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Micky Rakotondrabe, Philippe Lutz

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# Force Estimation in a Piezoelectric Cantilever using the Inverse-Dynamics-Based UIO Technique

Micky Rakotondrabe, *Member, IEEE* and Philippe Lutz, *Member, IEEE*

**Abstract**—This paper presents the estimation of the force applied by a piezocantilever dedicated to micromanipulation/microassembly. Relative to previous works, the presented method avoids the reliance on the force dynamics on the characteristics of the micro-objects. Furthermore, the estimation is a closed-loop kind technique so that convergency can be ensured efficiently. To perform these, we consider the force at the tip of a piezocantilever as an unknown input and we use an Unknown Input Observation technique. We especially use the Inverse-Dynamics-Based UIO technique because it is well suited for a piezocantilever model. The experiments show that the performances of the observer are convenient for micromanipulation/microassembly tasks.

## I. INTRODUCTION

In micromanipulation/microassembly, piezoelectric materials are very widespread because of the rapidity and the high resolution that they can offer. There are many piezoelectric materials based applications in micromanipulation/microassembly but one of the most used to manipulate a micro-object is microgripper (examples in [1][2]). A piezoelectric microgripper is composed of two piezoelectric cantilevers (piezocantilevers) (Fig. 1). When an electrical excitation is applied to each piezocantilever, it bends. The main objective is to pick the micro-object, transport or place it precisely to an other position thanks to the deflections of the cantilevers. If the manipulated object

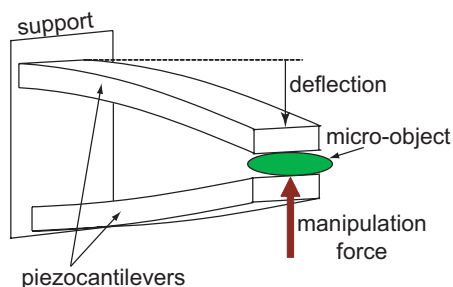


Fig. 1. A piezoelectric microgripper made up of two piezocantilevers.

is fragile (optical objects, biological objects, etc.), it is important to measure and control the manipulation

FEMTO-st Institute,  
UMR CNRS-6174 / UFC / ENSMM / UTBM  
Automatic Control and Micro-Mechatronics Systems department  
(AS2M department)  
25000 Besançon - France  
{mrakoton, plutz}@femto-st.fr

force. Conversely, if the manipulated object is rigid, the piezocantilevers might break if the force is not controlled. Therefore, a concern application is the accurate positioning of the object with the first piezocantilever while the force measurement and control with the second one [3].

Unfortunately, while the modelling and control of the deflection or strain of piezoelectric materials are enough present in the litterature ([4] gives a survey), those of the force are not. This is mainly due to the lack of force sensors convenient for microsystems and micromanipulation/microassembly either on performances or on sizes. Hence, instead of measuring the force, an alternative technique is to observe it using the available signals. In his PhD thesis [5], *Haddab* uses the linear relation between the induced charges and the force to reconstruct the latter. However, the technique requires a good conditioning setup to well recover a good signal. In addition, the technique limits the piezocantilevers to only works in sensing, not in both sensing/actuating. To overcome these limitations, the same litterature [5] proposes to consider the force as a state variable with a null derivative and to observe it thanks to a linear discrete observer. The input signals of the observer are the applied voltage and the measured deflection. Despite the convergency of such closed-loop kind technique, it requires the dynamic part of the force. Although, it has been demonstrated that the dynamic part of the manipulation force always depends on the characteristic of the manipulated object either for classical manipulators [6] or for piezocantilevers manipulators [7]. It is obvious that identifying them for each change of objects is not cogent. To avoid the use of the object characteristics in the modelling and observation, an open-loop kind observer has been proposed in our previous work [8]. As the technique requires a high accurate model, it is not enough robust relative to model errors and does not ensure convergency.

