $r_{in}$ , the tile outer one is  $r_{out}$  and its height is h. The angular width of 57 the tile is  $\theta_2 - \theta_1$ . The axis z is an axis of symmetry. The coulombian model of a permanent magnet is 58 used. Consequently, the tile permanent magnet is represented by two curved planes which correspond to 59 the upper (z = h) and lower (z = 0) faces of the ring sector. The upper one is charged with a surface 60 magnetic pole density  $+\sigma^*$ ; the lower one is charged with the opposite surface magnetic pole density 61  $-\sigma^*$ . All the illustrative calculations are done with  $\sigma^* = \vec{J} \cdot \vec{n} = 1T$ . In order to simplify the calculations, 62 the upper face only is taken into account to determine the three magnetic field components. However, 63 the total magnetic field can be calculated by the application of the linear superposition principle to both 64 faces. 65

Let us consider a point P on the ring sector upper face. The magnetic field  $\vec{H}$  created by the source point  $P(r, \theta_s, z)$  at any observation point  $M(r, \theta, z)$  of the space is given by (1).

$$\vec{H}(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \int_{\theta_1}^{\theta_2} \int_{r_{in}}^{r_{out}} \frac{\overrightarrow{PM}}{\left|\overrightarrow{PM}\right|^3} r_1 dr_1 d\theta_s \tag{1}$$

where  $\mu_0$  is the vacuum magnetic permeability ( $\mu_0 = 4\pi . 10^{-7} SI$ ) and  $\sigma^*$  is the fictitious magnetic pole



Fig. 1. Axially magnetized tile permanent magnet : parameter definition.

<sup>69</sup> surface density in tesla. Equation (1) can be written as follows:

$$\vec{H}(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \int_{\theta_1}^{\theta_2} \int_{r_{in}}^{r_{out}} \frac{(r-r_1\cos(\theta-\theta_s))\vec{u}_r - r_1\sin(\theta_s-\theta)\vec{u}_\theta + (z-z_1)\vec{u}_z}{(r_1^2 + r^2 + (z-z_1)^2 - 2r_1r\cos(\theta-\theta_s))^{\frac{3}{2}}} r_1 dr_1 d\theta_s \quad (2)$$

<sup>70</sup> B. Components along the three directions  $\vec{u}_r$ ,  $\vec{u}_{\theta}$ ,  $\vec{u}_z$ 

The integration of (2) leads to the magnetic field components along the three axes defined  $H_r(r, \theta, z)$ ,  $H_{\theta}(r, \theta, z), H_z(r, \theta, z)$ .

<sup>73</sup> 1) Azimuthal component  $H_{\theta}(r, \theta, z)$ : The magnetic field azimuthal component  $H_{\theta}(r, \theta, z)$  created by <sup>74</sup> the upper face is given by (3).

$$H_{\theta}(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \left(\eta(\theta,\theta_1) - \eta(\theta,\theta_2)\right) \tag{3}$$

75 with

$$\eta(\theta, \theta_{i}) = \frac{\sqrt{r^{2} + r_{in}^{2} + (z - h)^{2} - 2rr_{in}\cos(\theta_{i} - \theta)}}{r} + \cos(\theta_{i} - \theta)\log\left[r_{in} - r\cos(\theta_{i} - \theta) + \sqrt{r^{2} + r_{in}^{2} + (z - h)^{2} - 2rr_{in}\cos(\theta_{i} - \theta)}\right] - \frac{\sqrt{r^{2} + r_{out}^{2} + (z - h)^{2} - 2rr_{out}\cos(\theta_{i} - \theta)}}{r}$$



Fig. 2. Field azimuthal component  $H_{\theta}(r, \theta, z)$  versus the radial distance r of the observation point; h = 3mm,  $r_{in} = 25mm$ ,  $r_{out} = 28mm$ ,  $\theta_1 = 0$ ,  $\theta_2 = \frac{\pi}{2}$ ,  $\theta = \frac{\pi}{2}$ 

$$-\cos(\theta_i - \theta)\log\left[r_{out} - r\cos(\theta_i - \theta) + \sqrt{r^2 + r_{out}^2 + (z - h)^2 - 2rr_{out}\cos(\theta_i - \theta)}\right]$$
(4)

Equation (3) is valid for any observation point  $M(r, \theta, z)$  with  $0 \le \theta \le 2\pi$ . Figure 2 represents the azimuthal component  $H_{\theta}(r, \theta, z)$  versus the radial distance r of the observation point. The parameter values are h = 3 mm,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = 0 \text{ rad}$ ,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ .

