On the Concept of Time-Frequency Distributions Based on Complex-Lag Moments

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Work context: Analysis of Signals with complex Time-Frequency behaviour

Signals issued from diverse physical phenomena:

- Very challenging to analyse in Time-Frequency (TF) domain
- Signals composed of several TF components characterized by various non-linear contents
 - \Rightarrow inner interferences due to the TF non-linearity
 - \Rightarrow cross terms due to the multi-component structures

Outline

Time-Frequency Distribution Based on Complex-Lag Arguments

- 2 Complex-Lag Moment for Multi-Component Signals
- 3 Results



1. Time-Frequency Distribution Based on Complex-Lag Arguments

Concept of complex-lag distributions

• A way for inner interferences reduction with respect of Wigner distribution.



$$WVD(t,\omega) = \mathfrak{F}_{\tau} \left[s(t+\frac{\tau}{2})s^*(t-\frac{\tau}{2}) \right]$$

$$WVD(t,\omega) = \delta \left(\omega - \phi'(t) \right) *_{\omega} \mathfrak{F}_{\tau} \left[e^{jQ_{wv}(t,\tau)} \right]$$

with Spread Factor $Q_{wv}(t,\tau) = \phi^{(3)}(t) \frac{\tau^3}{2^2 \cdot 3!} + \phi^{(5)}(t) \frac{\tau^5}{2^4 \cdot 5!} + \dots$

Lj. Stankovic, 2002: Consideration of lags on imaginary axis



$$CTD(t,\omega) = \mathfrak{F}_{\tau} \left[s(t+\frac{\tau}{4})s^*(t-\frac{\tau}{4})s^{-j}(t+j\frac{\tau}{4})s^j(t-j\frac{\tau}{4}) \right]$$

$$CTD(t,\omega) = \delta \left(\omega - \phi'(t) \right) *_{\omega} \mathfrak{F}_{\tau} \left[e^{jQ_{ct}(t,\tau)} \right]$$

with Spread Factor $Q_{ct}(t,\tau) = \phi^{(5)}(t)\frac{\tau^5}{4^4 \cdot 5!} + \phi^{(9)}(t)\frac{\tau^9}{4^8 \cdot 9!} + \dots$



C. Cornu and all., 2007. "Time-frequency distributions with generalized complex lag argument"

• Cauchy's integral formula \Rightarrow computation of the K^{th} order derivative of ϕ at instant t as

$$\phi^{(K)}(t) = \frac{K!}{2\pi j} \oint_{\gamma} \frac{\phi(z)}{(z-t)^{K+1}} \, dz$$



$$\phi^{(K)}(t) = \frac{K!}{N\tau^K} \sum_{p=0}^{N-1} \phi\left(t + \tau e^{j\frac{2\pi p}{N}}\right) e^{-j\frac{2\pi pK}{N}} + \varepsilon$$

$$GCD_N^K[s](t,\omega) = \mathfrak{F}_{\tau} \left[GCM_N^K[s](t,\tau) \right] \\ = \delta \left(\omega - \phi^{(K)}(t) \right) *_{\omega} \mathfrak{F}_{\tau} \left[e^{jQ(t,\tau)} \right]$$

where
$$\operatorname{GCM}_{N}^{K}[s](t,\tau) = \prod_{p=0}^{N-1} s^{\omega_{N,p}^{N-K}} \left(t + \omega_{N,p} \sqrt[K]{\tau \frac{K!}{N}} \right)$$

with $\omega_{N,p} = e^{j2\pi p/N}$ and $Q(t,\tau) = N \sum_{p=1}^{+\infty} \phi^{(Np+K)}(t) \frac{\tau \frac{Np}{K}+1}{(Np+K)!} \left(\frac{K!}{N}\right)^{\frac{Np}{K}+1}$

"Inner interferences reduction" property of GCD

Increasing **parameter N** ($N \Leftrightarrow$ number of complex lags) leads to a reduction of Spread Factor and so attenuation of inner interferences.