The core of this paper is the observation of the manipulation force between a piezocantilever and manipulated objects. To avoid the use of the object characteristics and to ensure convergency, we consider the force as an unknown input and we use an Unknown Input Observation (UIO) technique. We especially use the Inverse-Dynamics-Based UIO proposed by *Liu and Peng* [9] because their analyzed class of models are well suited for piezocantilevers. Basically, the goal is the estimation of the state and of the unknown input (force) of the

piezocantilever.

First, the UIO observer technique used in this paper is presented. Then, we give the model of the piezocantilever. After that, the computation of the observer is presented. Finally, some experimental results of the force estimation end the paper. In the sequel, the term observation and estimation are alternated to mean the same signification.

## II. THE INVERSE-DYNAMICS-BASED UIO TECHNIQUE

We present in this section the Inverse-Dynamics-Based UIO technique to estimate both the state and the unknown input of a system [9].

### A. Problem statement

Consider the following class of models:

$$\begin{aligned} \dot{x} &= Ax + \Gamma(u, y) + Bd \\ y &= Cx \end{aligned} \quad (1)$$

where  $x \in R^n$  denotes the state vector,  $y \in R^{n_y}$  denotes the output vector,  $u \in R^{n_u}$  denotes the input vector,  $\Gamma(u, y) \in R^n$  is a known nonlinear function of  $u$  and  $y$ ,  $d \in R^{n_d}$  is the unknown input,  $A \in R^{n \times n}$  is the state matrix,  $C \in R^{n_y \times n}$  is the output matrix and  $B \in R^{n \times n_d}$  is the disturbance input matrix. For SISO problem, we have  $n_y = n_d = 1$ .  $n_u$  may be different from one. Indeed the studied problem is a disturbance estimation rather than a control problem. In the sequel, the unknown input is also called disturbance.

The following assumptions are assumed:

- the matrices  $A$ ,  $B$  and  $C$  are known,
- $B$  has full column rank,
- $(A, C)$  is observable.

The objective is to simultaneously estimate  $x$  and  $d$  from the known signals  $y$  and  $u$ .

### B. The equations of the observer

The equation of the state observer is:

$$\begin{aligned} \dot{\hat{x}} &= A\hat{x} + \Gamma(u, y) + B\hat{d} + K(y - \hat{y}) \\ \hat{y} &= C\hat{x} \end{aligned} \quad (2)$$

and the equation of the disturbance observer is:

$$\dot{\hat{d}} = F_1 y + F_2 \dot{y} + G_1 \hat{x} + G_2 \dot{\hat{x}} + G_3 \Gamma(u, y) \quad (3)$$

where:

- the symbol  $\hat{\cdot}$  denotes estimation,
- $K$  is the gain of the state observer,
- $F_1 \in R^{n_d \times n_y}$ ,  $F_2 \in R^{n_d \times n_y}$ ,  $G_1 \in R^{n_d \times n}$ ,  $G_2 \in R^{n_d \times n}$  and  $G_3 \in R^{n_d \times n_y}$  are gains of the disturbance observer.

The error dynamics is obtained using (equ 1), (equ 2) and (equ 3):

$$\dot{e} = (A - KC)e + Be_d \quad (4)$$

with  $e = x - \hat{x}$  and  $e_d = d - \hat{d}$ .

The issue is to find the convenient gains  $K$ ,  $F_1$ ,  $F_2$ ,  $G_1$ ,  $G_2$  and  $G_3$  to make  $e$  and  $e_d$  converge exponentially

to zero. Depending on whether there exists  $F_2$  or not such as  $F_2 CB - I_{n_d \times n_d} = 0$ , two observer schemes were proposed.