<sup>79</sup> 2) Radial component  $H_r(r, \theta, z)$ : The field radial component  $H_r(r, \theta, z)$  created by the upper face is <sup>80</sup> given by (5).

$$H_r(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \left( \alpha(u_2,r,z) - \alpha(u_1,r,z) \right)$$
(5)

81 where

$$u_i = \cos(\theta - \theta_i) \tag{6}$$

82 and

$$\begin{aligned} \alpha(u_{i},r,z) &= F_{1}(u_{i},r,z) \left( G_{1}(u_{i},r,z) \mathbf{E}^{*} \left[ \arcsin\left[\frac{\sqrt{r^{2} + r_{out}^{2} - 2rr_{out}u_{i} + (z-h)^{2}}}{(r+r_{out})^{2} + (z-h)^{2}} \right], \frac{(r+r_{out})^{2} + (z-h)^{2}}{(r-r_{out})^{2} + (z-h)^{2}} \right] \right) \\ &+ F_{1}(u_{i},r,z) \Pi^{*} \left[ \arcsin\left[\frac{\sqrt{r^{2} + r_{out}^{2} - 2rr_{out}u_{i} + (z-h)^{2}}}{(r+r_{out})^{2} + (z-h)^{2}} \right], \frac{(r+r_{out})^{2} + (z-h)^{2}}{(r-r_{out})^{2} + (z-h)^{2}} \right] \\ &- F_{2}(u_{i},r,z) \left( G_{2}(u_{i},r,z) \mathbf{E}^{*} \left[ \arcsin\left[\frac{\sqrt{r^{2} + r_{out}^{2} - 2rr_{in}u_{i} + (z-h)^{2}}}{(r+r_{in})^{2} + (z-h)^{2}} \right], \frac{(r+r_{in})^{2} + (z-h)^{2}}{(r-r_{in})^{2} + (z-h)^{2}} \right] \right) \end{aligned}$$

$$-F_{2}(u_{i},r,z)\Pi^{*}\left[ \arcsin\left[\frac{\sqrt{r^{2}+r_{in}^{2}-2rr_{in}u_{i}+(z-h)^{2}}}{(r+r_{in})^{2}+(z-h)^{2}}\right],\frac{(r+r_{in})^{2}+(z-h)^{2}}{(r-r_{in})^{2}+(z-h)^{2}}\right] + \frac{(u_{i}^{2}-1)}{\sqrt{1-u_{i}^{2}}}\log\left[r_{out}-ru_{i}+\sqrt{r^{2}+r_{out}^{2}-2rr_{out}u_{i}+(z-h)^{2}}\right] - \frac{(u_{i}^{2}-1)}{\sqrt{1-u_{i}^{2}}}\log\left[r_{in}-ru_{i}+\sqrt{r^{2}+r_{in}^{2}-2rr_{in}u_{i}+(z-h)^{2}}\right]$$

83 with :

$$F_1(u_i, r, z) = \frac{1}{\sqrt{1 - u_i^2}} \frac{2r_{out}(1 + u_i)\sqrt{\frac{rr_{out}(u_i - 1)}{(r - r_{out})^2 + (z - h)^2}}\sqrt{\frac{r^2 + r_{out}^2 - 2rr_{out}u_i + (z - h)^2}{(r + r_{out})^2 + (z - h)^2}}}{\sqrt{\frac{rr_{out}(1 + u_i)}{(r + r_{out})^2 + (z - h)^2}}}\sqrt{r^2 + r_{out}^2 - 2rr_{out}u_i + (z - h)^2}}$$
(8)

84

85

86

$$F_2(u_i, r, z) = \frac{1}{\sqrt{1 - u_i^2}} \frac{2r_{in}(1 + u_i)\sqrt{\frac{rr_{in}(u_i - 1)}{(r - r_{in})^2 + (z - h)^2}}\sqrt{\frac{r^2 + r_{in}^2 - 2rr_{in}u_i + (z - h)^2}{(r + r_{in})^2 + (z - h)^2}}}{\sqrt{\frac{rr_{in}(1 + u_i)}{(r + r_{in})^2 + (z - h)^2}}}\sqrt{r^2 + r_{in}^2 - 2rr_{in}u_i + (z - h)^2}}$$
(9)

$$G_1(r,z) = \frac{(r-r_{in})^2 + (z-h)^2}{2rr_{in}}$$
(10)

$$G_2(r,z) = \frac{(r - r_{out})^2 + (z - h)^2}{2rr_{out}}$$
(11)

87

$$E^*[k] = \int_0^{\phi = \frac{\pi}{2}} \sqrt{1 - k^2 \sin(\theta)^2} d\theta$$
 (12)