Illustration

• Spread Factors in some Time-Frequency Distributions

Distribution	Spread factor
$STFT \Leftrightarrow \operatorname{GCD}_1^1$	$Q(t,\tau) = \phi^{(2)}(t)\frac{\tau^2}{2!} + \phi^{(3)}(t)\frac{\tau^3}{3!} + \dots$
$WVD \Leftrightarrow \operatorname{GCD}_2^1$	$Q(t,\tau) = \phi^{(3)}(t)\frac{\tau^3}{2^2 3!} + \phi^{(5)}(t)\frac{\tau^5}{2^4 5!} + \dots$
$CTD \Leftrightarrow \operatorname{GCD}_4^1$	$Q(t,\tau) = \phi^{(5)}(t)\frac{\tau^5}{4^4 5!} + \phi^{(9)}(t)\frac{\tau^9}{4^8 9!} + \dots$
GCD_6^1	$Q(t,\tau) = \phi^{(7)}(t)\frac{\tau^{7}}{6^{6}7!} + \phi^{(13)}(t)\frac{\tau^{13}}{6^{12}13!} + \dots$

• Example for a test signal $s(t) = e^{j(6\cos(\pi t) + \frac{2}{3}\cos(3\pi t) + \frac{4}{3}\cos(5\pi t))}$



$GCD \Leftrightarrow Time-Phase Derivative Representation$

- GCD offers representations for derivatives of any order ${\bf K}$ of the instantaneous phase law of the signal.
- Example for a 4^{th} order polynomial phase signal



The theoretical derivatives (top) are correctly represented by GCDs of corresponding order (bottom)

2. Complex-Lag Moment for Multi-Component Signals

Consideration of Cross Terms

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- The GCM has a complicate non-linear form
- Multi-component case ⇒ high level of cross terms very difficult to calculate, as follows:

or
$$s(t) = s_1(t) + s_2(t) = e^{j\phi_1(t)} + e^{j\phi_2(t)}$$

$$GCM_N^K [s_1 + s_2](t, \tau)$$

$$= \prod_{p=0}^{N-1} \left[s_1 \left(t + \omega_{N,p} \sqrt[K]{\tau \frac{K!}{N}} \right) + s_2 \left(t + \omega_{N,p} \sqrt[K]{\tau \frac{K!}{N}} \right) \right]^{\omega_{N,p}^{N-K}}$$

$$= e^{j\phi_1^{(K)}(t)\tau + jQ_1(t,\tau)} + e^{j\phi_2^{(K)}(t)\tau + jQ_2(t,\tau)} + CT_{\phi_1,\phi_2}(N,K)$$



Case of a two-component signal: cross terms have strong energetic level and corrupt the visibility of auto-terms

Reduction of Cross Terms: "Multi-lag sets" principle

• Auto-terms **don't** depend of number of lags N



- First step: Compute GCMs for several lags sets N_i \Rightarrow auto terms keep the same structures \Rightarrow cross terms depend of N and are consequently differently located
- Second step: Sum the different GCMs

 $mlsGCM_{\{N_i\}}^{K}\left[s\right]\left(t,\tau\right)$

$$= \sum_{i=1}^{P} GCM_{N_{i}}^{K}[s](t,\tau)$$

= $P \sum_{l} e^{j\phi_{l}^{(K)}(t)\tau + jQ_{l}(t,\tau)} + \sum_{i=1}^{P} CT_{\phi_{l}}(N_{i},K)$

Highlighting of auto-terms / Decreasing of cross-terms



Case of a two-component signal: cross terms level reduction using mlsGCD

3. Results

GCD and mlsGCD can perform in many applications dealing with complex TF modulations.

Ex.1: Removal of quasi-stationary corrupting components



Separation of a Sinusoidal Frequency Modulation corrupted by chirp, tone and noise.

Ex.2: Application to Transient signals

- Infinite derivability of transient signals phase law
- Extraction, via derivability property, of transient impulses corrupted by both noise and coherent perturbations



Transient impulses extraction



Noise reduction by mlsGCD

4. Conclusion

We achieved...

Distribution able to:

- focus on arbitrary derivate of phase laws
- in multi-component context
- **③** with highly concentrated and cross-terms reduced representations

Prospects

- Signal-dependent choice of lag sets
- 2D extension of complex lag concept

End

Thank you for your Attention!!!