### C. First observer scheme

There exists  $F_2$  so that  $F_2 CB - I_{n_d \times n_d} = 0$ . For SISO problem, this is satisfied if and only if  $CB \neq 0$ . Thus:

- (i)  $F_2$  is chosen to satisfy

$$F_2 CB - I_{n_d \times n_d} = 0 \quad (5)$$

- (ii)  $F_1$  and  $K$  are selected such as

$$A - B(F_1 C + F_2 CA) - KC \quad (6)$$

is *Hurwitz*

- (iii) and

$$\begin{aligned} G_1 &= -(F_1 C + F_2 CA) \\ G_2 &= 0 \\ G_3 &= -F_2 C \end{aligned} \quad (7)$$

Indeed, using these gains, the disturbance error becomes dependent of the state error:

$$e_d = -(F_1 C + F_2 CA)e \quad (8)$$

and the error dynamic described by (equ 4) becomes:

$$\dot{e} = (A - B(F_1 C + F_2 CA) - KC)e \quad (9)$$

Because of the (ii) condition, the error dynamic is exponentially stable and  $e$  tends towards zero.

### D. Second observer scheme

Many physical systems fail to satisfy the condition required for the precedent observer scheme. Hence, if for any  $F_2$  one cannot satisfy  $F_2 CB - I_{n_d \times n_d} = 0$ , a second observer scheme was proposed.

Let  $B^+$  be the *Penrose-Moore* inverse of  $B$ . If one chooses:

$$\begin{aligned} G_1 &= -(F_1 C + B^+ A) \\ G_2 &= -(F_2 C - B^+) \\ G_3 &= -B^+ \end{aligned} \quad (10)$$

the error dynamic described by (equ 4) becomes:

$$M_e \dot{e} = A_e e \quad (11)$$

where  $M_e = I + B(F_2 C - B^+)$  and  $A_e = A - B(F_1 C + B^+ A) - KC$ .

If  $M_e$  is nonsingular, the gains  $F_1$ ,  $F_2$  and  $K$  should be selected such as  $M_e^{-1} A_e$  is *Hurwitz*. However if  $M_e$  is singular, the singular value decomposition (SVD) is used. Let:

$$\begin{aligned} M_e &= U \Sigma V^t \\ \Sigma &= \begin{bmatrix} \sigma & 0 \\ 0 & 0 \end{bmatrix} \end{aligned} \quad (12)$$

where  $U \in R^{n \times n}$  and  $V \in R^{n \times n}$  are unitary matrices, and  $\sigma \in R^{n_m \times n_m}$  ( $n_m \leq n$ ) is a positive-definite diagonal matrix. Using the (equ 11) and (equ 12), we obtain:

$$\begin{bmatrix} \sigma & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{z}_1 \\ \dot{z}_2 \end{bmatrix} = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \end{bmatrix} \quad (13)$$

with:

$$\begin{aligned} z &= \begin{bmatrix} z_1 \\ z_2 \end{bmatrix} \equiv V^t e \\ \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} &\equiv U^t A_e V \end{aligned} \quad (14)$$

Then, the error dynamic finally becomes:

$$\begin{aligned} \dot{z}_1 &= \sigma^{-1} A_{11} z_1 + \sigma^{-1} A_{12} z_2 \\ 0 &= \sigma^{-1} A_{21} z_1 + \sigma^{-1} A_{22} z_2 \end{aligned} \quad (15)$$

The core of the second observer scheme is to ensure the exponential stability of (equ 15). For that,  $K$ ,  $F_1$  and  $F_2$  should be selected such as  $A_{22}$  and  $A_{11} - A_{12}A_{22}^{-1}A_{21}$  are *Hurwitz*.

### III. MODEL OF THE PIEZOCANTILEVER

#### A. Presentation of the setup

The experimented piezocantilever is made up of a piezolayer and a Copper layer. Its sizes are:  $15mm \times 2mm \times 300\mu m$  (active length, width and height). The identification and the estimation experiments are performed using the Matlab-Simulink © software and a computer-DSPACE hardware with  $0.2ms$  of sampling time. A *Keyence* optical sensor is used to measure the deflection. It has  $10nm$  of resolution and better than  $400\mu m$  of accuracy. As there is no convenient sensor to check the force, the experiments are performed with known weights applied at the tip of the piezocantilever (Fig. 2).