Equation (5) is valid for any observation point  $M(r, \theta, z)$  with  $\theta \neq \theta_i$  and  $0 \leq \theta < 2\pi$ . This expression remains valid for ring permanent magnets, for which the angular width is  $2\pi$  ( $\theta_2 - \theta_1 = 2\pi$ ). It leads to the expression of the radial component already given by the authors for ring magnets [32]. Figure 3 represents the field radial component  $H_r(r, \theta, z)$  versus the radial distance r. The used parameters are  $h = 3 \text{ mm}, \theta_1 = 0 \text{ rad}, \theta_2 = \frac{\pi}{2} \text{ rad}, \theta = 0 \text{ rad}, r_{in} = 25 \text{ mm}, r_{out} = 28 \text{ mm}.$ 

<sup>93</sup> 3) Axial component  $H_z(r, \theta, z)$ : The field axial component  $H_z(r, \theta, z)$  created by the upper face is <sup>94</sup> given by (13).

$$H_z(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \left(\gamma(\theta,\theta_2) - \gamma(\theta,\theta_1)\right) \tag{13}$$

6

(7)



Fig. 3. Field radial component  $H_r(r, \theta, z)$  versus the radial distance r of the observation point; h = 3mm,  $r_{in} = 25mm$ ,  $r_{out} = 28mm$ ,  $\theta_1 = 0$  rad,  $\theta_2 = \frac{\pi}{2}$  rad,  $\theta = \frac{\pi}{2}$  rad

95 with

$$\begin{split} \gamma(\theta, \theta_i) &= h_1 \eta_1(\theta, \theta_i) \mathbf{\Pi}^* \left[ \frac{2(c_1 + d_1)f}{2c_1 f - \sqrt{2}\sqrt{d_1^2 f(-e+f)}}, i \sinh^{-1} \left[ \sqrt{\frac{-1}{c_1 + d_1}} \sqrt{c_1 - d_1 \cos(\theta - \theta_i)} \right], \frac{c_1 + d_1}{c_1 - d_1} \right] \\ &+ h_2 \eta_1(\theta, \theta_i) \mathbf{\Pi}^* \left[ \frac{2(c_1 + d_1)f}{2c_1 f + \sqrt{2}\sqrt{d_1^2 f(-e+f)}}, i \sinh^{-1} \left[ \sqrt{\frac{-1}{c_1 + d_1}} \sqrt{c_1 - d_1 \cos(\theta - \theta_i)} \right], \frac{c_1 + d_1}{c_1 - d_1} \right] \\ &- h_3 \eta_2(\theta, \theta_i) \mathbf{\Pi}^* \left[ \frac{2(c_2 + d_2)f}{2c_2 f - \sqrt{2}\sqrt{d_2^2 f(-e+f)}}, i \sinh^{-1} \left[ \sqrt{\frac{-1}{c_2 + d_2}} \sqrt{c_2 - d_2 \cos(\theta - \theta_i)} \right], \frac{c_2 + d_2}{c_2 - d_2} \right] \\ &- h_4 \eta_2(\theta, \theta_i) \mathbf{\Pi}^* \left[ \frac{2(c_2 + d_2)f}{2c_2 f + \sqrt{2}\sqrt{d_2^2 f(-e+f)}}, i \sinh^{-1} \left[ \sqrt{\frac{-1}{c_2 + d_2}} \sqrt{c_2 - d_2 \cos(\theta - \theta_i)} \right], \frac{c_2 + d_2}{c_2 - d_2} \right] \end{split}$$

$$(14)$$

96 with

$$\eta_1(\theta, \theta_i) = \frac{\left(-i\sqrt{\frac{d_1(-1+\cos(\theta-\theta_i))}{c_1-d_1}}\sqrt{\frac{d_1(1+\cos(\theta-\theta_i))}{c_1+d_1}}\frac{1}{\cos(\theta-\theta_i)}\right)}{\left(2\sqrt{\frac{-1}{c_1+d_1}}\sqrt{d_1^2f(-e+d_1)}(d_1^2(e-f))+2c_1^2f\right)}$$
(15)

97

$$\eta_2(\theta, \theta_i) = \frac{\left(-i\sqrt{\frac{d_2(-1+\cos(\theta-\theta_i))}{c_2-d_2}}\sqrt{\frac{d_2(1+\cos(\theta-\theta_i))}{c_2+d_2}}\frac{1}{\cos(\theta-\theta_i)}\right)}{\left(2\sqrt{\frac{-1}{c_2+d_2}}\sqrt{d_2^2f\left(-e+d_2\right)}(d_2^2(e-f))+2c_2^2f\right)}$$
(16)

 $h_1 = 2ad_1(\sqrt{2}c_1f + \sqrt{d_1^2f(-e+f)}) + b_1(\sqrt{2}d_1^2(e-f) - 2c_1\sqrt{d_1^2f(-e+f)})$ 