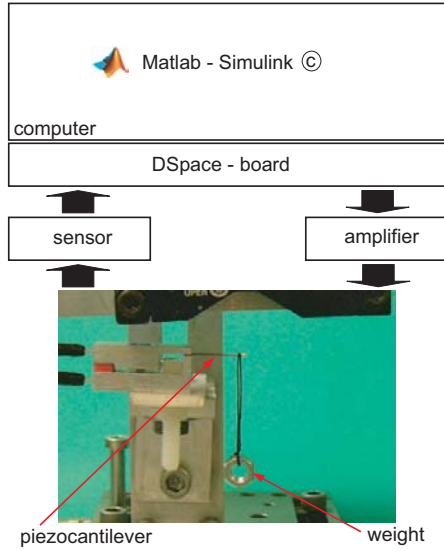


Fig. 2. The setup.

#### B. The state-space model

Let  $\delta$  be the deflection at the tip of the cantilever,  $U$  the applied voltage and  $F = d$  the unknown applied

force. When the applied electrical field is low, the piezo-cantilever behavior is linear (example in [3]). We have:

$$\begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} a_1 & a_2 \\ a_3 & a_4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \alpha \\ 0 \end{bmatrix} U + \begin{bmatrix} \beta \\ 0 \end{bmatrix} F \\ y = \delta = \begin{bmatrix} 0 & c \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \end{cases} \quad (16)$$

where the identified parameters are given by the Table I. The identification process was as follow:

- a step voltage input  $U$  is first applied to the piezo-cantilever,
- afterwards, a mass with known weight  $F$  is hung on at its tip,
- using the input data  $U$  and the read-out  $\delta$ , an ARMAX transfer function model is first identified and then transformed into the state-space model described by (equ 16). The identified model is without the elastic gain  $\beta$ . The latter is identified using the input data  $F$  and the corresponding output  $\delta$ .

In our model, we have  $B = \begin{bmatrix} \beta & 0 \end{bmatrix}^t$ .

TABLE I  
Identified parameters.

parameters	values	parameters	values
$a_1$	-1333.3	$\alpha$	3.856
$a_2$	-162.8	$\beta$	56.8
$a_3$	2048	$c$	10.2
$a_4$	0		

Fig. 3 presents the simulation of the model and the experimental results when a step voltage with 10V of amplitude is applied.

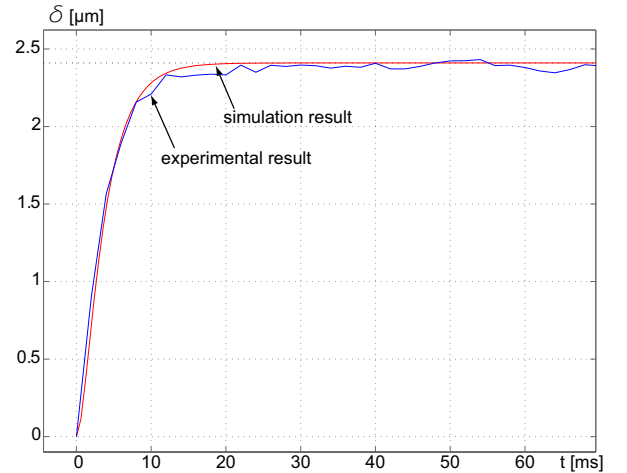


Fig. 3. Step response of the piezocantilever.

#### IV. COMPUTATION OF THE OBSERVER

##### A. Assumptions verification

The necessary conditions to use the Inverse-Dynamics-Based UIO technique are satisfied. Indeed:

(i) comparing the model in (equ 1) and the piezocantilever model in (equ 16), the matrices  $A$ ,  $B$  and  $C$  are known.

(ii)  $\text{rank}(B) = 1$  is full.

(iii)  $\text{rank} \left( \begin{bmatrix} C \\ CA \end{bmatrix} \right) = 2$  is full.