98

(17)

Parameters	
a	$2(z-h)(r^2+(z-h)^2)$
$b_1$	$2(z-h)rr_{out}$
$c_1$	$r^2 + r_{out}^2 + (z - h)^2$
$d_1$	$-2rr_{out}$
e	$-r^2 - 2(z-h)^2$
f	$r^2$
$b_2$	$2(z-h)rr_{in}$
<i>c</i> <sub>2</sub>	$r^2 + r_{in}^2 + (z - h)^2$
$d_2$	$-2rr_{in}$

 TABLE I

 Definition of the parameters used in (13)

99

$$h_2 = 2ad_1(-\sqrt{2}c_1f + \sqrt{d_1^2f(-e+f)}) + b_1(\sqrt{2}d_1^2(-e+f) - 2c_1\sqrt{d_1^2f(-e+f)})$$
(18)

100

$$h_3 = 2ad_2(\sqrt{2}c_2f + \sqrt{d_2^2f(-e+f)}) + b_2(\sqrt{2}d_2^2(e-f) - 2c_2\sqrt{d_2^2f(-e+f)})$$
(19)

$$h_4 = 2ad_2(-\sqrt{2}c_2f + \sqrt{d_2^2f(-e+f)}) + b_2(\sqrt{2}d_2^2(-e+f) - 2c_2\sqrt{d_2^2f(-e+f)})$$
(20)

where  $\Pi^*[n, \phi, m]$  is given in terms of the incomplete elliptic integral of the third kind by (21).

$$\mathbf{\Pi}^*[n,\phi,m] = \int_0^\phi \frac{1}{(1-n\sin(\theta)^2)\sqrt{1-m\sin(\theta)^2}} d\theta$$
(21)

Although the result  $H_z(r, \theta, z)$  is a real number, equation (14) contains the imaginary number i ( $i^2 = -1$ ) 103 because we did not succeed in obtaining a real expression for the axial component  $H_z(r, \theta, z)$ . The 104 parameters used in (14) are defined in Table I. However, as the imaginary part is the consequence of 105 numerical noise and nearly equals zero, when the expression (14) is used in symbolic mathematical tools 106 such as Mathematica or Maple, the real part of  $Hz(r, \theta, z)$  only has to be considered. Equation (13) is 107 valid for any observation point  $M(r, \theta, z)$  with  $\theta \neq \theta_i$  and  $0 \leq \theta < 2\pi$ . Here again, this expression 108 remains valid for ring permanent magnets, i.e. when the angular width is  $2\pi$  ( $\theta_2 - \theta_1 = 2\pi$ ). It also 109 leads to the expression of the axial component already given by the authors for ring magnets [32]. Figure 110 4 represents the axial component  $H_z(r, \theta, z)$  versus the radial distance r of the observation point. The 111 parameter values are h = 3 mm,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = 0 \text{ rad}$ ,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ . 112



Fig. 4. Field axial component  $H_z(r, \theta, z)$  versus the radial distance r of the observation point; h = 3mm,  $r_{in} = 25mm$ ,  $r_{out} = 28mm$ ,  $\theta_1 = 0$  rad,  $\theta_2 = \frac{\pi}{2}$  rad,  $\theta = \frac{\pi}{2}$  rad

#### **III. RADIALLY MAGNETIZED TILES**

#### 114 A. Notation and geometry

113

The geometry and its parameters are shown in Fig.(5). The axis z is an axis of symmetry. Again, the coulombian model of permanent magnets is used. The permanent magnet ring sector is thus represented by two curved planes which correspond here to the inner and outer faces of the ring. The inner face is charged with a magnetic pole surface density  $+\sigma^*$ ; the outer one is charged with the opposite magnetic pole surface density  $-\sigma^*$ . We only consider the inner face to simplify the analytical calculation. As stated previously, the total magnetic field can be calculated by the application of the linear superposition principle to both faces.

The magnetic pole volume density is not taken into account in this paper. This means that the total sum 122 of all the charges in the model does not equal zero. Indeed, as the magnetization is radial, the magnetic 123 pole surface density of the curved planes is uniform. The charge on the outer plane is thus greater than 124 the charge on the inner plane, as the surfaces of these planes. The volume charge density, linked to the 125 magnetization divergence, appears in fact to set the global charge to zero. If the radial width of the tile is 126 small, which also means that the tile is thin, then the difference between the inner and outer plane surface 127 is small, and so is the magnetic pole volume density: its neglecting is an acceptable approximation. This 128 approximation becomes less and less valid when the thickness of the tile increases. This paper presents 129 expressions for radially magnetized thin tiles. 130

Let us consider a point P on the tile inner face. The magnetic field  $\vec{H}$  created by the source point