##### B. Computation of the gains

As  $CB = 0$ , the second observer scheme is used in this paper. Consider the *Penrose-Moore* inverse of  $B$ :

$$B^+ = \begin{bmatrix} \frac{1}{\beta} & \rho \end{bmatrix} \quad (17)$$

where any  $\rho \in R$  is convenient. Hence, the matrices  $M_e$  and  $A_e$  can be deduced:

$$M_e = \begin{bmatrix} 0 & \beta(F_2c - \rho) \\ 0 & 0 \end{bmatrix} \quad (18)$$

$$A_e = \begin{bmatrix} -\rho\beta a_3 & -k_1c - \beta F_1c - \rho\beta a_4 \\ a_3 & a_4 - k_2c \end{bmatrix}$$

Because  $M_e$  is singular, we consider its SVD. We have:

$$U = \frac{1}{\sqrt{\beta^2(F_2c - \rho)^2 + 1}} \begin{bmatrix} \beta(F_2c - \rho) & 1 \\ 1 & -\beta(F_2c - \rho) \end{bmatrix}$$

$$\Sigma = \begin{bmatrix} \sqrt{\beta^2(F_2c - \rho)^2 + 1} & 0 \\ 0 & 0 \end{bmatrix}$$

$$V = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \quad (19)$$

So:

$$U^t A_e V = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} = \frac{1}{\sqrt{\beta^2(F_2c - \rho)^2 + 1}} \begin{bmatrix} \theta_{11} & \theta_{12} \\ \theta_{21} & \theta_{22} \end{bmatrix}$$

$$\begin{aligned} \theta_{11} &= -\beta(F_2c - \rho)(k_1 + \beta F_1c + \rho\beta a_4) + a_4 - k_2c \\ \theta_{12} &= -\rho\beta^2 a_3(F_2c - \rho) + a_3 \\ \theta_{21} &= -(k_1 + \beta F_1c + \rho\beta a_4) - \beta(F_2c - \rho)(a_4 - k_2c) \\ \theta_{22} &= -\beta F_2ca_3 \end{aligned} \quad (20)$$

The matrices  $A_{22}$  and  $A_{11} - A_{12}A_{22}^{-1}A_{21}$  are respectively Hurwitz if:

$$\beta F_2ca_3 > 0 \quad (21)$$

and

$$\begin{aligned} &\rho^3\beta^3k_2c + \rho^2\beta^2(\beta F_1c - 2\beta F_2k_2c^2 + k_1) \\ &+ \rho\beta c(\beta^2 F_2^2k_2c^2 - 2\beta^2 F_1F_2c - 2\beta F_2k_1 + k_2) \\ &+ (\beta^3 F_1F_2^2c^3 + \beta^2 F_2^2k_1c^2 + \beta F_1c + k_1) > 0 \end{aligned} \quad (22)$$

In addition to  $K$ ,  $F_1$  and  $F_2$ ,  $\rho$  is an extra design degree of freedom. There is no systematic method to directly choose optimal values of these parameters.

Moreover, there may be gains that satisfy (inequ 22) and (inequ 21) but which derive an instability of the observer. So, a simulation verification (with Simulink ©) is used to enforce their choice Fig. 4. In the figure, the PIEZOCANTILEVER bloc contains the state-space model described by (equ 16), the STATE-observer bloc contains the (equ 2) and the FORCE-observer bloc contains the (equ 3). Because of an algebraic loop problem, we introduce a filter with a large bandwidth in the observer connection.

The simulation is processed as follow. At  $t = 0$ , a step voltage  $U = 10V$  is applied to the piezocantilever. Then, at respectively  $t = 20ms$  and  $t = 40ms$ , a force of  $F = 2mN$  and  $F = 6mN$  are applied. The simulation results (Fig. 5) show that the estimate deflection and force immediately coincide with the output deflection and applied force. Besides the conditions in (inequ 22) and (inequ 21), the gains are chosen such as the convergence (settling) time of the force observer ( $\approx 9ms$ ) is less or equal to the settling time of the piezocantilever ( $\approx 10ms$ ). The vibrations in the estimate force (Fig. 5-b) can be reduced by adjusting the observer gain but the settling time would increase.

We find:

$$\begin{aligned} \rho &= 0.045 & K &= \begin{bmatrix} 47 \\ 0 \end{bmatrix} & F_1 &= 0.335 \\ F_2 &= 8.18 \times 10^{-3} & G_1 &= \begin{bmatrix} -68.69 & -0.542 \end{bmatrix} \\ G_2 &= \begin{bmatrix} 0.0176 & -0.038 \end{bmatrix} \\ G_3 &= \begin{bmatrix} -0.0176 & -0.045 \end{bmatrix} \end{aligned} \quad (23)$$

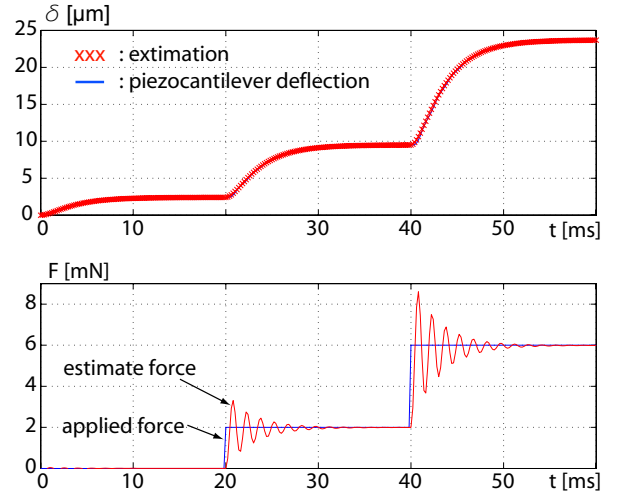


Fig. 5. Simulation results of the force estimation.

#### V. EXPERIMENTS

The observer has been implemented in the computer-DSPACE material. The aim of the experiments is to estimate a weight applied at the piezocantilever tip for different voltages.

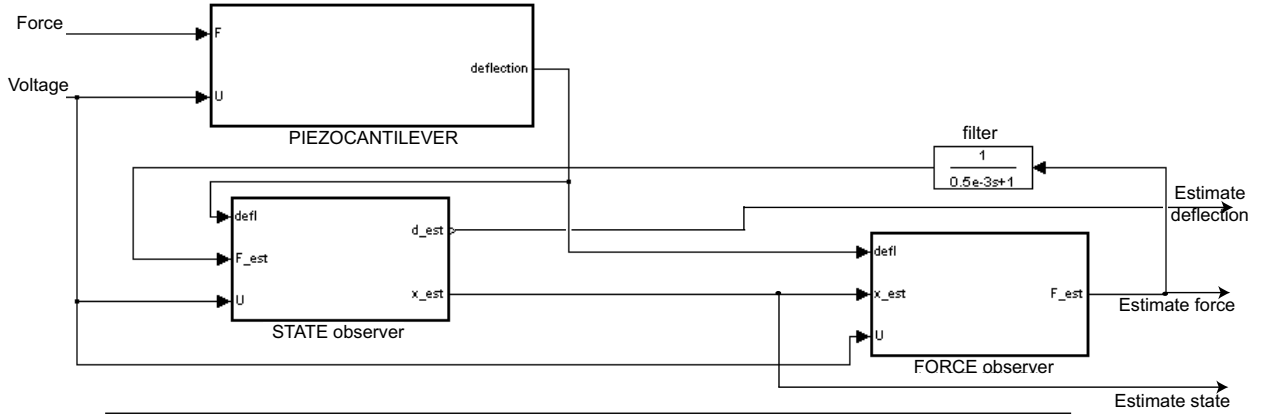


Fig. 4. Simulation of the observer.