Fig. 5. Radially magnetized permanent magnet tile: parameter definition

 $_{\mbox{\tiny 132}}\quad P(r,\theta_s,z)$  at any observation point  $M(r,\theta,z)$  of the space is given by (22).

$$\vec{H}(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \int_{\theta_1}^{\theta_2} \int_0^h \frac{\overrightarrow{PM}}{\left|\overrightarrow{PM}\right|^3} r_1 dz_1 d\theta_s$$
(22)

where  $\mu_0$  is the magnetic permeability of the vacuum ( $\mu_0 = 4\pi . 10^{-7} SI$ ) and  $\sigma^*$  is the fictitious magnetic pole surface density given in tesla. Equation (22) can be written as follows:

$$\vec{H}(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \int_{\theta_1}^{\theta_2} \int_0^h \frac{(r-r_1\cos(\theta-\theta_s))\vec{u}_r - r_1\sin(\theta_s-\theta)\vec{u}_\theta + (z-z_1)\vec{u}_z}{(r_1^2 + r^2 + (z-z_1)^2 - 2r_1r\cos(\theta-\theta_s))^{\frac{3}{2}}} r_1dz_1d\theta_s$$
(23)

<sup>135</sup> B. Components along the three directions  $\vec{u}_r$ ,  $\vec{u}_{\theta}$ ,  $\vec{u}_z$ 

The integration of (23) leads to the magnetic field components created by the inner face along the three axes defined  $H_r(r, \theta, z)$ ,  $H_{\theta}(r, \theta, z)$ ,  $H_z(r, \theta, z)$ .

138 1) Azimuthal component  $H_{\theta}(r, \theta, z)$ : The field azimuthal component  $H_{\theta}(r, \theta, z)$  is given by (24).

$$H_{\theta}(r,\theta,z) = \frac{\sigma}{4\pi\mu_0} \left(\beta(\theta,\theta_1) - \beta(\theta,\theta_2)\right) \tag{24}$$



Fig. 6. Field azimuthal component  $H_{\theta}(r, \theta, z)$  versus the radial distance r of the observation point; h = 3 mm,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ ,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = 0 \text{ rad}$ 

139 with

$$\beta(\theta, \theta_i) = \frac{r_{in}(b-z)}{r\sqrt{-(b-z)^2}} \arctan\left[\frac{\sqrt{r^2 + r_{in}^2 + (b-z)^2 - 2rr_{in}\cos(\theta_i - \theta)}}{\sqrt{-(b-z)^2}}\right] - \frac{r_{in}}{r} \tanh^{-1}\left[\frac{\sqrt{r^2 + r_{in}^2 + z^2 - 2rr_{in}\cos(\theta_i - \theta)}}{z}\right]$$
(25)

Equation (24) is valid for any observation point  $M(r, \theta, z)$  with  $0 \le \theta \le 2\pi$ . Figure 6 represents the azimuthal component  $H_{\theta}(r, \theta, z)$  versus the radial distance r of the observation point. The parameters values are  $h = 3mm, \theta_1 = 0, \theta_2 = \frac{\pi}{2}, \theta = 0, r_{in} = 25mm, r_{out} = 28mm$ .

143 2) Radial component  $H_r(r, \theta, z)$ : The radial component of the field  $H_r(r, \theta, z)$  is given by (26).

$$H_r(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \left(\beta(u_1) - \beta(u_2)\right) \tag{26}$$

144 with

$$u_i = \cos(\theta - \theta_i) \tag{27}$$

145 and

$$\beta(u_i) = \left(\frac{2i(1+u_i)\sqrt{\frac{d(-1+u_i)}{c+e_1+du_i}}\left(-(a_1d+b_1(c+e_1))\right)\mathbf{F}^*\left[i\sinh^{-1}\left[\frac{\sqrt{-c+d-e_1}}{\sqrt{c+e_1+du_i}}\right],\frac{c+d+e_1}{c-d+e_1}\right]}{d\sqrt{-c+d-e_1}e_1\sqrt{\frac{d(1+u_i)}{c+e_1+du_i}}\sqrt{1-u_i^2}}\right)$$

Parameters	
$a_1$	$r_{in}rz$
$b_1$	$-r_{in}^2z$
С	$r^2 + r_{in}^2$
d	$-2rr_{in}$
$e_1$	$z^2$
$a_2$	$-r_{in}r(z-h)$
$b_2$	$r_{in}^2(z-h)$
$e_2$	$(z-h)^2$