### A. Estimation of a null force

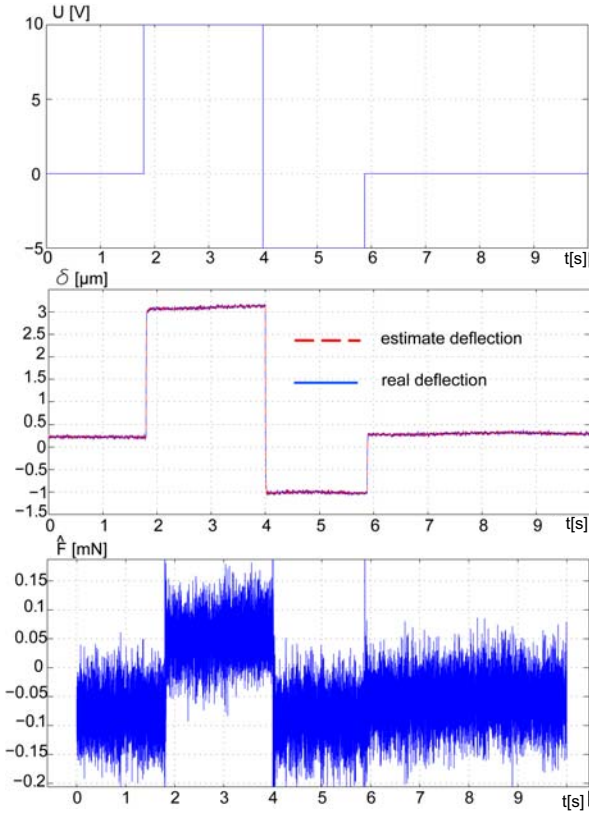


Fig. 6. Estimation of a null force. a: the applied voltage. b: the deflection (real and estimate). c: the estimate force.

First, no weight is put at the tip of the piezocantilever while two electrical steps are applied at different times. The results (Fig. 6) show the results. On the one hand, the estimate deflection and the real deflection well coincide. On the other hand, besides the accuracy of the sensor, the influence of the voltage into the estimate force is insignificant. Indeed, the corresponding accuracy is better than  $0.015 \left[ \frac{mN}{V} \right]$  (computed using the figures

### Fig. 6-a and Fig. 6-c).

### B. Estimation of a non-null force

Next, a weight with  $1.6mN$  amplitude is put at the tip. In the same way, different steps voltage are applied. The results are shown in the Fig. 7. Because of the difficulty to put brusquely the weight, the force signal is not a real step. Hence, there are low dynamic and vibration either in the resulting deflection or in the estimate force. These vibrations can be seen in the deflection and force when putting the mass at nearly  $t = 4s$  (Fig. 7). Despite this conditioning matter, the relative accuracy of the estimate force is interesting:  $6.25\%$  ( $\approx \frac{0.15[mN]}{1.6[mN]}$ ), computed from the Fig. 7-c and Fig. 7-d. Indeed, the provided accuracy is still sub-milliNewton even covering the forces used in micromanipulation/microassembly tasks that are up to a few tens of milliNewton. However, it is clear that the resolution will be limited by the deflection sensor and the stiffness of the piezocantilever.

### C. Estimation of a non-null force with harmonic electrical excitation

The last experiment concerns the use of harmonic electrical input. The aim is to evaluate the invariancy of the estimate force relative to variable input voltage. To perform the experiment, a weight ( $F = 1.6mN$ ) is hung at the tip while a sine voltage is applied. The results are given in Fig. 8. The deflection magnitude is given by  $20 \log \left( \frac{\delta_{amp}}{U_{amp}} \right)$  and the force magnitude is defined as  $20 \log \left( \frac{\hat{F}_{amp}}{F} \right)$ , where the subscript *amp* means amplitude of the sine signal. It clearly appears that the force observer is independent of the frequency of the input voltage, at least up to the concerned bandwidth. The fact that the force magnitude is non exactly equal to  $0dB$  is due to the accuracy as spoken in the previous subsection.

## VI. CONCLUSION

This paper has presented the estimation of the force applied at the tip of a piezocantilever. The final objective

is to observe and control the manipulation force between a microgripper and a micro-object during micromanipulation/microassembly tasks.