 TABLE II

 Definition of the parameters used in (26)

$$+ \left(\frac{2i(1+u_{i})\sqrt{\frac{d(-1+u_{i})}{c+e_{1}+du_{i}}}(b_{1}c-a_{1}d)\mathbf{\Pi}^{*}\left[\frac{e_{1}}{c-d+e_{1}},i\sinh^{-1}\left[\frac{\sqrt{-c+d+e_{1}}}{c+e_{1}+du_{i}}\right],\frac{c+d+e_{1}}{c-d+e_{1}}\right]}{d\sqrt{-c+d-e_{1}}e_{1}\sqrt{\frac{d(1+u_{i})}{c+e_{1}+du_{i}}}\sqrt{1-u_{i}^{2}}}\right)$$

$$+ \left(\frac{2i(1+u_{i})\sqrt{\frac{d(-1+u_{i})}{c+e_{2}+du_{i}}}(-(a_{2}d+b_{2}(c+e_{2})))\mathbf{F}^{*}\left[i\sinh^{-1}\left[\frac{\sqrt{-c+d-e_{2}}}{\sqrt{c+e_{2}+du_{i}}}\right],\frac{c+d+e_{2}}{c-d+e_{2}}\right]}{d\sqrt{-c+d-e_{2}}e_{2}\sqrt{\frac{d(1+u_{i})}{c+e_{2}+du_{i}}}\sqrt{1-u_{i}^{2}}}\right)$$

$$+ \left(\frac{2i(1+u_{i})\sqrt{\frac{d(-1+u_{i})}{c+e_{2}+du_{i}}}(b_{2}c-a_{2}d)\mathbf{\Pi}^{*}\left[\frac{e_{2}}{c-d+e_{2}},i\sinh^{-1}\left[\frac{\sqrt{-c+d+e_{2}}}{c+e_{2}+du_{i}}\right],\frac{c+d+e_{2}}{c-d+e_{2}}\right]}{d\sqrt{-c+d-e_{2}}e_{2}\sqrt{\frac{d(1+u_{i})}{c+e_{2}+du_{i}}}\sqrt{1-u_{i}^{2}}}\right)$$

$$(28)$$

where  $F^*[\phi,m]$  is defined by (31) and  $\Pi^*[n,\phi,m]$  is given by (21). Here again, Eq.(28) contains the 146 imaginary number i ( $i^2 = -1$ ) although the result  $H_r(r, z)$  is a real number because we did not succeed in 147 obtaining a real expression for the radial component  $H_r(r, \theta, z)$ . The parameters used in (28) are defined 148 in Table II. Of course, when using the expression (28) in tools like Mathematica or Mapple, the imaginary 149 part of  $H_r(r, \theta, z)$  has to be dropped, as it only corresponds to numerical noise and nearly equals zero. 150 Equation (26) can be used to calculate the radial component of the magnetic field at any observation 151 point  $M(r, \theta, z)$  with  $\theta \neq \theta_i$  and  $0 \leq \theta < 2\pi$ . When the tile angular width becomes  $2\pi (\theta_2 - \theta_1 = 2\pi)$ , 152 so for ring magnets, the expression becomes the one already given in a previous paper [32]. Figure 7 153 represents the field radial component  $H_r(r, \theta, z)$  versus the radial distance r of the observation point. The 154 parameter values are h = 3 mm,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = 0 \text{ rad}$ ,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ . 155



Fig. 7. Field radial component  $H_r(r, \theta, z)$  versus the radial distance r of the observation point; h = 3 mm,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ ,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = \frac{\pi}{2} \text{ rad}$ 

156 3) Axial component  $H_z(r, \theta, z)$ : The field axial component  $H_z(r, \theta, z)$  is given by (29):

$$H_z(r,\theta,z) = \frac{\sigma^*}{4\pi\mu_0} \left(\alpha(\theta,\theta_2) - \alpha(\theta,\theta_1)\right]$$
(29)

157 with

$$\alpha(\theta, \theta_i) = \frac{2r_{in}}{(r - r_{in})^2 + (z - h)^2} \mathbf{F}^* \left[ \frac{\theta - \theta_i}{2}, -\frac{4rr_{in}}{(r - r_{in})^2 + (z - h)^2} \right] - \frac{2r_{in}}{(r - r_{in})^2 + z^2} \mathbf{F}^* \left[ \frac{\theta - \theta_i}{2}, -\frac{4rr_{in}}{(r - r_{in})^2 + z^2} \right]$$
(30)

where  $\mathbf{F}^*[\phi, m]$  is given in terms of the elliptic integral of the first kind by (31).

$$\mathbf{F}^*[\phi|m] = \int_{\theta=0}^{\theta=\phi} \frac{1}{\sqrt{1-m\sin(\theta)^2}} d\theta \tag{31}$$

Equation (29) is valid for any observation point  $M(r, \theta, z)$  with  $0 \le \theta \le 2\pi$ . Figure 8 represents the field axial component  $H_z(r, \theta, z)$  versus the radial distance r of the observation point. The parameter values which are taken for the calculation are h = 3 mm,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = 0 \text{ rad}$ ,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ .



Fig. 8. Field axial component  $H_z(r, \theta, z)$  versus the radial distance r of the observation point; h = 3 mm,  $r_{in} = 25 \text{ mm}$ ,  $r_{out} = 28 \text{ mm}$ ,  $\theta_1 = 0 \text{ rad}$ ,  $\theta_2 = \frac{\pi}{2} \text{ rad}$ ,  $\theta = \frac{\pi}{2} \text{ rad}$ 

#### **IV. CONCLUSION**

This paper gives the analytical expressions of the three components of the magnetic field created by tile 164 permanent magnets whose magnetization is either radial or axial. When the tiles are axially magnetized, 165 the magnetic field can be calculated using these expressions at any point in space. For radially magnetized 166 tiles, the expressions given correspond to the field created by thin tiles, whose inner and outer radii are 167 not too different, and can be used at any point of the space as well. The magnetic field is directly 168 calculated, without the help of the magnetic potentials. The utility of such analytical expressions for the 169 three components of the magnetic field created by tile permanent magnets lies in the fact that the related 170 calculations have a low computational cost, especialy with regard to methods using finite elements or 171 finite differences, and that they allow efficient parametric optimization studies of devices, which is very 172 important for all applications. 173

The Mathematica files containing the analytical expressions used to calculate the three components of the magnetic field for axial and radial magnetizations are available line [33].

176

163

#### References

- B. Azzerboni and G. Saraceno, "Three-dimensional calculation of the magnetic field created by current-carrying massive disks,"
   *IEEE Trans. Magn.*, vol. 34, no. 5, pp. 2601–2604, 1998.
- [2] K. Kim, E. Levi, Z. Zabar, and L. Birenbaum, "Restoring force between two noncoaxial circular coils," *IEEE Trans. Magn.*,
   vol. 32, no. 32, pp. 478–484, 1996.

- [3] K. Kim, E. Levi, Z. Zabar, and L. Birenbaum, "Mutual inductance of noncoaxial circular coils with constant current density," 181 IEEE Trans. Magn., vol. 33, no. 32, pp. 4303-4309, 1997. 182
- [4] O. M. Kwon, C. Surussavadee, M. V. K. Chari, S. Salon, and K. S. Vasubramaniam, "Analysis of the far field of permanent 183 magnet motors and effects of geometric asymmetries and unbalance in magnet design," IEEE Trans. Magn., vol. 40, no. 3, 184 pp. 435-442, 2004. 185
- [5] J. P. Selvaggi, S. Salon, O. M. Kwon, and M. V. K. Chari, "Calculating the external magnetic field from permanent magnets 186 in permanent-magnet motors - an alternative method," IEEE Trans. Magn., vol. 40, no. 5, pp. 3278-3285, 2004. 187
- [6] S. Babic and C. Akyel, "Improvement in the analytical calculation of the magnetic field produced by permanent magnet rings," 188 Progress in Electromagnetics Research C, vol. 5, pp. 71-82, 2008. 189
- [7] S. Babic and C. Akyel, "An improvement in the calculation of the magetic field for an arbitrary geometry coil with rectangular 190 cross section," International Journal of Numerical Modelling: Electronic Networks, Devices and Fields, vol. 18, pp. 493-504, 191 192 November 2005.
- [8] S. Babic and C. Akyel, "Magnetic force calculation between thin coaxial circular coils in air," IEEE Trans. Magn., vol. 44, 193 194 no. 4, pp. 445-452, 2008.
- 195 [9] J. P. Selvaggi, S. Salon, O. M. Kwon, and M. V. K. Chari, "Computation of the three-dimensional magnetic field from solid permanent-magnet bipolar cylinders by employing toroidal harmonics," IEEE Trans. Magn., vol. 43, no. 10, pp. 3833-3839, 196 2007. 197
- [10] J. Conway, "Inductance calculations for noncoaxial coils using bessel functions," IEEE Trans. Magn., vol. 43, no. 3, pp. 1023-198 1034, 2007. 199
- [11] J. Conway, "Noncoaxial inductance calculations without the vector potential for axisymmetric coils and planar coils," IEEE 200 Trans. Magn., vol. 44, no. 10, pp. 453-462, 2008. 201
- [12] E. Perigo, R. Faria, and C. Motta, "General expressions for the magnetic flux density produced by axially magnetized toroidal 202 permanent magnets," IEEE Trans. Magn., vol. 43, no. 10, pp. 3826-3832, 2008. 203
- [13] Y. Zhilichev, "Calculation of magnetic field of tubular permanent magnet assemblies in cylindrical bipolar coordinates," IEEE 204 Trans. Magn., vol. 43, no. 7, pp. 3189-3195, 2007. 205
- [14] E. P. Furlani, S. Reznik, and A. Kroll, "A three-dimensional field solution for radially polarized cylinders," IEEE Trans. Magn., 206 vol. 31, no. 1, pp. 844-851, 1995. 207
- [15] E. P. Furlani, "A two-dimensional analysis for the coupling of magnetic gears," IEEE Trans. Magn., vol. 33, no. 3, pp. 2317-208 2321 1997 209
- 210 [16] E. P. Furlani, "Field analysis and optimization of ndfeb axial field permanent magnet motors," IEEE Trans. Magn., vol. 33, no. 5, pp. 3883-3885, 1997. 211
- [17] H. L. Rakotoarison, J. P. Yonnet, and B. Delinchant, "Using coulombian approach for modeling scalar potential and magnetic 212
- field of a permanent magnet with radial polarization," IEEE Trans. Magn., vol. 43, no. 4, pp. 1261–1264, 2007. 213
- [18] K. Halbach, "Design of permanent multiple magnets with oriented rec material," Nucl. Inst. Meth., vol. 169, pp. 1–10, 1980. 214
- [19] J. Bird and A. Lipo, "Characteristics of an electrodynamic wheel using a 2-d steady-state model," IEEE Trans. Magn., vol. 43, 215