In our approach, the force has been considered as an Unknown Input. Thus, we have used an Unknown Input Observation (UIO) technique to estimate it. Indeed, this avoids the use of the characteristics of the manipulated objects which are difficult and time-consuming to obtain. The Inverse-Dynamics-Based UIO technique was especially used because it is well suited for the model of a piezocantilever. The experimental results shown that the performances provided by the proposed estimation is suitable for micromanipulation/microassembly tasks. Among others, the accuracy is sub-milliNewton when the

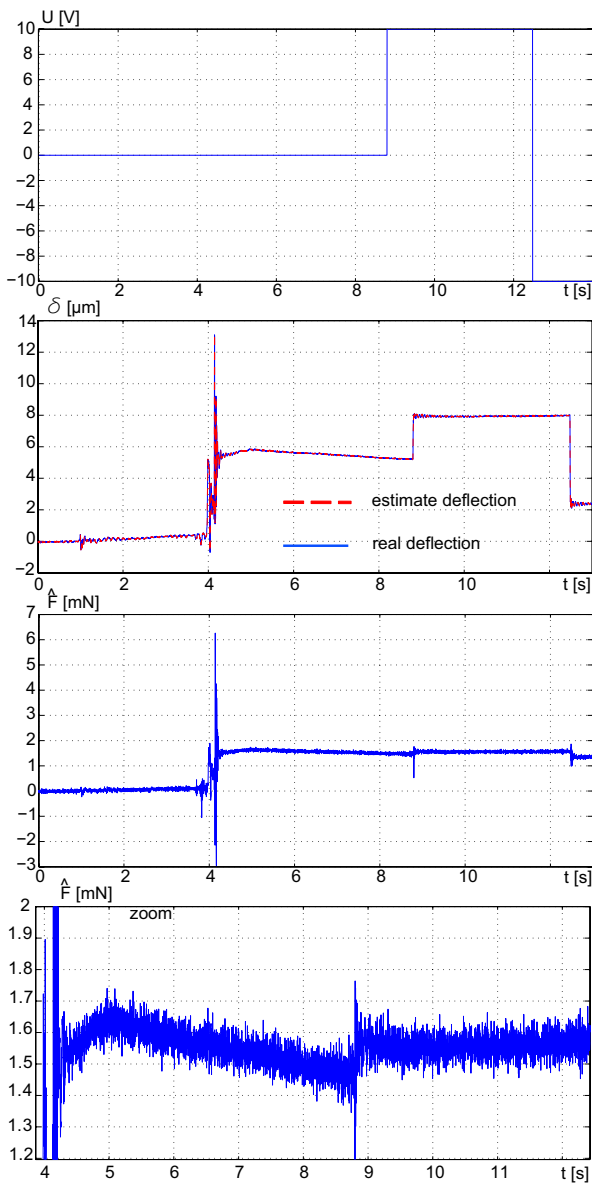


Fig. 7. Estimation of a non-null force ( $1.6mN$ ). a: the applied voltage. b: the deflection (real and estimate). c: the estimate force. d: zoom of the estimate force.

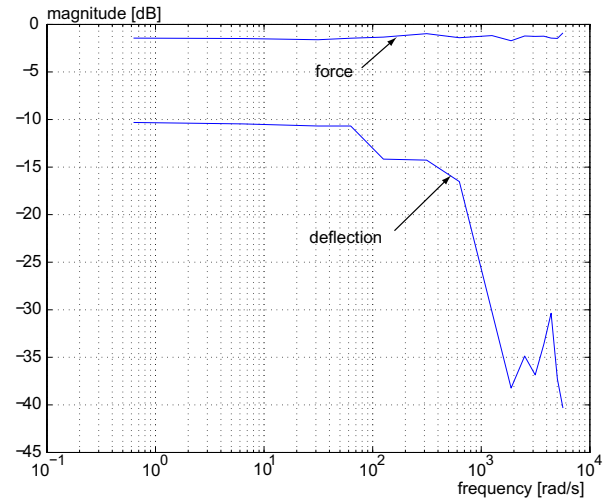


Fig. 8. Experiment with harmonic input voltage.

range of estimation is some tens of milliNewton.

The future works will be the integration of the observer into a complete microgripper and the control of force/position through the measured/estimated signals.

#### ACKNOWLEDGMENT

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