15

no. 8, pp. 3396-3405, 2007. 216

- [20] J. Hilton and S. McMurry, "Halbach cylinders with improved field homogeneity and tailored gradient fields," *IEEE Trans. Magn.*, vol. 43, no. 5, pp. 1898–1902, 2007.
- [21] Z. Zhu, Z. Xia, and D. Howe, "Comparison of halbach magnetized brushless machines based on discrete magnet segments or
   a single ring magnet," *IEEE Trans. Magn.*, vol. 38, no. 9, pp. 2997–2999, 2002.
- [22] M. Aydin, Z. Zhu, T. Lipo, and D. Howe, "Minimization of cogging torque in axial-flux permanent-magnet machines: design
   concepts," *IEEE Trans. Magn.*, vol. 43, no. 9, pp. 3614–3622, 2007.
- [23] Y. Li, J. Zou, and Y. Lu, "Optimum design of magnet shape in permanent-magnet synchronous motors," IEEE Trans. Magn.,
- vol. 39, no. 11, pp. 3523–4205, 2003.
- [24] V. Lemarquand, J. F. Charpentier, and G. Lemarquand, "Nonsinusoidal torque of permanent-magnet couplings," *IEEE Trans. Magn.*, vol. 35, no. 5, pp. 4200–4205, 1999.
- [25] J. F. Charpentier and G. Lemarquand, "Calculation of ironless permanent magnet coupling using semi-numerical magnetic
   pole theory method," *COMPEL*, vol. 20, no. 1, pp. 72–89, 2001.
- [26] C. Blache and G. Lemarquand, "High magnetic field gradients in flux confining permanent magnet structures," *Journal of Magnetism and Magnetic Materials*, vol. 104, pp. 1111–1112, 1992.
- [27] G. Lemarquand, "Ironless loudspeakers," IEEE Trans. Magn., vol. 43, no. 8, pp. 3371–3374, 2007.
- [28] R. Ravaud, G. Lemarquand, V. Lemarquand, and C. Depollier, "Ironless loudspeakers with ferrofluid seals," *Archives of Acoustics*, vol. 33, no. 4, pp. 3–10, 2008.
- [29] R. Ravaud, G. Lemarquand, V. Lemarquand, and C. Depollier, "The three exact components of the magnetic field created by
   a radially magnetized tile permanent magnet.," *Progress in Electromagnetics Research, PIER* 88, pp. 307–319, 2008.
- [30] R. Ravaud, G. Lemarquand, V. Lemarquand, and C. Depollier, "Discussion about the analytical calculation of the magnetic
   field created by permanent magnets.," *Progress in Electromagnetics Research B*, vol. 11, pp. 281–297, 2009.
- [31] R. Ravaud, G. Lemarquand, V. Lemarquand, and C. Depollier, "Magnetic field produced by a tile permanent magnet whose polarization is both uniform and tangential," *Progress in Electromagnetics Research B*, vol. 12, 2009.
- [32] R. Ravaud, G. Lemarquand, V. Lemarquand, and C. Depollier, "Analytical calculation of the magnetic field created by
   permanent-magnet rings," *IEEE Trans. Magn.*, vol. 44, no. 8, pp. 1982–1989, 2008.
- 242 [33] http://www.univ-lemans.fr/~glemar.